ANALOG DEVICES

Circuit Note
CN-0179

Circuits from the Lab® Reference Designs

Devices Connected/Referenced

AD8657	18 V, Precision, Micropower CMOS Rail-to-Rail I/O Dual Operational Amplifier
ADR02	Ultracompact Precision 5 V Voltage Reference
AD5641	2.7 V to 5.5 V, <100 µA, 14-Bit *nano*DAC, SPI Interface

4-20 mA Low Power, 14-Bit, Process Control Current Loop Transmitter

EVALUATION AND DESIGN SUPPORT

Circuit Evaluation Boards
   CN0179 Circuit Evaluation Board (EVAL-CN0179-PMDZ)
   System Demonstration Platform (EVAL-SDP-CB1Z)
   SDP Interposer Board (SDP-PMD-1B1Z)
Design and Integration Files
   Schematics, Layout Files, Bill of Materials

CIRCUIT FUNCTION AND BENEFITS

The circuit in Figure 1 is a 4 mA-to-20 mA current loop transmitter for communication between a process control system and its actuator. Besides being cost effective, this circuit offers the industry's low power solution. The 4 mA-to-20 mA current loop has been used extensively in programmable logic controllers (PLCs) and distributed control systems (DCS's), with digital or analog inputs and outputs. Current loop interfaces are usually preferred because they offer the most cost effective approach to long distance noise immune data transmission. The combination of the low power AD8657 dual op amp, AD5641 DAC, and ADR02 reference allows more power

budget for higher power devices, such as microcontrollers and digital isolators. The circuit output is 0 mA to 20 mA of current, and it operates on a single supply from 8 V to 18 V. The 4 mA to 20 mA range is usually mapped to represent the input control range from the DAC or micro-controller, while the output current range of 0 mA to 4 mA is often used to diagnose fault conditions.

The 14-bit, 5 V AD5641 requires 75 µA typical supply current. The AD8657 is a rail-to-rail input/output dual op amp and is one of the lowest power amplifiers currently available in the industry (22 µA per amplifier over the full supply voltage and input common-mode range) with high operating voltage of up to 18 V. The ADR02 ultracompact precision 5 V voltage reference requires only 650 µA. Together, these three devices consume a typical supply current of 747 µA.

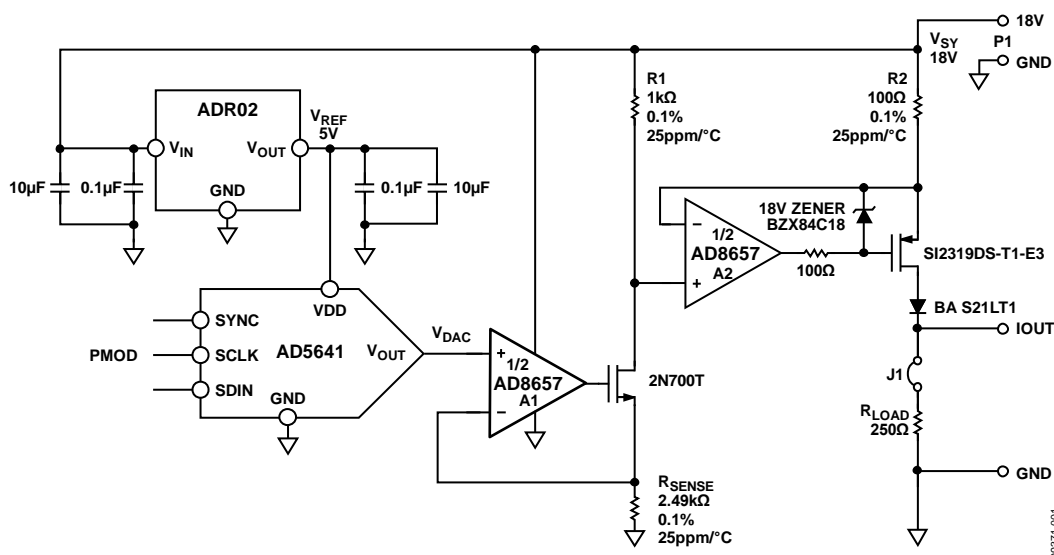The circuit has a 12-pin Pmod™ digital interface (Digilent specification).

Figure 1. Low Power 4 mA-to-20 mA Process Control Current Loop (Simplified Schematic: All Connections and Decoupling Not Shown)

Rev. A

CIRCUIT DESCRIPTION

For industrial and process control modules, 4 mA-to-20 mA current loop transmitters are used as a means of communication between the control unit and the actuator. Located at the control unit, the 14-bit AD5641 DAC produces an output voltage, V_{DAC}, between 0 V and 5 V as a function of the input code. The code is set via an SPI interface. The ideal relationship between the input code and output voltage is given by

$$V_{DAC} = V_{REF} \times (D/2^{14}) \tag{1}$$

where:

V_{REF} is the output of ADR02 and the power supply to the AD5641.
D is the decimal equivalent of the binary code that is loaded to the AD5641.

The DAC output voltage sets the current flowing through the sense resistor, R_{SENSE}, where

$$I_{SENSE} = V_{DAC}/R_{SENSE} \tag{2}$$

The current through R_{SENSE} varies from 0 mA to 2 mA as a function of V_{DAC}. This current develops a voltage across R1 and sets the voltage at the noninverting input of the AD8657 amplifier (A2). The A2 AD8657 closes the loop and brings the inverting input voltage to the same voltage as the noninverting input. Therefore, the current flowing through R1 is mirrored by a factor of 10 to R2. This is represented by Equation 3.

$$I_{OUT} = I_{R2} = (V_{DAC}/R_{SENSE}) \times (R1/R2) \tag{3}$$

With V_{DAC} ranging from 0 V to 5 V, the circuit generates a current output from 0 mA to 20 mA.

The AD5641 is a 14-bit DAC from the *nano*DAC family and operates from the 5 V output voltage of the ADR02 reference. It has an on-chip precision output buffer that is capable of swinging from rail-to-rail (within 10 mV), thus allowing a high dynamic output range. With a supply voltage of 5 V, AD5641 consumes a typical 75 µA of supply current.

In addition, this circuit solution requires a rail-to-rail input amplifier. The AD8657 dual op amp is an excellent choice, with low power and rail-to-rail features. The op amp operates with a typical supply current of 22 µA/amplifier over the specified supply voltage and input common-mode voltage. It also offers excellent noise and bandwidth per unit of current. The AD8657 is one of the lowest power amplifiers that operate on supplies of up to 18 V.

The ADR02 is an ultracompact, precision 5 V voltage reference. With an 18 V input voltage, quiescent current is only 650 µA, typical. It has an initial accuracy of 0.06% (B-grade) and 10 µV p-p voltage noise. Connecting a 0.1 µF ceramic capacitor to the output is highly recommended to improve stability and filter out low level voltage noise. An additional 1 µF to 10 µF electrolytic, tantalum, or ceramic capacitor in parallel can improve load transient response. A 1 µF to 10 µF electrolytic, tantalum or ceramic capacitor can also be connected to the input to improve transient response in applications where the supply voltage may fluctuate. An additional 0.1 µF ceramic capacitor should be connected in parallel to reduce supply noise.

Bypass capacitors (not shown in Figure 1) are required. In this case, a 10 µF tantalum capacitor in parallel with a 0.1 µF ceramic capacitor should be placed on each power pin of each dual op amp. Details of proper decoupling techniques can be found in Tutorial MT-101.

Figure 2 shows the linearity of the system, that is the measured output current from the circuit DAC input code from 0 to full-scale.
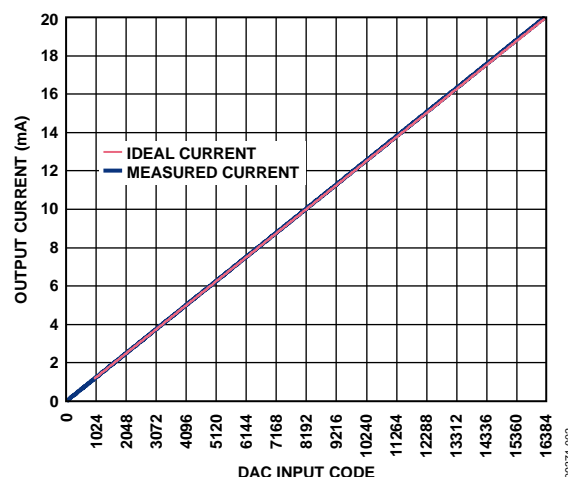

Figure 2. 0 mA to 20 mA Output Current

Figure 3 shows the output current error plot in percent full-scale range. The overall worst-case error is approximately 0.35% measured over the output range between Code 256 and Code 16,128.
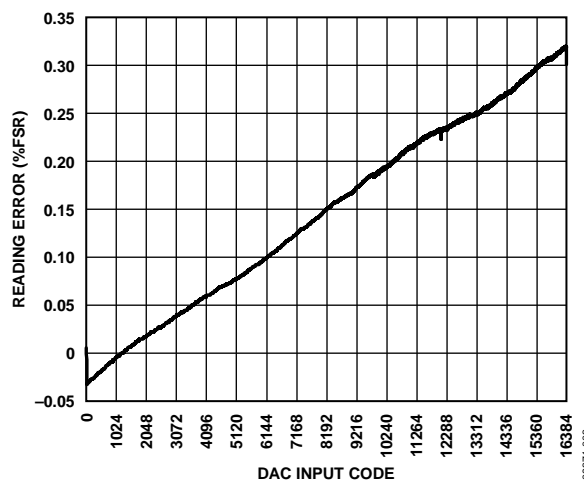

Figure 3. Output Current Error Plot

Figure 4 shows the calibrated output current error plot. Removing the gain and offset error from Figure 3, the accuracy is better than 0.05% measured over the output range between Code 256 and Code 16,128.
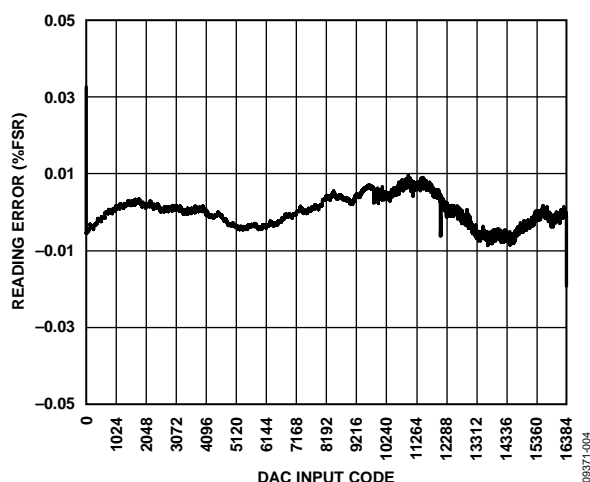

Figure 4. Calibrated Output Current Error Plot

The data in Figure 3 and Figure 4 shows larger errors at zero and full-scale because the output buffer of the AD5641 DAC limits when its output is within 10 mV of either supply rail. The region between Code 0 and Code 255 as well as the region between Code 16,129 and Code 16,384 are therefore excluded from the linearity specifications. This corresponds to approximately 0 V to 80 mV and 4.92 V to 5.00 V at the DAC voltage output; and 0 mA to 0.32 mA and 19.68 mA to 20.00 mA referenced to the current output.

The test data was taken using the board shown in Figure 6. Complete documentation for the system can be found in the CN-0179 Design Support package.

COMMON VARIATIONS

For a 16-bit resolution solution, consider the AD5660 or AD5662, respectively. The 16 V CMOS ADA4665-2 op amp is another option to replace the AD8657. It lower cost and has lower voltage noise at the expense of a higher supply current.

When selecting amplifiers for this application, always ensure that the input common-mode voltage range and the supply voltage are not exceeded.

CIRCUIT EVALUATION AND TEST

This circuit uses the EVAL-CN0179-PMDZ circuit board, the EVAL-SDP-CB1Z system demonstration platform (SDP) evaluation board and the SDP-PMD-IB1Z, a Pmod interposer board for the EVAL-SDP-CB1Z. The SDP and the SDP-PMD-IB1Z boards have 120-pin mating connectors, allowing the quick setup and evaluation of the circuit's performance. In order to evaluate the EVAL-CN0179-PMDZ board using the SDP-PMD-IB1Z and the SDP, the EVAL-CN0179-PMDZ is connected to the SDP-PMD-IB1Z by a standard 100 mil-spaced, 25 mil square, right angle 12 pin-Pmod header connector.

Information and details regarding how to use the evaluation software for data capturing and proper hardware installation can be found in the CN0179 Software User Guide.

Equipment Required

- PC with a USB port and Windows® XP, Windows® Vista (32-bit), or Windows® 7 (32-bit)
- EVAL-CN0179-PMDZ circuit evaluation board
- EVAL-SDP-CB1Z SDP evaluation board
- SDP-PMD-IB1Z
- CN0179 evaluation software
- Agilent E36311A dual dc power supply or equivalent
- Agilent 3458A multimeter or equivalent
- +6 V wall wart
- A GPIB-to-USB cable adapter (only required for capturing analog data from the output and transferring it to the PC)

Information and details regarding how to use the evaluation software for data capturing and proper hardware installation can be found in the CN0179 Software User Guide.

Test Setup and Measurements

The circuit was tested using the test setup in Figure 5. A photograph of the board is shown in Figure 6.

A jumper should not be connected to the J1 terminals when driving an external current loop. The jumper connects the internal 250 Ω load and should be used when making voltage measurements.
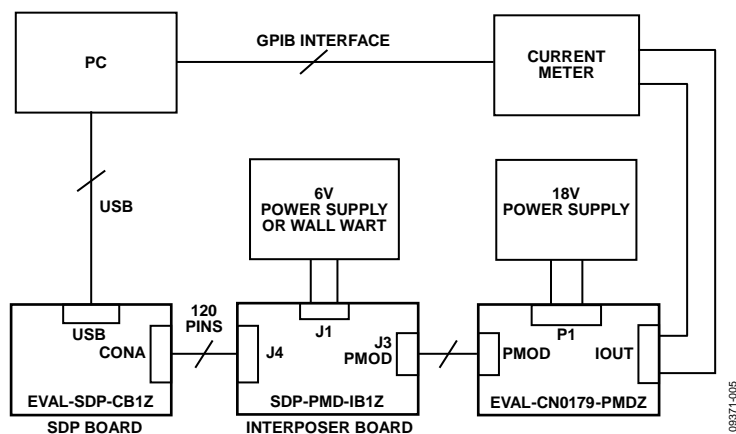
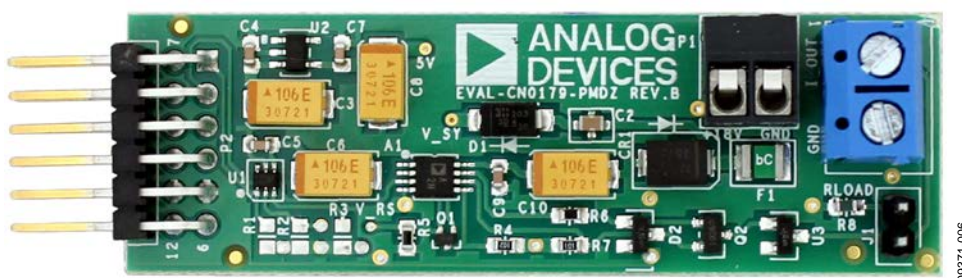*Figure 5. Functional Diagram of Test Setup*

Figure 6. Photo of EVAL-CN0179-PMDZ Board

LEARN MORE

CN0179 Design Support Package:
http://www.analog.com/CN0179-DesignSupport

AN-345 Application Note, *Grounding for Low- and High-Frequency Circuits*, Analog Devices.

AN-347 Application Note, *Shielding and Guarding: How to Exclude Interference-Type Noise,* Analog Devices.

Colm Slattery, Derrick Hartmann, and Li Ke, "PLC Evaluation Board Simplifies Design of Industrial Process Control Systems," *Analog Dialogue* (April 2009).

Jung, Walt. *Op Amp Applications*, Analog Devices. Also available as *Op Amp Applications Handbook*, Elsevier.

Kester, Walt. 2005. *The Data Conversion Handbook*. Chapters 3 and 7. Analog Devices.

MT-015 Tutorial, *Basic DAC Architectures II: Binary DACs*. Analog Devices.

MT-031 Tutorial, *Grounding Data Converters and Solving the Mystery of "AGND" and "DGND."* Analog Devices.

MT-035, *Op Amp Inputs, Outputs, Single-Supply, and Rail-to-Rail Issues*, Analog Devices.

MT-101 Tutorial, *Decoupling Techniques*. Analog Devices.

Data Sheets

AD8657 Data Sheet

ADR02 Data Sheet

AD5641 Data Sheet

AD5662 Data Sheet

AD5660 Data Sheet

ADA4665-2 Data Sheet

REVISION HISTORY

3/14—Rev. 0 to Rev. A

11/10—Revision 0: Initial Version

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