

FEATURES

Conversion loss: 7 dB

LO to RF and IF isolation: 40 dB

Input IP3: 17 dBm

RoHS compliant, 24-terminal, 4 mm × 4 mm LCC package

APPLICATIONS

Microwave and very small aperture terminal (VSAT) radios

Test equipment

Military electronic warfare (EW); electronic countermeasure (ECM); and command, control, communications and intelligence (C3I)

GENERAL DESCRIPTION

The HMC129ALC4 is a general-purpose, double-balanced monolithic microwave integrated circuit (MMIC) mixer housed in a leadless Pb-free, RoHS compliant LCC package, that can be used as an upconverter or downconverter in the 4 GHz to 8 GHz band. The HMC129ALC4 is ideally suited for applications where small size, no dc bias, and consistent IC performance are required. This mixer can operate over a wide LO drive input of 9 dBm to

18 dBm. It performs equally well as a biphas modulator or demodulator. The HMC129ALC4 eliminates the need for wire bonding, allowing use of surface-mount manufacturing techniques.

FUNCTIONAL BLOCK DIAGRAM

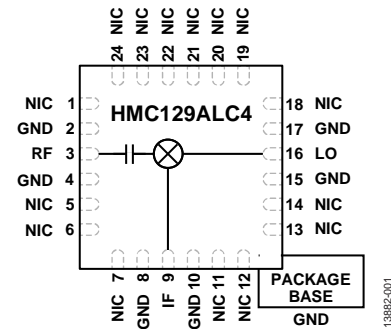


Figure 1.

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REVISION HISTORY

7/2017—Revision 0: Initial Version

SPECIFICATIONS

$T_A = 25^\circ\text{C}$, IF = 100 MHz, LO = 15 dBm, upper side band. All measurements performed as a downconverter, unless otherwise noted, on the evaluation printed circuit board (PCB).

Table 1.

| Parameter | Min | Typ | Max | Unit |
|---|-----|-----|-----|------|
| FREQUENCY | | | | |
| RF Pin | 4 | | 8 | GHz |
| IF Pin | DC | | 3 | GHz |
| LO Pin | 4 | | 8 | GHz |
| LO DRIVE LEVEL | 9 | 15 | 18 | dBm |
| RADIO FREQUENCY (RF) PERFORMANCE | | | | |
| Downconverter | | | | |
| Conversion Loss | | 7 | 9 | dB |
| Single Sideband (SSB) Noise Figure | | 7 | | dB |
| Input Third-Order Intercept (IIP3) | 15 | 17 | | dBm |
| Input 1 dB Compression Point (IP1dB) | | 10 | | dBm |
| Input Second-Order Intercept (IIP2) | | 50 | | dBm |
| RF to IF Isolation | | 20 | | dB |
| LO to RF Isolation | | 40 | | dB |
| LO to IF Isolation | 35 | 40 | | dB |
| Upconverter | | | | |
| Conversion Loss | | 7 | | dB |
| Input Third-Order Intercept (IIP3) | | 17 | | dBm |
| Input 1 dB Compression Point (IP1dB) | | 7 | | dBm |

ABSOLUTE MAXIMUM RATINGS

Table 2.

| Parameter | Rating |
|---|-----------------|
| RF Input Power | 25 dBm |
| LO Input Power | 27 dBm |
| IF Input Power | 25 dBm |
| IF Source/Sink Current | 6 mA |
| Reflow Temperature | 260°C |
| Maximum Junction Temperature | 175°C |
| Continuous Power Dissipation, P_{DISS} ($T_A = 85^\circ\text{C}$, Derate 5 mW/°C Above 85°C) | 450 mW |
| Operating Temperature Range | -40°C to +85°C |
| Storage Temperature Range | -65°C to +150°C |
| Electrostatic Discharge (ESD) Sensitivity | |
| Human Body Model (HBM) | 250 V |
| Field Induced Charged Device Model (FICDM) | 500 V |

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

THERMAL RESISTANCE

Thermal performance is directly linked to PCB design and operating environment. Careful attention to PCB thermal design is required.

θ_{JA} is the natural convection junction to ambient thermal resistance measured in a one cubic foot sealed enclosure. θ_{JC} is the junction to case thermal resistance.

Table 3. Thermal Resistance

| Package Type | θ_{JA} | θ_{JC} | Unit |
|---------------------|---------------|---------------|------|
| E-24-1 ¹ | 120 | 200 | °C/W |

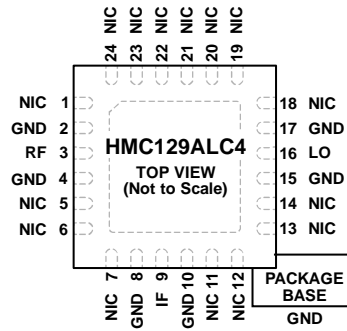
¹ Test Condition 1: JEDEC standard JESD51-2.

ESD CAUTION



ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS



NOTES
 1. NIC = NOT INTERNALLY CONNECTED. THESE PINS CAN BE CONNECTED TO RF/DC GROUND. PERFORMANCE IS NOT AFFECTED.
 2. EXPOSED PAD. THE EXPOSED PAD MUST BE CONNECTED TO RF/DC GROUND.

Figure 2. Pin Configuration

Table 4. Pin Function Descriptions

| Pin No. | Mnemonic | Description |
|--------------------------------|------------|--|
| 1, 5, 6, 7, 11 to 14, 18 to 24 | NIC | Not Internally Connected. These pins can be connected to RF/dc ground. Performance is not affected. |
| 2, 4, 8, 10, 15, 17 | GND | Ground. These pins and package bottom must be connect to RF/dc ground. |
| 3 | RF | Radio Frequency Port. This pin is ac-coupled and matched to 50 Ω. |
| 9 | IF | Intermediate Frequency Port. This pin is dc-coupled. For applications not requiring operation to dc, dc block this port externally using a series capacitor of a value chosen to pass the necessary IF frequency range. For operation to dc, this pin must not source/sink more than 6 mA of current or die malfunction and possible die failure may result. |
| 16 | LO EPAD | Local Oscillator Port. This pin is dc-coupled and matched to 50 Ω. Exposed Pad. The exposed pad must be connected to RF/dc ground. |

INTERFACE SCHEMATICS



Figure 3. GND Interface Schematic

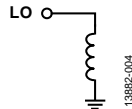


Figure 4. LO Interface Schematic

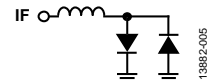


Figure 5. IF Interface Schematic

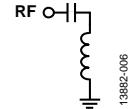


Figure 6. RF Interface Schematic

TYPICAL PERFORMANCE CHARACTERISTICS

DOWNCONVERTER PERFORMANCE

Downconverter performance at IF = 100 MHz, upper sideband (low-side LO).

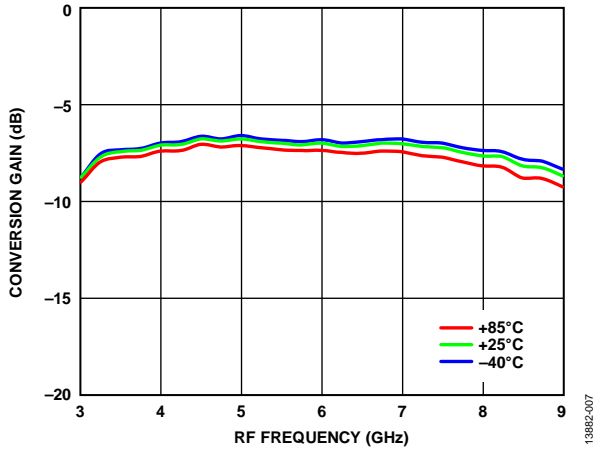


Figure 7. Conversion Gain vs. RF Frequency at Various Temperatures, LO = 15 dBm

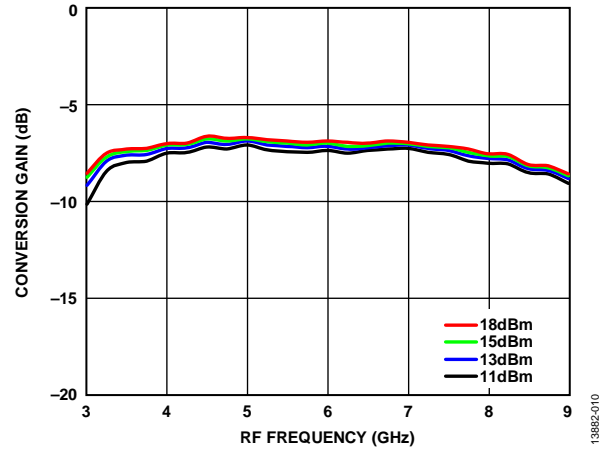


Figure 10. Conversion Gain vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

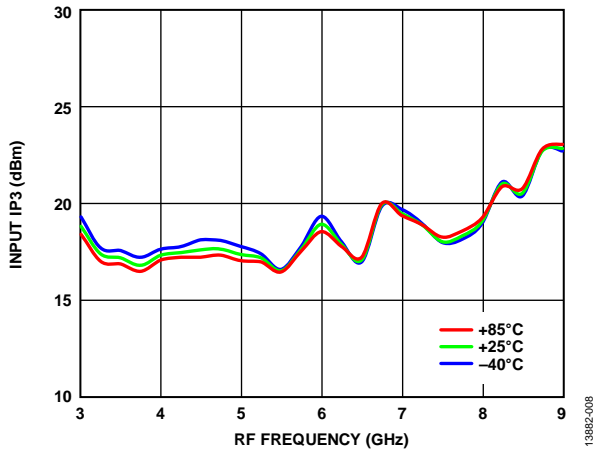


Figure 8. Input IP3 vs. RF Frequency at Various Temperatures, LO = 15 dBm

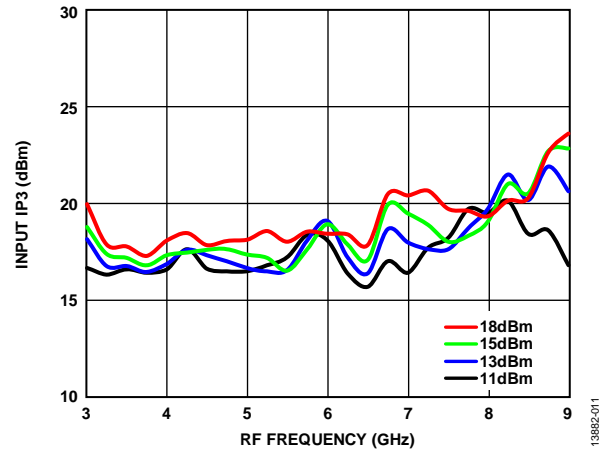


Figure 11. Input IP3 vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

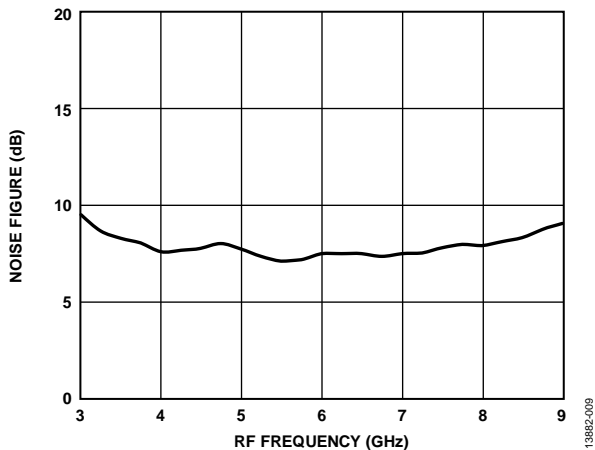


Figure 9. Noise Figure vs. RF Frequency at 25°C , LO = 15 dBm

Downconverter P1dB and IP2

IF = 100 MHz, upper sideband (low-side LO).

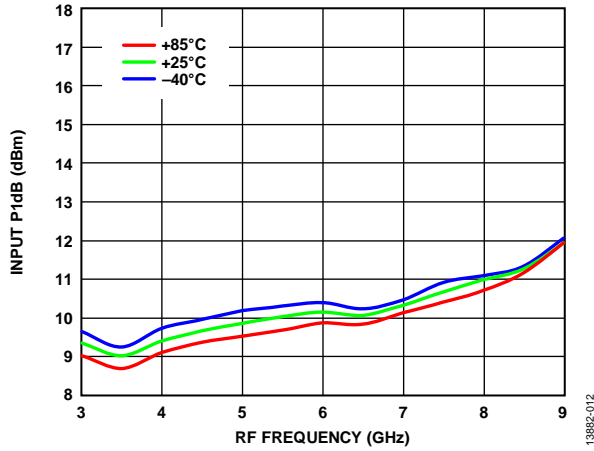


Figure 12. Input P1dB vs. RF Frequency at Various Temperatures, LO = 15 dBm

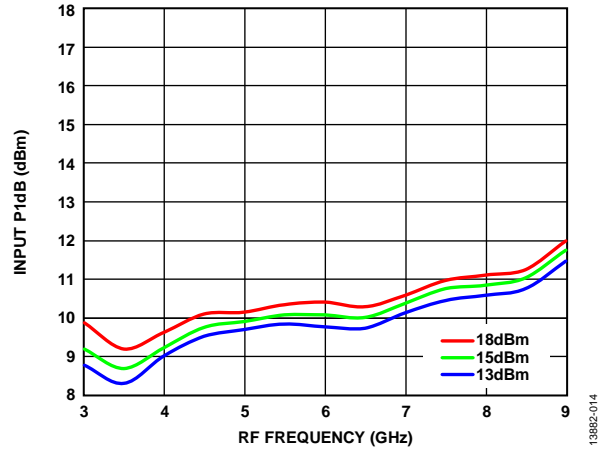


Figure 14. Input P1dB vs. RF Frequency at Various LO Power Levels, TA = 25°C

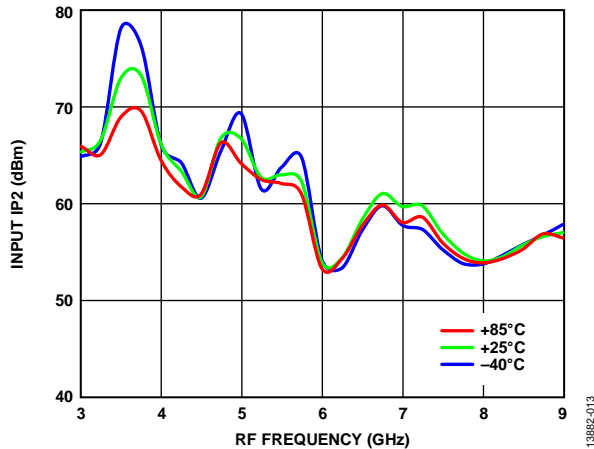


Figure 13. Input IP2 vs. RF Frequency at Various Temperatures, LO = 15 dBm

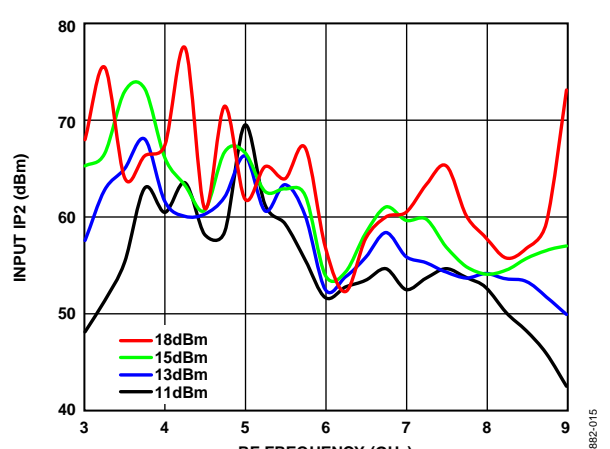


Figure 15. Input IP2 vs. RF Frequency at Various LO Power Levels, TA = 25°C

UPCONVERTER PERFORMANCE

Upconverter performance at IF = 100 MHz, upper sideband.

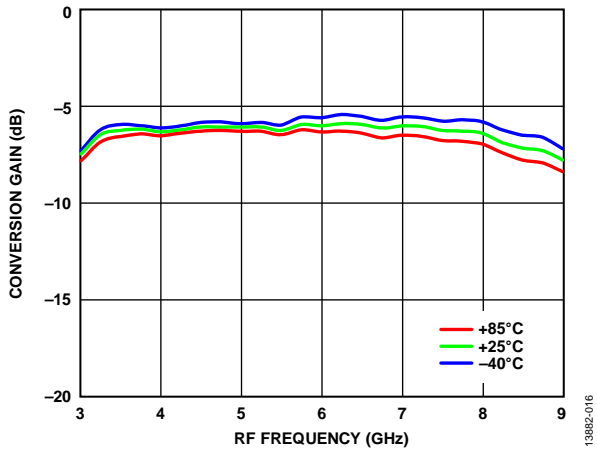


Figure 16. Conversion Gain vs. RF Frequency at Various Temperatures, LO = 15 dBm

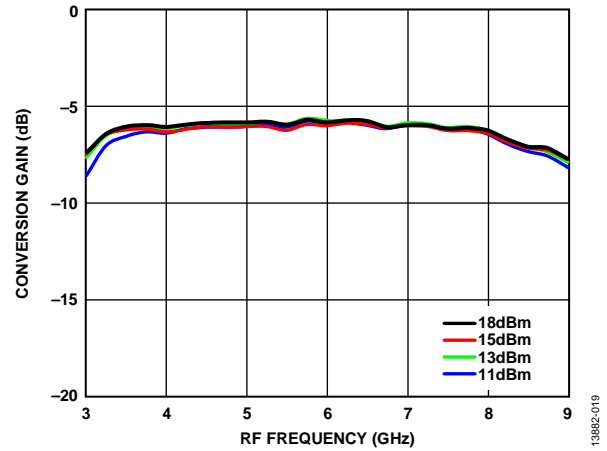


Figure 19. Conversion Gain vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

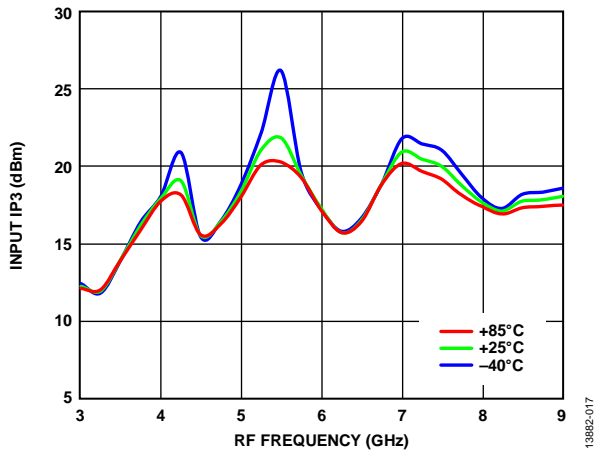


Figure 17. Input IP3 vs. RF Frequency at Various Temperatures, LO = 15 dBm

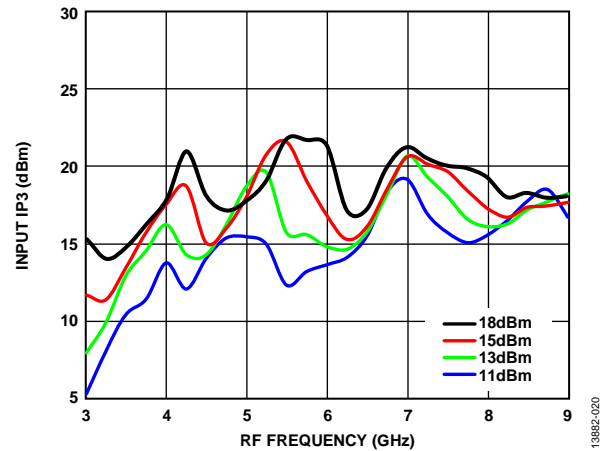


Figure 20. Input IP3 vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

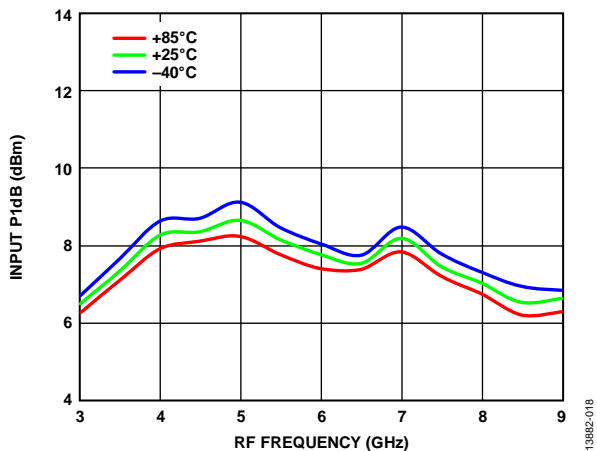


Figure 18. Input P1dB vs. RF Frequency at Various Temperatures, LO = 15 dBm

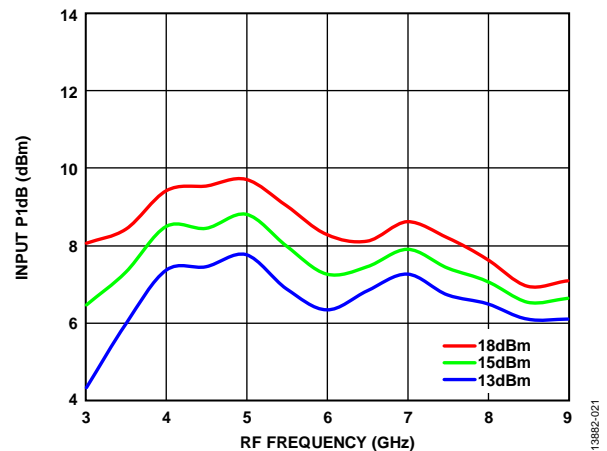


Figure 21. Input P1dB vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

ISOLATION AND RETURN LOSS

Downconverter performance at IF = 100 MHz, upper sideband.

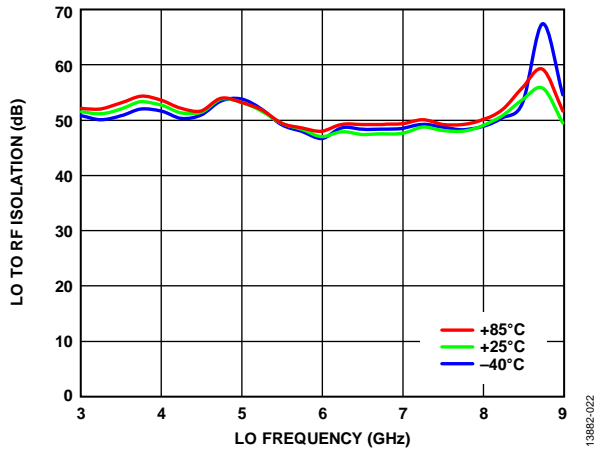


Figure 22. LO to RF Isolation vs. LO Frequency at Various Temperatures, LO = 15 dBm

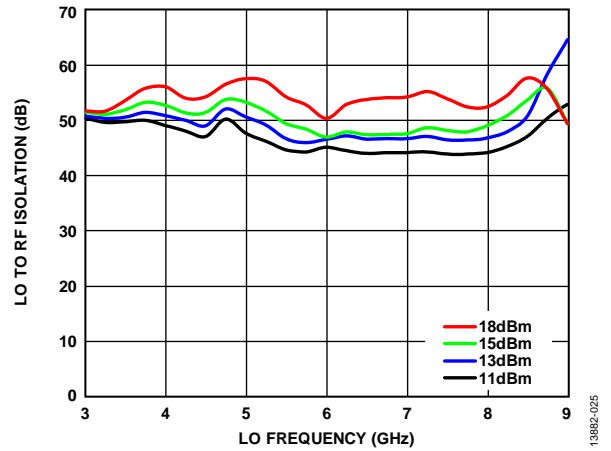


Figure 25. LO to RF Isolation vs. LO Frequency at Various LO Power levels, $T_A = 25^\circ\text{C}$

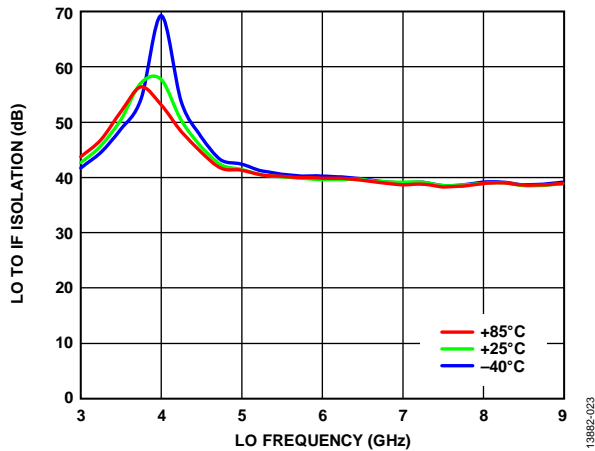


Figure 23. LO to IF Isolation vs. LO Frequency at Various Temperatures, LO = 15 dBm

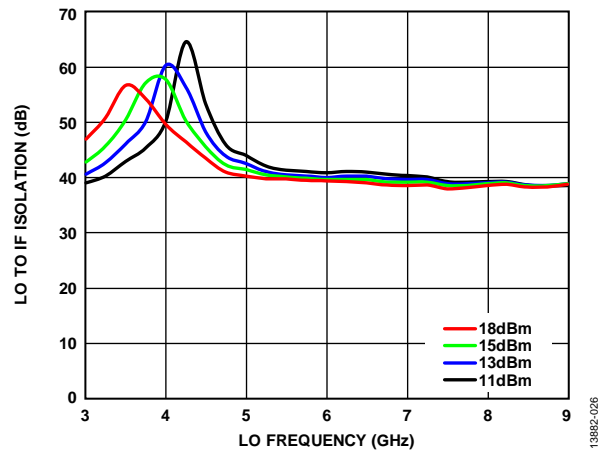


Figure 26. LO to IF Isolation vs. LO Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

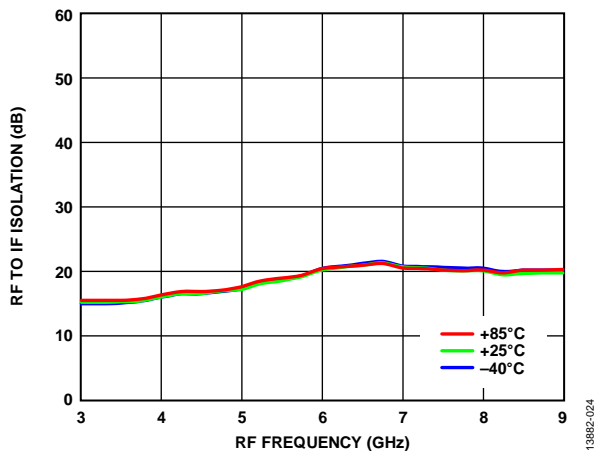


Figure 24. RF to IF Isolation vs. RF Frequency at Various Temperatures, LO = 15 dBm

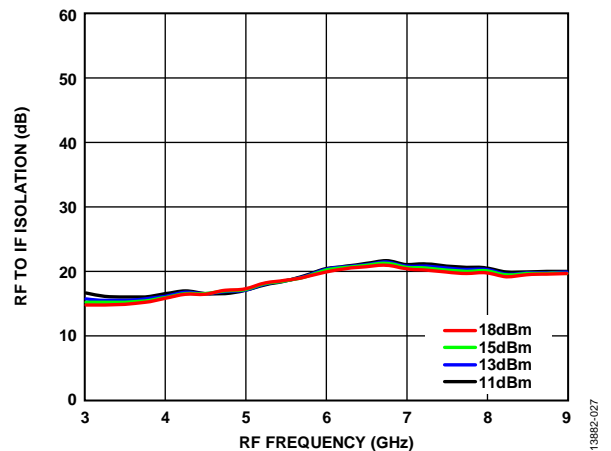


Figure 27. RF to IF Isolation vs. RF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

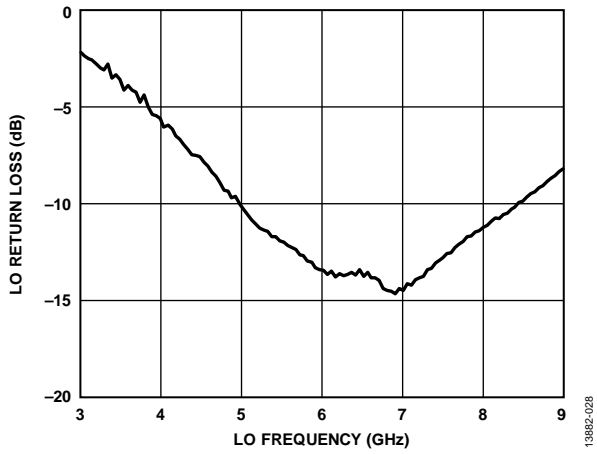


Figure 28. LO Return Loss vs. LO Frequency at 25°C, LO = 15 dBm

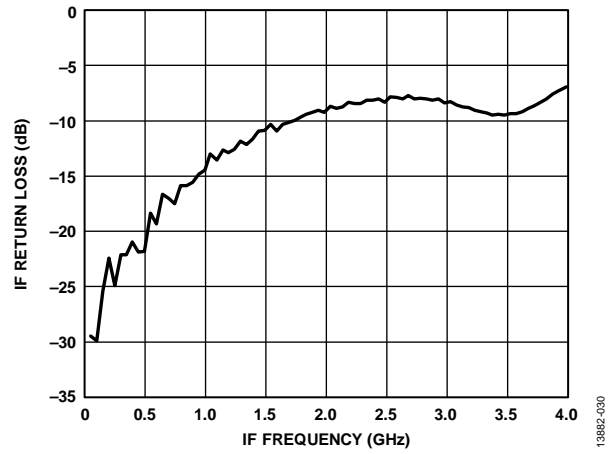


Figure 30. IF Return Loss vs. IF Frequency at 25°C, LO = 6 GHz, 15 dBm

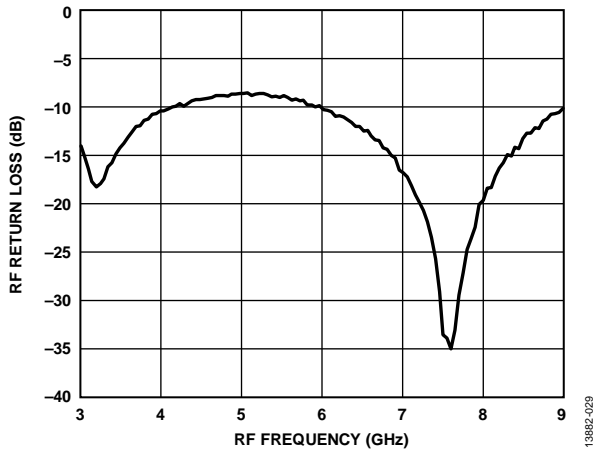


Figure 29. RF Return Loss vs. RF Frequency at 25°C, LO = 6 GHz, 15 dBm

IF BANDWIDTH—DOWNCONVERTER

Upper sideband, LO frequency = 6 GHz

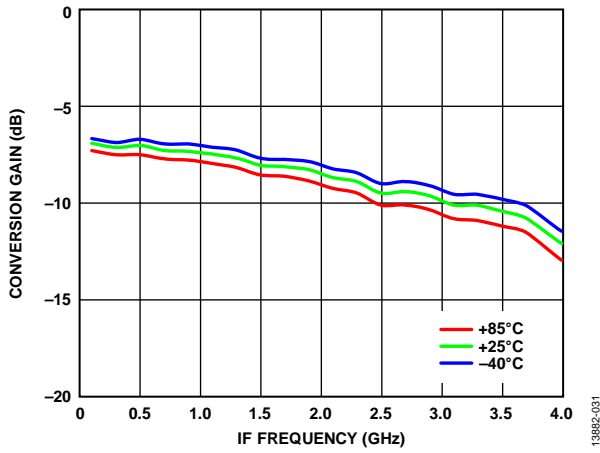


Figure 31. Conversion Gain vs. IF Frequency at Various Temperatures, LO = 15 dBm

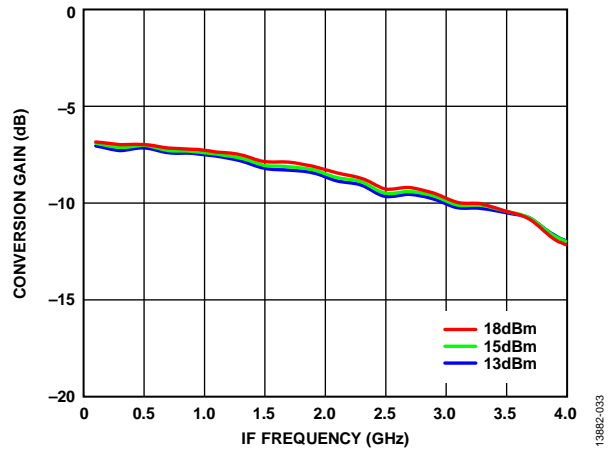


Figure 33. Conversion Gain vs. IF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

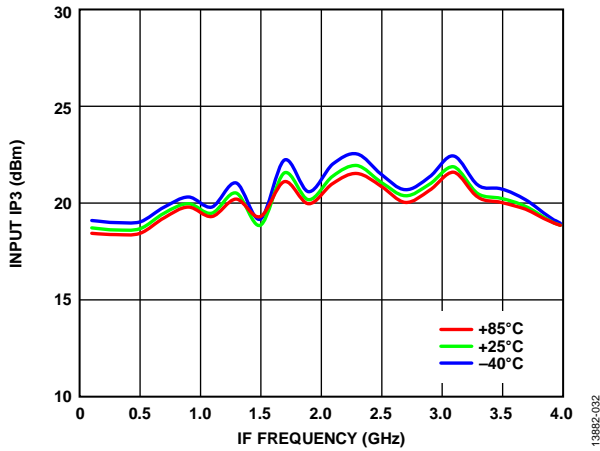


Figure 32. Input IP3 vs. IF Frequency at Various Temperatures, LO = 15 dBm

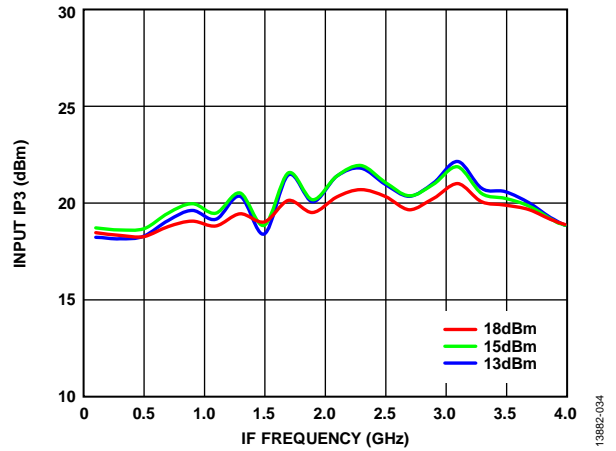


Figure 34. Input IP3 vs. IF Frequency at Various LO Power Levels, $T_A = 25^\circ\text{C}$

IF BANDWIDTH—UPCONVERTER

Upper sideband, LO frequency = 6 GHz.

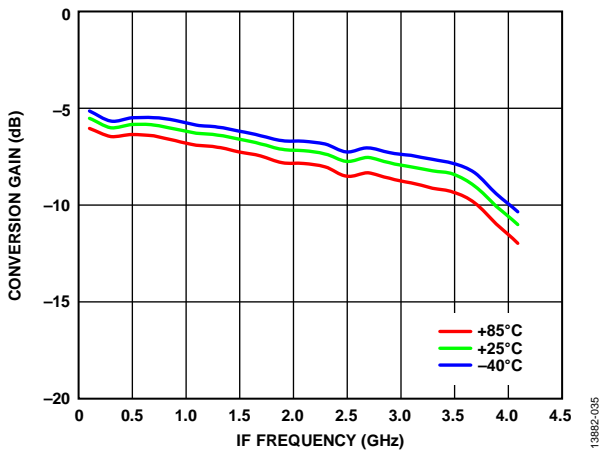


Figure 35. Conversion Gain vs. IF Frequency at Various Temperatures, LO = 15 dBm

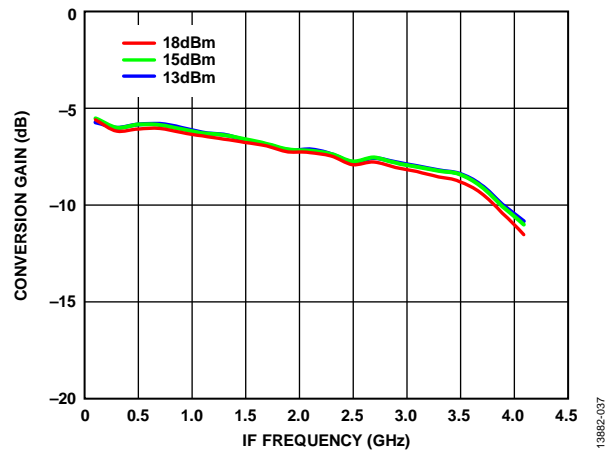


Figure 37. Conversion Gain vs. IF Frequency at Various LO Power Levels, TA = 25°C

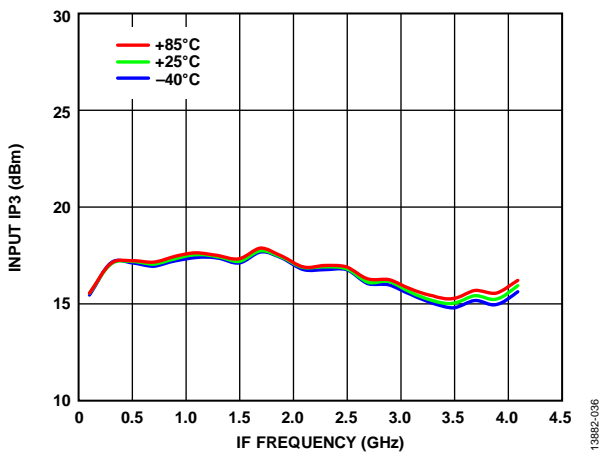


Figure 36. Input IP3 at vs. IF Frequency at Various Temperatures, LO = 15 dBm

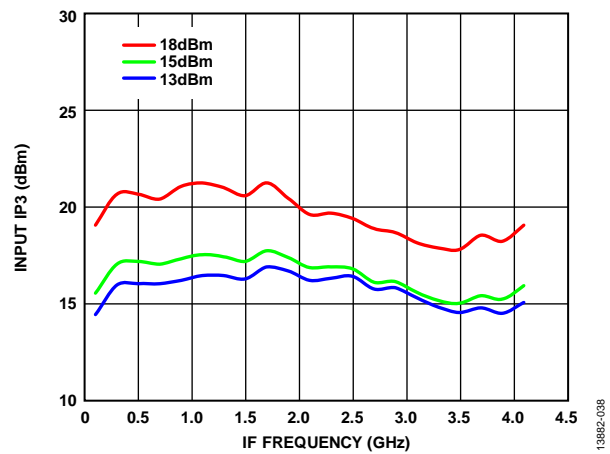


Figure 38. Input IP3 vs. IF Frequency at Various LO Power Levels, TA = 25°C

SPURIOUS AND HARMONICS PERFORMANCE

Mixer spurious products are measured in dBc from the IF output power level. N/A means not applicable.

Downconverter $M \times N$ Spurious Outputs

Spur values are $(M \times RF) - (N \times LO)$.

RF = 6.1GHz at -10 dBm , LO = 6 GHz at 15 dBm.

| | | N x LO | | | | |
|--------|---|--------|----|----|----|-----|
| | | 0 | 1 | 2 | 3 | 4 |
| M x RF | 0 | N/A | 7 | 34 | 30 | 53 |
| | 1 | 13 | 0 | 34 | 46 | 72 |
| | 2 | 85 | 66 | 68 | 65 | 88 |
| | 3 | 86 | 89 | 87 | 74 | 80 |
| | 4 | 84 | 85 | 89 | 92 | 101 |

Upconverter $M \times N$ Spurious Outputs

Spur values are $(M \times IF) + (N \times LO)$.

IF = 100 MHz at -10 dBm, LO = 6 GHz at 15 dBm.

| | | N x LO | | | | |
|--------|---|--------|----|----|----|----|
| | | 0 | 1 | 2 | 3 | 4 |
| M x IF | 0 | N/A | 24 | 15 | 29 | 39 |
| | 1 | 74 | 0 | 30 | 39 | 53 |
| | 2 | 103 | 67 | 61 | 72 | 84 |
| | 3 | 102 | 64 | 85 | 88 | 85 |
| | 4 | 101 | 93 | 93 | 89 | 86 |

THEORY OF OPERATION

The HMC129ALC4 is a general-purpose, double-balanced mixer that can be used as an upconverter or a downconverter from 4 GHz to 8 GHz.

When used as a downconverter, the HMC129ALC4 downconverts radio frequencies (RF) between 4 GHz and 8 GHz to intermediate frequencies (IF) between dc and 3 GHz.

When used as an upconverter, the mixer upconverts intermediate frequencies between dc and 3 GHz to radio frequencies between 4 GHz and 8 GHz.

APPLICATIONS INFORMATION

TYPICAL APPLICATION CIRCUIT

Figure 39 shows the typical application circuit for the HMC129ALC4. The HMC129ALC4 is a passive device and does not require any external components. The LO pin is internally ac-coupled. The RF and IF pins are internally dc-coupled. When IF operation to dc is not required, using an external series capacitor is recommended, of a value chosen to pass the necessary IF frequency range. When IF operation to dc is required, do not exceed the IF source and sink current rating specified in the Absolute Maximum Ratings section.

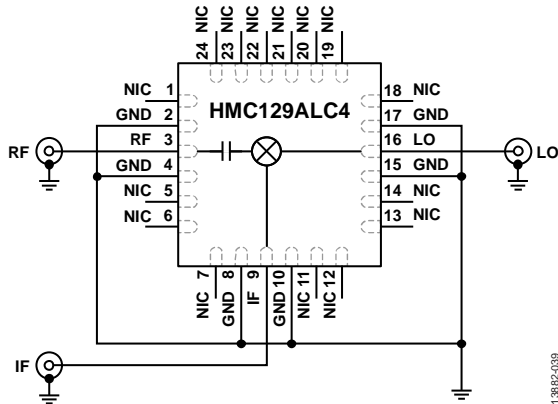


Figure 39. Typical Application Circuit

EVALUATION PCB INFORMATION

Use RF circuit design techniques for the circuit board used in the application. Ensure that signal lines have 50 Ω impedance, and connect the package ground leads and the exposed pad directly to the ground plane (see Figure 40). Use a sufficient number of via holes to connect the top and bottom ground planes. The evaluation circuit board shown in Figure 40 is available from Analog Devices, Inc., upon request.

Table 5. List of Materials for Evaluation PCB
EV1HMC129ALC4

| Item | Description |
|------------------|--|
| J1 to J3 | PCB mount SMA connector |
| U1 | HMC129ALC4 |
| PCB ¹ | 109726 evaluation board on Rogers 4350 |

¹ 109726 is the raw bare PCB identifier. Reference EV1HMC129ALC4 when ordering the complete evaluation PCB.

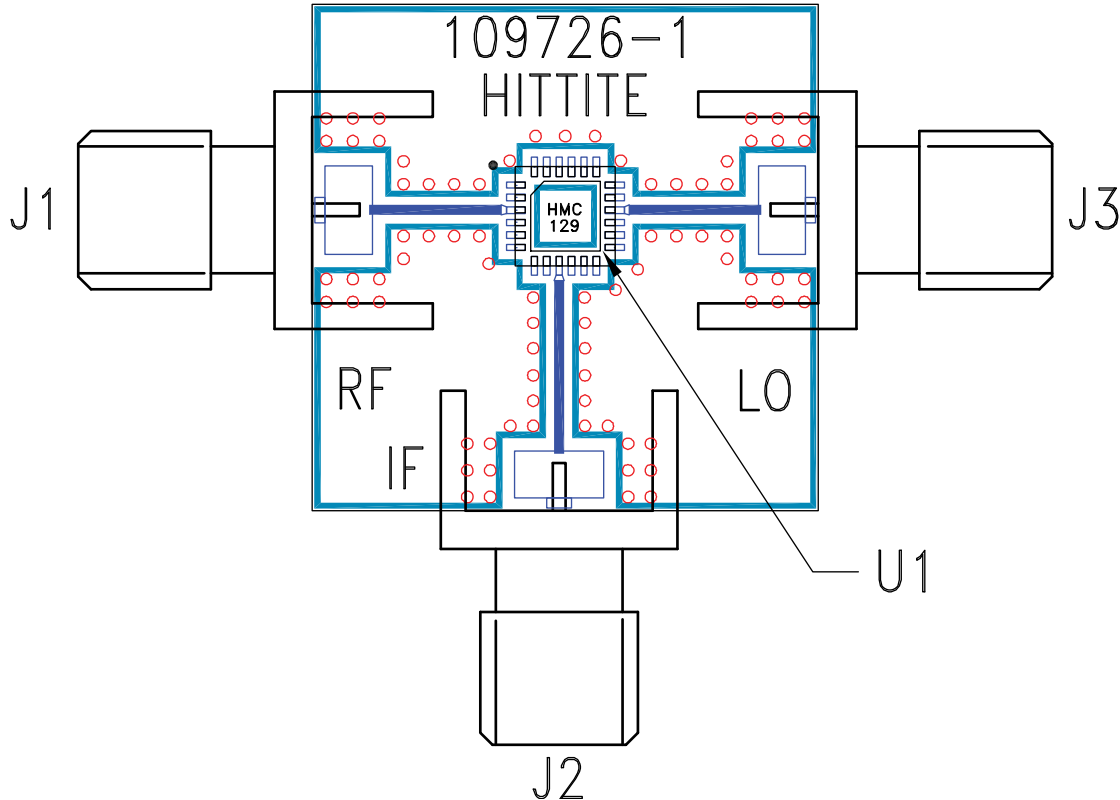


Figure 40. Evaluation PCB Top Layer

OUTLINE DIMENSIONS

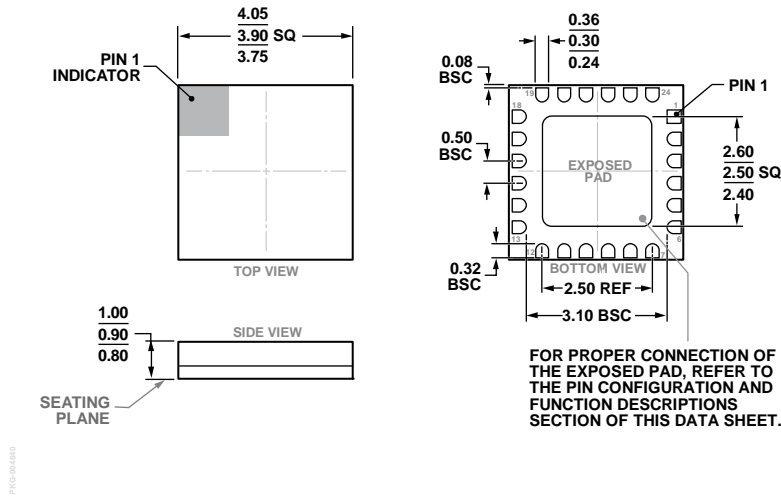


Figure 41. 24-Terminal Ceramic Leadless Chip Carrier (LCC)
(E-24-1)
Dimensions shown in millimeters

ORDERING GUIDE

| Model ¹ | Temperature Range | MSL Rating ² | Package Description | Package Option | Branding |
|--------------------|-------------------|-------------------------|-------------------------|----------------|---------------|
| HMC129ALC4 | -40°C to +85°C | 3 | 24-Terminal Ceramic LCC | E-24-1 | H129A XXXX |
| HMC129ALC4TR | -40°C to +85°C | 3 | 24-Terminal Ceramic LCC | E-24-1 | H129A XXXX |
| HMC129ALC4TR-R5 | -40°C to +85°C | 3 | 24-Terminal Ceramic LCC | E-24-1 | H129A XXXX |
| EV1HMC129ALC4 | | | Evaluation PCB Assembly | | |

¹ The HMC129ALC4, HMC129ALC4TR, and HMC129ALC4TR-R5 are RoHS compliant.
² The peak reflow temperature is 260°C. See the Absolute Maximum Ratings section, Table 2.

Данный компонент на территории Российской Федерации

Вы можете приобрести в компании MosChip.

Для оперативного оформления запроса Вам необходимо перейти по данной ссылке:

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Вы можете разместить у нас заказ для любого Вашего проекта, будь то серийное производство или разработка единичного прибора.

В нашем ассортименте представлены ведущие мировые производители активных и пассивных электронных компонентов.

Нашей специализацией является поставка электронной компонентной базы двойного назначения, продукции таких производителей как XILINX, Intel (ex.ALTERA), Vicor, Microchip, Texas Instruments, Analog Devices, Mini-Circuits, Amphenol, Glenair.

Сотрудничество с глобальными дистрибьюторами электронных компонентов, предоставляет возможность заказывать и получать с международных складов практически любой перечень компонентов в оптимальные для Вас сроки.

На всех этапах разработки и производства наши партнеры могут получить квалифицированную поддержку опытных инженеров.

Система менеджмента качества компании отвечает требованиям в соответствии с ГОСТ Р ИСО 9001, ГОСТ РВ 0015-002 и ЭС РД 009

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