



Low Power, 3.6 MHz, Low Noise, Rail-to-Rail Output, Operational Amplifiers

ADA4691-2/ADA4691-4/ADA4692-2/ADA4692-4

FEATURES

- Low power: 180 μ A typical**
- Very low input bias currents: 0.5 pA typical**
- Low noise: 16 nV/ $\sqrt{\text{Hz}}$ typical**
- 3.6 MHz bandwidth**
- Offset voltage: 500 μ V typical**
- Low offset voltage drift: 4 μ V/ $^{\circ}$ C maximum**
- Low distortion: 0.003% THD + N**
- 2.7 V to 5 V single supply or ± 1.35 V to ± 2.5 V dual supply**
- Available in very small 2 mm \times 2 mm LFCSP packages**

APPLICATIONS

- Photodiode amplifiers
- Sensor amplifiers
- Portable medical and instrumentation
- Portable audio: MP3s, PDAs, and smartphones
- Communications
- Low-side current sense
- ADC drivers
- Active filters
- Sample-and-hold

GENERAL DESCRIPTION

The ADA4691-2/ADA4692-2 are dual and the ADA4691-4/ADA4692-4 are the quad rail-to-rail output, single-supply amplifiers featuring low power, wide bandwidth, and low noise. The ADA4691-2 has two independent shutdown pins, allowing further reduction in supply current. The ADA4691-4 is a quad with dual shutdown pins each controlling a pair of amplifiers and is available in the 16-lead LFCSP. The ADA4692-4 is a quad version without shutdown.

These amplifiers are ideal for a wide variety of applications. Audio, filters, photodiode amplifiers, and charge amplifiers, all benefit from this combination of performance and features. Additional applications for these amplifiers include portable consumer audio players with low noise and low distortion that provide high gain and slew rate response over the audio band at low power. Industrial applications with high impedance sensors, such as piezoelectric and IR sensors, benefit from the high impedance and low 0.5 pA input bias, low offset drift, and enough bandwidth and response for low gain applications.

The ADA4691/ADA4692 family is fully specified over the extended industrial temperature range (-40°C to $+125^{\circ}\text{C}$). The ADA4691-2 is available in a 10-lead LFCSP and a 9-ball WLCSP. The ADA4692-2 is available in an 8-lead SOIC and 8-lead LFCSP. The ADA4691-4 is available in a 16-lead LFCSP. The ADA4692-4 is available in a 14-lead TSSOP. For pin configurations, see the [Pin Configurations](#) section.

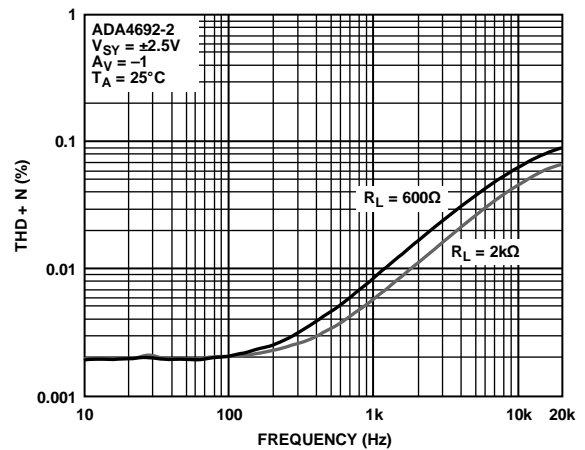


Figure 1. THD + Noise vs. Frequency

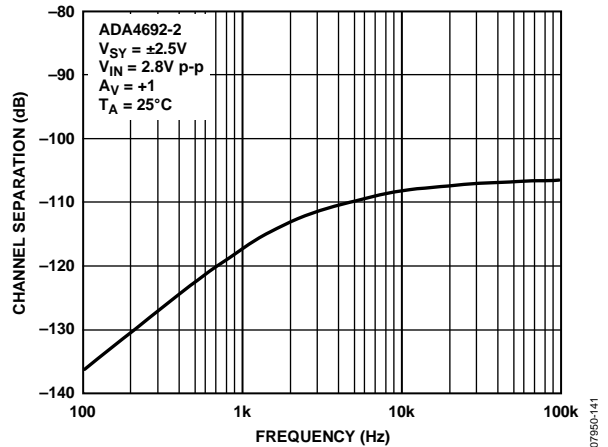


Figure 2. Channel Separation vs. Frequency

Table 1.

	Micropower	Low Power	Low Power with Shutdown	Standard Op Amp With Shutdown	High Bandwidth
Single	AD8613			AD8591	AD8691
Dual	AD8617	ADA4692-2	ADA4691-2	AD8592	AD8692
Quad	AD8619	ADA4692-4	ADA4691-4	AD8594	AD8694

Rev. C

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REVISION HISTORY

12/09—Rev. B to Rev. C

Added ADA4691-4, 16-Lead LFCSP	Throughout
Added Figure 1, Figure 2, and Table 1; Renumbered Sequentially	1
Changes to Applications Section and General Description Section.....	1
Changes to Table 1.....	3
Changes to Table 2.....	4
Changes to Table 4.....	6
Updated Outline Dimensions	17
Changes to Ordering Guide	20

9/09—Rev. A to Rev. B

Added ADA4691-2, 9-Ball WLCSP; ADA4692-2, 8-Lead LFCSP; and ADA4692-4, 14-Lead TSSOP	Throughout
Changes to General Description	1
Updated Outline Dimensions	16
Changes to Ordering Guide	17

6/09—Rev. 0 to Rev. A

Added ADA4691-2, 10 Lead LFCSP	Throughout
Changes to Table 1.....	3
Changes to Table 2.....	4
Changes to Captions for Figure 40, Figure 41, Figure 43, and Figure 44	13
Added Shutdown Operations Section	15
Updated Outline Dimensions	16
Changes to Ordering Guide	16

3/09—Revision 0: Initial Version

SPECIFICATIONS

ELECTRICAL CHARACTERISTICS—2.7 V OPERATION

$V_{SY} = 2.7\text{ V}$, $V_{CM} = V_{SY}/2$, $T_A = 25^\circ\text{C}$, unless otherwise specified.

Table 2.

Parameter	Symbol	Test Conditions/Comments	Min	Typ	Max	Unit
INPUT CHARACTERISTICS						
Offset Voltage	V_{OS}	$V_{CM} = -0.3\text{ V to }+1.6\text{ V}$		0.5	2.5	mV
Dual (ADA469x-2)		$V_{CM} = -0.1\text{ V to }+1.6\text{ V}; -40^\circ\text{C} < T_A < +125^\circ\text{C}$			3.5	mV
Quad (ADA469x-4)		$V_{CM} = -0.1\text{ V to }+1.6\text{ V}; -40^\circ\text{C} < T_A < +125^\circ\text{C}$			4.0	mV
Offset Voltage Drift	$\Delta V_{OS}/\Delta T$	$-40^\circ\text{C} < T_A < +125^\circ\text{C}$		1	4	$\mu\text{V}/^\circ\text{C}$
Input Bias Current	I_B	$-40^\circ\text{C} < T_A < +125^\circ\text{C}$		0.5	5	pA
Input Offset Current	I_{OS}	$-40^\circ\text{C} < T_A < +125^\circ\text{C}$		1	8	pA
Input Voltage Range		$-40^\circ\text{C} < T_A < +125^\circ\text{C}$	-0.3		+1.6	V
Common-Mode Rejection Ratio	CMRR	$V_{CM} = -0.3\text{ V to }+1.6\text{ V}$	70	90		dB
Large Signal Voltage Gain	A_{VO}	$V_{CM} = -0.1\text{ V to }+1.6\text{ V}; -40^\circ\text{C} < T_A < +125^\circ\text{C}$	62			dB
		$R_L = 2\text{ k}\Omega, V_{OUT} = 0.5\text{ V to }2.2\text{ V}$	90	100		dB
		$-40^\circ\text{C} < T_A < +85^\circ\text{C}$	80			dB
		$-40^\circ\text{C} < T_A < +125^\circ\text{C}$	63			dB
		$R_L = 600\ \Omega, V_{OUT} = 0.5\text{ V to }2.2\text{ V}$	85	95		dB
Input Capacitance	C_{IN}					
Differential Mode	C_{INDM}			2.5		pF
Common Mode	C_{INCM}			7		pF
Logic High Voltage (Enabled)	V_{IH}	$-40^\circ\text{C} < T_A < +125^\circ\text{C}$	1.6			V
Logic Low Voltage (Power-Down)	V_{IL}	$-40^\circ\text{C} < T_A < +125^\circ\text{C}$			0.5	V
Logic Input Current (Per Pin)	I_{IN}	$-40^\circ\text{C} < T_A < +125^\circ\text{C}, 0\text{ V} \leq V_{SD} \leq 2.7\text{ V}$			1	μA
OUTPUT CHARACTERISTICS						
Output Voltage High	V_{OH}	$R_L = 2\text{ k}\Omega\text{ to GND}$	2.65	2.67		V
		$-40^\circ\text{C} < T_A < +125^\circ\text{C}$	2.6			V
		$R_L = 600\ \Omega\text{ to GND}$	2.55	2.59		V
		$-40^\circ\text{C} < T_A < +125^\circ\text{C}$	2.5			V
Output Voltage Low	V_{OL}	$R_L = 2\text{ k}\Omega\text{ to }V_{SY}$		24	30	mV
		$-40^\circ\text{C} < T_A < +125^\circ\text{C}$			40	mV
		$R_L = 600\ \Omega\text{ to }V_{SY}$		78	95	mV
		$-40^\circ\text{C} < T_A < +125^\circ\text{C}$			130	mV
Short-Circuit Current	I_{SC}	$V_{OUT} = V_{SY}\text{ or GND}$		± 15		mA
Closed-Loop Output Impedance	Z_{OUT}	$f = 1\text{ MHz}, A_V = -100$		372		Ω
Output Pin Leakage Current		$-40^\circ\text{C} < T_A < +125^\circ\text{C}, \text{ shutdown active}, V_{SD} = V_{SS}$		10		nA
POWER SUPPLY						
Power Supply Rejection Ratio	PSRR	$V_S = 2.7\text{ V to }5.5\text{ V}$	80	90		dB
		$-40^\circ\text{C} < T_A < +125^\circ\text{C}$	75			dB
Supply Current Per Amplifier	I_{SY}	$V_{OUT} = V_{SY}/2$		165	200	μA
		$-40^\circ\text{C} < T_A < +125^\circ\text{C}$			240	μA
Supply Current Shutdown Mode	I_{SD}	All amplifiers shut down, $V_{SD} = V_{SS}$		10		nA
		$-40^\circ\text{C} < T_A < +125^\circ\text{C}$			2	μA

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Parameter	Symbol	Test Conditions/Comments	Min	Typ	Max	Unit
DYNAMIC PERFORMANCE						
Slew Rate	SR	$R_L = 600 \Omega, C_L = 20 \text{ pF}, A_V = +1$ $R_L = 2 \text{ k}\Omega, C_L = 20 \text{ pF}, A_V = +1$		1.1 1.4		V/ μ s V/ μ s
Settling Time to 0.1%	t_s	Step = 0.5 V, $R_L = 2 \text{ k}\Omega, 600 \Omega$		1		μ s
Gain Bandwidth Product	GBP	$R_L = 1 \text{ M}\Omega, C_L = 35 \text{ pF}, A_V = +1$		3.6		MHz
Phase Margin	Φ_M	$R_L = 1 \text{ M}\Omega, C_L = 35 \text{ pF}, A_V = +1$		49		Degrees
Turn-On/Turn-Off Time		$R_L = 600 \Omega$		1		μ s
NOISE PERFORMANCE						
Distortion	THD + N	$A_V = -1, R_L = 2 \text{ k}\Omega, f = 1 \text{ kHz}, V_{IN \text{ rms}} = 0.15 \text{ V rms}$ $A_V = -1, R_L = 600 \Omega, f = 1 \text{ kHz}, V_{IN \text{ rms}} = 0.15 \text{ V rms}$ $A_V = +1, R_L = 2 \text{ k}\Omega, f = 1 \text{ kHz}, V_{IN \text{ rms}} = 0.15 \text{ V rms}$ $A_V = +1, R_L = 600 \Omega, f = 1 \text{ kHz}, V_{IN \text{ rms}} = 0.15 \text{ V rms}$		0.009 0.01 0.006 0.009		% % % %
Voltage Noise	e_n p-p	$f = 0.1 \text{ Hz to } 10 \text{ Hz}$		3.1		μ V p-p
Voltage Noise Density	e_n	$f = 1 \text{ kHz}$ $f = 10 \text{ kHz}$		16 13		nV/ $\sqrt{\text{Hz}}$ nV/ $\sqrt{\text{Hz}}$

ELECTRICAL CHARACTERISTICS—5 V OPERATION

$V_{SY} = 5 \text{ V}, V_{CM} = V_{SY}/2, T_A = 25^\circ\text{C}$, unless otherwise specified.

Table 3.

Parameter	Symbol	Test Conditions/Comments	Min	Typ	Max	Unit
INPUT CHARACTERISTICS						
Offset Voltage	V_{OS}	$V_{CM} = -0.3 \text{ V to } +3.9 \text{ V}$ $V_{CM} = -0.1 \text{ V to } +3.9 \text{ V}; -40^\circ\text{C} < T_A < +125^\circ\text{C}$ $V_{CM} = -0.1 \text{ V to } +3.9 \text{ V}; -40^\circ\text{C} < T_A < +125^\circ\text{C}$		0.5	2.5 3.5 4.0	mV mV mV
Offset Voltage Drift	$\Delta V_{OS}/\Delta T$	$-40^\circ\text{C} < T_A < +125^\circ\text{C}$		1	4	$\mu\text{V}/^\circ\text{C}$
Input Bias Current	I_B	$-40^\circ\text{C} < T_A < +125^\circ\text{C}$		0.5	5 360	pA pA
Input Offset Current	I_{OS}	$-40^\circ\text{C} < T_A < +125^\circ\text{C}$		1	8 260	pA pA
Input Voltage Range		$-40^\circ\text{C} < T_A < +125^\circ\text{C}$	-0.3		+3.9	V
Common-Mode Rejection Ratio	CMRR	$V_{CM} = -0.3 \text{ V to } +3.9 \text{ V}$ $V_{CM} = -0.1 \text{ V to } +3.9 \text{ V}; -40^\circ\text{C} < T_A < +125^\circ\text{C}$	75	98		dB dB
Large Signal Voltage Gain	A_{VO}	$R_L = 2 \text{ k}\Omega, V_O = 0.5 \text{ V to } 4.5 \text{ V}, V_{CM} = 0 \text{ V}$ $-40^\circ\text{C} < T_A < +85^\circ\text{C}$ $-40^\circ\text{C} < T_A < +125^\circ\text{C}$ $R_L = 600 \Omega, V_O = 0.5 \text{ V to } 4.5 \text{ V}, V_{CM} = 0 \text{ V}$	95	110		dB dB dB dB
Input Capacitance						
Differential Mode	C_{INDM}			2.5		pF
Common Mode	C_{INCM}			7		pF
Logic High Voltage (Enabled)	V_{IH}	$-40^\circ\text{C} < T_A < +125^\circ\text{C}$	2.0			V
Logic Low Voltage (Power-Down)	V_{IL}	$-40^\circ\text{C} < T_A < +125^\circ\text{C}$			0.8	V
Logic Input Current (Per Pin)	I_{IN}	$-40^\circ\text{C} < T_A < +125^\circ\text{C}, 0 \text{ V} \leq V_{SD} \leq 2.7 \text{ V}$			1	μA

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Parameter	Symbol	Test Conditions/Comments	Min	Typ	Max	Unit
OUTPUT CHARACTERISTICS						
Output Voltage High	V_{OH}	$R_L = 2\text{ k}\Omega$	4.95	4.97		V
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	4.90			V
Output Voltage Low	V_{OL}	$R_L = 600\ \Omega$ to GND	4.85	4.88		V
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	4.80			V
		$R_L = 2\text{ k}\Omega$		30	35	mV
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$			50	mV
Short-Circuit Limit	I_{SC}	$R_L = 600\ \Omega$		100	110	mV
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$			155	mV
Short-Circuit Limit	I_{SC}	$V_{OUT} = V_{SY}$ or GND		± 55		mA
Closed-Loop Output Impedance	Z_{OUT}	ADA4691-2, $f = 1\text{ MHz}$, $A_V = -100$		364		Ω
		ADA4691-2, $f = 1\text{ MHz}$, $A_V = -100$		246		Ω
Output Pin Leakage Current		$-40^\circ\text{C} < T_A < +125^\circ\text{C}$, shutdown active, $V_{SD} = V_{SS}$		10		nA
POWER SUPPLY						
Power Supply Rejection Ratio	PSRR	$V_{SY} = 2.7\text{ V}$ to 5.5 V	80	90		dB
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$	75			dB
Supply Current Per Amplifier	I_{SY}	$V_{OUT} = V_{SY}/2$		180	225	μA
Supply Current Shutdown Mode	I_{SD}	$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$			275	μA
		All amplifiers shut down, $V_{SD} = V_{SS}$		10		nA
		$-40^\circ\text{C} \leq T_A \leq +125^\circ\text{C}$			2	μA
DYNAMIC PERFORMANCE						
Slew Rate	SR	$R_L = 2\text{ k}\Omega$, $600\ \Omega$, $C_L = 20\text{ pF}$, $A_V = +1$		1.3		V/ μs
Settling Time to 0.1%	t_s	$V_{IN} = 2\text{ V}$ step, $R_L = 2\text{ k}\Omega$ or $600\ \Omega$		1.5		μs
Gain Bandwidth Product	GBP	$R_L = 1\text{ M}\Omega$, $C_L = 35\text{ pF}$, $A_V = +1$		3.6		MHz
Phase Margin	Φ_M	$R_L = 1\text{ M}\Omega$, $C_L = 35\text{ pF}$, $A_V = +1$		52		Degrees
Turn-On/Turn-Off Time		$R_L = 600\ \Omega$		1		μs
NOISE PERFORMANCE						
Distortion	THD + N	$A_V = -1$, $R_L = 2\text{ k}\Omega$, $f = 1\text{ kHz}$, $V_{IN\text{ rms}} = 0.8\text{ V rms}$		0.006		%
		$A_V = -1$, $R_L = 600\ \Omega$, $f = 1\text{ kHz}$, $V_{IN\text{ rms}} = 0.8\text{ V rms}$		0.008		%
		$A_V = +1$, $R_L = 2\text{ k}\Omega$, $f = 1\text{ kHz}$, $V_{IN\text{ rms}} = 0.8\text{ V rms}$		0.001		%
		$A_V = +1$, $R_L = 600\ \Omega$, $f = 1\text{ kHz}$, $V_{IN\text{ rms}} = 0.8\text{ V rms}$		0.003		%
Voltage Noise	e_n p-p	$f = 0.1\text{ Hz}$ to 10 Hz		3.2		$\mu\text{V p-p}$
Voltage Noise Density	e_n	$f = 1\text{ kHz}$		16		nV/ $\sqrt{\text{Hz}}$
		$f = 10\text{ kHz}$		13		nV/ $\sqrt{\text{Hz}}$

ABSOLUTE MAXIMUM RATINGS

Table 4.

Parameter	Rating
Supply Voltage	6 V
Input Voltage	$V_{SS} - 0.3 \text{ V}$ to $V_{DD} + 0.3 \text{ V}$
Input Current ¹	$\pm 10 \text{ mA}$
Shutdown Pin Rise/Fall Times	50 μs maximum
Differential Input Voltage ²	$\pm V_{SY}$
Output Short-Circuit Duration to GND	Indefinite
Temperature	
Storage Temperature Range	-65°C to $+150^\circ\text{C}$
Operating Temperature Range	-40°C to $+125^\circ\text{C}$
Junction Temperature Range	-65°C to $+150^\circ\text{C}$
Lead Temperature (Soldering, 60 sec)	300°C

¹ Input pins have clamp diodes to the supply pins. Limit the input current to 10 mA or less whenever the input signal exceeds the power supply rail by 0.3 V.

² Differential input voltage is limited to 5 V or the supply voltage, whichever is less.

Stresses above those listed under Absolute Maximum Ratings may cause permanent damage to the device. This is a stress rating only; functional operation of the device at these or any other conditions above those indicated in the operational section of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

THERMAL RESISTANCE

θ_{JA} is specified for the worst-case conditions, that is, a device soldered in a circuit board for surface-mount packages and measured using a standard 4-layer board, unless otherwise specified.

Table 5. Thermal Resistance

Package Type	θ_{JA}	θ_{JC}	Unit
8-Lead SOIC_N (R-8)	120	45	$^\circ\text{C}/\text{W}$
8-Lead LFCSP (CP-8-6)	125	40	$^\circ\text{C}/\text{W}$
9-Ball WLCSP (CB-9-3)	77	N/A ¹	$^\circ\text{C}/\text{W}$
10-Lead LFCSP (CP-10-11)	115	40	$^\circ\text{C}/\text{W}$
16-Lead LFCSP (CP-16-22)	75	12	$^\circ\text{C}/\text{W}$
14-Lead TSSOP (RU-14)	112	35	$^\circ\text{C}/\text{W}$

¹ N/A = not applicable.

ESD CAUTION



ESD (electrostatic discharge) sensitive device.

Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PIN CONFIGURATIONS

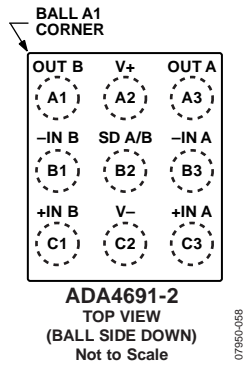


Figure 3. 9-Ball Wafer Level Chip Scale WLCSP (CB-9-3)

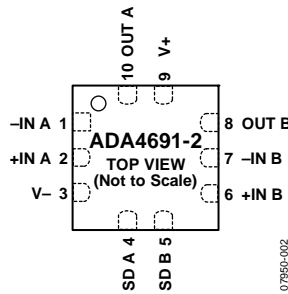
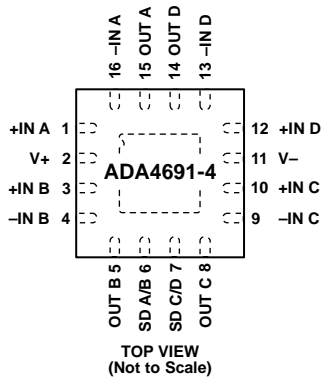


Figure 4. 10-Lead, 2 mm x 2 mm LFCSP (CP-10-11)



NOTES
1. IT IS RECOMMENDED THAT THE EXPOSED PAD BE CONNECTED TO V-.

Figure 5. 16-Lead, 3 mm x 3 mm LFCSP (CP-16-22)

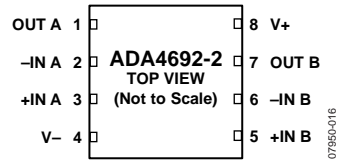


Figure 6. 8-Lead, 2 mm x 2 mm LFCSP (CP-8-6)

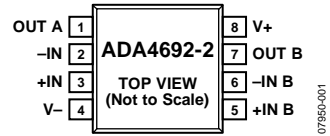


Figure 7. 8-Lead SOIC_N (R-8)

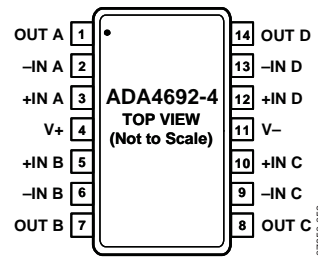


Figure 8. 14-Lead TSSOP (RU-14)

TYPICAL PERFORMANCE CHARACTERISTICS

T_A = 25°C, unless otherwise noted.



Figure 9. Input Offset Voltage Distribution



Figure 12. Input Offset Voltage Distribution

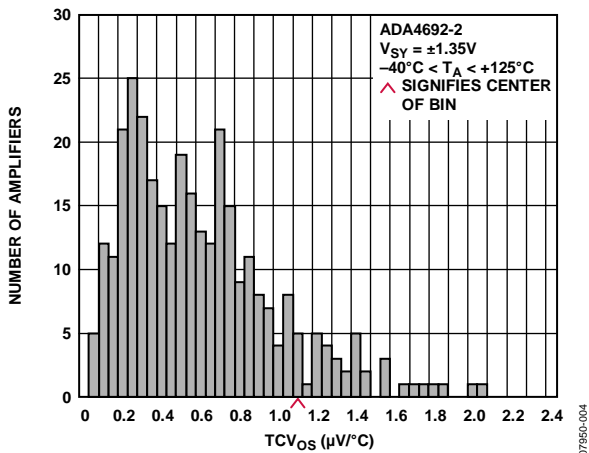


Figure 10. Input Offset Voltage Drift Distribution

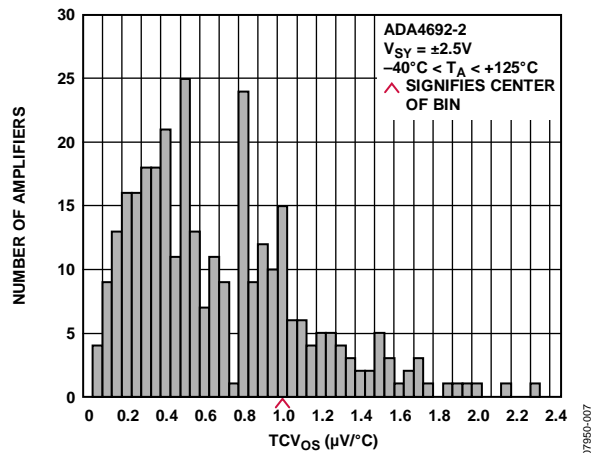


Figure 13. Input Offset Voltage Drift Distribution

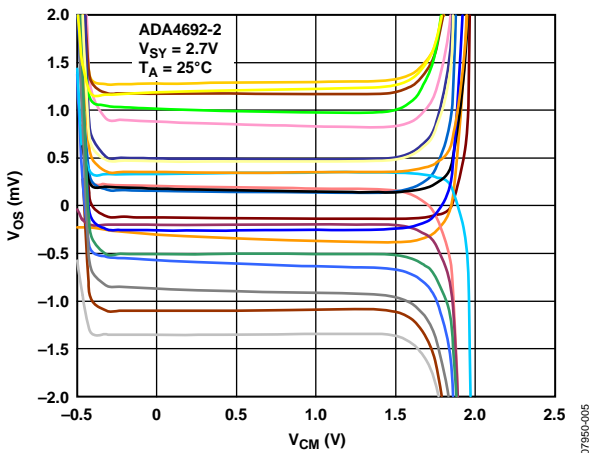


Figure 11. Input Offset Voltage vs. Common-Mode Voltage

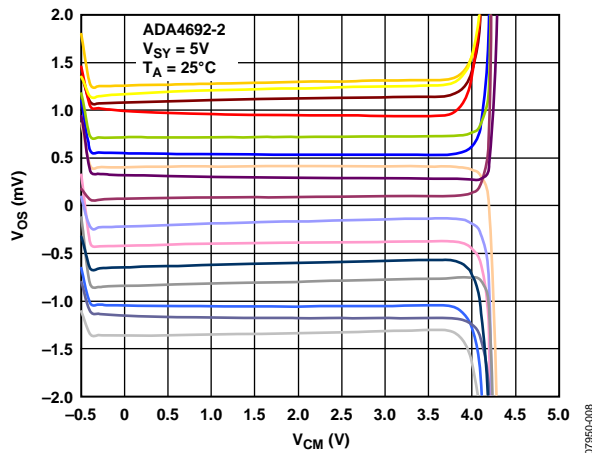


Figure 14. Input Offset Voltage vs. Common-Mode Voltage

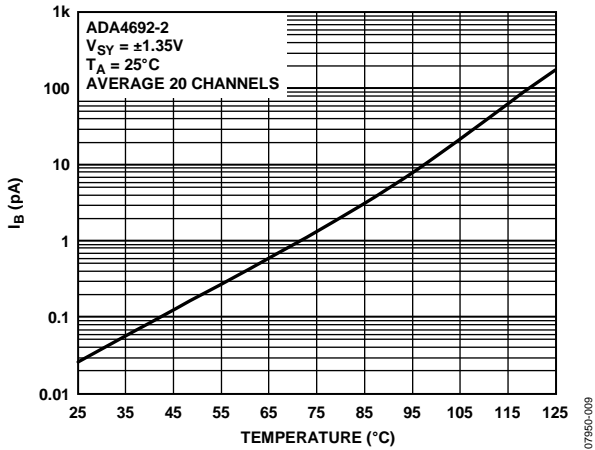


Figure 15. Input Bias Current vs. Temperature

07950-009

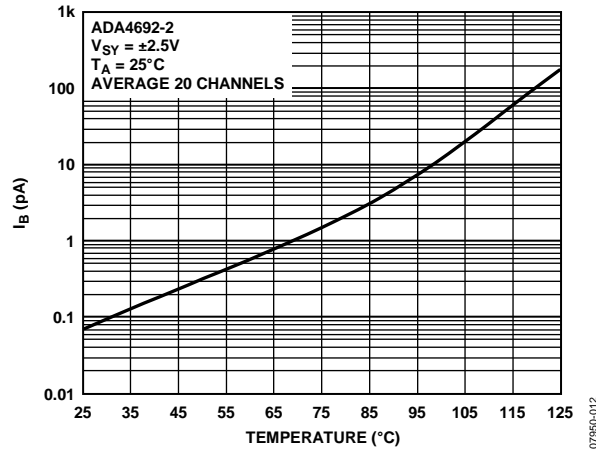


Figure 18. Input Bias Current vs. Temperature

07950-012

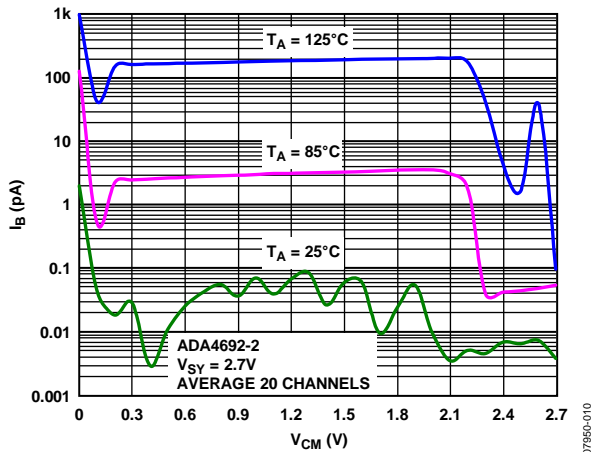


Figure 16. Input Bias Current vs. Common-Mode Voltage

07950-010

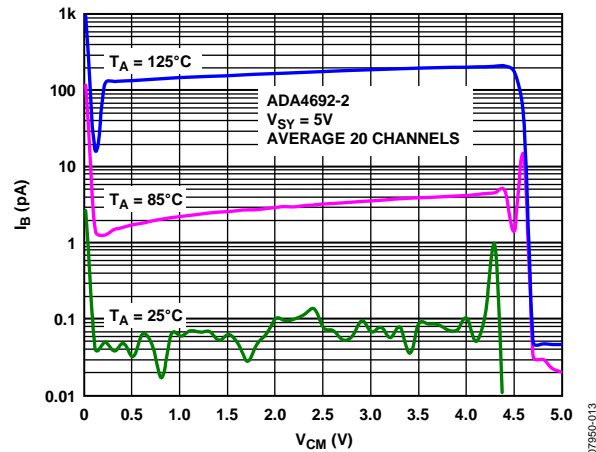


Figure 19. Input Bias Current vs. Common-Mode Voltage

07950-013

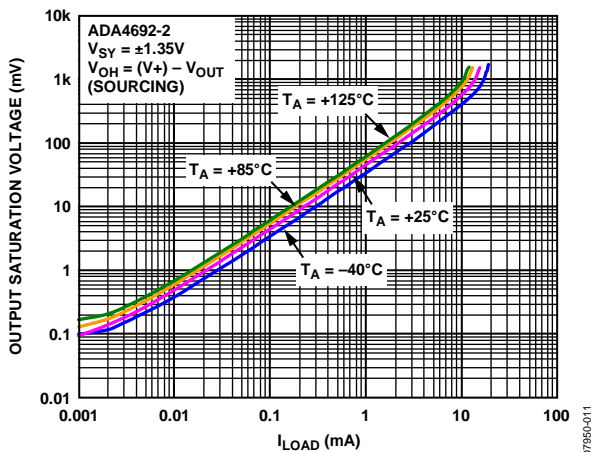


Figure 17. Output Voltage (V_{OH}) to Supply Rail vs. Load Current

07950-011

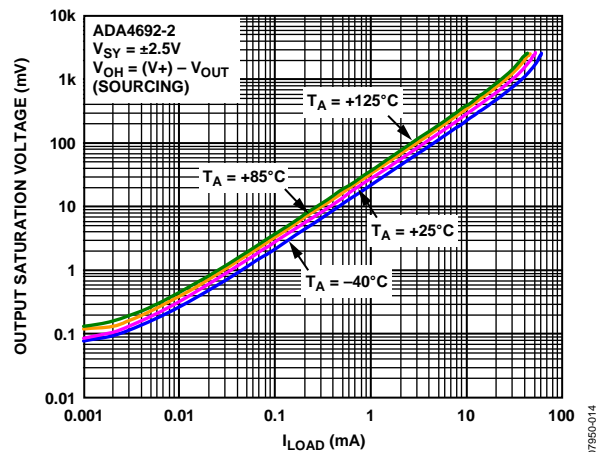


Figure 20. Output Voltage (V_{OH}) to Supply Rail vs. Load Current

07950-014

ADA4691-2/ADA4691-4/ADA4692-2/ADA4692-4

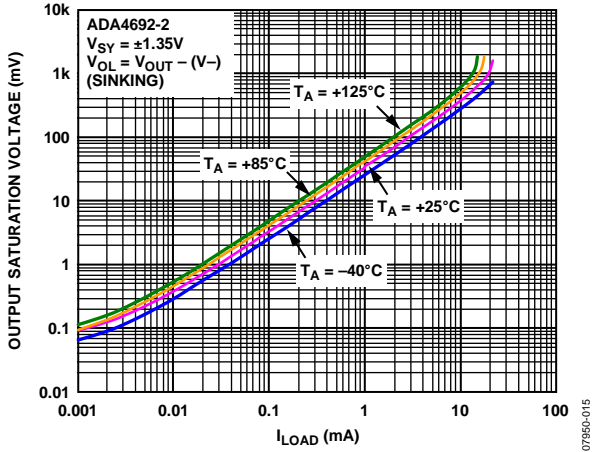


Figure 21. Output Voltage (V_{OL}) to Supply Rail vs. Load Current

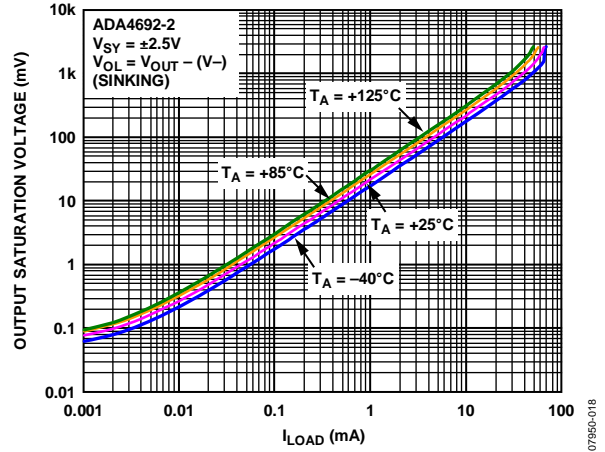


Figure 24. Output Voltage (V_{OL}) to Supply Rail vs. Load Current

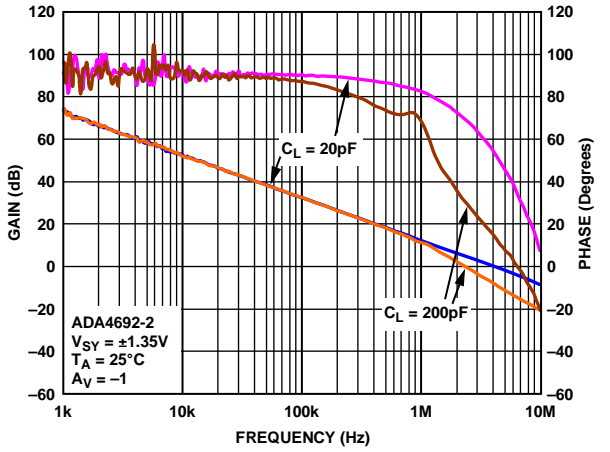


Figure 22. Open-Loop Gain and Phase vs. Frequency

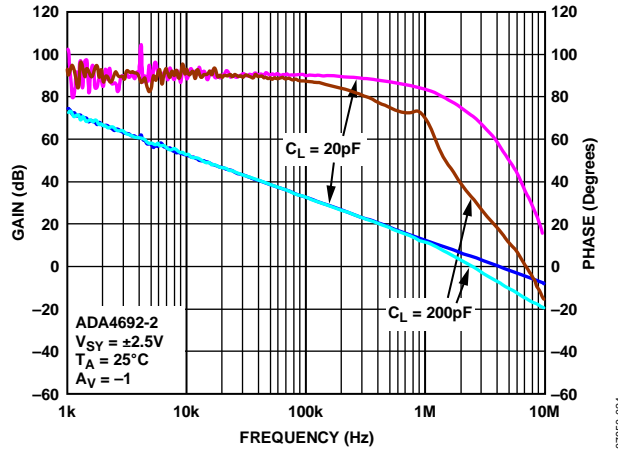


Figure 25. Open-Loop Gain and Phase vs. Frequency

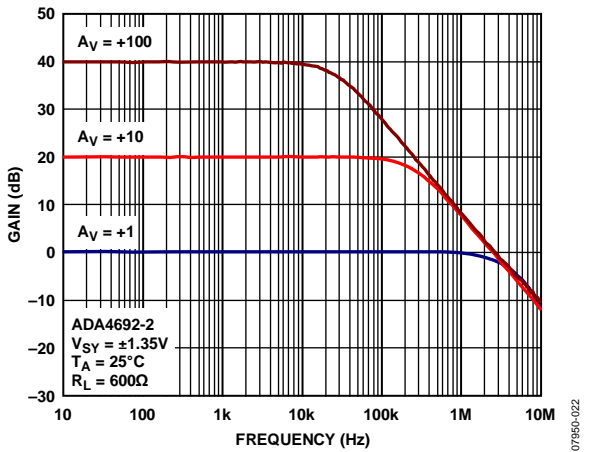


Figure 23. Closed-Loop Gain vs. Frequency

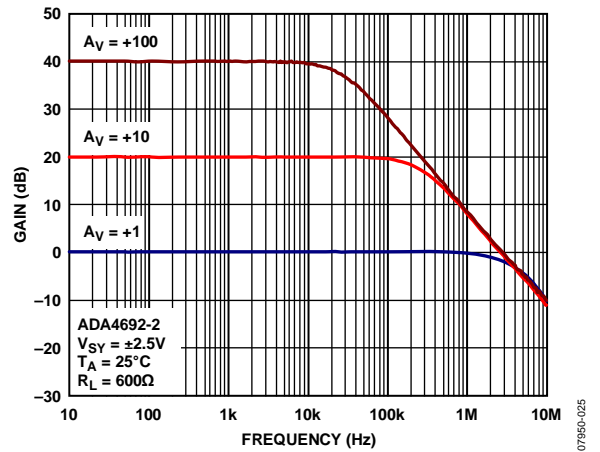


Figure 26. Closed-Loop Gain vs. Frequency

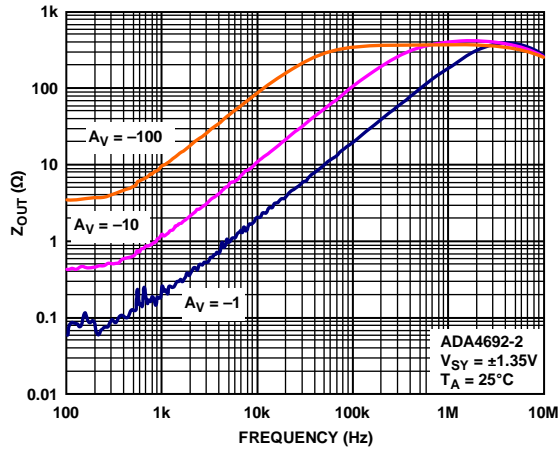


Figure 27. Output Impedance vs. Frequency

07950-023

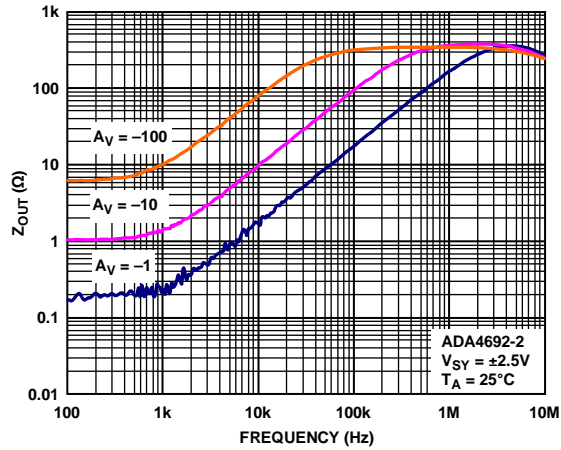


Figure 30. Output Impedance vs. Frequency

07950-026

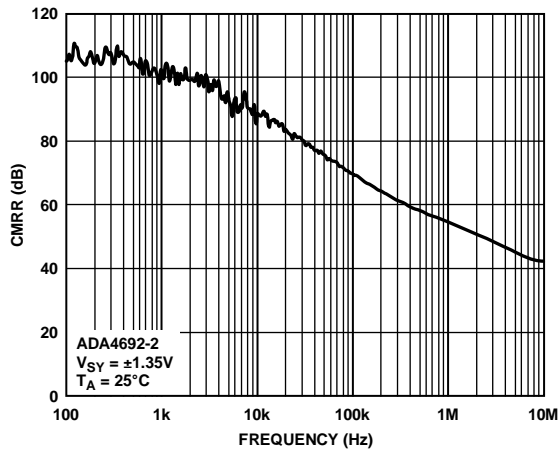


Figure 28. CMRR vs. Frequency

07950-027

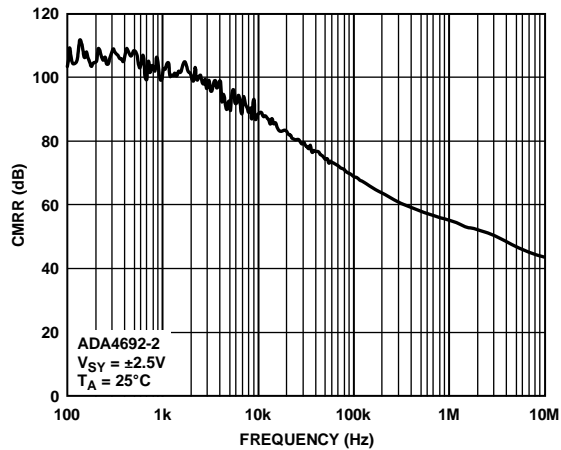


Figure 31. CMRR vs. Frequency

07950-030

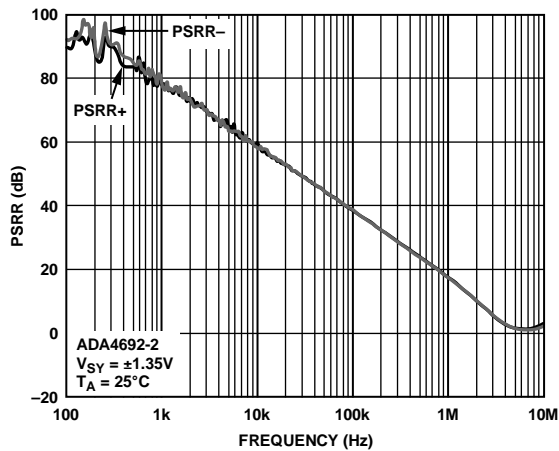


Figure 29. PSRR vs. Frequency

07950-028

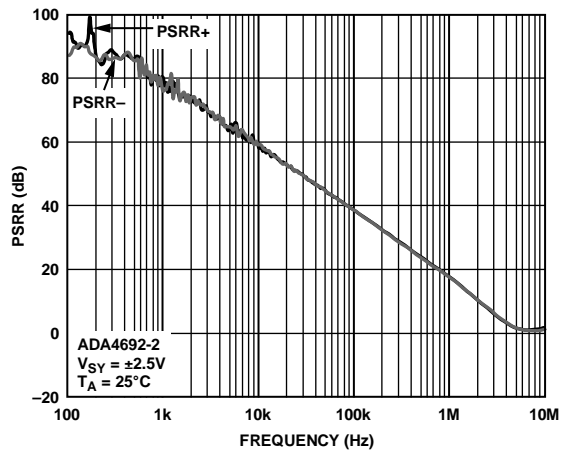


Figure 32. PSRR vs. Frequency

07950-031

ADA4691-2/ADA4691-4/ADA4692-2/ADA4692-4

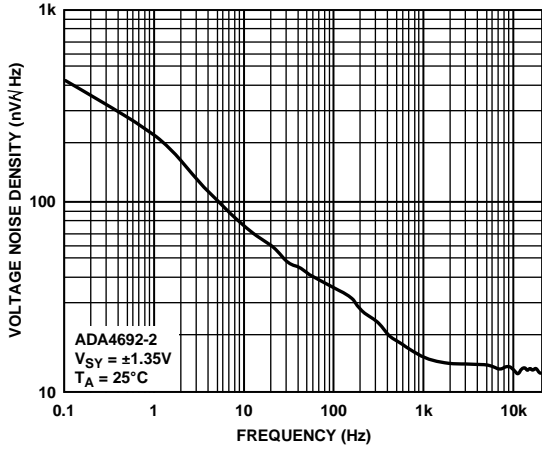


Figure 33. Voltage Noise Density vs. Frequency

07950-029

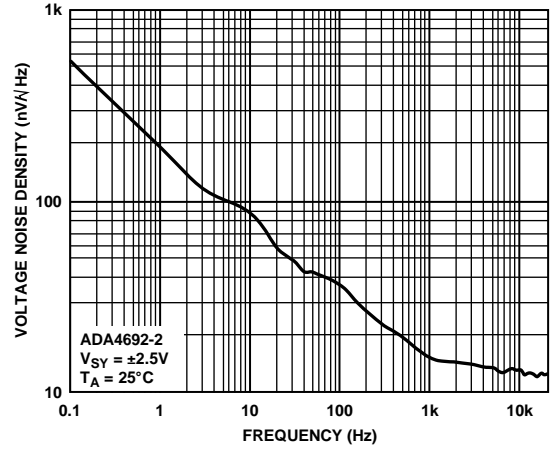


Figure 36. Voltage Noise Density vs. Frequency

07950-032

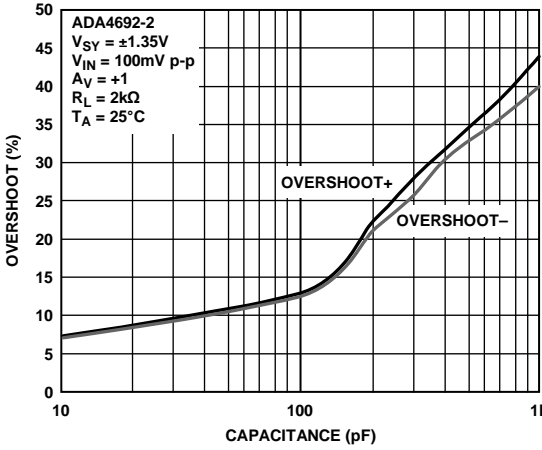


Figure 34. Small Signal Overshoot vs. Load Capacitance

07950-033

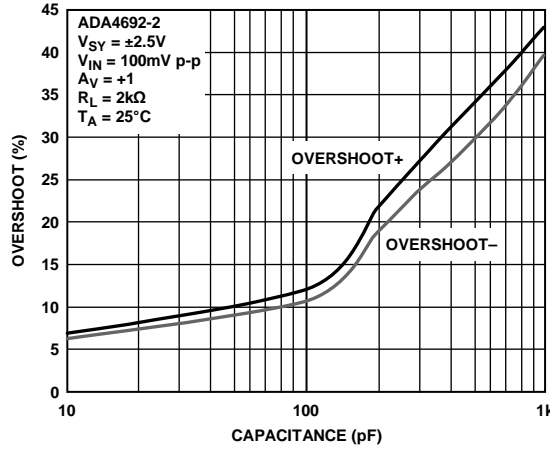


Figure 37. Small Signal Overshoot vs. Load Capacitance

07950-036

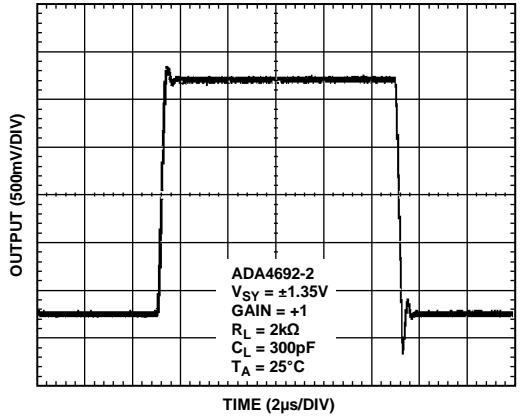


Figure 35. Large Signal Transient Response

07950-034

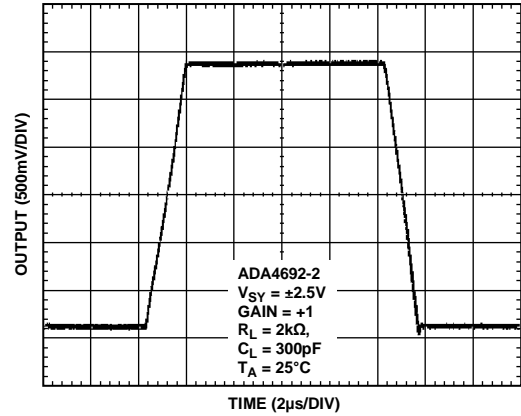


Figure 38. Large Signal Transient Response

07950-037

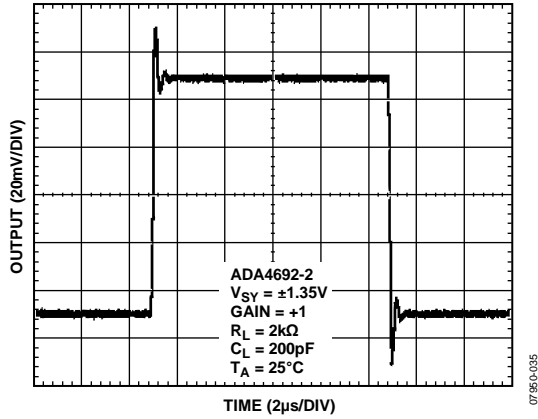


Figure 39. Small Signal Transient Response

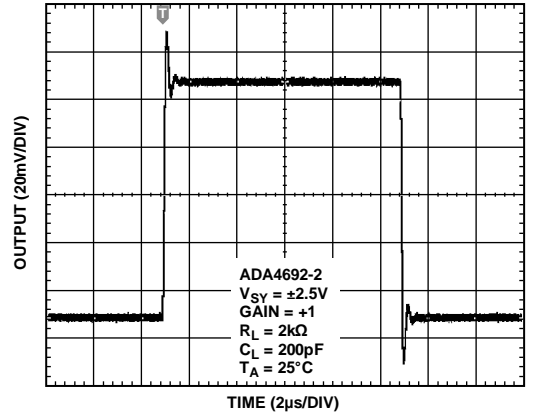


Figure 42. Small Signal Transient Response

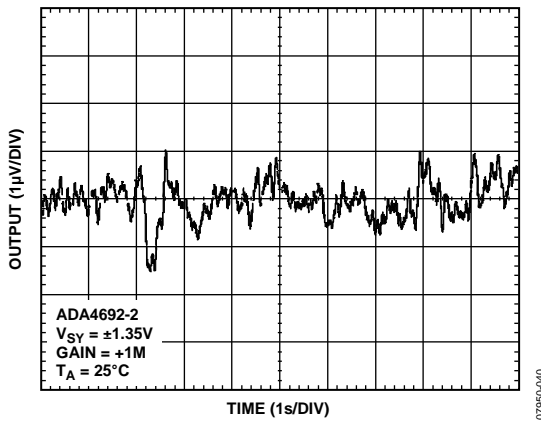


Figure 40. 0.1 Hz to 10 Hz Noise

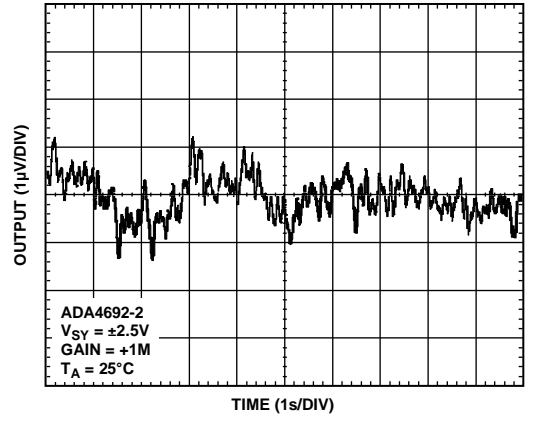


Figure 43. 0.1 Hz to 10 Hz Noise

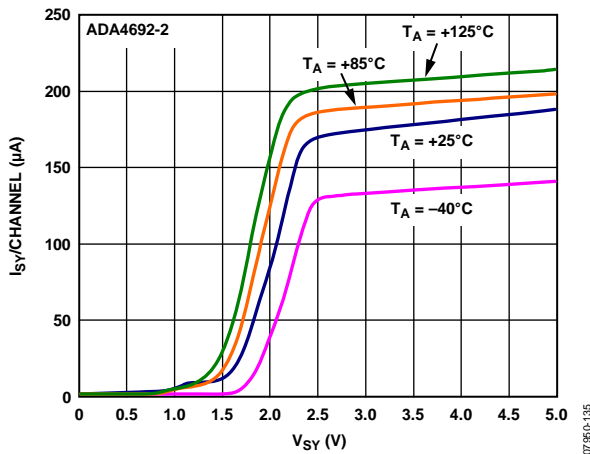


Figure 41. Supply Current per Amplifier vs. Supply Voltage

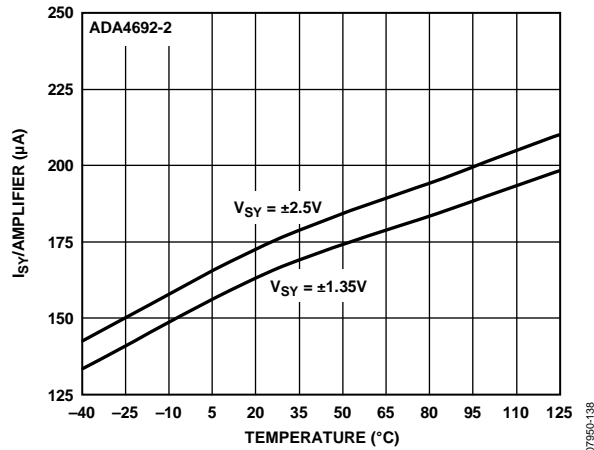


Figure 44. Supply Current per Channel vs. Temperature

ADA4691-2/ADA4691-4/ADA4692-2/ADA4692-4

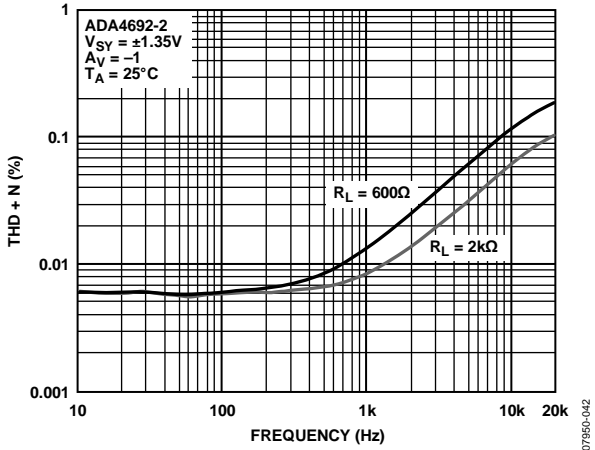


Figure 45. THD + Noise vs. Frequency

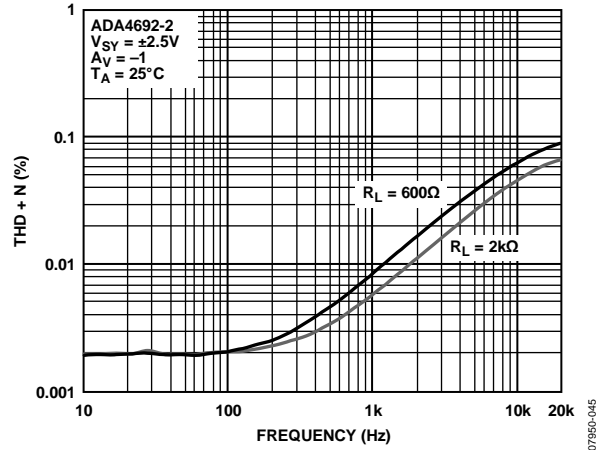


Figure 48. THD + Noise vs. Frequency

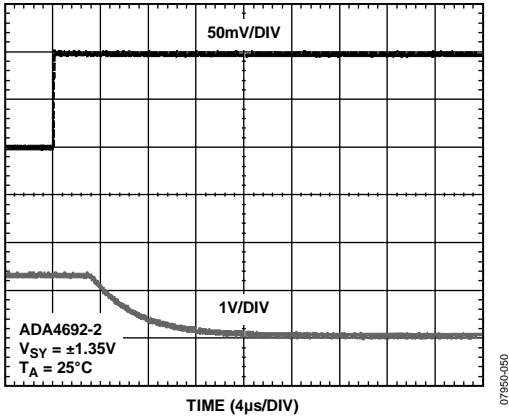


Figure 46. Positive Overload Recovery

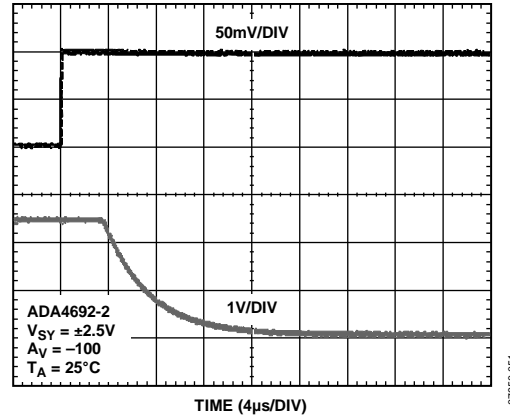


Figure 49. Positive Overload Recovery

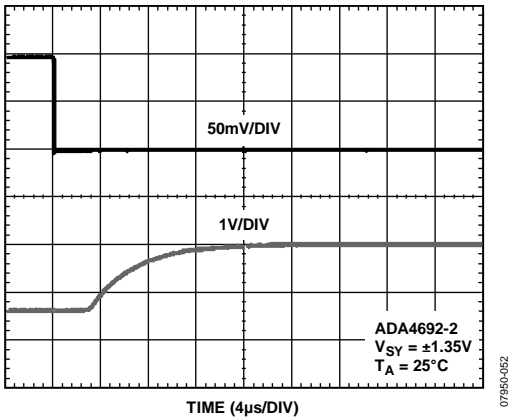


Figure 47. Negative Overload Recovery

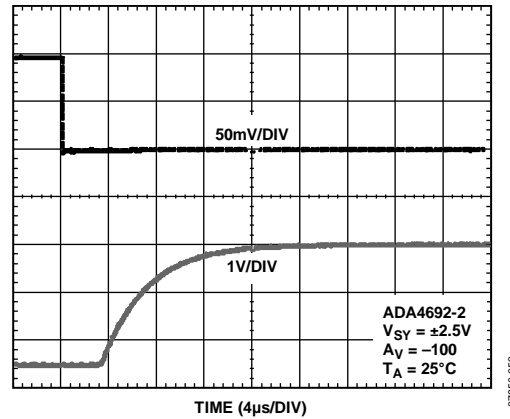
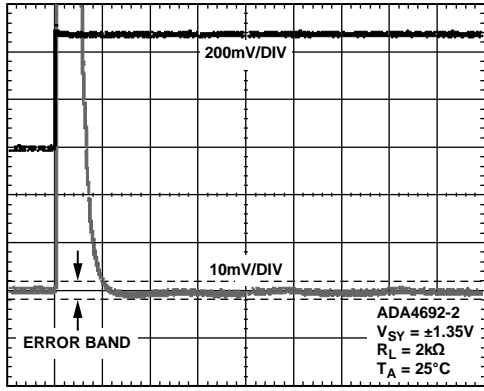


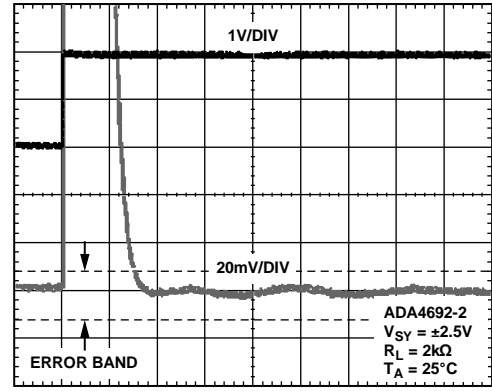
Figure 50. Negative Overload Recovery



TIME (1μs/DIV)

Figure 51. Positive Settling Time to 0.1%

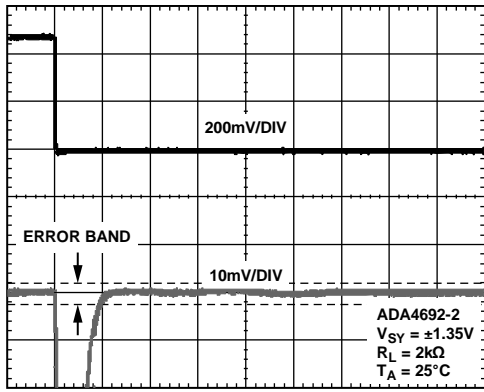
07950-054



TIME (1μs/DIV)

Figure 54. Positive Settling Time to 0.1%

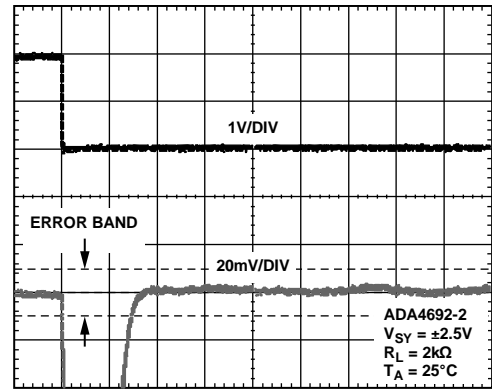
07950-055



TIME (1μs/DIV)

Figure 52. Negative Settling Time to 0.1%

07950-056



TIME (1μs/DIV)

Figure 55. Negative Settling Time to 0.1%

07950-057

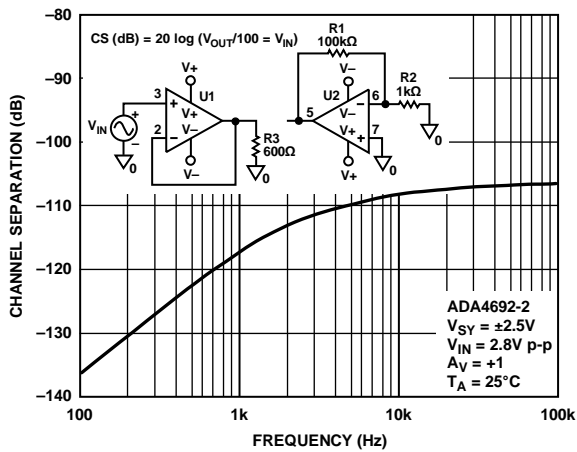


Figure 53. Channel Separation (CS) vs. Frequency

07950-140

SHUTDOWN OPERATION

INPUT PIN CHARACTERISTICS

The ADA4691-2 has a classic CMOS logic inverter input for each shutdown pin, as shown in Figure 56.

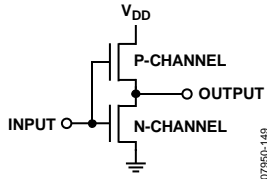


Figure 56. CMOS Inverter

With slowly changing inputs, the top transistor and bottom transistor may be slightly on at the same time, increasing the supply current. This can be avoided by driving the input with a digital logic output having fast rise and fall times. Figure 57 through Figure 59 show the supply current for both sections switching simultaneously with rise times of 1 μ s, 10 μ s, and 1 ms. Clearly, the rise and fall times should be faster than 10 μ s. Using an RC time constant to enable/disable shutdown is not recommended.

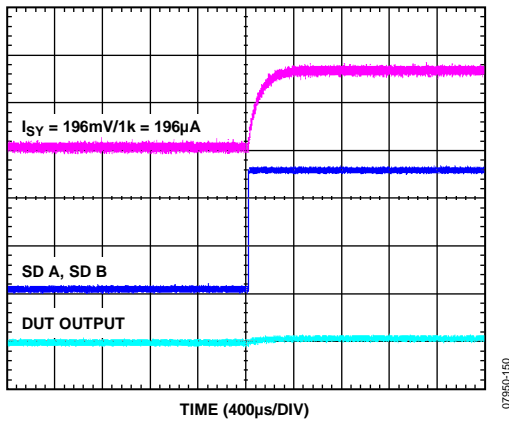


Figure 57. Shutdown Pin Rise Time = 1 μ s

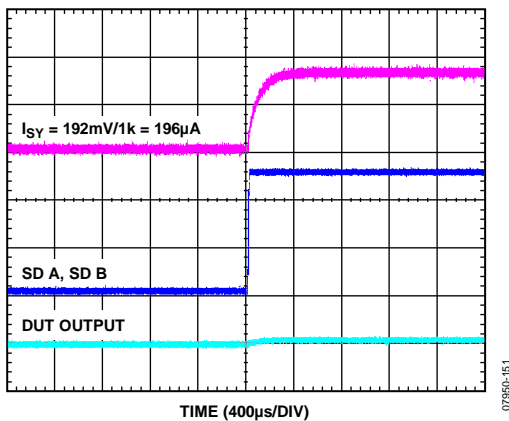


Figure 58. Shutdown Pin Rise Time = 10 μ s

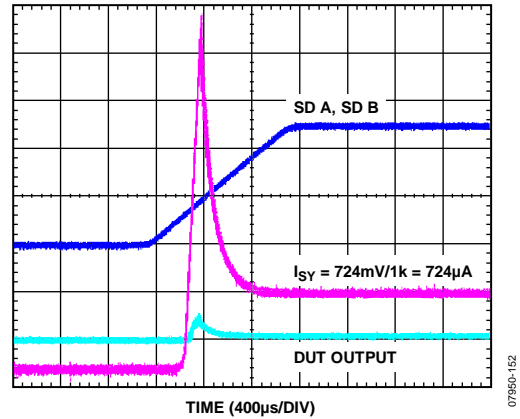


Figure 59. Shutdown Pin Rise Time = 1 ms

INPUT THRESHOLD

The input threshold is approximately 1.2 V above the V- pin when operating on ground and 5 V and 0.9 V when operating on 2.7 V (see Figure 60 and Figure 61). The threshold is relatively stable over temperature. For operation on split supplies, the logic swing may have to be level shifted.

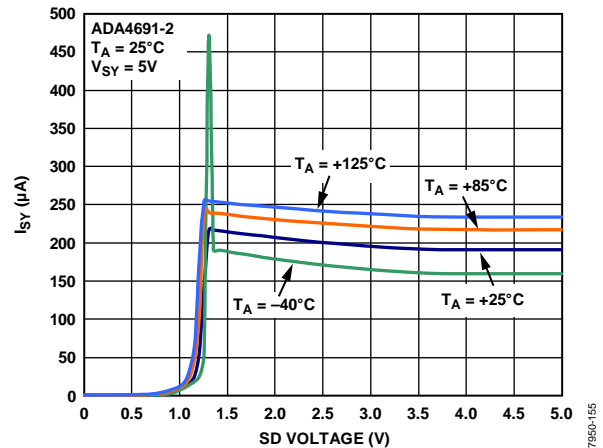


Figure 60. Supply Current vs. Temperature, $V_{SY} = 5V$

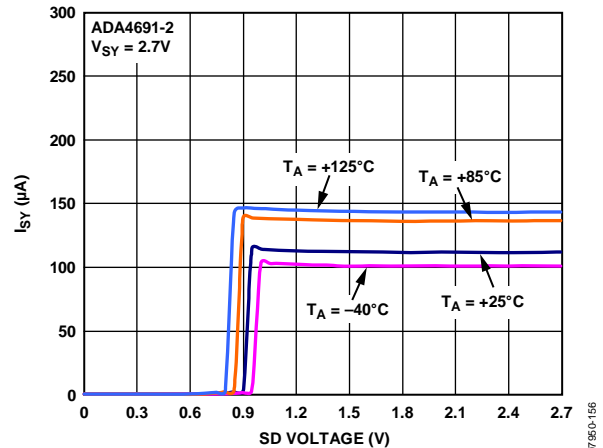
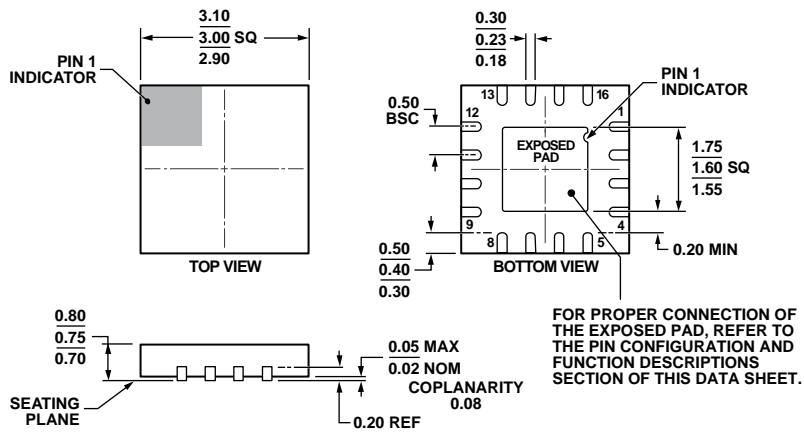


Figure 61. Supply Current vs. Temperature, $V_{SY} = 2.7V$

OUTLINE DIMENSIONS



COMPLIANT TO JEDEC STANDARDS MO-220-WEED.

Figure 62. 16-Lead Lead Frame Chip Scale Package [LFCSP_WQ]
3 mm × 3 mm Body, Very Very Thin Quad
(CP-16-22)

Dimensions shown in millimeters

070209-C

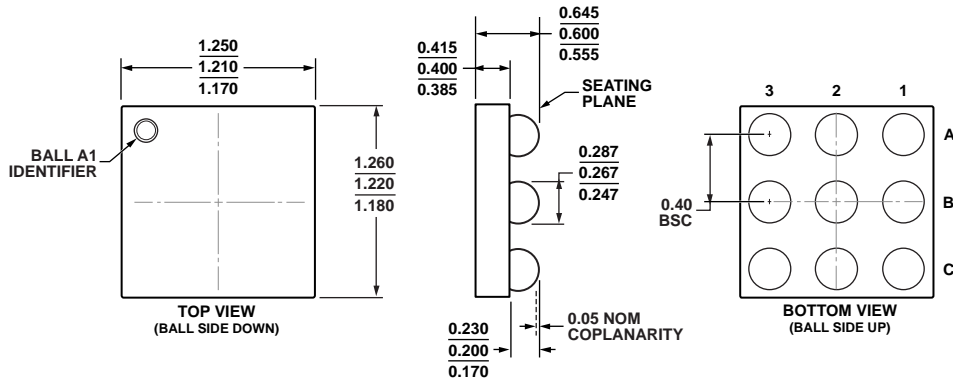


Figure 63. 9-Ball Wafer Level Chip Scale Package [WLCSP]
(CB-9-3)

Dimensions shown in millimeters

091709-A

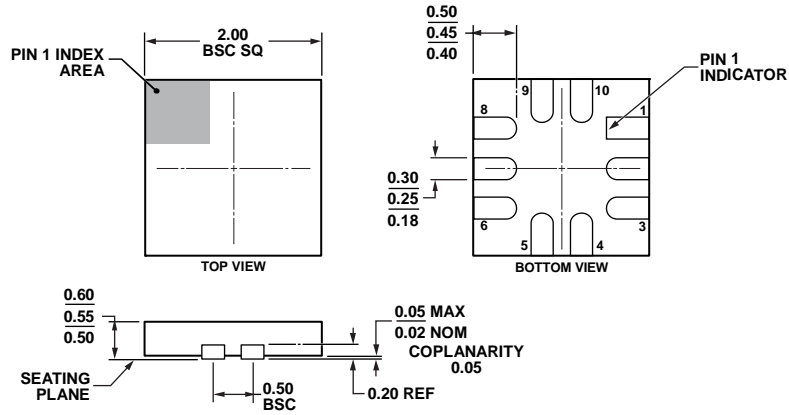


Figure 64. 10-Lead Lead Frame Chip Scale Package [LFCSP_UQ]
 2 mm x 2 mm Body, Ultra Thin Quad
 (CP-10-11)
 Dimensions shown in millimeters

081308-D

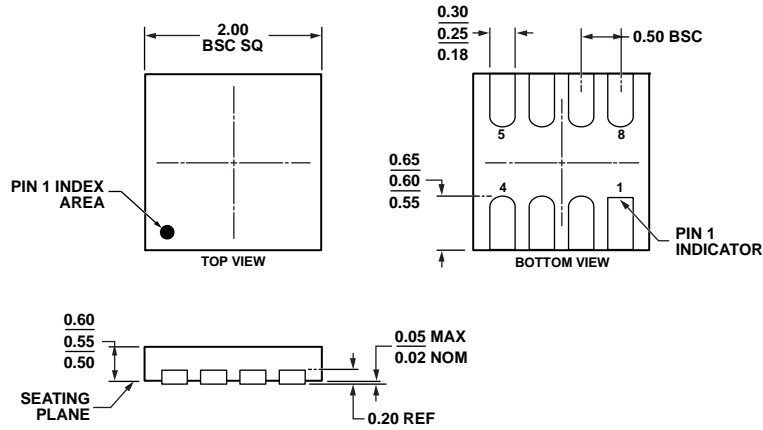
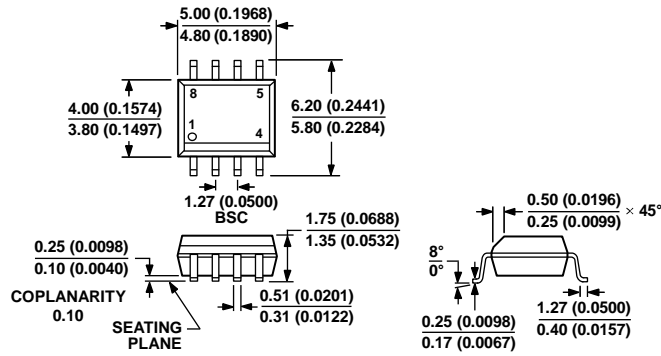


Figure 65. 8-Lead Lead Frame Chip Scale Package [LFCSP_UD]
 2 mm x 2 mm Body, Ultra Thin, Dual Lead
 (CP-8-6)
 Dimensions shown in millimeters

062409-A

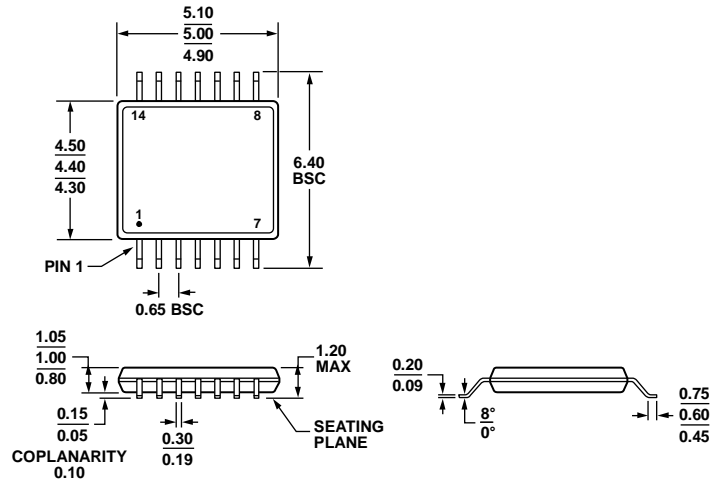


COMPLIANT TO JEDEC STANDARDS MS-012-A A
 CONTROLLING DIMENSIONS ARE IN MILLIMETERS; INCH DIMENSIONS
 (IN PARENTHESES) ARE ROUNDED-OFF MILLIMETER EQUIVALENTS FOR
 REFERENCE ONLY AND ARE NOT APPROPRIATE FOR USE IN DESIGN.

012A07-A

Figure 66. 8-Lead Standard Small Outline Package [SOIC_N]
 Narrow Body
 (R-8)

Dimensions shown in millimeters and (inches)



COMPLIANT TO JEDEC STANDARDS MO-153-AB-1

Figure 67. 14-Lead Thin Shrink Small Outline Package [TSSOP]
 (RU-14)

Dimensions shown in millimeters

061906-A

ADA4691-2/ADA4691-4/ADA4692-2/ADA4692-4

ORDERING GUIDE

Model ¹	Temperature Range	Package Description	Package Option	Branding
ADA4691-2ACBZ-R7	-40°C to +125°C	9-Ball WLCSP	CB-9-3	A2C
ADA4691-2ACBZ-RL	-40°C to +125°C	9-Ball WLCSP	CB-9-3	A2C
ADA4691-2ACPZ-R7	-40°C to +125°C	10-Lead LFCSP_UQ	CP-10-11	A2
ADA4691-2ACPZ-RL	-40°C to +125°C	10-Lead LFCSP_UQ	CP-10-11	A2
ADA4691-4ACPZ-R2	-40°C to +125°C	16-Lead LFCSP_WQ	CP-16-22	A2P
ADA4691-4ACPZ-R7	-40°C to +125°C	16-Lead LFCSP_WQ	CP-16-22	A2P
ADA4691-4ACPZ-RL	-40°C to +125°C	16-Lead LFCSP_WQ	CP-16-22	A2P
ADA4692-2ACPZ-R7	-40°C to +125°C	8-Lead LFCSP_UD	CP-8-6	A3
ADA4692-2ACPZ-RL	-40°C to +125°C	8-Lead LFCSP_UD	CP-8-6	A3
ADA4692-2ARZ	-40°C to +125°C	8-Lead SOIC_N	R-8	
ADA4692-2ARZ-R7	-40°C to +125°C	8-Lead SOIC_N	R-8	
ADA4692-2ARZ-RL	-40°C to +125°C	8-Lead SOIC_N	R-8	
ADA4692-4ARUZ	-40°C to +125°C	14-Lead TSSOP	RU-14	
ADA4692-4ARUZ-RL	-40°C to +125°C	14-Lead TSSOP	RU-14	

¹ Z = RoHS Compliant Part.

Данный компонент на территории Российской Федерации

Вы можете приобрести в компании MosChip.

Для оперативного оформления запроса Вам необходимо перейти по данной ссылке:

<http://moschip.ru/get-element>

Вы можете разместить у нас заказ для любого Вашего проекта, будь то серийное производство или разработка единичного прибора.

В нашем ассортименте представлены ведущие мировые производители активных и пассивных электронных компонентов.

Нашей специализацией является поставка электронной компонентной базы двойного назначения, продукции таких производителей как XILINX, Intel (ex.ALTERA), Vicor, Microchip, Texas Instruments, Analog Devices, Mini-Circuits, Amphenol, Glenair.

Сотрудничество с глобальными дистрибьюторами электронных компонентов, предоставляет возможность заказывать и получать с международных складов практически любой перечень компонентов в оптимальные для Вас сроки.

На всех этапах разработки и производства наши партнеры могут получить квалифицированную поддержку опытных инженеров.

Система менеджмента качества компании отвечает требованиям в соответствии с ГОСТ Р ИСО 9001, ГОСТ РВ 0015-002 и ЭС РД 009

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