

Step-up DC/DC Converter with Shutdown Function

NO.EA-284-130409

OUTLINE

The R1204x Series are low supply current CMOS-based PWM step-up DC/DC converters. Internally, a single IC consists of an NMOS FET, an oscillator, a PWM comparator, a voltage reference unit, an error amplifier, a current limit circuit, an under voltage lockout circuit (UVLO), an over-voltage protection circuit (OVP), a soft-start circuit, a maximum duty cycle limit circuit, and a thermal shutdown protection circuit. By simply using an inductor, a resistor, capacitors and a diode as external components, a high-efficiency step-up DC/DC converter can be easily configured.

As protection functions, the IC contains a thermal shutdown protection circuit, a current limit circuit, an OVP circuit, and an UVLO circuit. A thermal shutdown circuit detects overheating of the ICs and stops the operation to protect it from damage. A current limit circuit limits the peak current of Lx, and an OVP circuit detects the over voltage of output, and an UVLO circuit detects the low input voltage.

The R1204x Series provide the R1204xxxA/D/G/H versions, which are optimized for serial driving of white LEDs with constant current, and the R1204xxxB/C/E/F versions, which are optimized for constant voltage driving. Among the R1204xxxB/C/E/F versions, only the R1204xxxC/F versions are equipped with PWM/VFM auto switching controls. The LED current can be determined by the value of current setting resistor. The brightness of the LEDs can be quickly adjusted by applying a PWM signal (200Hz to 300kHz) to the CE pin. The R1204x Series are available in DFN(PLP)1820-6 and TSOT-23-6 packages.

FEATURES

- Input Voltage Range 2.3V to 5.5V
- Supply Current..... Typ. 800 μ A
- Standby Current..... Max. 5 μ A
- Feedback Voltage..... 0.2V \pm 10mV (R1204xxxxA/D)
0.4V \pm 10mV(R1204xxxxG/H)
1.0V \pm 15mV (R1204xxxxB/C/E/F)
- Lx Current Limit Function Min. 700mA
- Over Voltage Protection 23V, 33V, 42V (\pm 1.5V)
- Oscillator Frequency Typ. 1.0MHz (R1204xxxxA/B/C/G)
Typ. 750kHz (R1204xxxxD/E/F/H)
- Maximum Duty Cycle Min. 91% (R1204xxxxA/B/C/G)
Min. 92% (R1204xxxxD/E/F/H)
- FET ON Resistance..... Typ. 0.8 Ω
- UVLO Function
- Thermal Protection Function
- LED Dimming Control for R1204xxxxA/D by external PWM signal (200Hz to 300kHz frequency)
- Packages..... DFN(PLP)1820-6, TSOT-23-6
- Recommended Bypass Capacitor..... 1.0 μ F

APPLICATIONS

- Constant voltage power source for hand-held equipment
- OLED power supply for hand-held equipment
- White LED driver for hand-held equipment

BLOCK DIAGRAMS

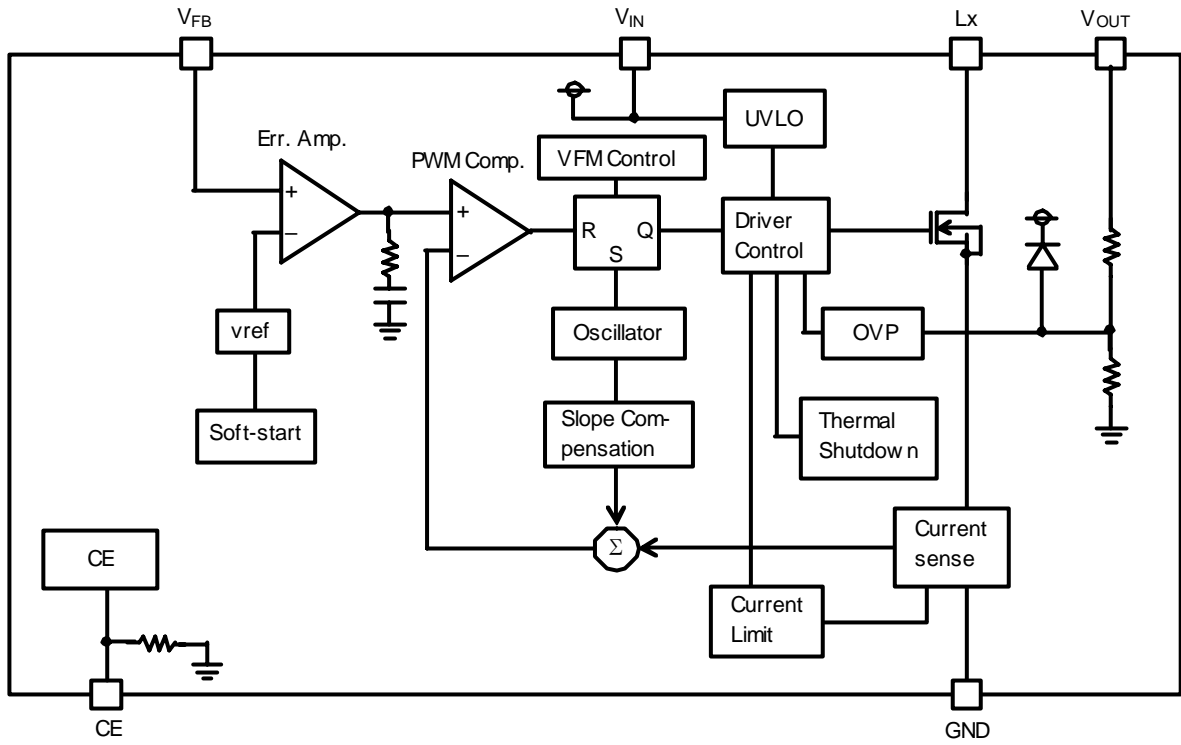
R1204xxxxA/D/G/H



R1204xxxxB/E



R1204xxxxC/F



SELECTION GUIDE

The package type, the OVP detector threshold, the feedback voltage and the PWM/VFM auto switching control are user-selectable options as described below.

| Product Name | Package | Quantity per Reel | Pb Free | Halogen Free |
|------------------|----------------|-------------------|---------|--------------|
| R1204Kxy2z-TR | DFN(PLP)1820-6 | 5,000 pcs | Yes | Yes |
| R1204Nxy3z-TR-FE | TSOT-23-6 | 3,000 pcs | Yes | Yes |

x: Designation of OVP detector threshold

- (1) 23V
- (2) 33V
- (3) 42V

y: Designation of current limit

- (1) Typ. 900mA

z: Designation of the feedback voltage, the oscillator frequency, the PWM/VFM auto switching control

| z | Feedback Voltage | Oscillator Frequency | PWM/VFM Auto Switching Control |
|-----|------------------|----------------------|--------------------------------|
| (A) | Typ. 0.2V | Typ. 1MHz | No |
| (B) | Typ. 1V | | No |
| (C) | Typ. 1V | | Yes |
| (D) | Typ. 0.2V | Typ. 750kHz | No |
| (E) | Typ. 1V | | No |
| (F) | Typ. 1V | | Yes |
| (G) | Typ. 0.4V | Typ. 1MHz | No |
| (H) | | Typ. 750kHz | |

PIN CONFIGURATIONS



PIN DESCRIPTIONS

• DFN(PLP)1820-6

| Pin No | Symbol | Description |
|--------|-----------|-----------------------------------|
| 1 | V_{OUT} | Output Pin |
| 2 | L_X | Switching Pin (Open Drain Output) |
| 3 | GND | Ground Pin |
| 4 | V_{IN} | Input Pin |
| 5 | CE | Chip Enable Pin ("H" Active) |
| 6 | V_{FB} | Feedback Pin |

The exposed tab is substrate level (GND). It is recommended that the exposed tab be connected to the ground plane on the board or otherwise be left open.

• TSOT-23-6

| Pin No | Symbol | Description |
|--------|-----------|-----------------------------------|
| 1 | L_X | Switching Pin (Open Drain Output) |
| 2 | GND | Ground Pin |
| 3 | V_{FB} | Feedback Pin |
| 4 | CE | Chip Enable Pin ("H" Active) |
| 5 | V_{OUT} | Output Pin |
| 6 | V_{IN} | Input Pin |

ABSOLUTE MAXIMUM RATINGS

(GND=0V)

| Symbol | Item | Rating | Unit | |
|-----------|-----------------------------------------------------------------|----------------|------|----|
| V_{IN} | V_{IN} Pin Voltage | -0.3 to 6.5 | V | |
| V_{CE} | CE Pin Voltage | -0.3 to 6.5 | V | |
| V_{FB} | V_{FB} Pin Voltage | -0.3 to 6.5 | V | |
| V_{OUT} | V_{OUT} Pin Voltage | -0.3 to 48 | V | |
| V_{LX} | L_X Pin Voltage | -0.3 to 48 | V | |
| I_{LX} | L_X Pin Current | 1200 | mA | |
| P_D | Power Dissipation (Standard Test Land Pattern) ^{*1} | DFN(PLP)1820-6 | 880 | mW |
| | | TSOT-23-6 | 460 | |
| T_a | Operating Temperature Range | -40 to +85 | °C | |
| T_{stg} | Storage Temperature Range | -55 to +125 | °C | |

^{*1} For more information about Power Dissipation and Standard Land Pattern, please refer to *PACKAGE INFORMATION*.

ABSOLUTE MAXIMUM RATINGS

Electronic and mechanical stress momentarily exceeded absolute maximum ratings may cause the permanent damages and may degrade the lifetime and safety for both device and system using the device in the field. The functional operation at or over these absolute maximum ratings is not assured.

RECOMMENDED OPERATING CONDITIONS (ELECTRICAL CHARACTERISTICS)

All of electronic equipment should be designed that the mounted semiconductor devices operate within the recommended operating conditions. The semiconductor devices cannot operate normally over the recommended operating conditions, even if when they are used over such conditions by momentary electronic noise or surge. And the semiconductor devices may receive serious damage when they continue to operate over the recommended operating conditions.

ELECTRICAL CHARACTERISTICS

(Ta=25°C)

| Symbol | Item | Conditions | Min. | Typ. | Max. | Unit | |
|--------------|---------------------------------------|------------------------------------------------|----------------------|-----------------|----------------|-------------|-----|
| V_{IN} | Operating Input Voltage | | 2.3 | | 5.5 | V | |
| I_{DD} | Supply Current | $V_{IN}=5.5V$, $V_{FB}=0V$, L_X at no load | | 0.8 | | mA | |
| Istandby | Standby Current | $V_{IN}=5.5V$, $V_{CE}=0V$ | | 1.0 | 5.0 | μA | |
| V_{UVLO1} | UVLO Detector Threshold | V_{IN} falling | 1.9 | 2.0 | 2.1 | V | |
| V_{UVLO2} | UVLO Released Voltage | V_{IN} rising | | $V_{UVLO1}+0.1$ | | V | |
| V_{CEH} | CE Input Voltage "H" | $V_{IN}=5.5V$ | 1.5 | | | V | |
| V_{CEL} | CE Input Voltage "L" | $V_{IN}=2.3V$ | | | 0.5 | V | |
| R_{CE} | CE Pull Down Resistance | $V_{IN}=3.6V$ | R1204xxxxA/B/D/E/G/H | | 1200 | k Ω | |
| | | | R1204xxxxC/F | | 600 | | |
| V_{FB} | V_{FB} Voltage Accuracy | $V_{IN}=3.6V$ | R1204xxxxA/D | 0.19 | 0.2 | 0.21 | V |
| | | | R1204xxxxG/H | 0.39 | 0.4 | 0.41 | |
| | | | R1204xxxxB/C/E/F | 0.985 | 1.000 | 1.015 | |
| I_{FB} | V_{FB} Input Current | $V_{IN}=5.5V$, $V_{FB}=0V$ or $5.5V$ | -0.1 | | 0.1 | μA | |
| Tstart | Soft-start Time | $V_{IN}=3.6V$, R1204xxxxB/C/E/F | | 10 | | ms | |
| R_{ON} | FET ON Resistance | $V_{IN}=3.6V$, $I_{LX}=100mA$ | | 0.8 | | Ω | |
| I_{LXLEAK} | FET Leakage Current | $V_{LX}=40V$ | | | 3.0 | μA | |
| I_{LXLIM} | FET Current Limit | $V_{IN}=3.6V$ | 700 | 900 | 1100 | mA | |
| fosc | Oscillator Frequency | $V_{IN}=3.6V$, $V_{FB}=0V$ | R1204xxxxA/B/C/G | 0.9 | 1.0 | 1.1 | MHz |
| | | | R1204xxxxD/E/F/H | 675 | 750 | 825 | kHz |
| Maxduty | Oscillator Maximum Duty Cycle | $V_{IN}=3.6V$, $V_{FB}=0V$ | R1204xxxxA/B/C/G | 91 | | | % |
| | | | R1204xxxxD/E/F/H | 92 | | | % |
| V_{OVP1} | OVP Detector Threshold | $V_{IN}=3.6V$, V_{OUT} rising | R1204x1xxx | 22 | 23 | 24.0 | V |
| | | | R1204x2xxx | 31.5 | 33 | 34.5 | |
| | | | R1204x3xxx | 40.2 | 42 | 43.8 | |
| V_{OVP2} | OVP Released Voltage | $V_{IN}=3.6V$, V_{OUT} falling | R1204x1xxx | | $V_{OVP1}-0.6$ | | V |
| | | | R1204x2xxx | | $V_{OVP1}-1.2$ | | |
| | | | R1204x3xxx | | $V_{OVP1}-2.4$ | | |
| T_{TSD} | Thermal Shutdown Temperature | $V_{IN}=3.6V$ | | 150 | | $^{\circ}C$ | |
| T_{TSR} | Thermal Shutdown Released Temperature | $V_{IN}=3.6V$ | | 100 | | $^{\circ}C$ | |

All test items listed under Electrical Characteristics are done under the pulse load condition ($T_j \approx T_a = 25^{\circ}C$).

TYPICAL APPLICATIONS AND NOTES CONCERNING EXTERNAL COMPONENTS

• Typical Applications

R1204xxxxA/D/G/H



Figure 1.



Figure 2.

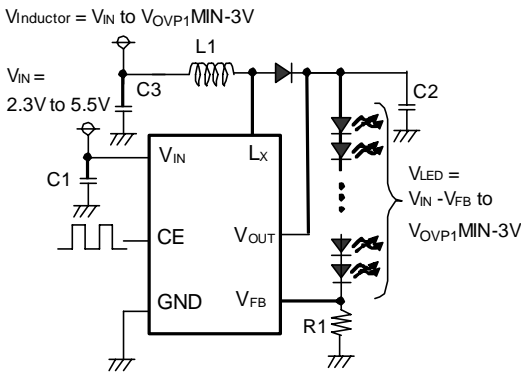


Figure 3.

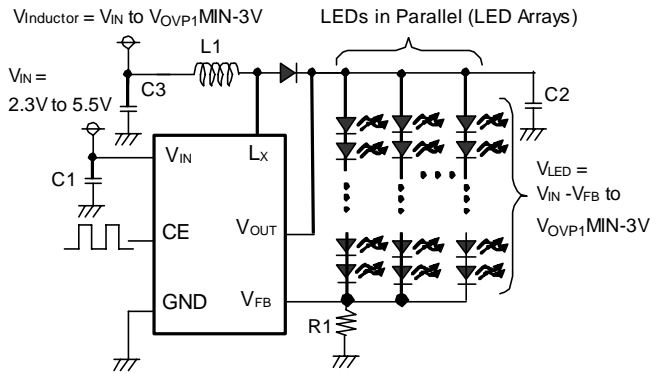


Figure 4.

R1204xxxxB/C/E/F



Figure 5.



Figure 6.

Table 1. Recommended Inductors

| L1 (μH) | Parts No. | Rated Current (mA) | Size (mm) | Versions |
|----------------------|------------------|--------------------|-------------|------------------|
| 10 | VLS252010ET-100M | 550 | 2.5×2.0×1.0 | R1204xxxxA/B/C/G |
| 10 | VLF302512MT-100M | 620 | 3.0×2.5×1.2 | |
| 10 | VLF403212MT-100M | 900 | 4.0×3.2×1.2 | |
| 22 | VLF302512MT-220M | 430 | 3.0×2.5×1.2 | R1204xxxxD/E/F/H |
| 22 | VLF403212MT-220M | 540 | 4.0×3.2×1.2 | |
| 22 | VLF504012MT-220M | 800 | 5.0×4.0×1.2 | |

Table 2. Recommended Components for R1204xxxxA/D/G/H

| Symbol | Rated Voltage (V) | Parts No. |
|--------|-------------------|--------------------------------------------------------------------|
| D1 | 60 | CRS12 |
| C1 | 6.3 | CM105B105K06 |
| C2 | 50 | C2012X5R1H105K |
| | | R1204xxxxG/H ($I_{\text{LED}} > 22\text{mA}$): C2012X5R1H225K |
| C3 | 16 | C2012X5R1C105K |

| Symbol | R1204x | Values |
|--------|--------------|----------------------------------------------|
| R1 | R1204xxxxA/D | (10/ LED Arrays ^{*1}) (Ω) |
| | R1204xxxxG/H | (20/ LED Arrays ^{*1}) (Ω) |

*1 LED Arrays indicate the number of parallel LEDs in series.

Table 3. Recommended Components for R1204xxxxB/C/E/F

| Symbol | Rated Voltage (V) | Parts No. |
|--------|-------------------|----------------|
| D1 | 60 | CRS12 |
| C1 | 6.3 | CM105B105K06 |
| C2 | 16 | C2012X5R1C475K |
| | 25 | C2012X5R1E105K |
| | 50 | C2012X5R1H105K |
| C4 | 16 | C2012X5R1C105K |

Table 4. Recommended Component Values for R1204xxxxB/C/E/F

| V_{SET} (V) | $7 < V_{\text{SET}} \leq 10$ | $10 < V_{\text{SET}} \leq 25$ | $25 < V_{\text{SET}}$ |
|----------------------|----------------------------------|----------------------------------|----------------------------------|
| R1 (k Ω) | 10 | 10 | 10 |
| R2 (k Ω) | $(V_{\text{SET}} - 1) \times R1$ | $(V_{\text{SET}} - 1) \times R1$ | $(V_{\text{SET}} - 1) \times R1$ |
| R3 (Ω) | 0 | 0 | 0 |
| C1 (μF) | 1.0 | 1.0 | 1.0 |
| C2 (μF) | 4.7 | 1.0×2 | 1.0 |
| C3 (pF) | 10 | 10 | 10 |
| C4 (μF) | 1.0 | 1.0 | 1.0 |

Notes Concerning External Components

- **LED Current Setting (R1204xxxxA/D/G/H)**

The LED current (I_{LED}) when a "H" signal is applied to the CE pin (Duty=100%) can be determined by the value of feedback resistor (R1).

$$I_{LED} = 0.2 / R1 \quad (\text{R1204xxxxA/D})$$

$$I_{LED} = 0.4 / R1 \quad (\text{R1204xxxxG/H})$$

- **LED Dimming Control (R1204xxxxA/D/G/H)**

The brightness of the LEDs can be adjusted by applying a PWM signal to the CE pin. By inputting "L" voltage for a certain period of time (Typ. 9ms(R1204xxxxA/G) / 12ms(R1204xxxxD/H) or more), the IC goes into standby mode and turns off LEDs. I_{LED} can be controlled by the duty of a PWM signal for the CE pin.

The relation between the high-duty of the CE pin (Hduty) and I_{LED} is calculatable by the following formula.

$$I_{LED} = Hduty \times V_{FB} / R1$$

The frequency range of a PWM signal should be set within the range of 200Hz to 300kHz.

In the case of using a 20kHz or less PWM signal for dimming the LEDs, the increasing or decreasing of the inductor current (I_L) may make noise in the audible band. In this case, a high frequency PWM signal should be used.

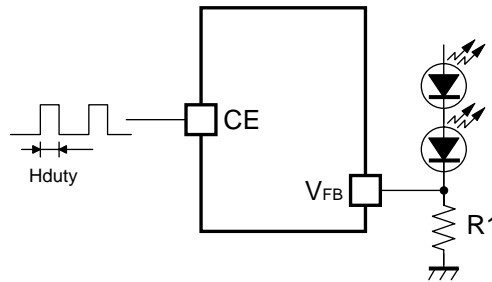


Figure 7. Dimming Control by CE Pin

- **Low luminance Dimming Accuracy (R1204xxxxG/H)**

Low luminance Dimming filtered V_{FB} voltage tolerance depends on the offset voltage of the internal DC/DC converter. By this offset voltage, some voltage difference may be generated between V_{REF} voltage and V_{FB} voltage. Low luminance Dimming Accuracy is shown in Table.5

Table 5. Low luminance Dimming Accuracy for R1204xxxxG/H (R1=20ohm)

| The duty of a PWM signal for the CE pin | I_{LED} Min. | I_{LED} Max. |
|-----------------------------------------|---------------------|--------------------|
| 3.5% (Frequency = 20kHz ~ 300kHz) | 0.01mA ² | 2.1mA ² |

² guaranteed by design engineering (Ta=25 °C)

• Output Voltage Setting (R1204xxxxB/C/E/F)

The relation between the output voltage (V_{SET}) and the resistors ($R1$, $R2$) is calculatable by the following formula.

$$V_{SET} = V_{FB} \times (R1 + R2) / R1$$

The sum of $R1$ and $R2$ should be 300k Ω or less.

Ensure the V_{IN} and GND lines are sufficiently robust. If their impedances are too high, noise pickup or unstable operation may result. Set a capacitor ($C2$) with a suitable voltage resistance (more than 1.5 times of V_{SET}) between the V_{IN} and GND pins, and as close as possible to the pins.

• PWM/VFM Auto Switching Control (R1204xxxxC/F)

In low output current, the IC automatically switches to high-efficiency VFM mode. The minimum Onduty (D_{ON_MIN}) of VFM mode is set to approximately 30% and is fixed inside the IC. If the difference between the voltages of the input and the output is small, or the Onduty in continuous mode (D_{ON_CON}) becomes lower than D_{ON_MIN} , the IC will not shift to PWM mode but will stay with VFM mode instead even in high output current, as a result, the ripple current will be increased. D_{ON_MIN} should be 70% or more ($V_{SET} > V_{IN} \times 3.33$).

• Current Limit Function

If the peak current of inductor (IL_{max}) exceeds the current limit, current limit function turns the driver off and turns it on in every switching cycle to continually monitor the driver current.

• Soft-Start Function

(R1204xxxxA/D/G/H)

Unless otherwise V_{OUT} is beyond the threshold ($V_f \times$ number of LED lights), current will not flow through LEDs, as a result, V_{FB} voltage will not increase. The IC increases V_{OUT} by controlling the output of error amplifier to "H" and turning the L_x switch on and off for a certain period of time (until the current flow). At the mean time, the inrush current is controlled by gradually increasing the current limit. If V_{OUT} is over the threshold (the current flows), the IC controls the soft-start function by gradually increasing the reference voltage of error amplifier.

(R1204xxxxB/C/E/F)

The IC controls the soft-start function by gradually increasing the reference voltage of error amplifier. Soft-start begins when the output voltage of error amplifier is 0V and ends when it reaches the constant voltage.

• Under Voltage Lock Out Function (UVLO)

UVLO function stops DC/DC operation and prevents malfunction when the supply voltage falls below the UVLO detector threshold.

• Over Voltage Protection Circuit (OVP)

OVP circuit monitors the V_{OUT} pin voltage and if it reaches the OVP voltage it will stop oscillation. When the V_{OUT} pin voltage decreases it will restart oscillation, but if the cause of the excess V_{OUT} pin voltage is not removed the OVP circuit will operate repeatedly so as to restrict the V_{OUT} pin voltage.

• Thermal Shutdown Function

If the junction temperature exceeds the thermal shutdown temperature, thermal shutdown function turns the driver off. If the junction temperature becomes lower than the thermal shutdown released temperature, the thermal shutdown function resets the IC to restart the operation.

- **Selection of Capacitor**

Set a 1 μ F or more input capacitor (C1) between the V_{IN} and GND pins as close as possible to the pins.

Set a 1 μ F to 10 μ F output capacitor (C2) between the V_{OUT} and GND pins.

R1204xxxxA/D: Set C2 with 1 μ F to 4.7 μ F between the V_{OUT} and GND pins.

R1204xxxxB/C/E/F: Set C2 with 1 μ F to 10 μ F between the V_{OUT} and GND pins.

- **Selection of Inductor**

IL_{max} under steady operation can be calculated by the formula below.

$$IL_{max} = 1.25 \times I_{LED} \times V_{OUT} / V_{IN} + 0.5 \times V_{IN} \times (V_{OUT} - V_{IN}) / (L \times V_{OUT} \times f_{osc})$$

When starting up the IC or adjusting the brightness of LED lights, a large transient current may flow into an inductor (L1). IL_{max} of L1 should be equal or smaller than the current limit of the IC. L with 10 μ H to 22 μ H is recommended.

- **Other External Components Settings**

Set a capacitor (C3) between the V_{OUT} and V_{FB} pins to improve the response of DC/DC converter by giving high-frequency voltage feedback. Please note that C3 operation could be different from the theory of operation depending on component layouts and parasitic capacitances.

OPERATION OF STEP-UP DC/DC CONVERTER AND OUTPUT CURRENT

Figure 8. Basic Circuits



Figure 9. The inductor current (IL) flowing through the inductor (L)



Discontinuous Mode

Continuous Mode

There are two operation modes in the PWM step-up DC/DC converter: continuous mode and discontinuous mode.

When a transistor is in the On-state, the voltage to be applied to L is described as V_{IN} . An increase in the inductor current (i_1) can be written as follows:

$$\Delta i_1 = V_{IN} \times t_{on} / L \dots\dots\dots \text{Formula 1}$$

In the step-up circuit, the energy accumulated during the On-state is transferred into the capacitor even in the Off-state. A decrease in the inductor current (i_2) can be written as follows:

$$\Delta i_2 = (V_{OUT} - V_{IN}) \times t_{open} / L \dots\dots\dots \text{Formula 2}$$

In the PWM switching control, i_1 and i_2 become continuous when $t_{open}=t_{off}$, which is called continuous mode.

When the IC is in the continuous mode and operates in steady-state conditions, the variations of i_1 and i_2 are same:

$$V_{IN} \times t_{on} / L = (V_{OUT} - V_{IN}) \times t_{off} / L \dots\dots\dots \text{Formula 3}$$

Therefore, the duty cycle in the continuous mode is:

$$\text{Duty} = t_{on} / (t_{on} + t_{off}) = (V_{OUT} - V_{IN}) / V_{OUT} \dots\dots\dots \text{Formula 4}$$

When $t_{open}=t_{off}$, the average of I_L is:

$$I_L (\text{Ave.}) = V_{IN} \times t_{on} / (2 \times L) \dots\dots\dots \text{Formula 5}$$

If the input voltage (V_{IN}) is equal to V_{OUT} , the output current (I_{OUT}) is:

$$I_{OUT} = V_{IN}^2 \times t_{on} / (2 \times L \times V_{OUT}) \dots\dots\dots \text{Formula 6}$$

If I_{OUT} is larger than Formula 6, the IC switches to the continuous mode.

I_{Lmax} flowing through L is:

$$I_{Lmax} = I_{OUT} \times V_{OUT} / V_{IN} + V_{IN} \times t_{on} / (2 \times L) \dots\dots\dots \text{Formula 7}$$

$$I_{Lmax} = I_{OUT} \times V_{OUT} / V_{IN} + V_{IN} \times T \times (V_{OUT} - V_{IN}) / (2 \times L \times V_{OUT}) \dots\dots\dots \text{Formula 8}$$

As a result, I_{Lmax} becomes larger compared to I_{OUT} .

When considering the input and output conditions or selecting the external components, please pay attention to I_{Lmax} .

- ★ The above calculations are based on the ideal operation of the ICs in the continuous mode. They do not include the losses caused by the external components or L_x switch. The actual maximum output current will be 50% to 80% of the above calculation results. Especially, if I_L is large or V_{IN} is low, it may cause the switching losses. As for V_{OUT} , please consider V_F of the diode (approximately 0.8V).

TYPICAL CHARACTERISTICS

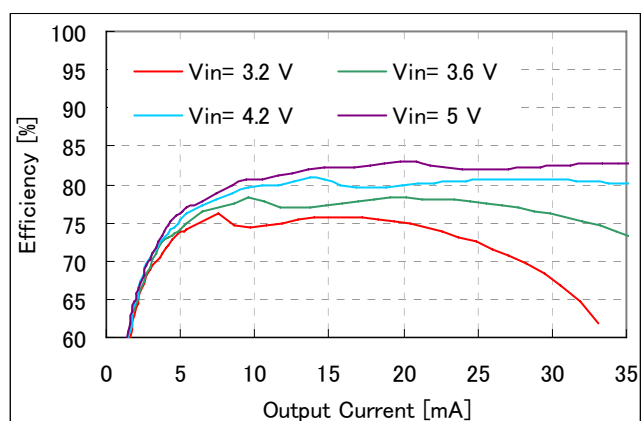
1) Efficiency vs. Output Current of R1204xxxxA/D/G/H

(LED used: NSSW208A (NICHIA) $V_f=3.0V$ ($I_{LED}=20mA$))

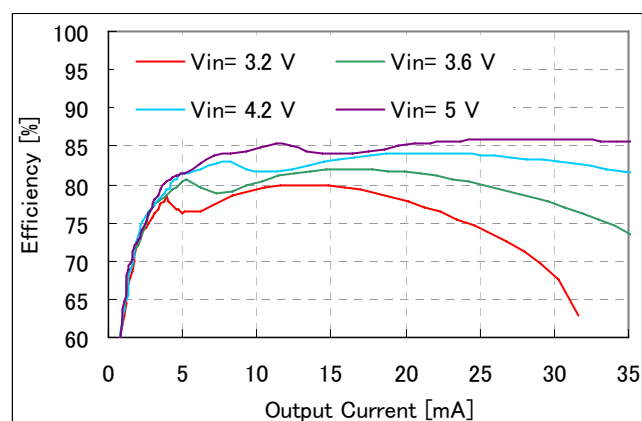
1-1) Efficiency vs. Output Current with Different Output Voltages

10 LEDs in Series ($V_{OUT}=30V$ ($I_{OUT}=20mA$))

R1204xxxxA/G, $L=10\mu H$ (VLF302512MT-100M)

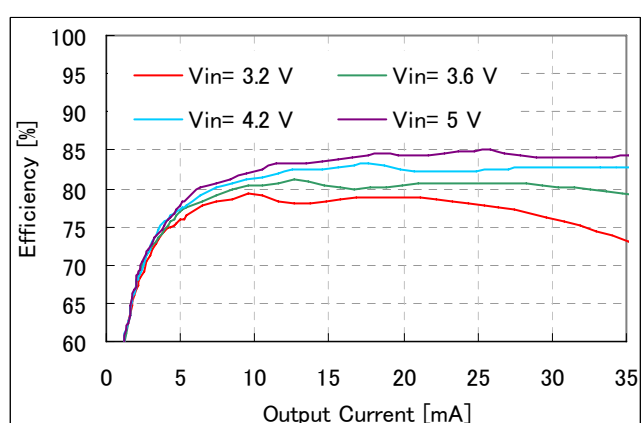


R1204xxxxD/H, $L=22\mu H$ (VLF302512MT-220M)

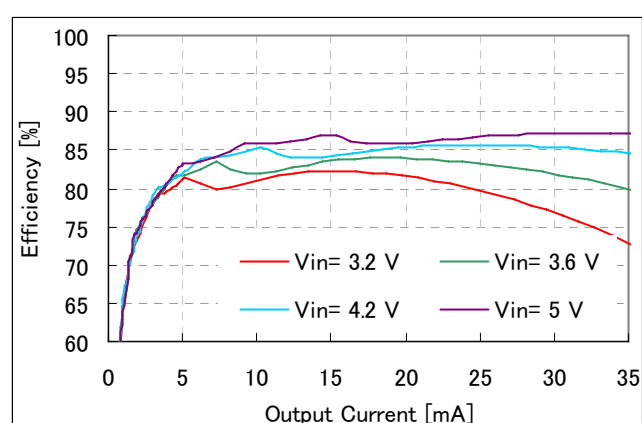


8 LEDs in Series ($V_{OUT}=24V$ ($I_{OUT}=20mA$))

R1204xxxxA/G, $L=10\mu H$ (VLF302512MT-100M)

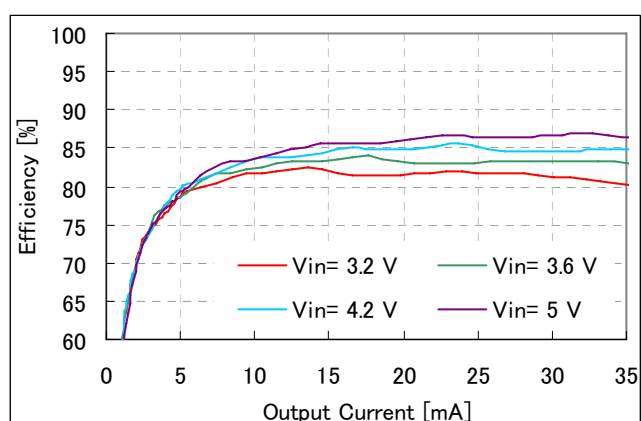


R1204xxxxD/H, $L=22\mu H$ (VLF302512MT-220M)

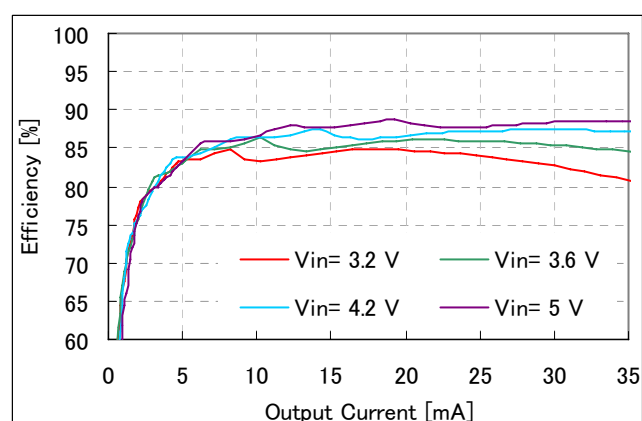


6 LEDs in Series ($V_{OUT}=18V$ ($I_{OUT}=20mA$))

R1204xxxxA/G, $L=10\mu H$ (VLF302512MT-100M)



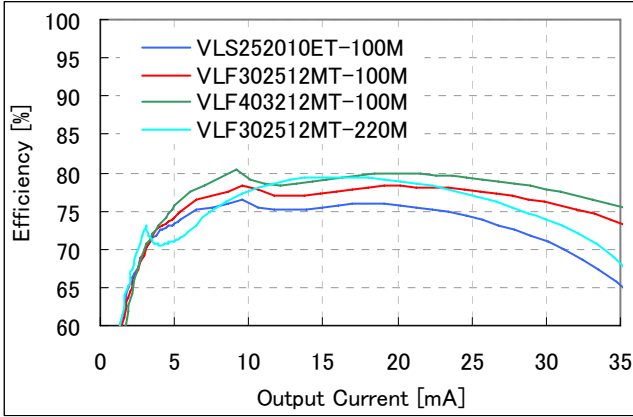
R1204xxxxD/H, $L=22\mu H$ (VLF302512MT-220M)



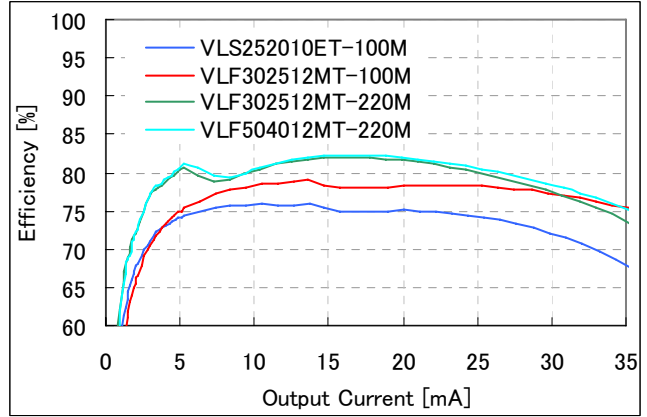
1-2) Efficiency vs. Output Current with Different Inductors ($V_{IN}=3.6V$)

10 LEDs in Series ($V_{OUT}=30V$ ($I_{OUT}=20mA$))

R1204xxxxA/G

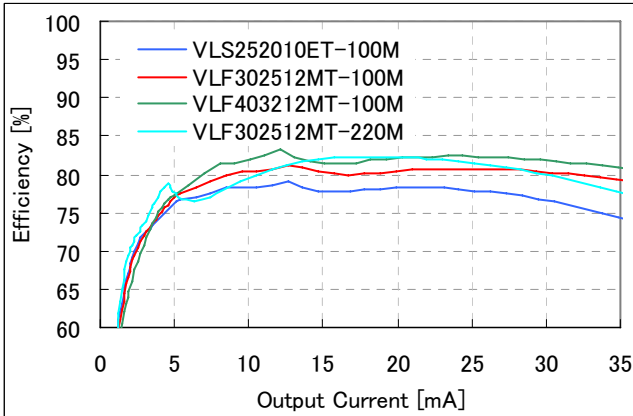


R1204xxxxD/H

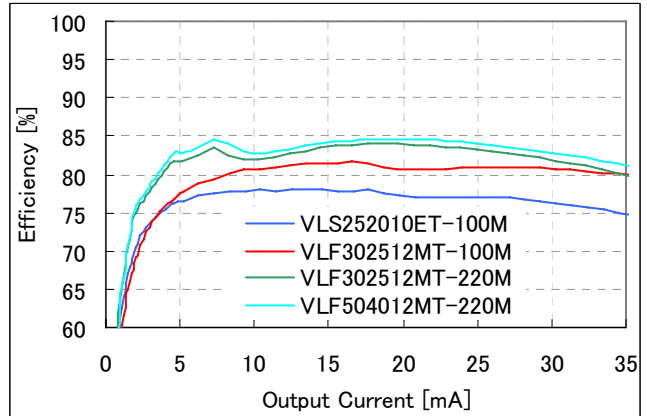


8 LEDs in Series ($V_{OUT}=24V$ ($I_{OUT}=20mA$))

R1204xxxxA/G



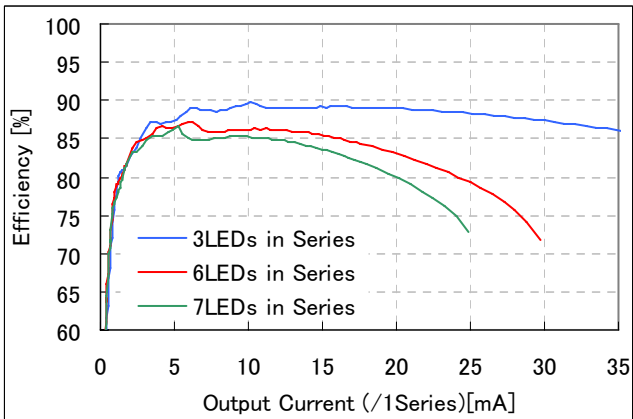
R1204xxxxD/H



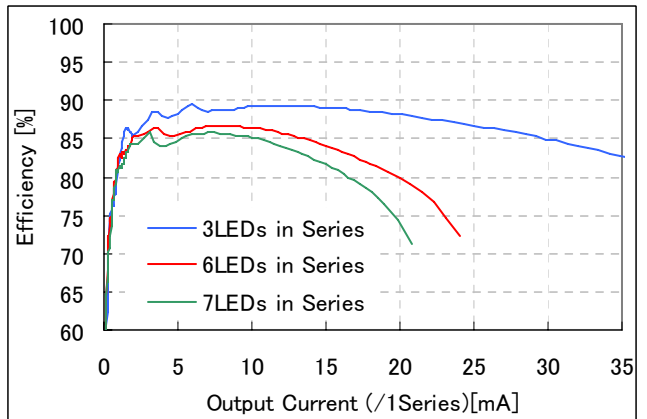
1-3) Efficiency vs. Output Current with Different Numbers of LEDs

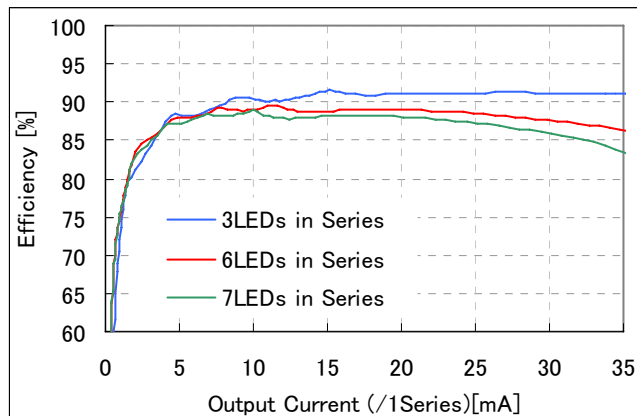
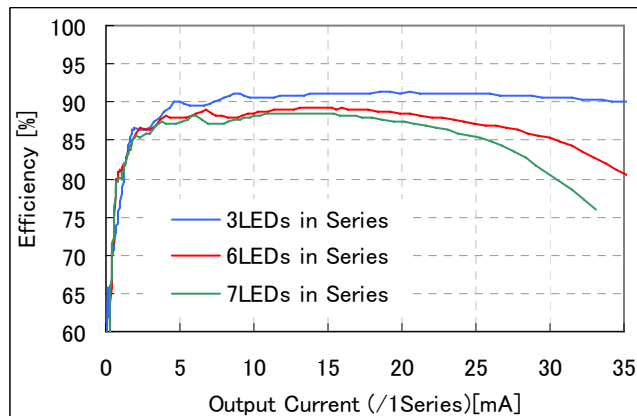
LEDs in 3 Parallels ($V_{IN}=3.6V$)

R1204xxxxA/G, $L=10\mu H$ (VLF302512MT-100M)

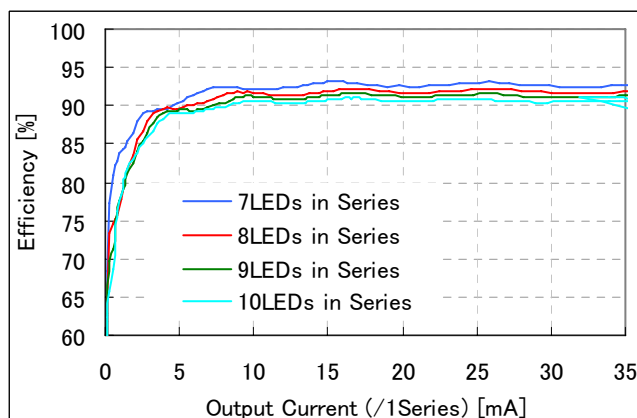
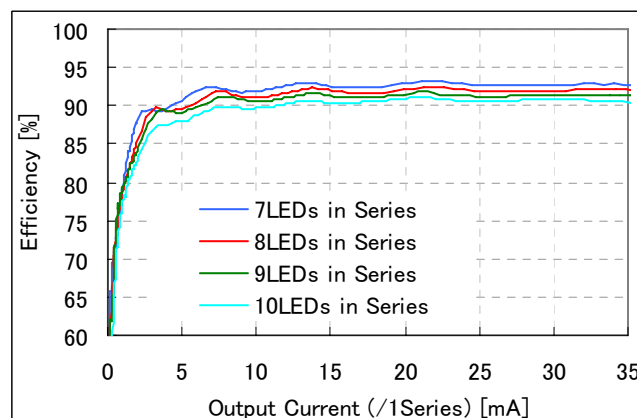
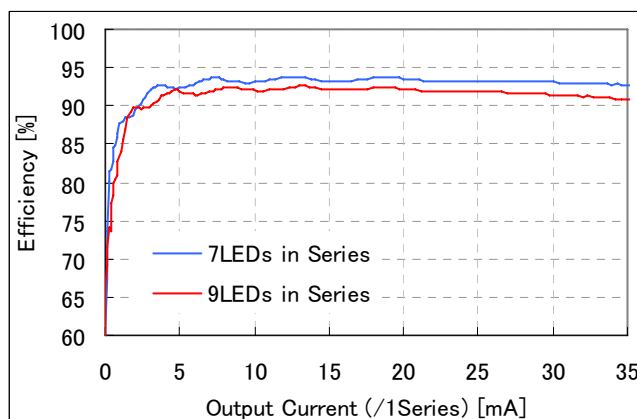
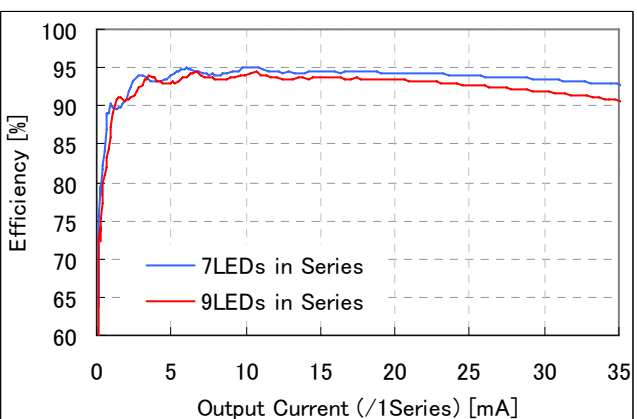


R1204xxxxD/H, $L=22\mu H$ (VLF302512MT-220M)



LEDs in 3 Parallels ($V_{IN}=5.0V$)R1204xxxxA/G, $L=10\mu H$ (VLF302512MT-100M)R1204xxxxD/H, $L=22\mu H$ (VLF302512MT-220M)

1-4) Efficiency vs. Output Current with Different Numbers of LEDs

LEDs in 3 Parallels ($V_{IN}=3.6V$, Inductor Voltage=12.0V)R1204xxxxA/G, $L=10\mu H$ (VLF302512MT-100M)R1204xxxxD/H, $L=22\mu H$ (VLF302512MT-220M)LEDs in 6 Parallels ($V_{IN}=3.6V$, Inductor Voltage=12.0V)R1204xxxxA/G, $L=10\mu H$ (VLF302512MT-100M)R1204xxxxD/H, $L=22\mu H$ (VLF302512MT-220M)

2) Efficiency vs. Output Current of R1204xxxxB/C/E/F

2-1) Efficiency vs. Output Current with Different Output Voltages

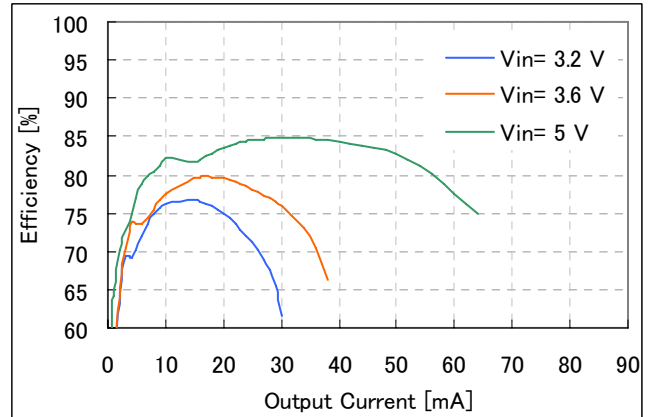
$V_{SET}=31V$

V_{IN} = Inductor Voltages

R1204xxxxC, L=10 μ H (VLF302512MT-100M)

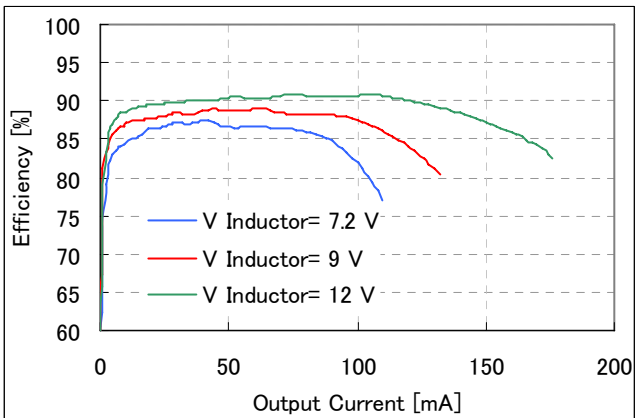


R1204xxxxF, L=22 μ H (VLF302512MT-220M)



Different V_{IN} / Inductor Voltages ($V_{IN}=3.6V$)

R1204xxxxC, L=10 μ H (VLF302512MT-100M)



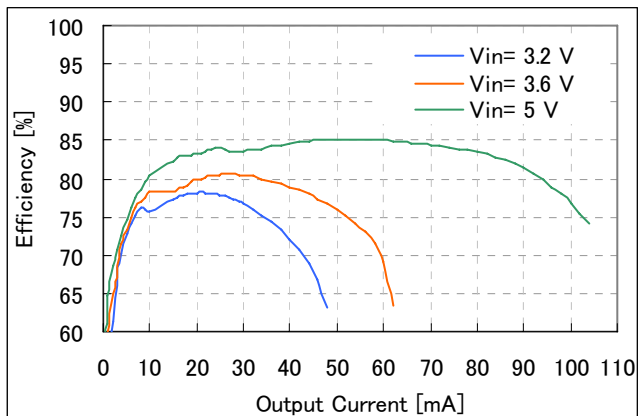
R1204xxxxF, L=22 μ H (VLF302512MT-220M)



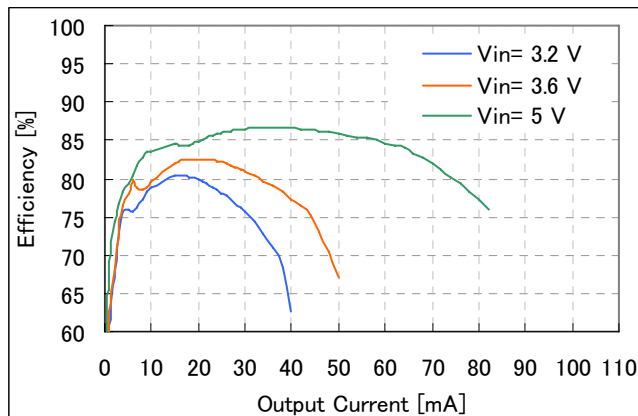
$V_{SET}=25V$

V_{IN} = Inductor Voltages

R1204xxxxC, L=10 μ H (VLF302512MT-100M)

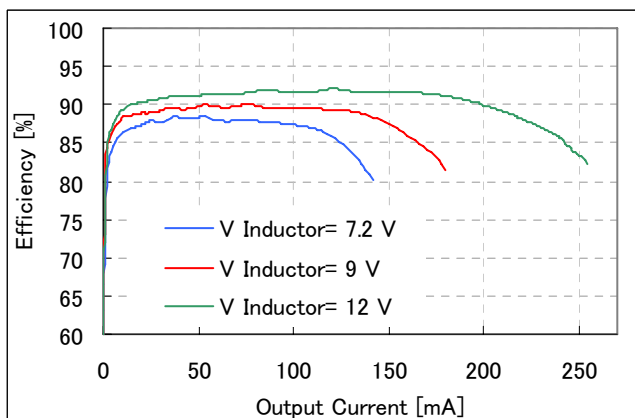


R1204xxxxF, L=22 μ H (VLF302512MT-220M)

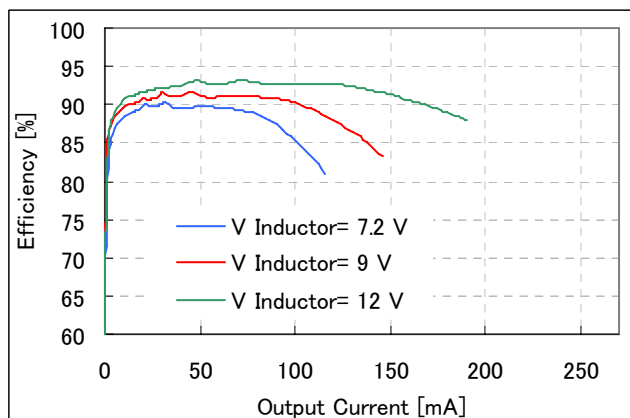


Different V_{IN} / Inductor Voltages ($V_{IN}=3.6V$)

R1204xxxxC, L=10 μ H (VLF302512MT-100M)



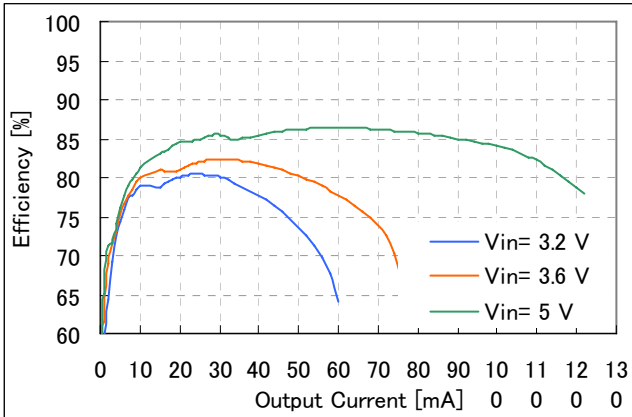
R1204xxxxF, L=22 μ H (VLF302512MT-220M)



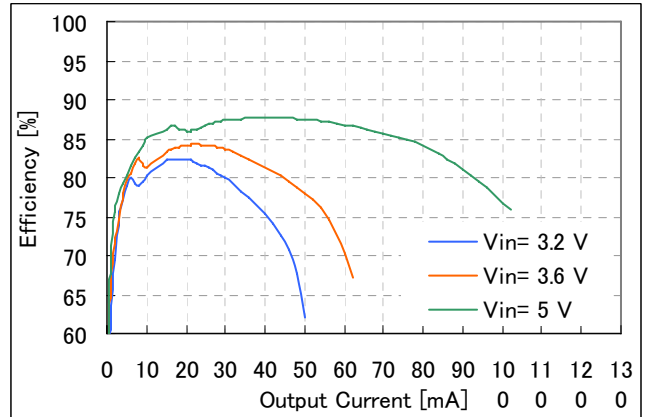
$V_{SET}=21V$

V_{IN} =Inductor Voltages

R1204xxxxC, L=10 μ H (VLF302512MT-100M)

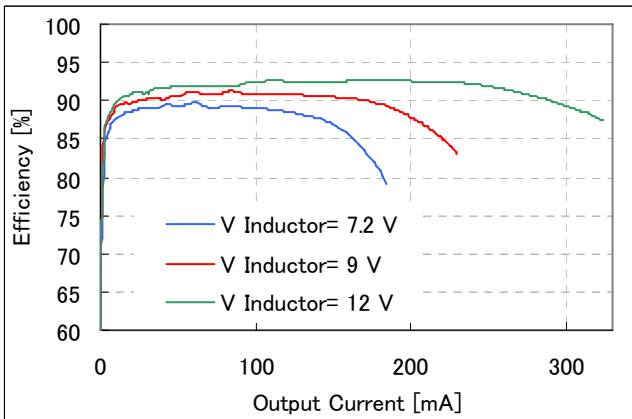


R1204xxxxF, L=22 μ H (VLF302512MT-220M)

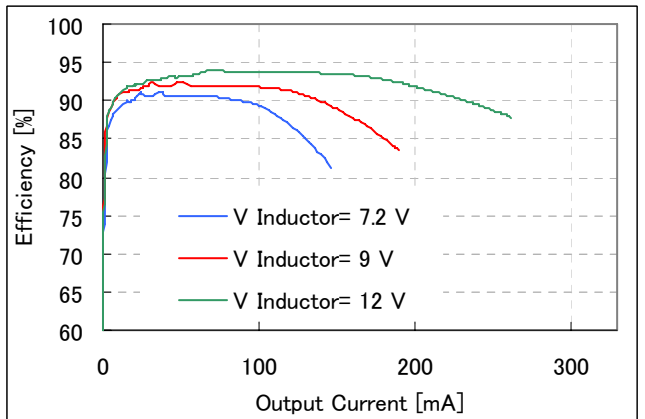


Different V_{IN} / Inductor Voltages ($V_{IN}=3.6V$)

R1204xxxxC, L=10 μ H (VLF302512MT-100M)

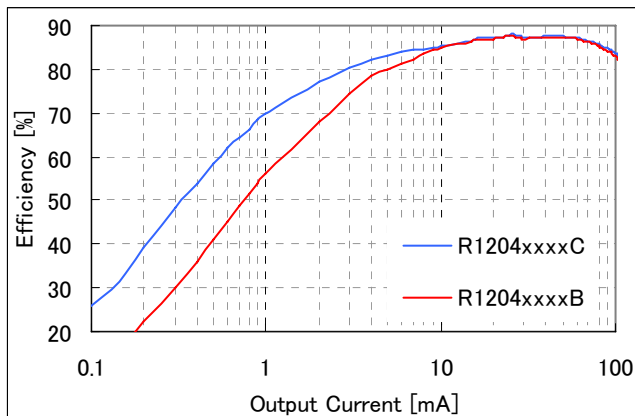


R1204xxxxF, L=22 μ H (VLF302512MT-220M)

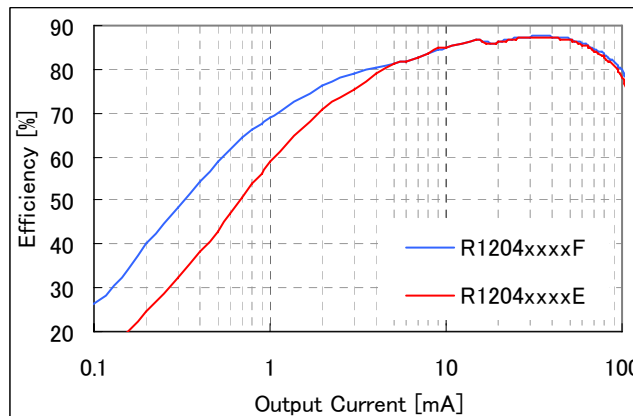


2-2) Efficiency vs. Output Current with PWM Control and PWM/VFM Auto Switching Control
 ($V_{IN}=3.6V$, $V_{SET}=12V$)

R1204xxxxB/C, L=10 μ H (VLF302512MT-100M)

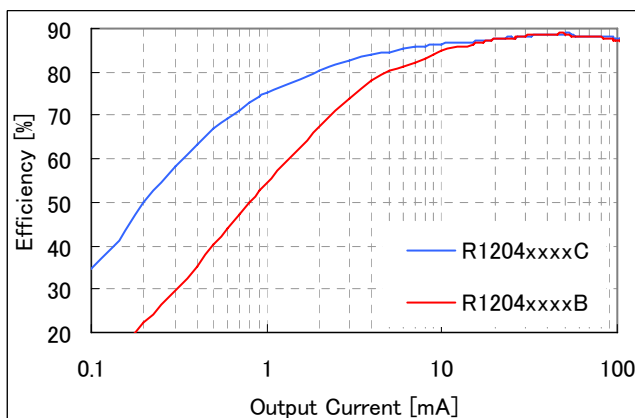


R1204xxxxE/F, L=22 μ H (VLF302512MT-220M)



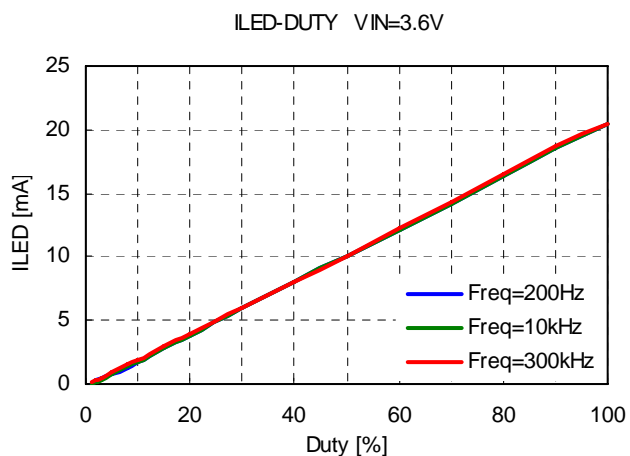
Inductor Voltage=7.2V, $V_{SET}=25V$

R1204xxxxB/C, L=10 μ H (VLF302512MT-100M)



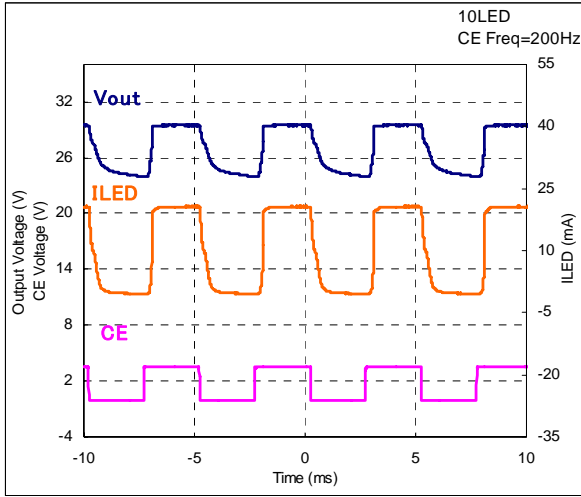
3) Maxduty vs. ILED

(R1204xxxxA/D/G/H, 10 LEDs in Series, $V_{IN}=3.6V$)

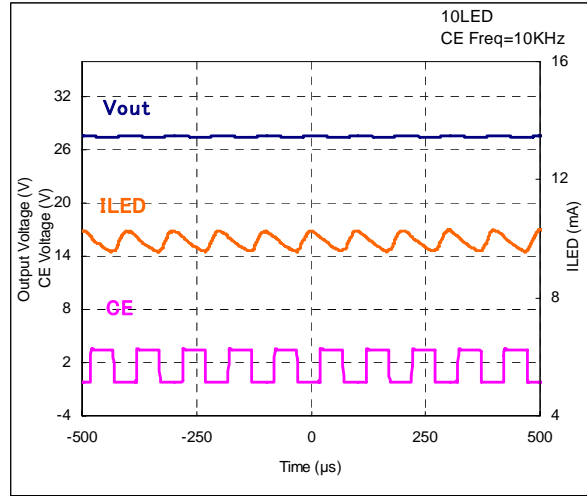


4) V_{OUT} / I_{LED} Ripple of R1204xxxxA/D/G/H When Dimming
 (10 LEDs in Series, $L=10\mu\text{H}$ (VLF302512MT-100M))

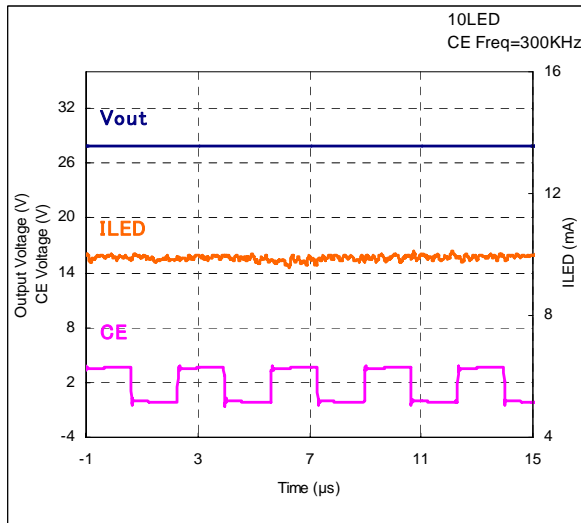
CE Freq=200Hz



CE Freq=10kHz

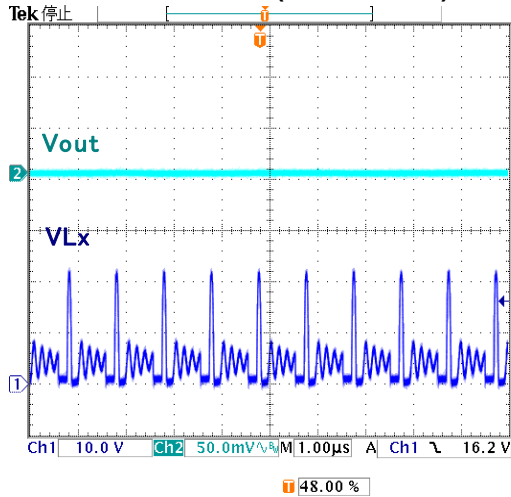


CE Freq=300kHz

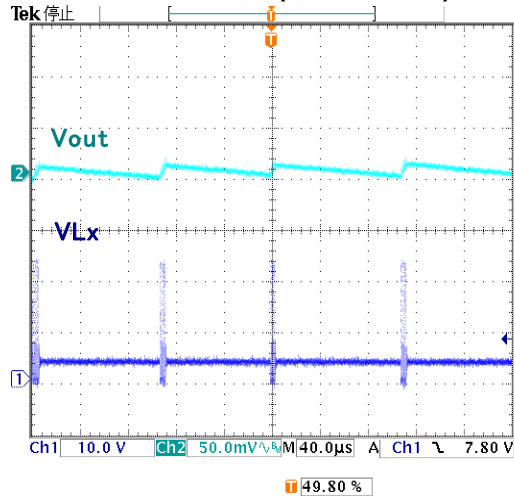


5) V_{OUT} Ripple ($V_{IN}=3.6\text{V}$, $V_{SET}=21\text{V}$, $I_{OUT}=0\text{mA}$, $L=10\mu\text{H}$ (VLF302512MT-100M))

PWM Control (R1204xxxxB)



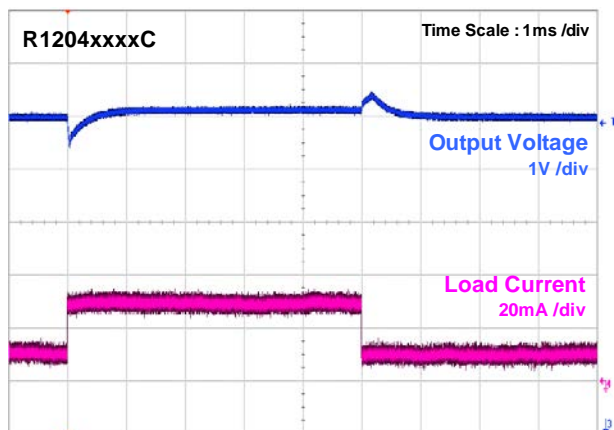
VFM Control (R1204xxxxC)



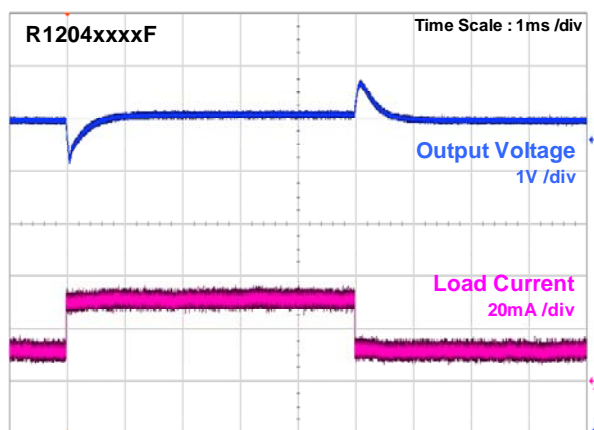
6) Load Transient Response

($V_{IN}=3.6V$, $V_{SET}=25V$, $L=10\mu H$ (VLF302512MT-100M), $I_{OUT}=10mA \leftrightarrow 30mA$, $T_r=T_f=0.5\mu s$)

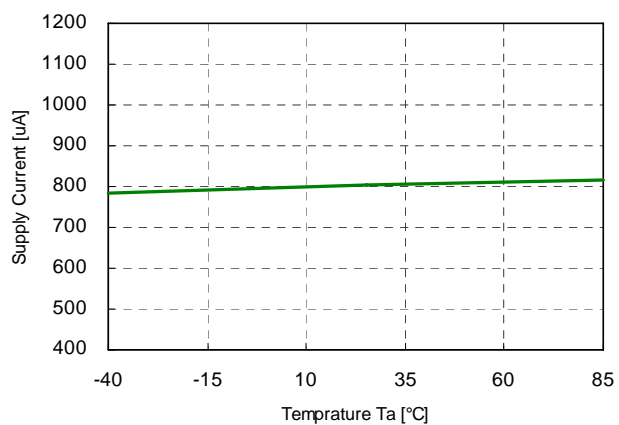
R1204xxxxC



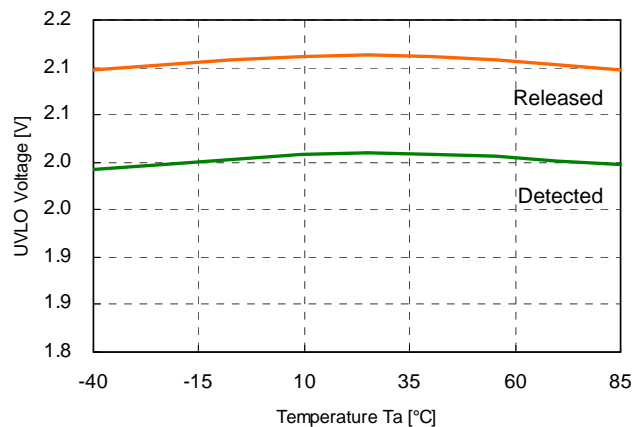
R1204xxxxF



7) Supply Current vs. Ambient Temperature

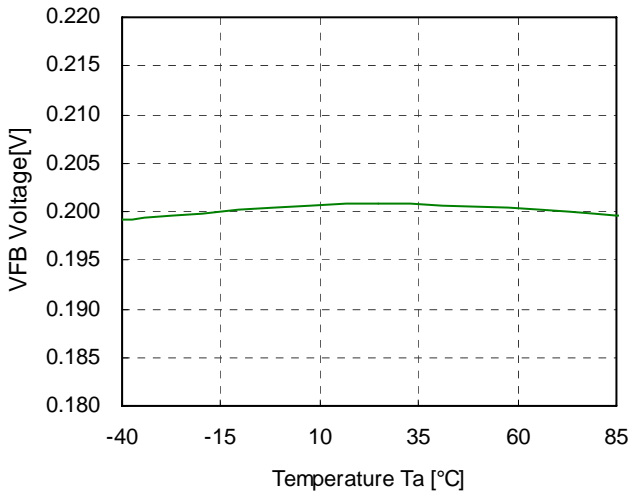


8) UVLO vs. Ambient Temperature

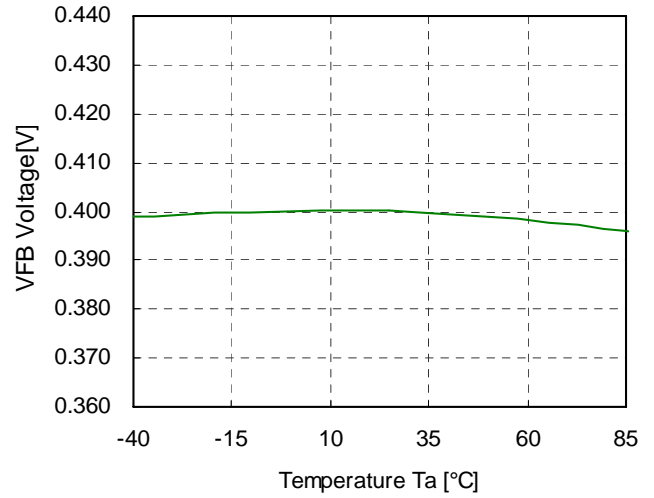


9) V_{FB} Voltage vs. Ambient Temperature

R1204xxxxA/D



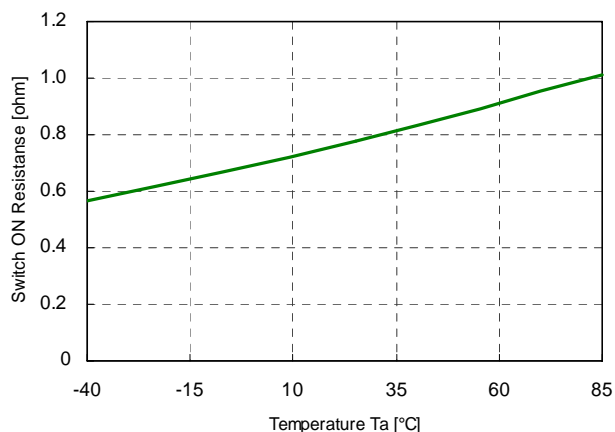
R1204xxxxG/H



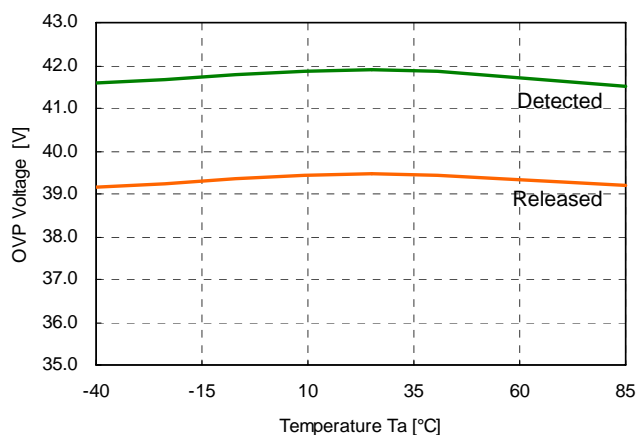
R1204xxxxB/C/E/F



10) Switch ON Resistance vs. Ambient Temperature



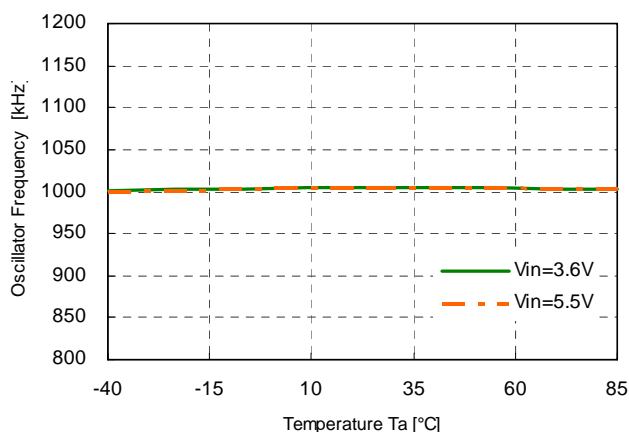
11) OVP Voltage vs. Ambient Temperature R1204x3xxx



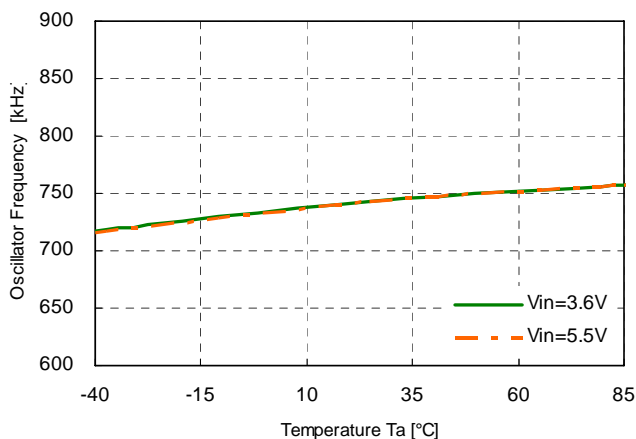
12) L_xCurrent Limit vs. Ambient Temperature



13) Oscillator Frequency vs. Ambient Temperature R1204xxxxA/B/C/G



R1204xxxxD/E/F/H





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