VNQ830E-E

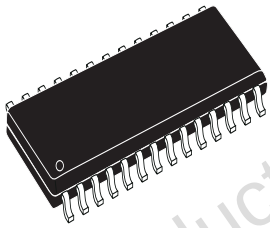
Quad channel high-side driver

Features

Type	$R_{DS(on)}$	I_{OUT}	V_{CC}
VNQ830E-E	65 mΩ(1)	9.5 A(1)	36 V

1. Per each channel.

- Output current: 9.5 A
- CMOS compatible inputs
- On-state open-load detection
- Off-state open-load detection
- Output stuck to V_{CC} detection
- Open drain status outputs
- Undervoltage shutdown
- Overvoltage clamp
- Thermal shutdown
- Current and power limitation
- Very low standby current
- Protection against loss of ground and loss of V_{CC}
- Reverse battery protection
- Very low electromagnetic susceptibility
- Optimized electromagnetic emission



SO-28 (double island)

Description

The VNQ830E-E is a quad HSD formed by assembling two VND830E-E chips in the same SO-28 package. The VND830E-E is a monolithic device made using STMicroelectronics™ VIPower™ M0-3 technology.

It is intended for driving resistive or inductive loads with one side connected to ground. Active V_{CC} pin voltage clamp protects the device against low energy spikes (see ISO7637 transient compatibility table).

The device detects open-load condition both in on-state and off-state. Output shorted to V_{CC} is detected in the off-state. Output current limitation protects the device in overload condition. In case of long duration overload, the device limits the dissipated power to safe level up to thermal shutdown intervention. Thermal shutdown with automatic restart allows the device to recover normal operation as soon as fault condition disappears.

Table 1. Device summary

Package	Order codes	
	Tube	Tape and reel
SO-28	VNQ830E-E	VNQ830ETR-E

Contents

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1 Block diagram and pin description

Figure 1. Block diagram

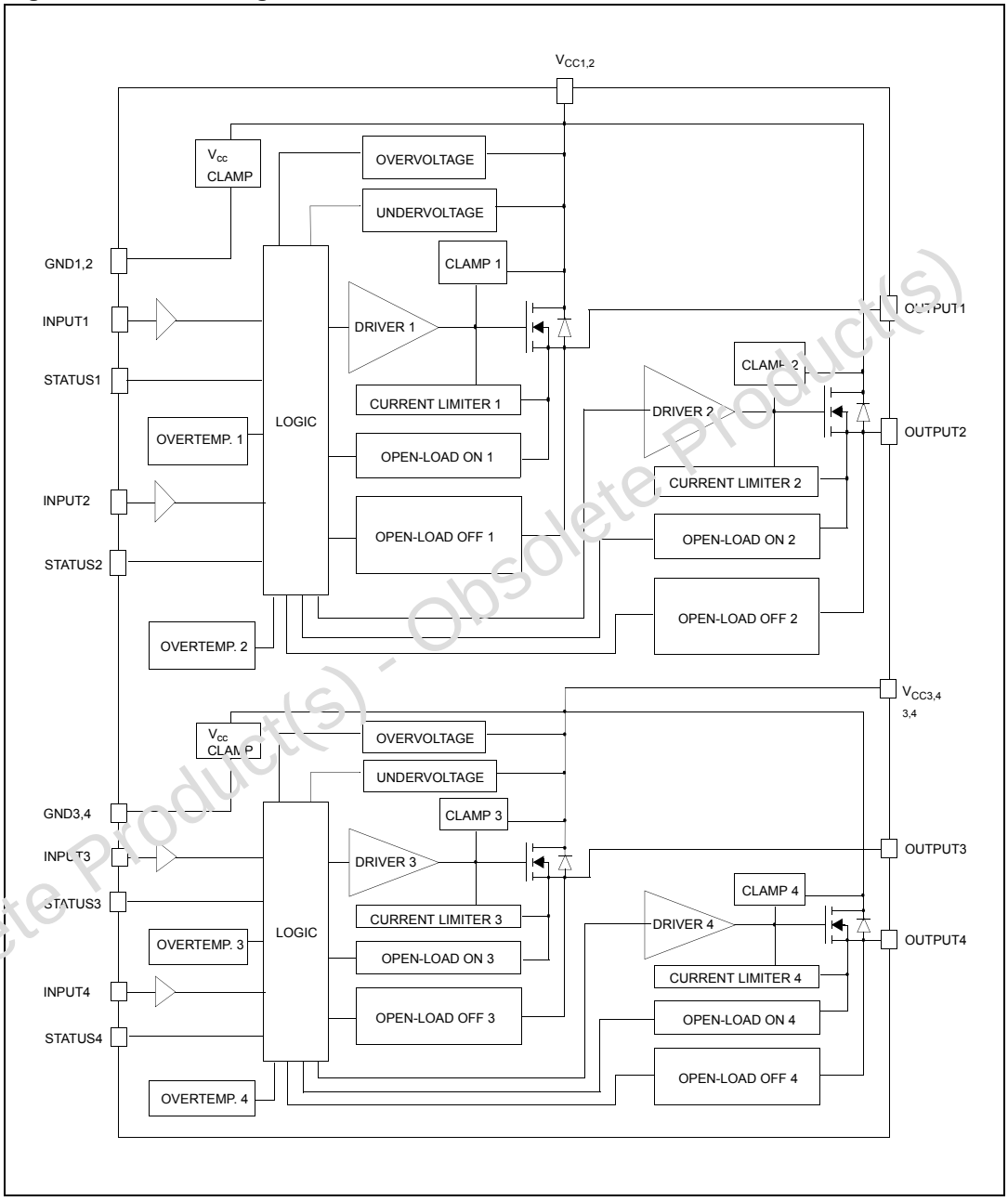


Figure 2. Configuration diagram (top view)

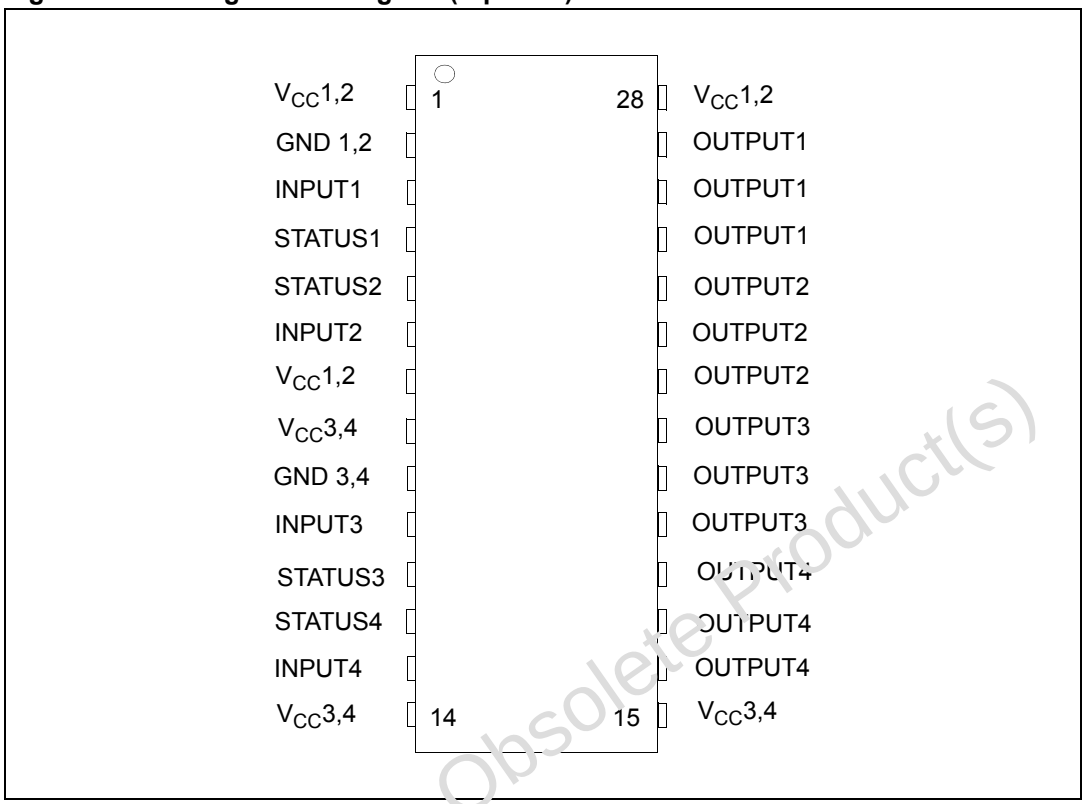


Table 2. Suggested connections for unused and not connected pins

Connection / pin	Status	N.C.	Output	Input
Floating	X	X	X	X
To ground		X		Through 10 KΩ resistor

2 Electrical specifications

2.1 Absolute maximum ratings

Stressing the device above the rating listed in *Table 3* may cause permanent damage to the device. These are stress ratings only and operation of the device at these or any other conditions above those indicated in the operating sections of this specification is not implied. Exposure to Absolute Maximum Rating conditions for extended periods may affect device reliability. Refer also to the STMicroelectronics SURE Program and other relevant quality document.

Table 3. Absolute maximum rating

Symbol	Parameter	Value	Unit
V_{CC}	DC supply voltage	41	V
- V_{CC}	Reverse DC supply voltage	-0.3	V
- I_{GND}	DC reverse ground pin current	-200	mA
I_{OUT}	DC output current	Internally limited	A
- I_{OUT}	Reverse DC output current	-6	A
I_{IN}	DC input current	+/-10	mA
I_{STAT}	DC status current	+/-10	mA
V_{ESD}	Electrostatic discharge (Human Body Model: R=1.5 KΩ; C = 100 pF) - INPUT - STATUS - OUTPUT - V_{CC}	 4000 4000 5000 5000	 V V V V
E_{MAX}	Maximum switching energy L = 0.5 mH; R_L = 0 Ω; V_{bat} = 13.5 V; T_{jstart} = 150 °C; I_L = 13.5 A)	64	mJ
P_{tot}	Power dissipation (per island) at T_{lead} = 25 °C	6.25	W
T_j	Junction operating temperature	Internally limited	°C
T_c	Case operating temperature	- 40 to 150	°C
T_{stg}	Storage temperature	- 55 to 150	°C

2.2 Thermal data

Table 4. Thermal data (per island)

Symbol	Parameter	Value		Unit
$R_{thj\text{-}lead}$	Thermal resistance junction-lead per chip	20		°C/W
$R_{thj\text{-}amb}$	Thermal resistance junction-ambient (one chip ON)	60(1)	45(2)	°C/W
$R_{thj\text{-}amb}$	Thermal resistance junction-ambient (two chips ON)	46(1)	30(2)	°C/W

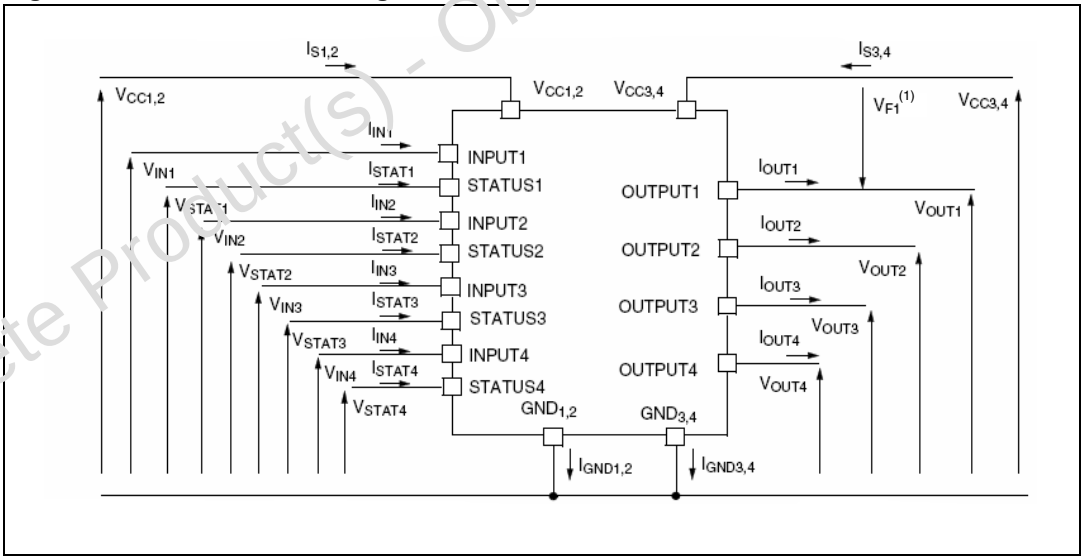
1. When mounted on a standard single-sided FR-4 board with 0.5 cm^2 of Cu (at least 35 µm thick) connected to all V_{CC} pins. Horizontal mounting and no artificial air flow.
2. When mounted on a standard single-sided FR-4 board with 6 cm^2 of Cu (at least 35 µm thick) connected to all V_{CC} pins. Horizontal mounting and no artificial air flow.

2.3 Electrical characteristics

Values specified in this section are for 8 V < V_{CC} < 36 V; -40 °C < T_j < 150 °C, unless otherwise stated.

(Per each channel)

Figure 3. Current and voltage conventions



1. $V_{Fn} = V_{CCn} - V_{OUTn}$ during reverse battery condition.

Table 5. Power output

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_{CC}(1)	Operating supply voltage		5.5	13	36	V
V_{USD}(1)	Undervoltage shutdown		3	4	5.5	V
V_{OV}(1)	Overvoltage shutdown		36			V

Table 5. Power output (continued)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
R_{ON}	On-state resistance	I_{OUT} = 2 A; T_j = 25 °C I_{OUT} = 2 A; V_{CC} > 8 V			65 130	mΩ mΩ
I_S(1)	Supply current	Off-state; V_{CC} = 13 V; V_{IN} = V_{OUT} = 0 V Off-state; V_{CC} = 13 V; V_{IN} = V_{OUT} = 0 V; T_j = 25°C On-state; V_{CC} = 13 V; V_{IN} = 5 V; I_{OUT} = 0 A		12 12 5	40 25 7	µA µA mA
$I_{L(off1)}$	Off-state output current	V_{IN} = V_{OUT} = 0 V	0		50	µA
$I_{L(off2)}$	Off-state output current	V_{IN} = 0 V; V_{OUT} = 3.5 V	-75		0	µA
$I_{L(off3)}$	Off-state output current	V_{IN} = V_{OUT} = 0 V; V_{CC} = 13 V; T_j = 125 °C			5	µA
$I_{L(off4)}$	Off-state output current	V_{IN} = V_{OUT} = 0 V; V_{CC} = 13 V; T_j =25 °C			3	µA

1. Per island.

Table 6. Switching (per each channel) (V_{CC} = 13 V)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
$t_{d(on)}$	Turn-on delay time	R_L = 6.5 Ω from V_{IN} rising edge to V_{OUT} = 1.3 V	-	50	-	µs
$t_{d(off)}$	Turn-off delay time	R_L = 6.5 Ω from V_{IN} falling edge to V_{OUT} = 11.7 V	-	50	-	µs
$dV_{OUT}/dt_{(on)}$	Turn-on voltage slope	R_L = 6.5 Ω from V_{OUT} = 1.3 V to V_{OUT} = 10.4 V	-	See *Figure 21*	-	V/µs
$dV_{OUT}/dt_{(off)}$	Turn-off voltage slope	R_L = 6.5 Ω from V_{OUT} = 11.7 V to V_{OUT} = 1.3 V	-	See *Figure 22*	-	V/µs

Table 7. V_{CC} - output diode

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_F	Forward on voltage	- I_{OUT} = 1.2 A; T_j = 150 °C	-	-	0.6	V

Table 8. Status pin (per each channel)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_{STAT}	Status low output voltage	I_{STAT} = 1.6 mA			0.5	V
I_{LSTAT}	Status leakage current	Normal operation; V_{STAT} = 5 V			10	µA
C_{STAT}	Status pin input capacitance	Normal operation; V_{STAT} = 5 V			100	pF
V_{SCL}	Status clamp voltage	I_{STAT} = 1 mA I_{STAT} = - 1 mA	6	6.8 - 0.7	8	V V

Table 9. Protections (per each channel)(1)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
T_{TSD}	Shutdown temperature		150	175	200	°C
T_R	Reset temperature		135			°C
T_{hyst}	Thermal hysteresis		7	15		°C
t_{SDL}	Status delay in overload conditions	$T_j > T_{TSD}$			20	µs
I_{lim}	Current limitation	V_{CC} = 13 V 5.5 V < V_{CC} < 36 V	9.5	13.5	18 18	A A
V_{demag}	Turn-off output clamp voltage	I_{OUT} = 2 A; L = 6 mH	V_{CC}-41	V_{CC}-48	V_{CC}-55	V

1. To ensure long term reliability under heavy overload or short circuit conditions, protection and related diagnostic signals must be used together with a proper software strategy. If the device is subjected to abnormal conditions, this software must limit the duration and number of activation cycles.

Table 10. Logic input (per each channel)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
V_{IL}	Low level input voltage				1.25	V
V_{IH}	High level input voltage		3.25			V
$V_{I(hyst)}$	Input hysteresis voltage		0.5			V
I_{IL}	Low level input current	V_{IN} = 1.5 V	1			µA
I_{IH}	High level input current	V_{IN} = 3.25 V			10	µA
V_{ICL}	Input clamp voltage	I_{IN} = 1 mA I_{IN} = -1 mA	6	6.8 - 0.7	8	V V

Table 11. Open-load detection (per each channel)

Symbol	Parameter	Test conditions	Min.	Typ.	Max.	Unit
I_{OL}	Open-load on-state detection threshold	V_{IN} = 5 V	50	115	200	mA
$t_{DOL(on)}$	Open-load on-state detection delay	I_{OUT} = 0 A			200	µs
V_{OL}	Open-load off-state voltage detection threshold	V_{IN} = 0 V	1.5	2.9	3.5	V
$t_{DOL(off)}$	Open-load detection delay at turn-off				1000	µs

Figure 4. Status timings

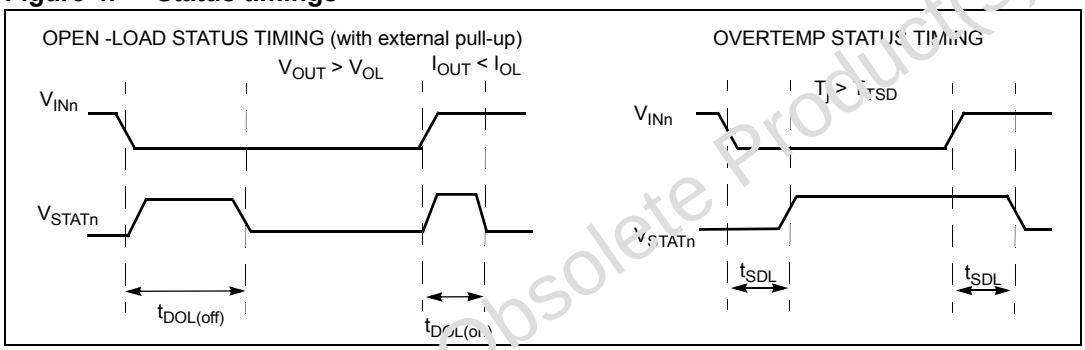


Figure 5. Switching time waveforms

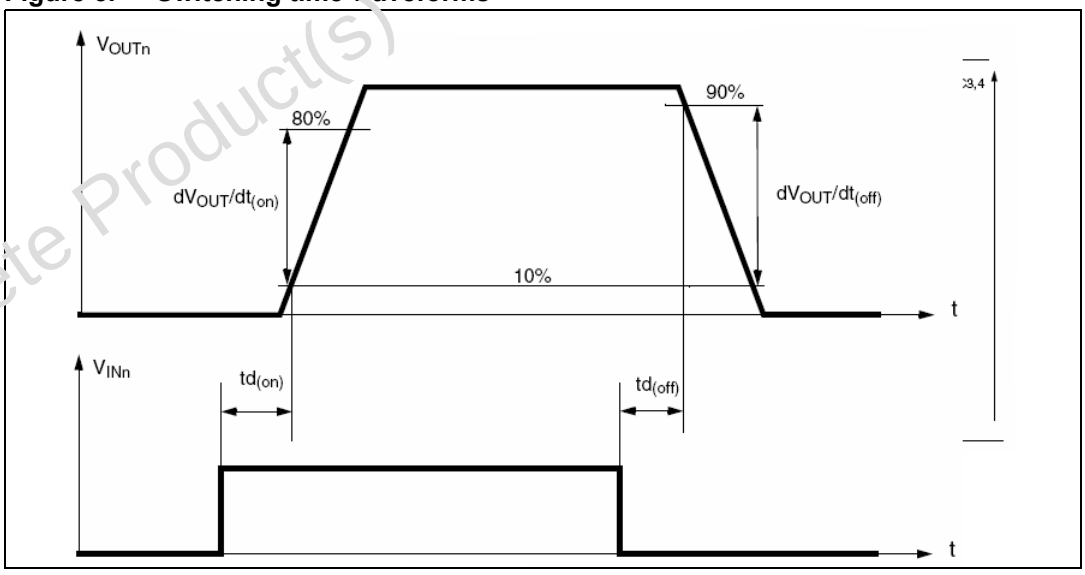


Table 12. Truth table

Conditions	Input	Output	Status
Normal operation	L H	L H	H H
Current limitation	L H H	L X X	H ($T_j < T_{TSD}$) H ($T_j > T_{TSD}$) L
Overtemperature	L H	L L	H L
Undervoltage	L H	L L	X X
Overvoltage	L H	L L	H H
Output voltage > V_{OL}	L H	H H	L H
Output current < I_{OL}	L H	L H	H L

Table 13. Electrical transient requirements on V_{CC} pin (part 1)

ISO T/R 7637/1 Test pulse	Test levels				
	I	II	III	IV	Delays and impedance
1	- 25 V	- 50 V	- 75 V	- 100 V	2 ms, 10 Ω
2	+ 25 V	+ 50 V	+ 75 V	+ 100 V	0.2 ms, 10 Ω
3a	- 25 V	- 50 V	- 100 V	- 150 V	0.1 µs, 50 Ω
3b	+ 25 V	+ 50 V	+ 75 V	+ 100 V	0.1 µs, 50 Ω
4	- 4 V	- 5 V	- 6 V	- 7 V	100 ms, 0.01 Ω
5	+ 26.5 V	+ 46.5 V	+ 66.5 V	+ 86.5 V	400 ms, 2 Ω

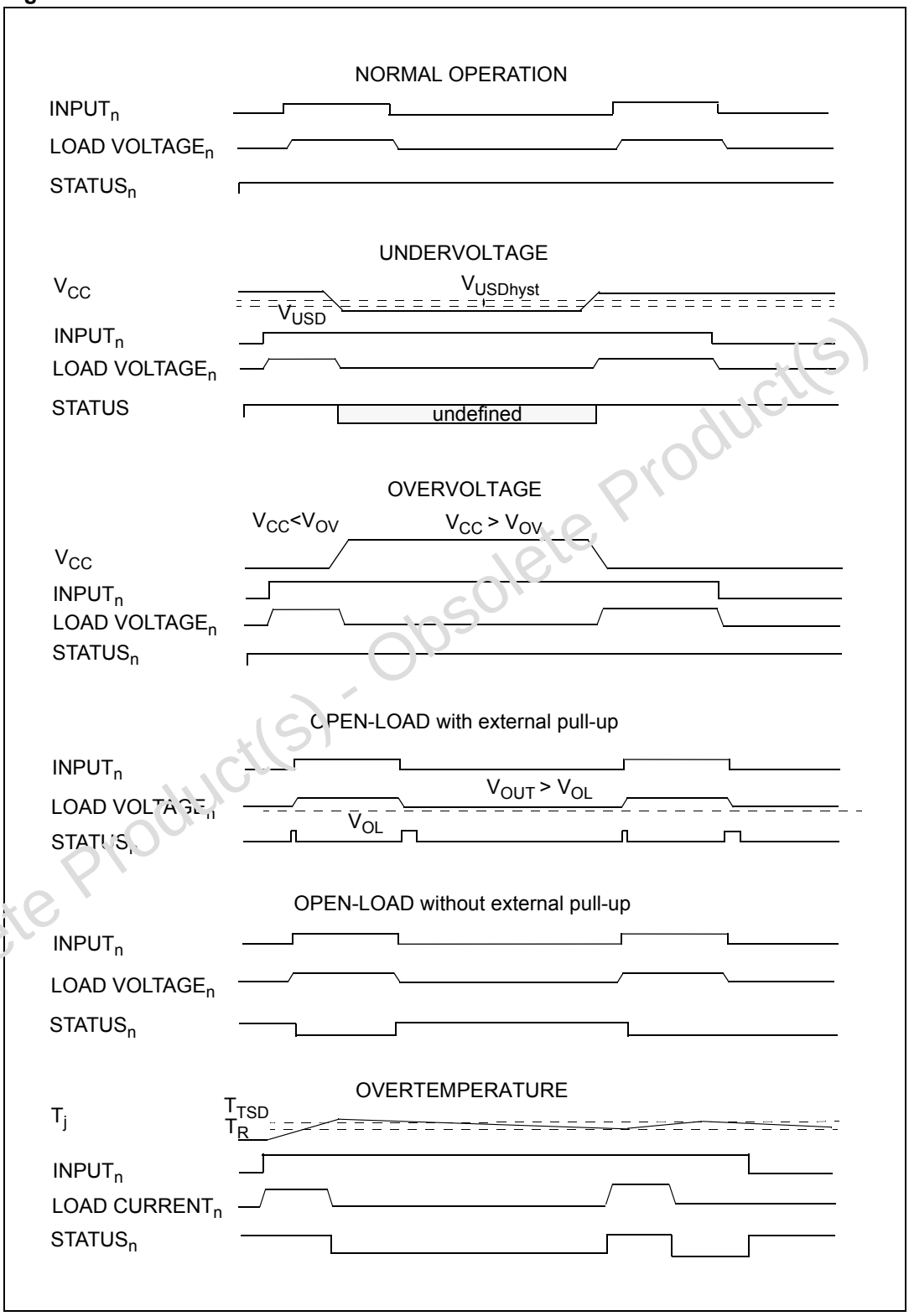
Table 14. Electrical transient requirements on V_{CC} pin (part 2)

ISO T/R 7637/1 Test pulse	Test levels results			
	I	II	III	IV
1	C	C	C	C
2	C	C	C	C
3a	C	C	C	C
3b	C	C	C	C
4	C	C	C	C
5	C	E	E	E

Table 15. Electrical transient requirements on V_{CC} pin (part 3)

Class	Contents
C	All functions of the device are performed as designed after exposure to disturbance.
E	One or more functions of the device is not performed as designed after exposure and cannot be returned to proper operation without replacing the device.

Figure 6. Waveforms



2.4 Electrical characteristics curves

Figure 7. Off-state output current

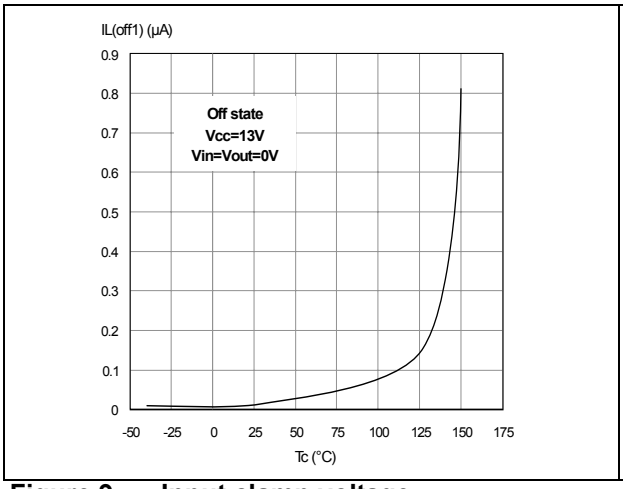


Figure 8. High level input current

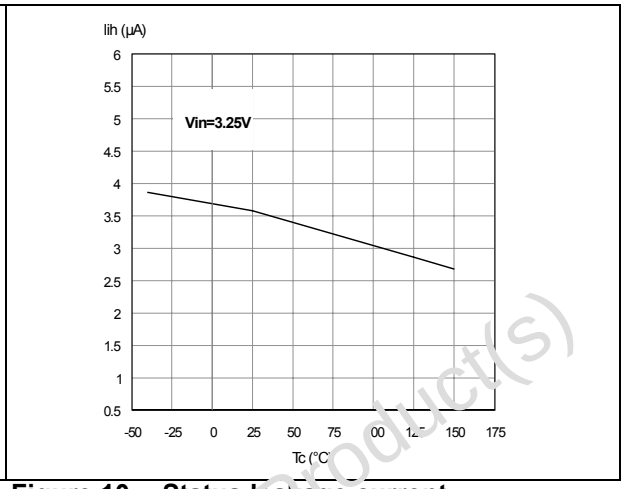


Figure 9. Input clamp voltage

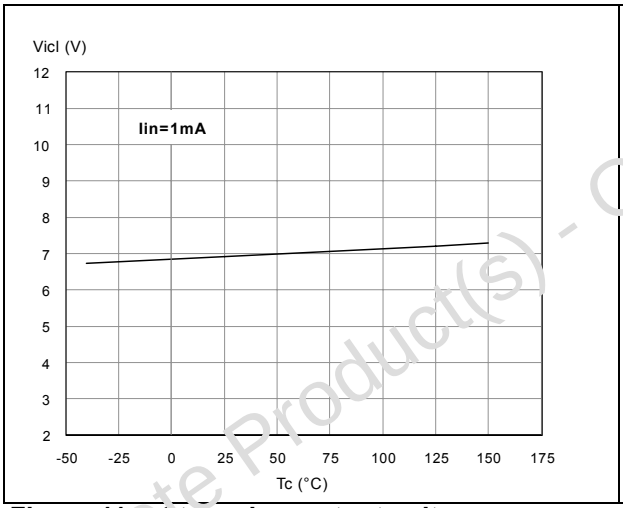


Figure 10. Status leakage current

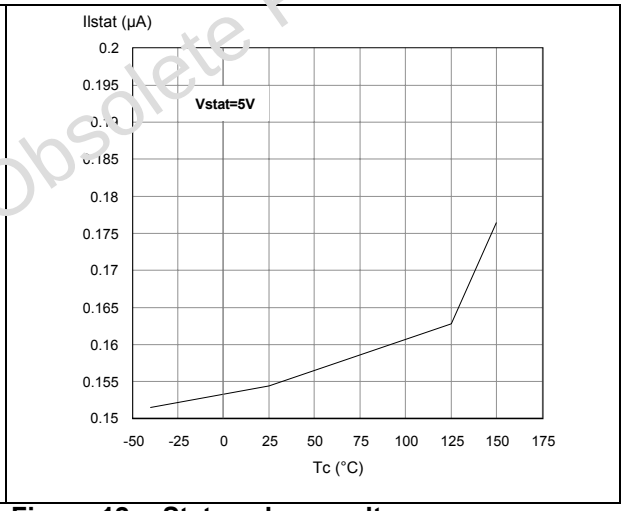


Figure 11. Status low output voltage

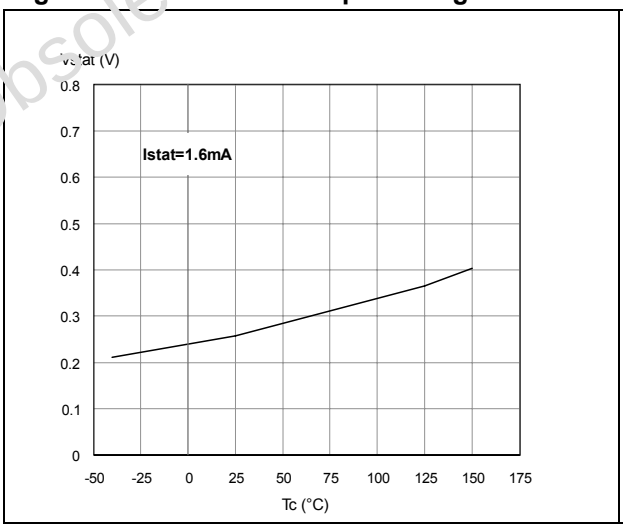


Figure 12. Status clamp voltage

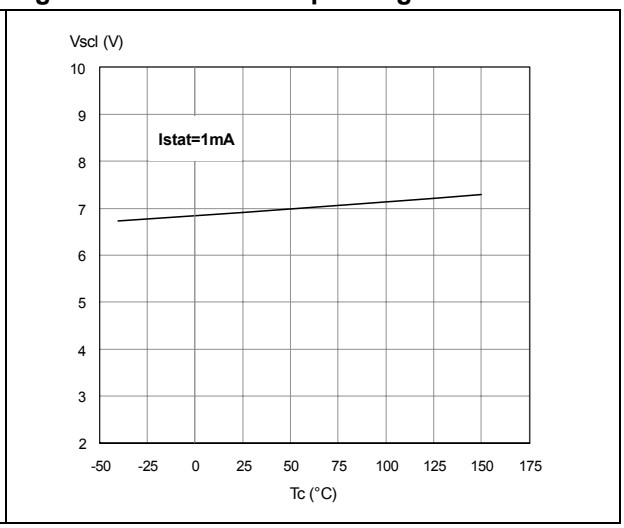


Figure 13. On-state resistance vs T_{case}

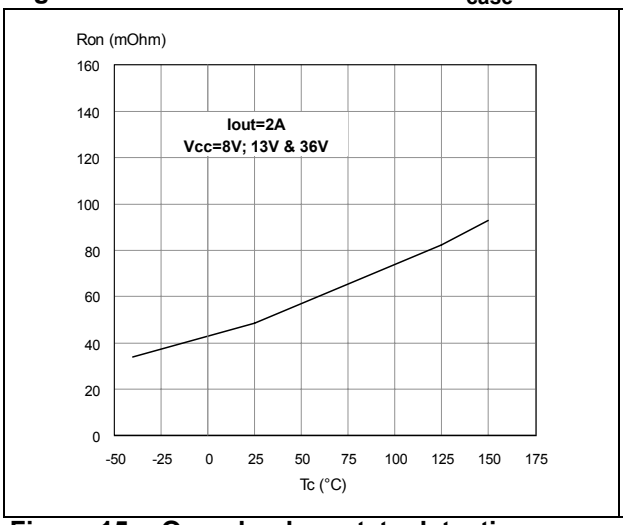


Figure 14. On-state resistance vs V_{CC}

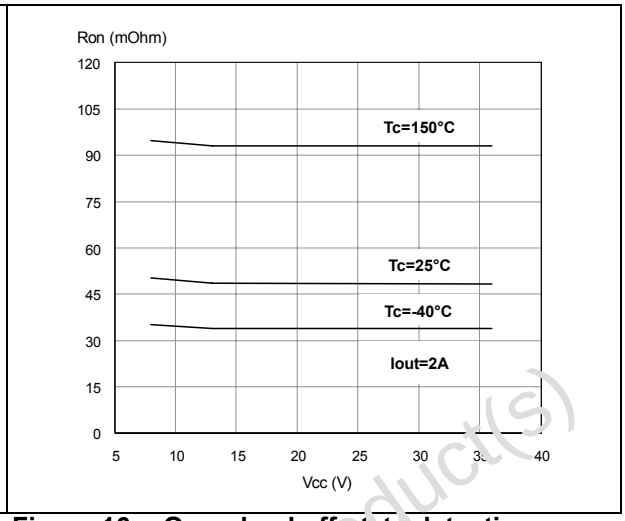


Figure 15. Open-load on-state detection threshold

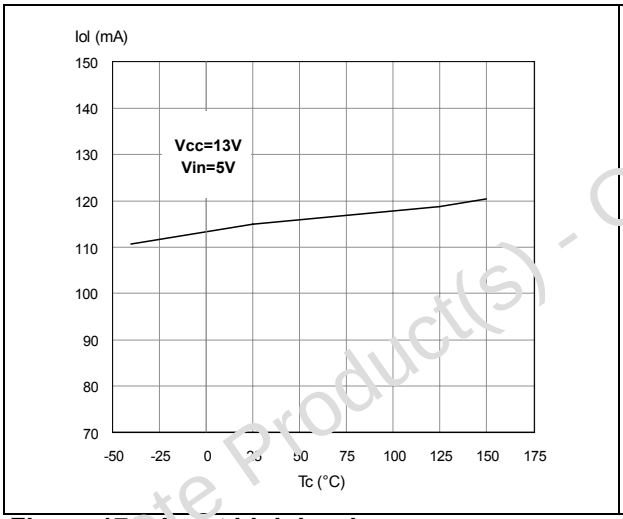


Figure 16. Open-load off-state detection threshold

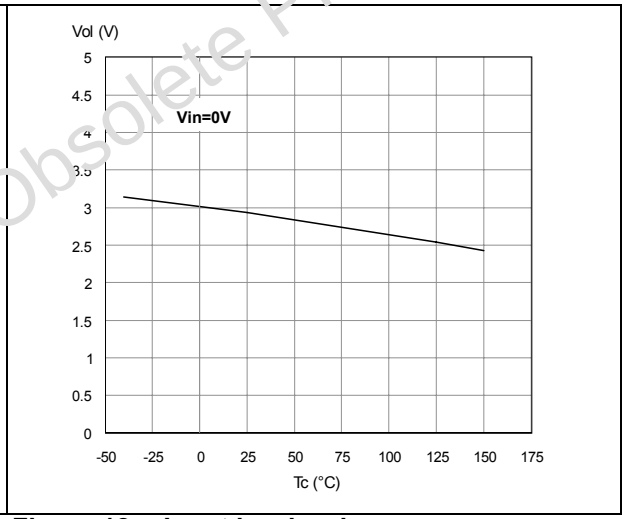


Figure 17. Input high level

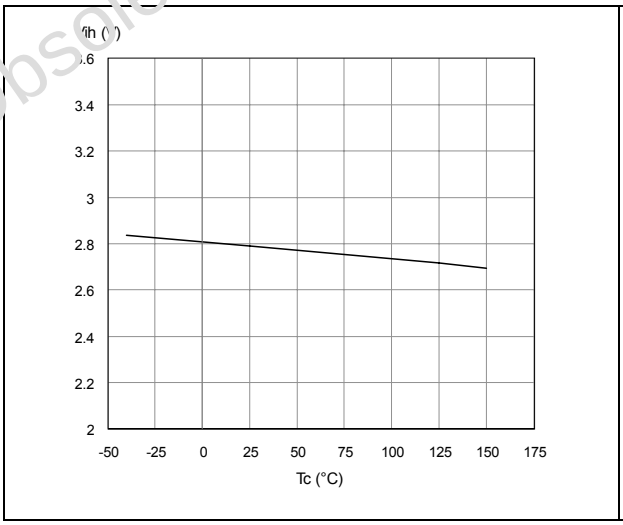


Figure 18. Input low level

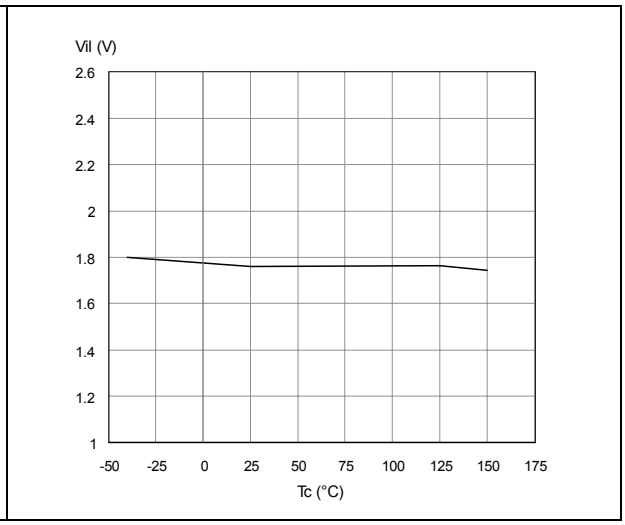


Figure 19. Input hysteresis voltage

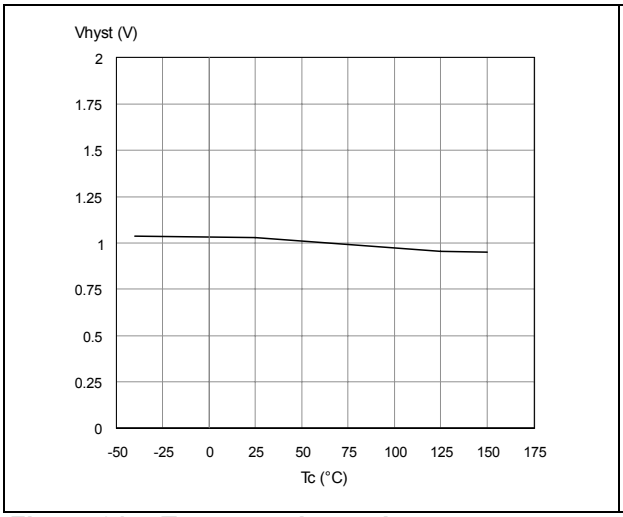


Figure 20. Overvoltage shutdown

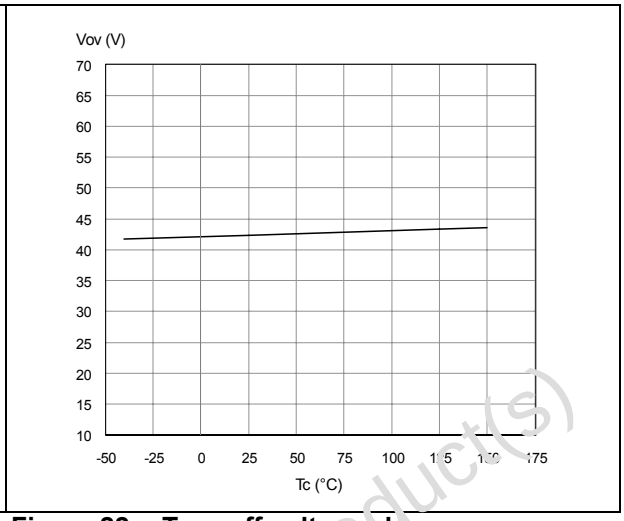


Figure 21. Turn-on voltage slope

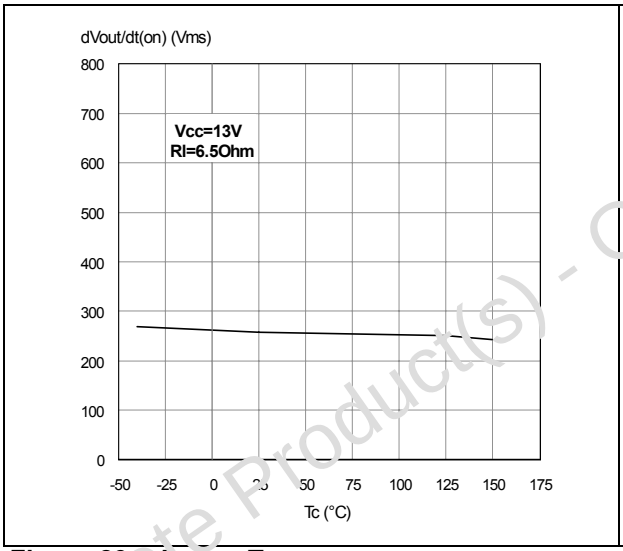


Figure 22. Turn-off voltage slope

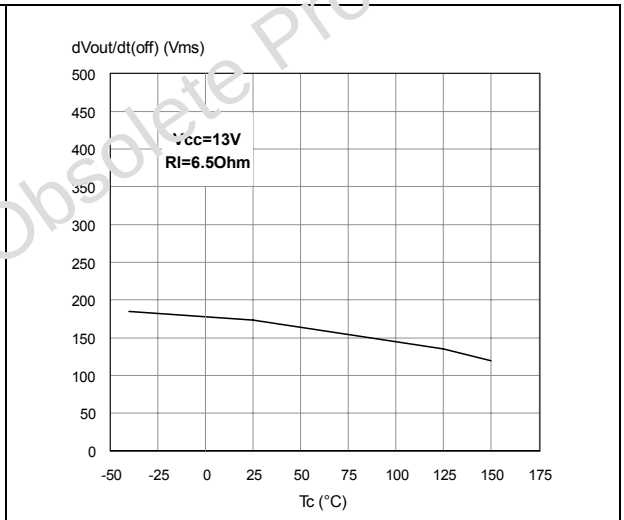


Figure 23. I_{LIM} vs T_{case}

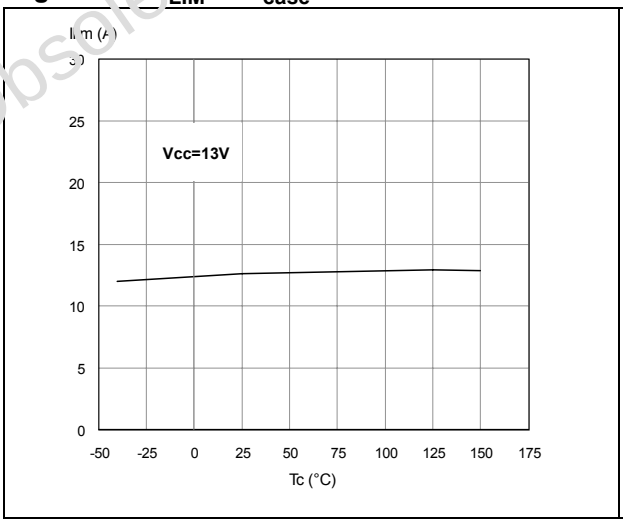
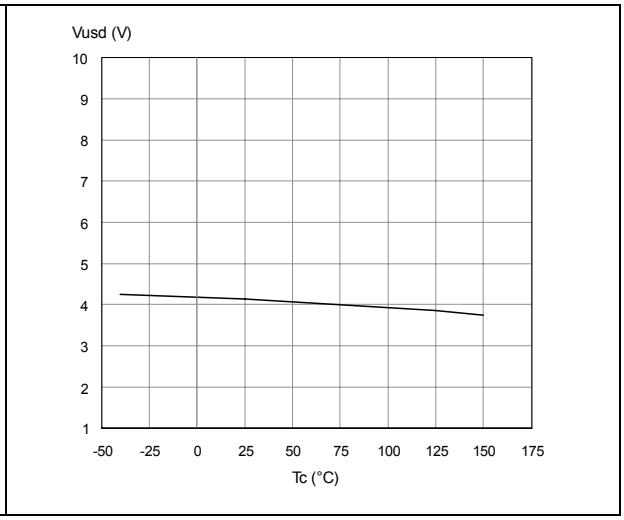
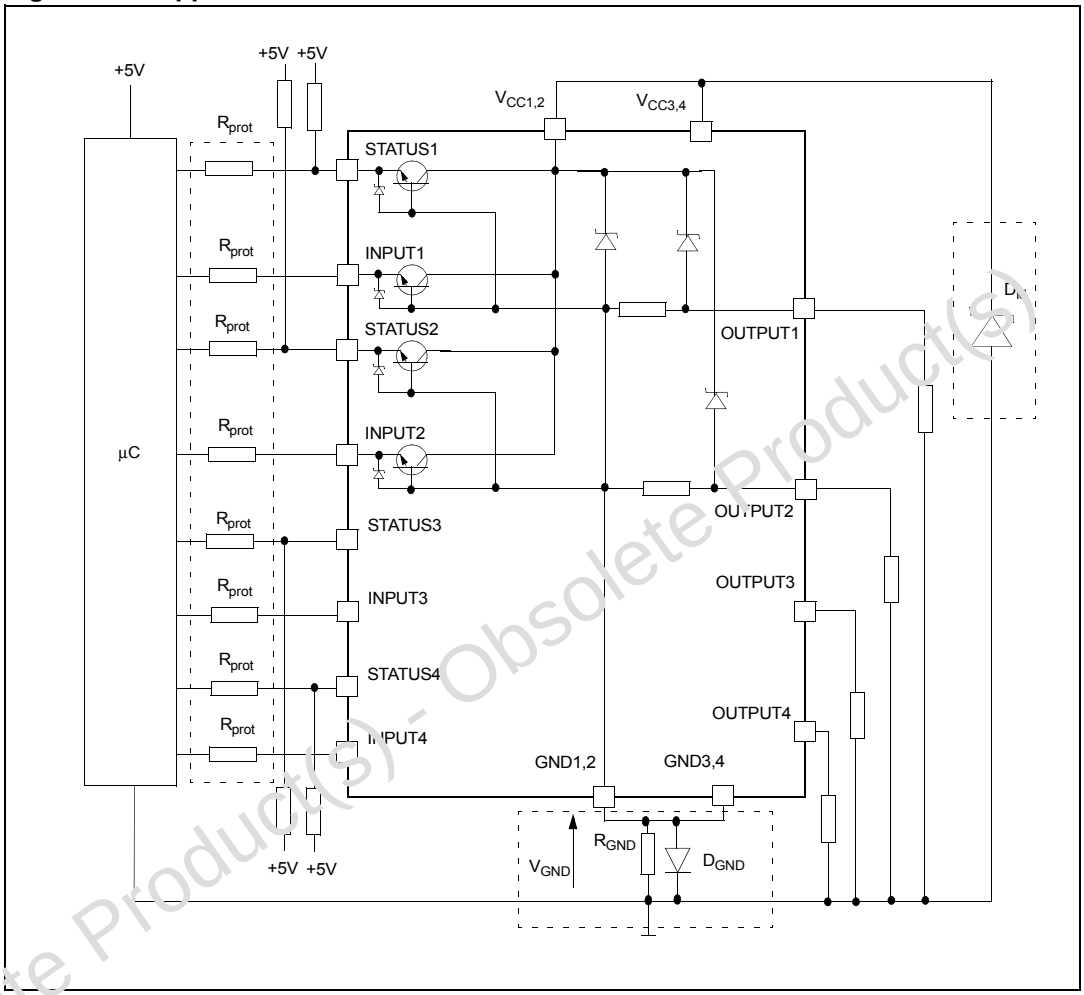


Figure 24. Undervoltage shutdown



3 Application information

Figure 25. Application schematic



Note: Channels 3 & 4 have the same internal circuit as channel 1 & 2.

3.1 GND protection network against reverse battery

This section provides two solutions for implementing a ground protection network against reverse battery.

3.1.1 Solution 1: a resistor in the ground line (R_{GND} only)

This can be used with any type of load.

The following show how to dimension the R_{GND} resistor:

1. $R_{GND} \leq 600m\ V / 2\ (I_{S(on)max})$
2. $R_{GND} \geq (-V_{CC}) / (-I_{GND})$

where - I_{GND} is the DC reverse ground pin current and can be found in the absolute maximum rating section of the device's datasheet.

Power dissipation in R_{GND} (when $V_{CC} < 0$ during reverse battery situations) is:

$$P_D = (- V_{CC})^2 / R_{GND}$$

This resistor can be shared amongst several different HSDs. Please note that the value of this resistor should be calculated with formula (1) where $I_{S(on)max}$ becomes the sum of the maximum on-state currents of the different devices.

Please note that, if the microprocessor ground is not shared by the device ground, then the R_{GND} produces a shift ($I_{S(on)max} * R_{GND}$) in the input thresholds and the status output values. This shift varies depending on how many devices are ON in the case of several high side drivers sharing the same R_{GND}.

If the calculated power dissipation requires the use of a large resistor, or several devices have to share the same resistor, then ST suggests using *Section 3.1.2* described below.

3.1.2 Solution 2: a diode (D_{GND}) in the ground line

A resistor (R_{GND} = 1 kΩ) should be inserted in parallel to D_{GND} if the device is driving an inductive load. This small signal diode can be safely shared amongst several different HSD. Also in this case, the presence of the ground network produces a shift (~600mV) in the input threshold and the status output values if the microprocessor ground is not common with the device ground. This shift does not vary if more than one HSD shares the same diode/resistor network.

3.2 Load dump protection

D_{ld} is necessary (Voltage Transient Suppressor) if the load dump peak voltage exceeds the V_{CC} maximum DC rating. The same applies if the device is subject to transients on the V_{CC} line that are greater than those shown in *Table 13*.

3.3 MCU I/O protection

If a ground protection network is used and negative transients are present on the V_{CC} line, the control pins are pulled negative. ST suggests to insert a resistor (R_{prot}) in line to prevent the microcontroller I/O pins from latching up.

The value of these resistors is a compromise between the leakage current of microcontroller and the current required by the HSD I/Os (Input levels compatibility) with the latch-up limit of microcontroller I/Os:

$$- V_{CCpeak} / I_{latchup} \le R_{prot} \le (V_{OH\mu C}-V_{IH}-V_{GND}) / I_{IHmax}$$

Example

For the following conditions:

V_{CCpeak}= - 100 V and $I_{latchup} \ge$ 20 mA; VOHμC ≥ 4.5 V

5 kΩ ≤ R_{prot} ≤ 65 kΩ.

Recommended values are:

R_{prot} = 10 kΩ

3.4 Open-load detection in off-state

Off-state open-load detection requires an external pull-up resistor (R_{PU}) connected between OUTPUT pin and a positive supply voltage (V_{PU}) like the +5 V line used to supply the microprocessor.

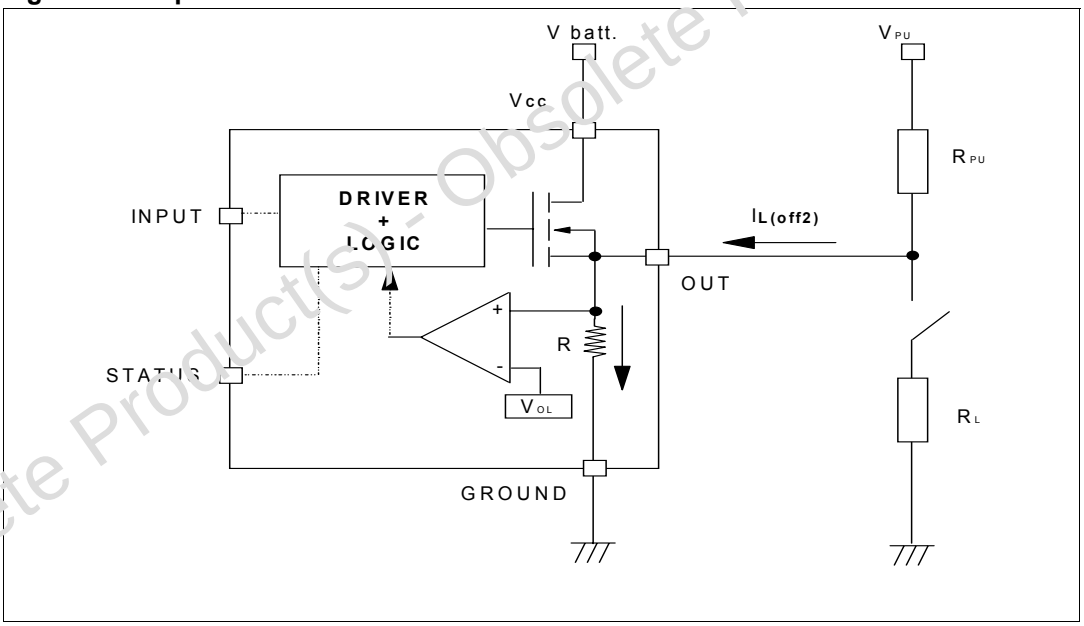
The external resistor has to be selected according to the following requirements:

1. No false open-load indication when load is connected: in this case we have to avoid V_{OUT} to be higher than V_{OLmin}; this results in the following condition $V_{OUT} = (V_{PU} / (R_L + R_{PU}))R_L < V_{OLmin}$.
2. No misdetection when load is disconnected: in this case the V_{OUT} has to be higher than V_{OLmax}; this results in the following condition $R_{PU} < (V_{PU} - V_{OLmax}) / I_{L(off2)}$

Because $I_{s(OFF)}$ may significantly increase if V_{out} is pulled high (up to several mA), the pull-up resistor R_{PU} should be connected to a supply that is switched OFF when the module is in standby.

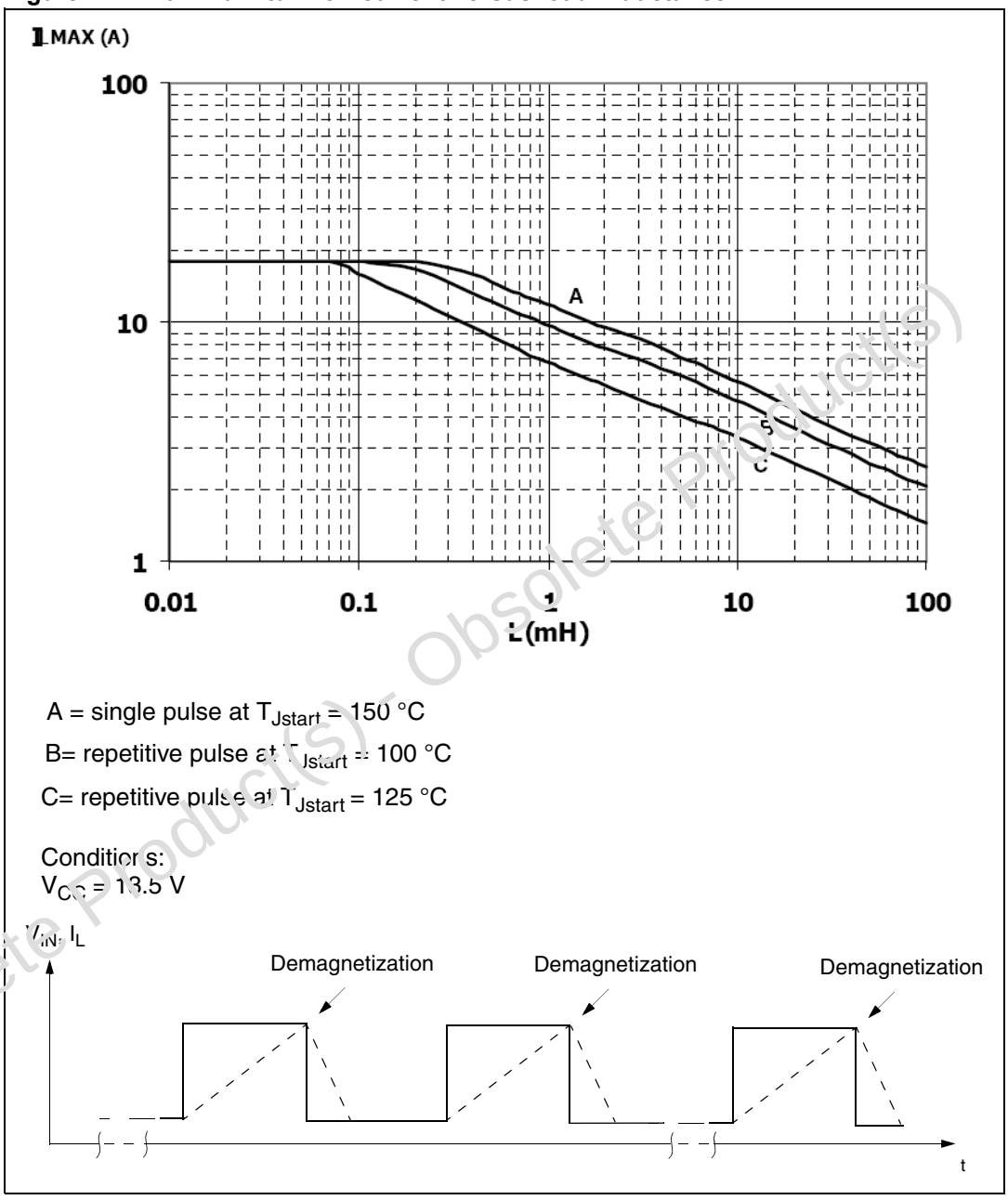
The values of V_{OLmin}, V_{OLmax} and $I_{L(off2)}$ are available in the *Chapter 2: Electrical specifications*.

Figure 26. Open-load detection in off-state



3.5 Maximum demagnetization energy

Figure 27. Maximum turn-off current versus load inductance



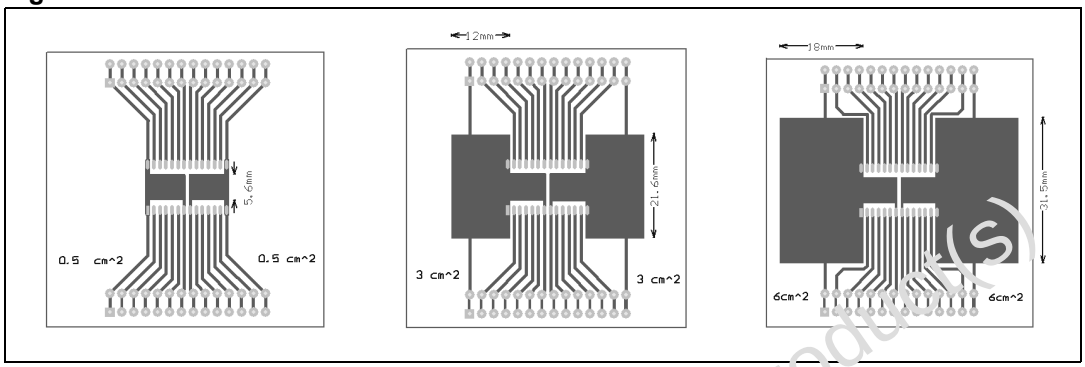
Note: Values are generated with $R_L = 0\ \Omega$.

In case of repetitive pulses, T_{jstart} (at beginning of each demagnetization) of every pulse must not exceed the temperature specified above for curves B and C.

4 Package and PCB thermal data

4.1 SO-28 thermal data

Figure 28. SO-28 PC board



Note: Layout condition of R_{th} and Z_{th} measurements (PCB FR4 area = 58 mm x 58 mm, PCB thickness = 2 mm, Cu thickness = 35 µm, Copper areas: 0.5 cm^2, 3 cm^2, 6 cm^2).

Table 16. Thermal calculation according to the PCB heatsink area

Chip 1	Chip 2	T_{jchip1}	T_{jchip2}	Note
On	Off	$R_{thA} \times P_{dchip1} + T_{amb}$	$R_{thC} \times P_{dchip1} + T_{amb}$	
Off	On	$R_{thC} \times P_{dchip2} + T_{amb}$	$R_{thA} \times P_{dchip2} + T_{amb}$	
On	On	$R_{thB} \times (P_{dchip1} + P_{dchip2}) + T_{amb}$	$R_{thB} \times (P_{dchip1} + P_{dchip2}) + T_{amb}$	$P_{dchip1} = P_{dchip2}$
On	On	$(R_{thA} \times P_{dchip1}) + R_{thC} \times P_{dchip2} + T_{amb}$	$(R_{thA} \times P_{dchip2}) + R_{thC} \times P_{dchip1} + T_{amb}$	$P_{dchip1} \neq P_{dchip2}$

R_{thA} = thermal resistance junction to ambient with one chip ON

R_{thB} = thermal resistance junction to ambient with both chips ON and $P_{dchip1} = P_{dchip2}$

R_{thC} = mutual thermal resistance

Figure 29. $R_{thj\text{-}amb}$ vs PCB copper area in open box free air condition

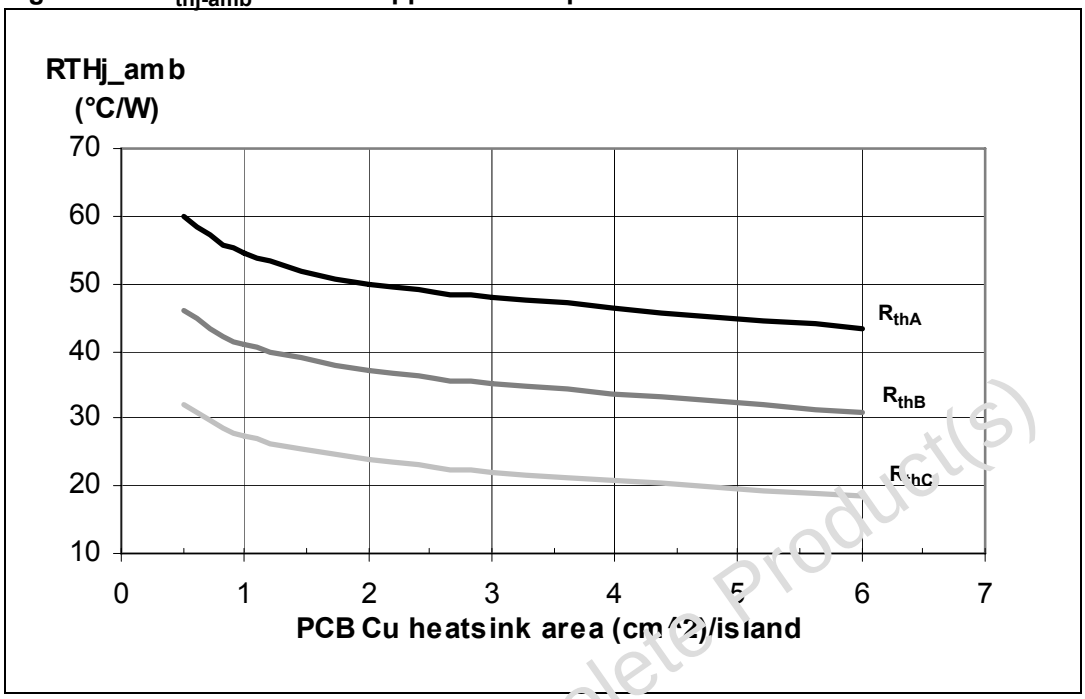
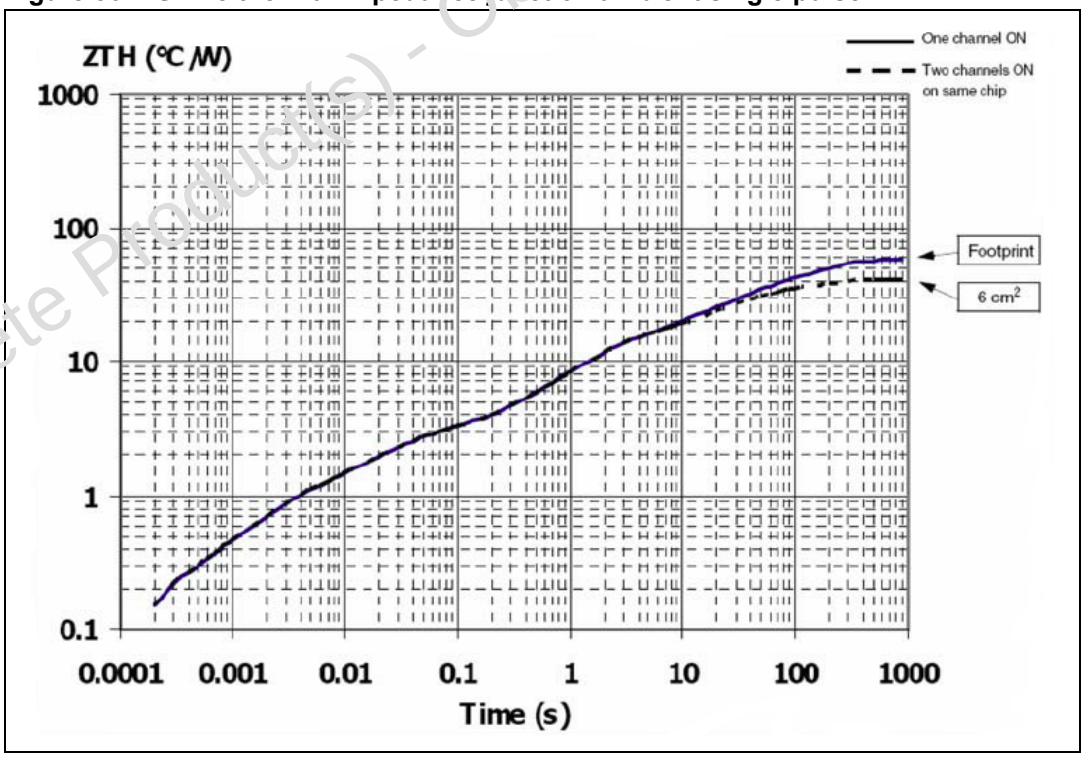


Figure 30. SP-28 thermal impedance junction ambient single pulse



Equation 1: pulse calculation formula

$$Z_{TH\delta} = R_{TH} \cdot \delta + Z_{THtp}(1 - \delta)$$

where $\delta = t_p / T$

Figure 31. Thermal fitting model of a quad channel HSD in SO-28

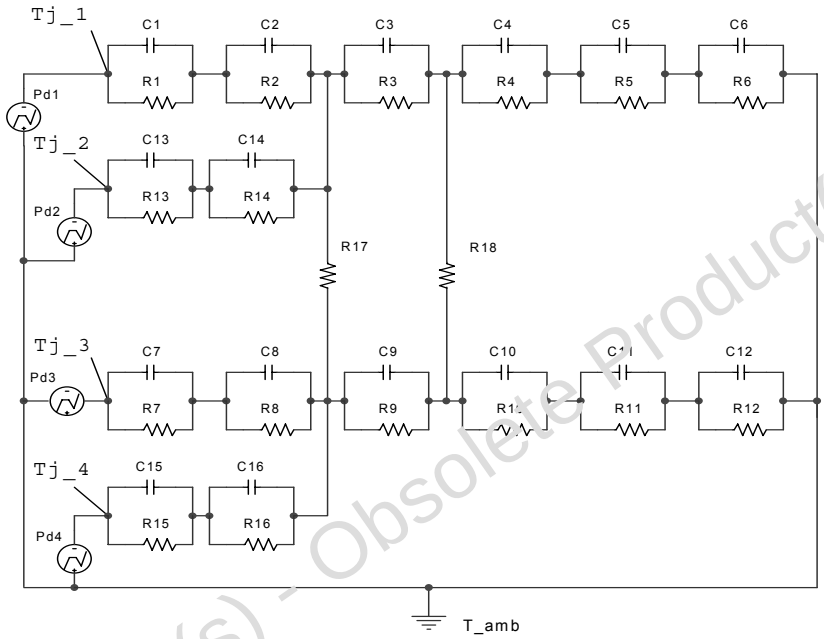


Table 17. Thermal parameters

Area / island (cm^2)	Footprint	6
R1 = R7 = R13 = R15 (°C/W)	0.15	
R2 = R8 = R14 = R16 (°C/W)	0.7	
R3 = R9 (°C/W)	1.8	
R4 = R10 (°C/W)	10	
R5 = R11 (°C/W)	15	
R6 = R12 (°C/W)	30	13
C1 = C7 = C13 = C15 (W.s/°C)	0.0005	
C2 = C8 = C14 = C16 (W.s/°C)	0.003	
C3 = C9 (W.s/°C)	0.015	
C4 = C10 (W.s/°C)	0.15	
C5 = C11 (W.s/°C)	1.5	
C6 = C12 (W.s/°C)	5	8
R17 = R18 (°C/W)	150	

5 Package and packing information

5.1 ECOPACK® packages

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK® packages, depending on their level of environmental compliance. ECOPACK® specifications, grade definitions and product status are available at: www.st.com. ECOPACK® is an ST trademark.

Figure 32. SO-28 package dimensions

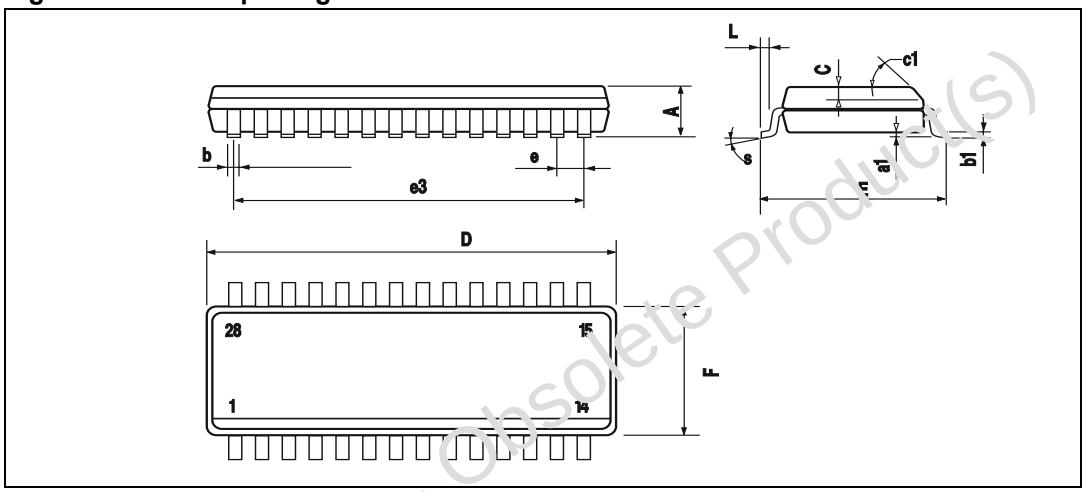
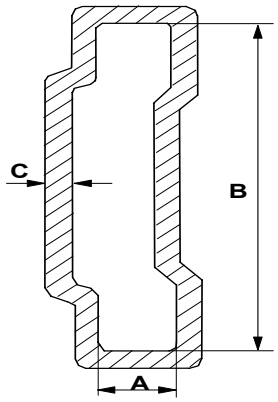


Table 18. SO-28 mechanical data

Symbol	Millimeters		
	Min.	Typ.	Max.
A			2.65
a1	0.10		0.30
b	0.35		0.49
b1	0.23		0.32
C		0.50	
c1	45° (typ.)		
D	17.7		18.1
E	10.00		10.65
e		1.27	
e3		16.51	
F	7.40		7.60
L	0.40		1.27
S	8° (max.)		

5.2 SO-28 packing information

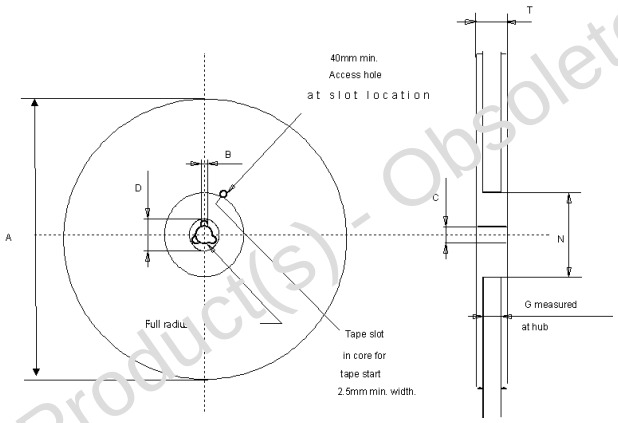
Figure 33. SO-28 tube shipment (no suffix)



Base Q.ty	28
Bulk Q.ty	700
Tube length (± 0.5)	532
A	3.5
B	13.8
C (± 0.1)	0.6

All dimensions are in mm.

Figure 34. SO-28 tape and reel shipment (suffix "13TR")



Reel dimensions

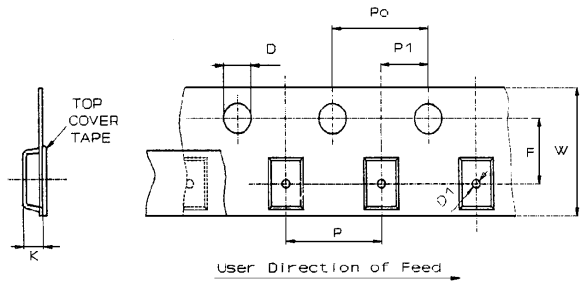
Base Q.ty	1000
Bulk Q.ty	1000
A (max)	330
B (min)	1.5
C (± 0.2)	13
F	20.2
G (+ 2 / -0)	16.4
N (min)	60
T (max)	22.4

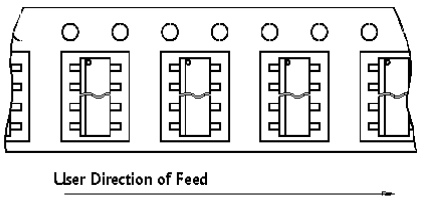
Tape dimensions

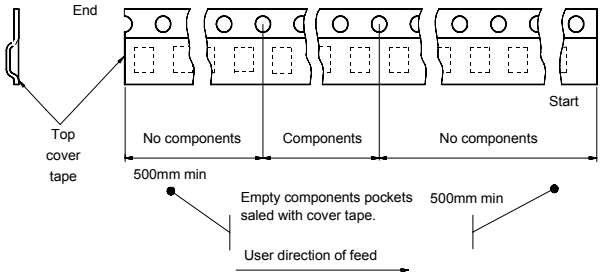
According to Electronic Industries Association (EIA) Standard 481 rev. A, Feb. 1986

Tape width	**W**	16
Tape hole spacing	**P0 (± 0.1)**	4
Component spacing	**P**	12
Hole diameter	**D (± 0.1/-0)**	1.5
Hole diameter	**D1 (min)**	1.5
Hole position	**F (± 0.05)**	7.5
Compartment depth	**K (max)**	6.5
Hole spacing	**P1 (± 0.1)**	2

All dimensions are in mm.







6 Revision history

Table 19. Document revision history

Date	Revision	Changes
03-May-2010	1	Initial release.

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