

135 μ A, 14nV/ $\sqrt{\text{Hz}}$, Rail-to-Rail Output Precision Op Amp with Shutdown

FEATURES

- 35 μ V Maximum Offset Voltage
- 110pA Maximum Input Bias Current
- 135 μ A Supply Current
- Rail-to-Rail Output Swing
- 12 μ A Supply Current in Shutdown
- 120dB Minimum Voltage Gain ($V_S = \pm 15\text{V}$)
- 0.8 $\mu\text{V}/^\circ\text{C}$ Maximum V_{OS} Drift
- 14nV/ $\sqrt{\text{Hz}}$ Input Noise Voltage
- 2.7V to $\pm 18\text{V}$ Supply Voltage Operation
- Operating Temperature Range: -40°C to 85°C
- Space Saving 3mm $\times$ 3mm DFN Package

APPLICATIONS

- Thermocouple Amplifiers
- Precision Photo Diode Amplifiers
- Instrumentation Amplifiers
- Battery-Powered Precision Systems

DESCRIPTION

The LT[®]6010 op amp combines low noise and high precision input performance with low power consumption and rail-to-rail output swing.

Input offset voltage is trimmed to less than 35 μV . The low drift and excellent long-term stability guarantee a high accuracy over temperature and over time. The 110pA maximum input bias current and 120dB minimum voltage gain further maintain this precision over operating conditions.

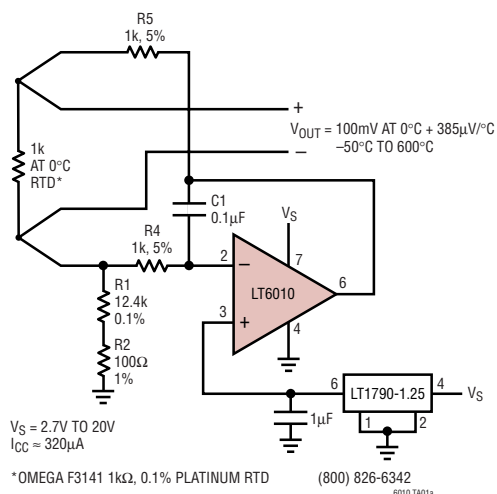
The LT6010 works on any power supply voltage from 2.7V to 36V, and draws only 135 μA of supply current on a 5V supply. A power saving shutdown feature reduces supply current to 12 μA . The output voltage swings to within 40mV of either supply rail, making the amplifier a good choice for low voltage single supply operation.

The LT6010 is fully specified at 5V and $\pm 15\text{V}$ supplies and from -40°C to 85°C . The device is available in SO-8 and space-saving 3mm $\times$ 3mm DFN packages. This op amp is also available in dual (LT6011) and quad (LT6012) packages.

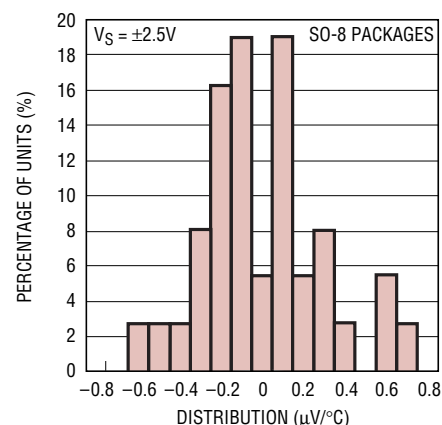
 LTC and LT are registered trademarks of Linear Technology Corporation.

TYPICAL APPLICATION

Single Supply Current Source for Platinum RTD



Distribution of Offset Voltage Drift

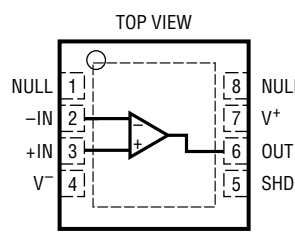
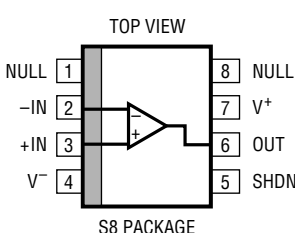


LT6010

ABSOLUTE MAXIMUM RATINGS (Note 1)

Total Supply Voltage (V^+ to V^-)	40V	Maximum Junction Temperature	
Differential Input Voltage (Note 2)	10V	DD Package	125°C
Input Voltage, Shutdown Voltage	V^+ to V^-	SO-8 Package	150°C
Input Current (Note 2)	$\pm 10\text{mA}$	Storage Temperature Range	
Output Short-Circuit Duration (Note 3)	Indefinite	DD Package	-65°C to 125°C
Operating Temperature Range (Note 4) ..	-40°C to 85°C	SO-8 Package	-65°C to 150°C
Specified Temperature Range (Note 5) ...	-40°C to 85°C	Lead Temperature (Soldering, 10 sec)	300°C

PACKAGE/ORDER INFORMATION

 DD PACKAGE 8-LEAD (3mm x 3mm) PLASTIC DFN $T_{JMAX} = 125^\circ\text{C}$, $\theta_{JA} = 160^\circ\text{C/W}$ UNDERSIDE METAL INTERNALLY CONNECTED TO V^- (PCB CONNECTION OPTIONAL)	ORDER PART NUMBER	 S8 PACKAGE 8-LEAD PLASTIC SO $T_{JMAX} = 150^\circ\text{C}$, $\theta_{JA} = 190^\circ\text{C/W}$	ORDER PART NUMBER
	LT6010CDD LT6010IDD LT6010ACDD LT6010AIDD		LT6010CS8 LT6010IS8 LT6010ACS8 LT6010AIS8
	DD PART MARKING*		S8 PART MARKING
	LADU		6010 6010I 6010A 6010AI

*Temperature grades are identified by a label on the shipping container.
Consult LTC Marketing for parts specified with wider operating temperature ranges.

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^\circ\text{C}$. $V_S = 5\text{V}$, 0V ; $V_{CM} = 2.5\text{V}$; R_L to 0V ; $\text{SHDN} = 0.2\text{V}$, unless otherwise specified. (Note 5)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage (Note 7)	LT6010AS8		10	35	μV
		$T_A = 0^\circ\text{C}$ to 70°C	●		60	μV
		$T_A = -40^\circ\text{C}$ to 85°C	●		75	μV
		LT6010S8		20	55	μV
		$T_A = 0^\circ\text{C}$ to 70°C	●		85	μV
		$T_A = -40^\circ\text{C}$ to 85°C	●		110	μV
		LT6010ADD		20	60	μV
		$T_A = 0^\circ\text{C}$ to 70°C	●		85	μV
$\Delta V_{OS}/\Delta T$	Input Offset Voltage Drift (Note 6)	LT6010AS8, LT6010S8	●	0.2	0.8	$\mu\text{V}/^\circ\text{C}$
		LT6010ADD, LT6010DD	●	0.2	1.3	$\mu\text{V}/^\circ\text{C}$

sn6010 6010fs

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^\circ\text{C}$. $V_S = 5\text{V}$, 0V ; $V_{CM} = 2.5\text{V}$; $R_L = 0\text{V}$; $\text{SHDN} = 0.2\text{V}$, unless otherwise specified. (Note 5)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
I_{OS}	Input Offset Current (Note 7)	LT6010AS8		20	110	pA
		$T_A = 0^\circ\text{C to } 70^\circ\text{C}$	●		150	pA
		$T_A = -40^\circ\text{C to } 85^\circ\text{C}$	●		200	pA
		LT6010S8		40	200	pA
		$T_A = 0^\circ\text{C to } 70^\circ\text{C}$	●		300	pA
		$T_A = -40^\circ\text{C to } 85^\circ\text{C}$	●		400	pA
		LT6010ADD		20	200	pA
		$T_A = 0^\circ\text{C to } 70^\circ\text{C}$	●		300	pA
I_B	Input Bias Current (Note 7)	$T_A = -40^\circ\text{C to } 85^\circ\text{C}$	●		400	pA
		LT6010AS8		20	± 110	pA
		$T_A = 0^\circ\text{C to } 70^\circ\text{C}$	●		± 150	pA
		$T_A = -40^\circ\text{C to } 85^\circ\text{C}$	●		± 200	pA
		LT6010S8		40	± 200	pA
		$T_A = 0^\circ\text{C to } 70^\circ\text{C}$	●		± 300	pA
		$T_A = -40^\circ\text{C to } 85^\circ\text{C}$	●		± 400	pA
		LT6010ADD		20	± 200	pA
	Input Noise Voltage	$T_A = 0^\circ\text{C to } 70^\circ\text{C}$	●		± 300	pA
		$T_A = -40^\circ\text{C to } 85^\circ\text{C}$	●		± 400	pA
		LT6010DD		40	± 300	pA
		$T_A = 0^\circ\text{C to } 70^\circ\text{C}$	●		± 400	pA
		$T_A = -40^\circ\text{C to } 85^\circ\text{C}$	●		± 500	pA
		Input Noise Voltage		400		nV _{p-p}
		Input Noise Voltage Density		14		nV/ $\sqrt{\text{Hz}}$
		Input Noise Current Density		0.1		pA/ $\sqrt{\text{Hz}}$
R_{IN}	Input Resistance	Common Mode, $V_{CM} = 1\text{V to } 3.8\text{V}$		10	120	G Ω
		Differential			20	M Ω
C_{IN}	Input Capacitance			4		pF
V_{CM}	Input Voltage Range (Positive)	Guaranteed by CMRR	●	3.8	4	V
	Input Voltage Range (Negative)	Guaranteed by CMRR	●		0.7	V
CMRR	Common Mode Rejection Ratio	$V_{CM} = 1\text{V to } 3.8\text{V}$	●	107	135	dB
	Minimum Supply Voltage	Guaranteed by PSRR	●		2.4	V
PSRR	Power Supply Rejection Ratio	$V_S = 2.7\text{V to } 36\text{V}$, $V_{CM} = 1/2V_S$	●	112	135	dB
A_{VOL}	Large-Signal Voltage Gain	$R_L = 10\text{k}$, $V_{OUT} = 1\text{V to } 4\text{V}$	●	300	2000	V/mV
		$R_L = 2\text{k}$, $V_{OUT} = 1\text{V to } 4\text{V}$	●	250	2000	V/mV
V_{OUT}	Maximum Output Swing (Positive, Referred to V^+)	No Load, 50mV Overdrive	●	35	55	mV
		$I_{SOURCE} = 1\text{mA}$, 50mV Overdrive	●	120	170	mV
	Maximum Output Swing (Negative, Referred to 0V)	No Load, 50mV Overdrive	●	40	55	mV
		$I_{SINK} = 1\text{mA}$, 50mV Overdrive	●	150	225	mV
			●		275	mV

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^\circ\text{C}$. $V_S = 5\text{V}$, 0V ; $V_{CM} = 2.5\text{V}$; R_L to 0V ; $\text{SHDN} = 0.2\text{V}$, unless otherwise specified. (Note 5)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
I_{SC}	Output Short-Circuit Current (Note 3)	$V_{OUT} = 0\text{V}$, 1V Overdrive (Source)	10 4	14		mA mA
		$V_{OUT} = 5\text{V}$, -1V Overdrive (Sink)	10 4	21		mA mA
SR	Slew Rate	$A_V = -10$, $R_F = 50\text{k}$, $R_G = 5\text{k}$	0.06	0.09		V/ μs
		$T_A = 0^\circ\text{C}$ to 70°C	0.05			V/ μs
		$T_A = -40^\circ\text{C}$ to 85°C	0.04			V/ μs
GBW	Gain Bandwidth Product	$f = 10\text{kHz}$	250 225	330		kHz kHz
t_s	Settling Time	$A_V = -1$, 0.01%, $V_{OUT} = 1.5\text{V}$ to 3.5V		45		μs
t_r , t_f	Rise Time, Fall Time	$A_V = 1$, 10% to 90%, 0.1V Step		1		μs
I_{SHDN}	SHDN Pin Current	$\text{SHDN} \leq V^- + 0.2\text{V}$ (On)			0.25	μA
		$\text{SHDN} = V^- + 2.0\text{V}$ (Off)		15	25	μA
t_{SHDN}	SHDN Turn-On, Turn-Off Time	$\text{SHDN} = V^-$ (On) to $V^- + 2.0\text{V}$ (Off)		25		μs
		$\text{SHDN} = V^- + 2.0\text{V}$ (Off) to V^- (On)		25		μs
I_S	Supply Current	$\text{SHDN} \leq V^- + 0.2\text{V}$ (On)		135	150	μA
		$T_A = 0^\circ\text{C}$ to 70°C			190	μA
		$T_A = -40^\circ\text{C}$ to 85°C			210	μA
		$\text{SHDN} = V^- + 2.0\text{V}$ (Off)		12	25	μA
					50	μA

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^\circ\text{C}$. $V_S = \pm 15\text{V}$, $V_{CM} = 0\text{V}$, R_L to 0V ; $\text{SHDN} = -14.8\text{V}$, unless otherwise specified. (Note 5)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{OS}	Input Offset Voltage (Note 7)	LT6010AS8		10	60	μV
		$T_A = 0^\circ\text{C}$ to 70°C			80	μV
		$T_A = -40^\circ\text{C}$ to 85°C			110	μV
		LT6010S8		20	85	μV
		$T_A = 0^\circ\text{C}$ to 70°C			120	μV
		$T_A = -40^\circ\text{C}$ to 85°C			160	μV
		LT6010ADD		20	85	μV
		$T_A = 0^\circ\text{C}$ to 70°C			105	μV
$\Delta V_{OS}/\Delta T$	Input Offset Voltage Drift (Note 6)	$T_A = -40^\circ\text{C}$ to 85°C			135	μV
		LT6010DD		30	110	μV
		$T_A = 0^\circ\text{C}$ to 70°C			145	μV
		$T_A = -40^\circ\text{C}$ to 85°C			185	μV
I_{OS}	Input Offset Current (Note 7)	LT6010AS8		20	110	pA
		$T_A = 0^\circ\text{C}$ to 70°C			150	pA
		$T_A = -40^\circ\text{C}$ to 85°C			200	pA
		LT6010S8		40	200	pA
		$T_A = 0^\circ\text{C}$ to 70°C			300	pA
		$T_A = -40^\circ\text{C}$ to 85°C			400	pA
		LT6010ADD		20	200	pA
		$T_A = 0^\circ\text{C}$ to 70°C			300	pA
		$T_A = -40^\circ\text{C}$ to 85°C			400	pA

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^\circ\text{C}$. $V_S = \pm 15\text{V}$, $V_{CM} = 0\text{V}$, R_L to 0V ; $\text{SHDN} = -14.8\text{V}$, unless otherwise specified. (Note 5)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
I_{OS}	Input Offset Current (Note 7)	LT6010DD		40	300	pA
		$T_A = 0^\circ\text{C to } 70^\circ\text{C}$	●		400	pA
		$T_A = -40^\circ\text{C to } 85^\circ\text{C}$	●		500	pA
I_B	Input Bias Current (Note 7)	LT6010AS8		20	± 110	pA
		$T_A = 0^\circ\text{C to } 70^\circ\text{C}$	●		± 150	pA
		$T_A = -40^\circ\text{C to } 85^\circ\text{C}$	●		± 200	pA
		LT6010S8		40	± 200	pA
		$T_A = 0^\circ\text{C to } 70^\circ\text{C}$	●		± 300	pA
		$T_A = -40^\circ\text{C to } 85^\circ\text{C}$	●		± 400	pA
		LT6010ADD		20	± 200	pA
		$T_A = 0^\circ\text{C to } 70^\circ\text{C}$	●		± 300	pA
		$T_A = -40^\circ\text{C to } 85^\circ\text{C}$	●		± 400	pA
		LT6010DD		40	± 300	pA
		$T_A = 0^\circ\text{C to } 70^\circ\text{C}$	●		± 400	pA
		$T_A = -40^\circ\text{C to } 85^\circ\text{C}$	●		± 500	pA
	Input Noise Voltage	0.1Hz to 10Hz		400		nV _{p-p}
e_n	Input Noise Voltage Density	$f = 1\text{kHz}$		13		nV/ $\sqrt{\text{Hz}}$
i_n	Input Noise Current Density	$f = 1\text{kHz}$		0.1		pA/ $\sqrt{\text{Hz}}$
R_{IN}	Input Resistance	Common Mode, $V_{CM} = \pm 13.5\text{V}$		50	400	G Ω
		Differential			20	M Ω
C_{IN}	Input Capacitance			4		pF
V_{CM}	Input Voltage Range	Guaranteed by CMRR	●	± 13.5	± 14	V
CMRR	Common Mode Rejection Ratio	$V_{CM} = -13.5\text{V to } 13.5\text{V}$		115	135	dB
			●	112		dB
	Minimum Supply Voltage	Guaranteed by PSRR	●	± 1.2	± 1.35	V
PSRR	Power Supply Rejection Ratio	$V_S = \pm 1.35\text{V to } \pm 18\text{V}$	●	112	135	dB
A_{VOL}	Large-Signal Voltage Gain	$R_L = 10\text{k}, V_{OUT} = -13.5\text{V to } 13.5\text{V}$		1000	2000	V/mV
			●	600		V/mV
		$R_L = 5\text{k}, V_{OUT} = -13.5\text{V to } 13.5\text{V}$		500	1500	V/mV
			●	300		V/mV
V_{OUT}	Maximum Output Swing (Positive, Referred to V^+)	No Load, 50mV Overdrive		45	80	mV
			●		100	mV
	$I_{SOURCE} = 1\text{mA}$, 50mV Overdrive			140	195	mV
			●		240	mV
V_{OUT}	Maximum Output Swing (Negative, Referred to 0V)	No Load, 50mV Overdrive		45	80	mV
			●		100	mV
	$I_{SINK} = 1\text{mA}$, 50mV Overdrive			150	250	mV
			●		300	mV
I_{SC}	Output Short-Circuit Current (Note 3)	$V_{OUT} = 0\text{V}$, 1V Overdrive (Source)		10	15	mA
			●	5		mA
		$V_{OUT} = 0\text{V}$, -1V Overdrive (Sink)		10	20	mA
			●	5		mA

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^\circ\text{C}$. $V_S = \pm 15\text{V}$, $V_{CM} = 0\text{V}$, R_L to 0V ; $\text{SHDN} = -14.8\text{V}$, unless otherwise specified. (Note 5)

SYMBOL	PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
SR	Slew Rate	$A_V = -10$, $R_F = 50\text{k}$, $R_G = 5\text{k}$	0.08	0.11		$\text{V}/\mu\text{s}$
		$T_A = 0^\circ\text{C}$ to 70°C	0.07			$\text{V}/\mu\text{s}$
		$T_A = -40^\circ\text{C}$ to 85°C	0.05			$\text{V}/\mu\text{s}$
GBW	Gain Bandwidth Product	$f = 10\text{kHz}$	275	350		kHz
			250			kHz
t_s	Settling Time	$A_V = -1$, 0.01% , $V_{OUT} = 0\text{V}$ to 10V		85		μs
t_r , t_f	Rise Time, Fall Time	$A_V = 1$, 10% to 90% , 0.1V Step		1		μs
I_{SHDN}	SHDN Pin Current	$\text{SHDN} \leq V^- + 0.2\text{V}$ (On)	●		0.25	μA
		$\text{SHDN} = V^- + 2.0\text{V}$ (Off)	●	15	25	μA
t_{SHDN}	SHDN Turn-On, Turn-Off Time	$\text{SHDN} = V^-$ (On) to $V^- + 2.0\text{V}$ (Off)		25		μs
		$\text{SHDN} = V^- + 2.0\text{V}$ (Off) to V^- (On)		25		μs
I_S	Supply Current	$\text{SHDN} \leq V^- + 0.2\text{V}$ (On)		260	330	μA
		$T_A = 0^\circ\text{C}$ to 70°C	●		380	μA
		$T_A = -40^\circ\text{C}$ to 85°C	●		400	μA
		$\text{SHDN} = V^- + 2.0\text{V}$ (Off)		18	50	μA

Note 1: Absolute Maximum Ratings are those beyond which the life of the device may be impaired.

Note 2: The inputs are protected by back-to-back diodes and internal series resistors. If the differential input voltage exceeds 10V , the input current must be limited to less than 10mA .

Note 3: A heat sink may be required to keep the junction temperature below absolute maximum ratings.

Note 4: Both the LT6010C and LT6010I are guaranteed functional over the operating temperature range of -40°C to 85°C .

Note 5: The LT6010C is guaranteed to meet the specified performance

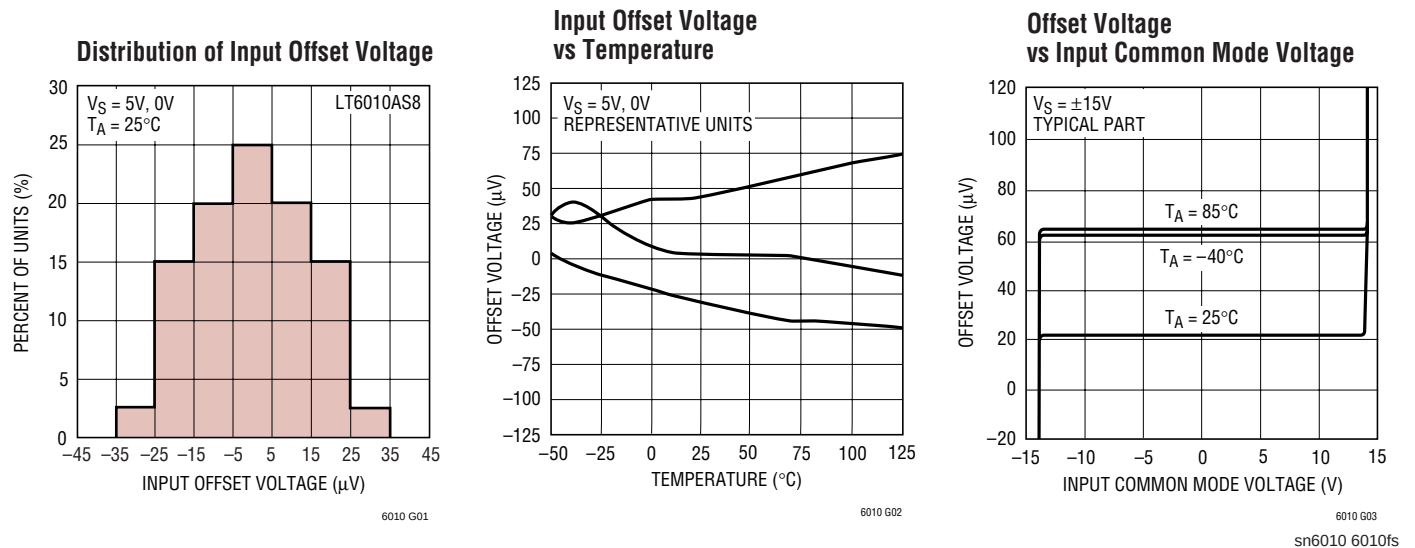
from 0°C to 70°C and is designed, characterized and expected to meet specified performance from -40°C to 85°C but is not tested or QA sampled at these temperatures. The LT6010I is guaranteed to meet specified performance from -40°C to 85°C .

Note 6: This parameter is not 100% tested.

Note 7: The specifications for V_{OS} , I_B and I_{OS} depend on the grade and on the package. The following table clarifies the notations used in the specification table:

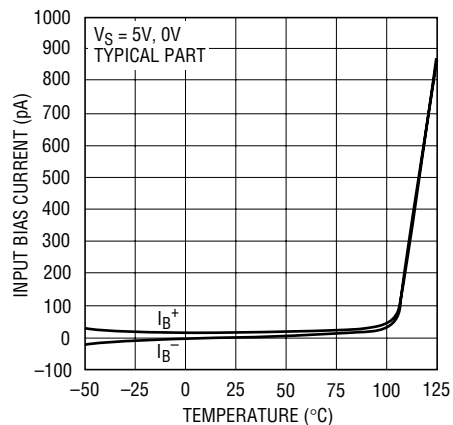
	Standard Grade	A Grade
S8 Package	LT6010S8	LT6010AS8
DFN Package	LT6010DD	LT6010ADD

TYPICAL PERFORMANCE CHARACTERISTICS



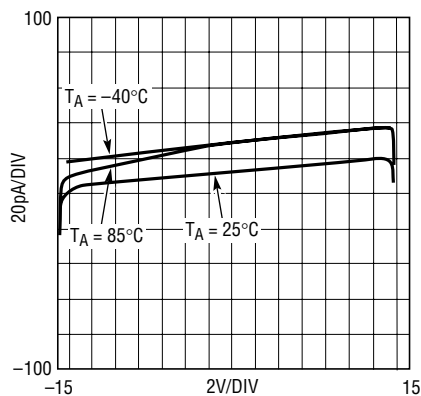
TYPICAL PERFORMANCE CHARACTERISTICS

Input Bias Current vs Temperature



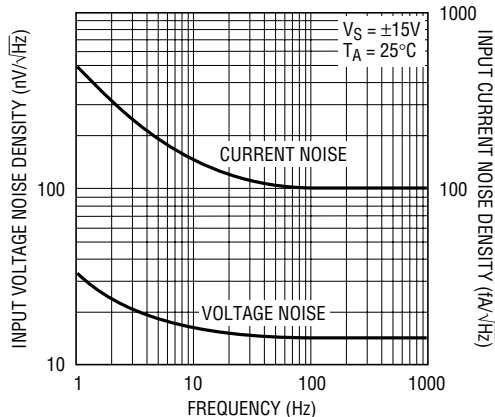
6010 G04

Input Bias Current vs Input Common Mode Voltage



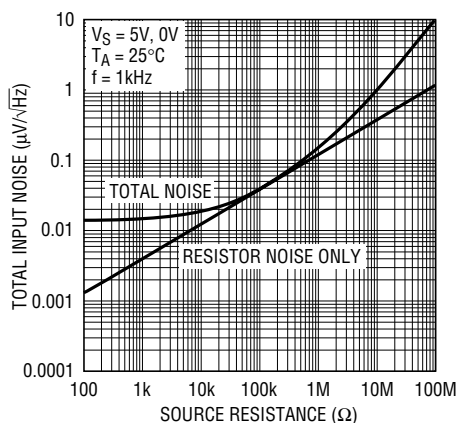
6010 G05

e_n, i_n vs Frequency



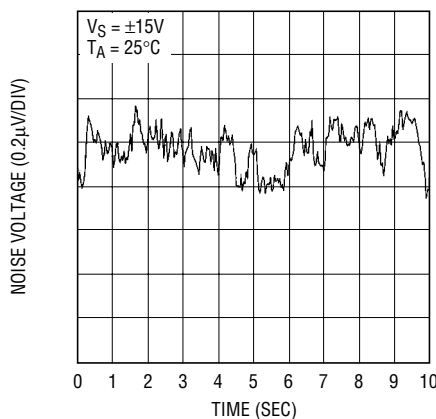
6010 G06

Total Input Noise vs Source Resistance



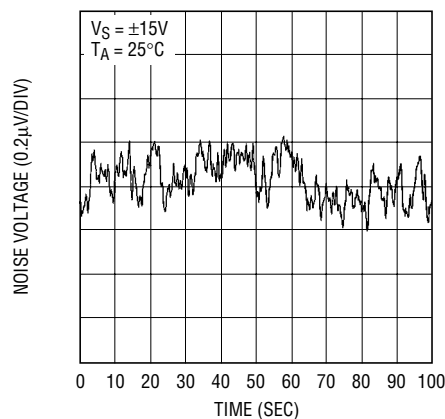
6010 G07

0.1Hz to 10Hz Noise



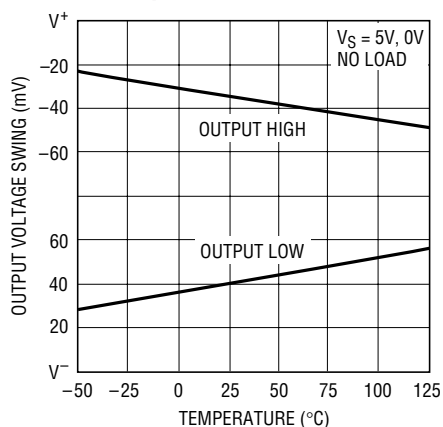
6010 G08

0.01Hz to 1Hz Noise



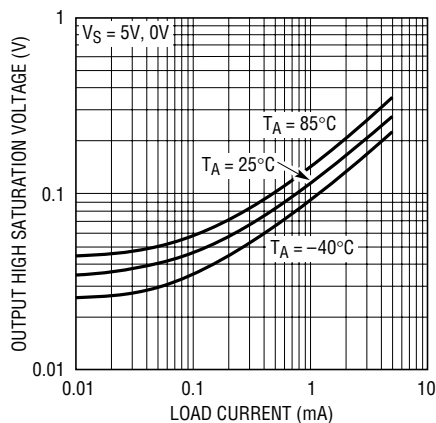
6010 G09

Output Voltage Swing vs Temperature



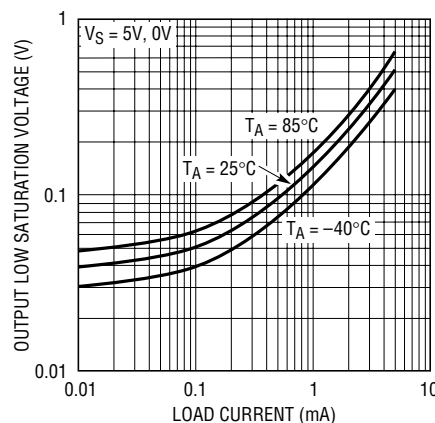
6010 G10

Output Saturation Voltage vs Load Current (Output High)



6010 G11

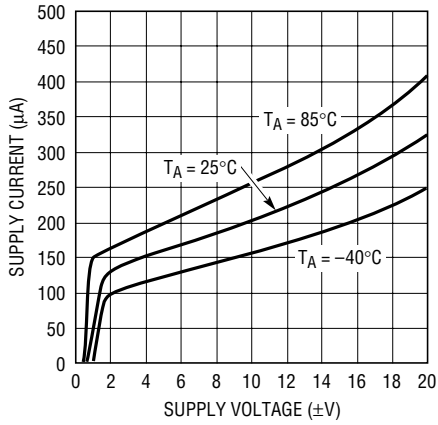
Output Saturation Voltage vs Load Current (Output Low)



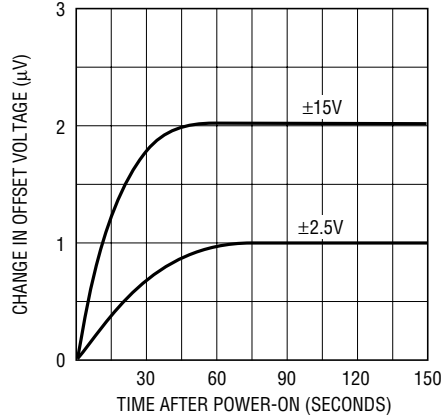
6010 G12

TYPICAL PERFORMANCE CHARACTERISTICS

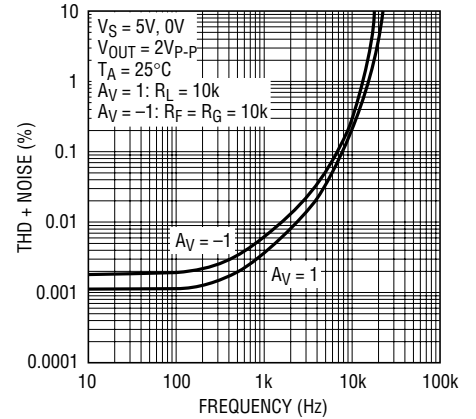
Supply Current vs Supply Voltage



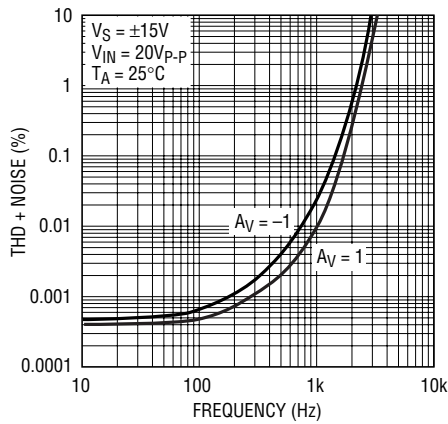
Warm-Up Drift



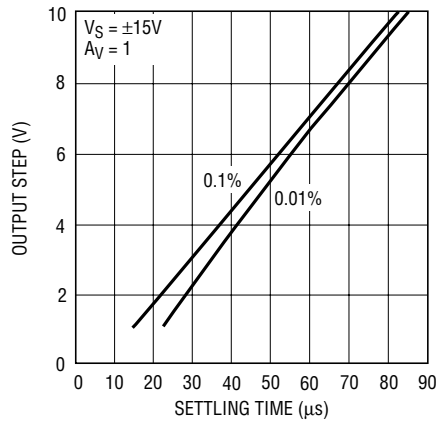
THD + Noise vs Frequency



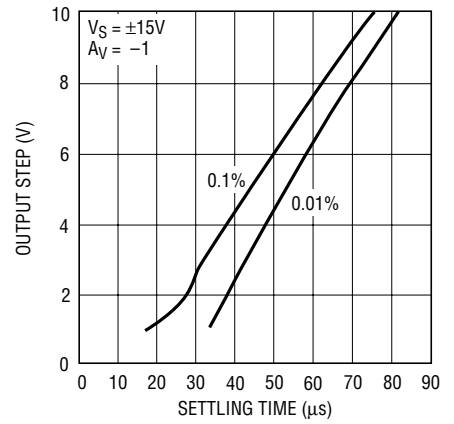
THD + Noise vs Frequency



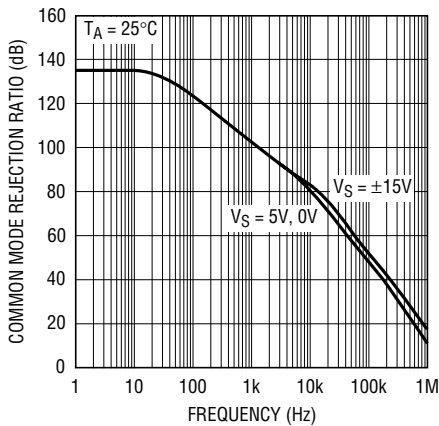
Settling Time vs Output Step



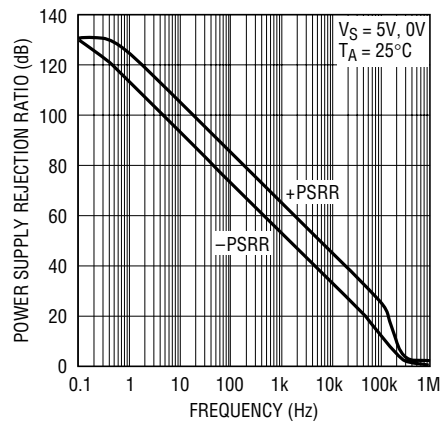
Settling Time vs Output Step



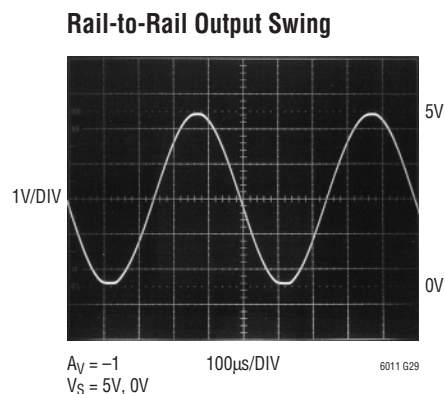
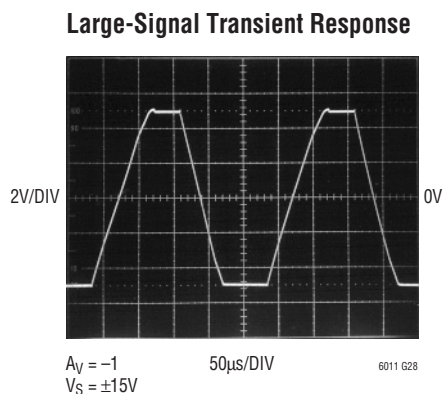
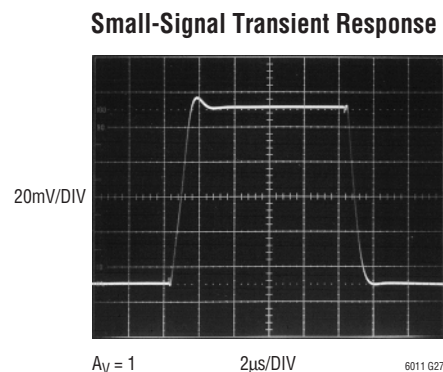
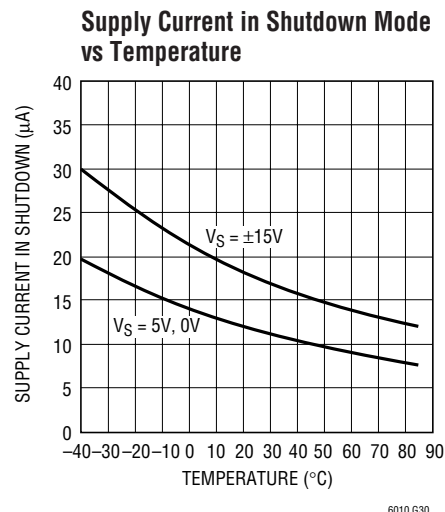
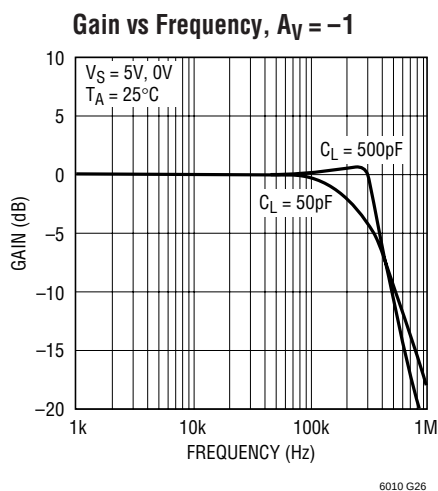
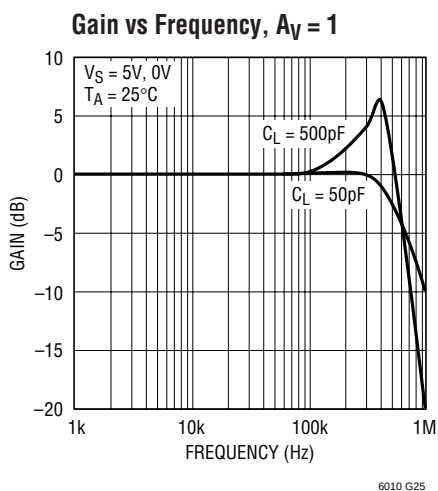
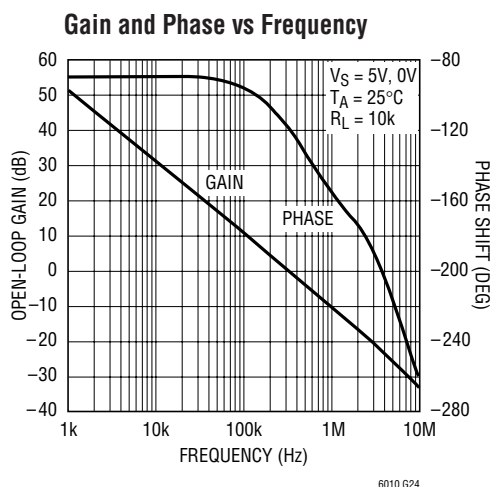
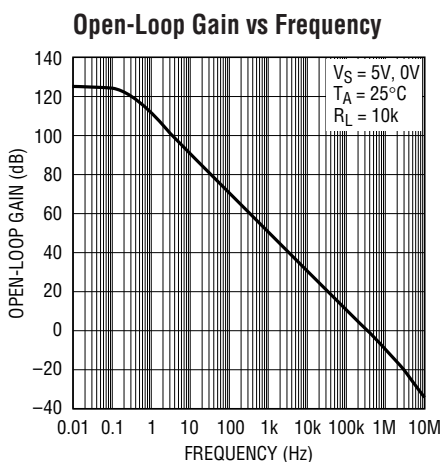
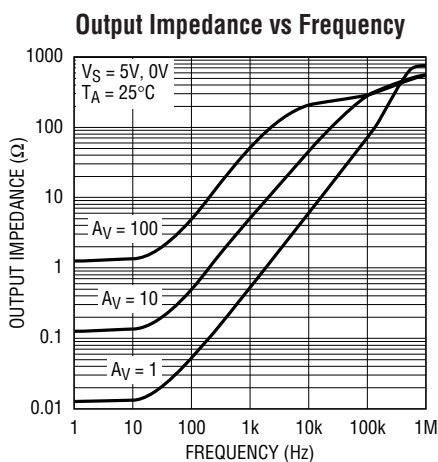
CMRR vs Frequency



PSRR vs Frequency



TYPICAL PERFORMANCE CHARACTERISTICS



APPLICATIONS INFORMATION

Preserving Input Precision

Preserving the input accuracy of the LT6010 requires that the applications circuit and PC board layout do not introduce errors comparable to or greater than the 20 μ V typical offset of the amplifier. Temperature differentials across the input connections can generate thermocouple voltages of 10's of microvolts, so the connections to the input leads should be short, close together, and away from heat dissipating components. Air currents across the board can also generate temperature differentials.

The extremely low input bias currents (20pA typical) allow high accuracy to be maintained with high impedance sources and feedback resistors. The LT6010 low input bias currents are obtained by a cancellation circuit on-chip. The input bias currents are permanently trimmed at wafer testing to a low level. Do not try to balance the input resistances in each input lead; instead, keep the resistance at either input as low as possible for maximum accuracy.

Leakage currents on the PC board can be higher than the LT6010's input bias current. For example, 10G Ω of leakage between a 15V supply lead and an input lead will generate 1.5nA! Surround the input leads by a guard ring, driven to the same potential as the input common mode, to avoid excessive leakage in high impedance applications.

Input Protection

The LT6010 features on-chip back-to-back diodes between the input devices, along with 500 Ω resistors in series with either input. This internal protection limits the input current to approximately 10mA (the maximum

allowed) for a 10V differential input voltage. Use additional external series resistors to limit the input current to 10mA in applications where differential inputs of more than 10V are expected. For example, a 1k resistor in series with each input provides protection against 30V differential voltage.

Input Common Mode Range

The LT6010 output is able to swing nearly to each power supply rail (rail-to-rail out), but the input stage is limited to operating between $V^- + 1V$ and $V^+ - 1.2V$. Exceeding this common mode range will cause the gain to drop to zero, however no phase reversal will occur.

Total Input Noise

The LT6010 amplifier contributes negligible noise to the system when driven by sensors (sources) with impedance between 20k Ω and 1M Ω . Throughout this range, total input noise is dominated by the 4kTR_S noise of the source. If the source impedance is less than 20k Ω , the input voltage noise of the amplifier starts to contribute with a minimum noise of 14nV/ $\sqrt{\text{Hz}}$ for very low source impedance. If the source impedance is more than 1M Ω , the input current noise of the amplifier, multiplied by this high impedance, starts to contribute and eventually dominate. Total input noise spectral density can be calculated as:

$$v_{n(\text{TOTAL})} = \sqrt{e_n^2 + 4kTR_S + (i_n R_S)^2}$$

where $e_n = 14\text{nV}/\sqrt{\text{Hz}}$, $i_n = 0.1\text{pA}/\sqrt{\text{Hz}}$ and R_S the total impedance at the input, including the source impedance.

APPLICATIONS INFORMATION

Offset Voltage Adjustment

The input offset voltage of the LT6010 and its drift with temperature are permanently trimmed at wafer testing to the low level as specified in the electrical characteristic. However, if further adjustment of V_{OS} is desired, nulling with a 50k potentiometer is possible and will not degrade drift with temperature. Trimming to a value other than zero

creates a drift of $(V_{OS}/300\mu V)\mu V/^{\circ}C$, e.g., if V_{OS} is adjusted to $300\mu V$, the change in drift will be $1\mu V/^{\circ}C$. The adjustment range with a 50k pot is approximately $\pm 0.9mV$ (see Figures 1A and 1B). The sensitivity and resolution of the nulling can be improved by using a smaller pot in conjunction with fixed resistors. The configuration shown has an approximate nulling range of $\pm 150\mu V$ (see Figures 2A and 2B).

Standard Adjustment

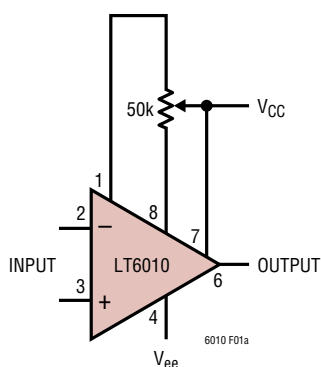


Figure 1A

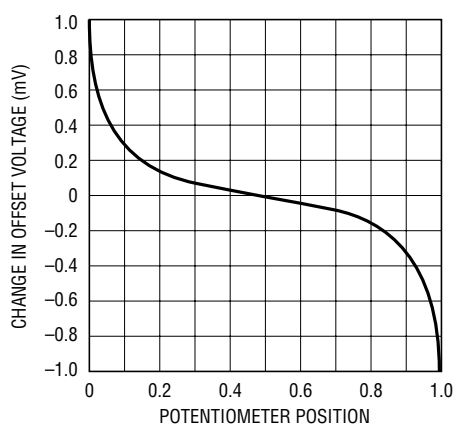


Figure 1B

Improved Sensitivity Adjustment

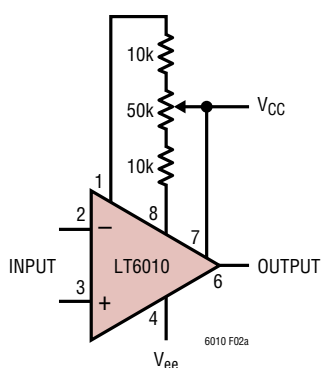


Figure 2A

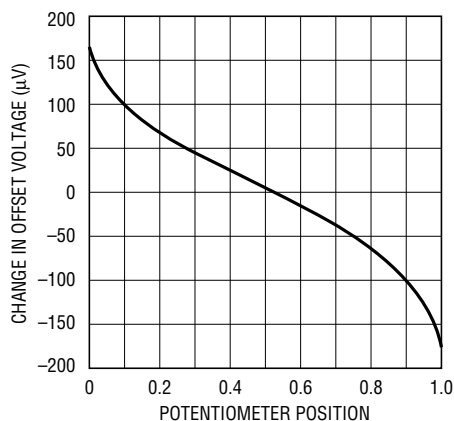


Figure 2B

APPLICATIONS INFORMATION

Shutdown

The LT6010 can be put into shutdown mode to conserve power. When the SHDN pin is biased at less than 0.2V above the negative supply, the part operates normally. When pulled 2V or more above V^- , the supply current drops to about 12 μ A, shutting down the op amp.

The output of the LT6010 op amp is not isolated from the inputs while in shutdown mode. Therefore, this shutdown feature cannot be used for multiplexing applications.

There is an internal 85k resistor at the SHDN pin. If the SHDN voltage source is more than 2V above the negative supply, an external series resistor can be placed between the source and SHDN pin to reduce SHDN pin current (see Figure 3). For an example of suggested values see Table 1. The resistors listed ensure that the voltage at the SHDN pin is 2V above the negative supply.

Table 1

V_{SHDN} (V)	R_{SHDN} (k Ω)
2	NONE
3	77k
4	153k
5	230k

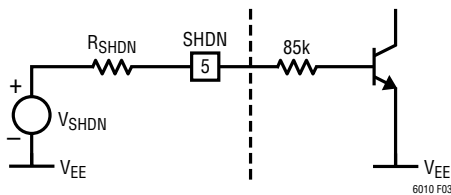


Figure 3

Capacitive Loads

The LT6010 can drive capacitive loads up to 500pF in unity gain. The capacitive load driving capability increases as the amplifier is used in higher gain configurations. A small series resistance between the output and the load further increases the amount of capacitance that the amplifier can drive.

Rail-to-Rail Operation

The LT6010 outputs can swing to within millivolts of either supply rail, but the inputs cannot. However, for most op amp configurations, the inputs need to swing less than the outputs. Figure 4 shows the basic op amp configurations, lists what happens to the op amp inputs and specifies whether or not the op amp must have rail-to-rail inputs. Select a rail-to-rail input op amp only when really necessary, because the input precision specifications are usually inferior.

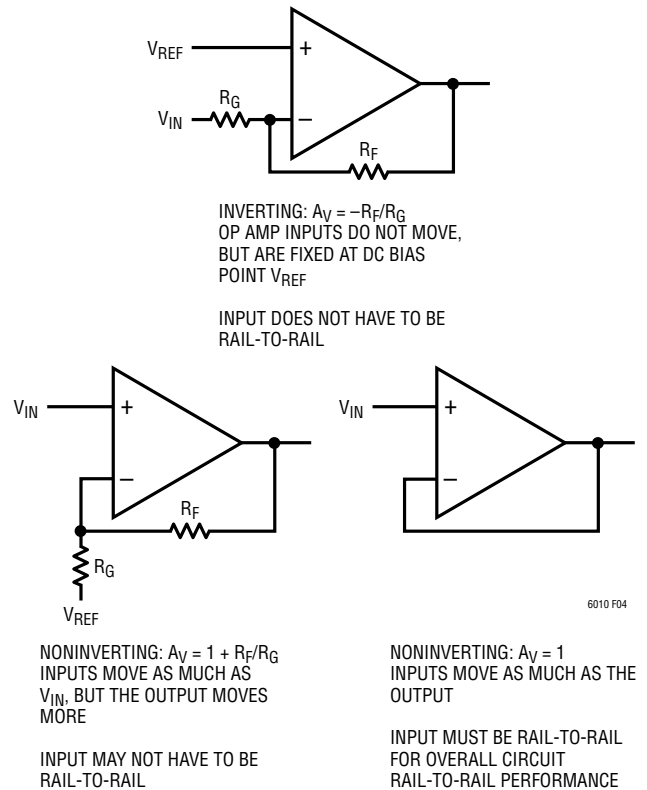
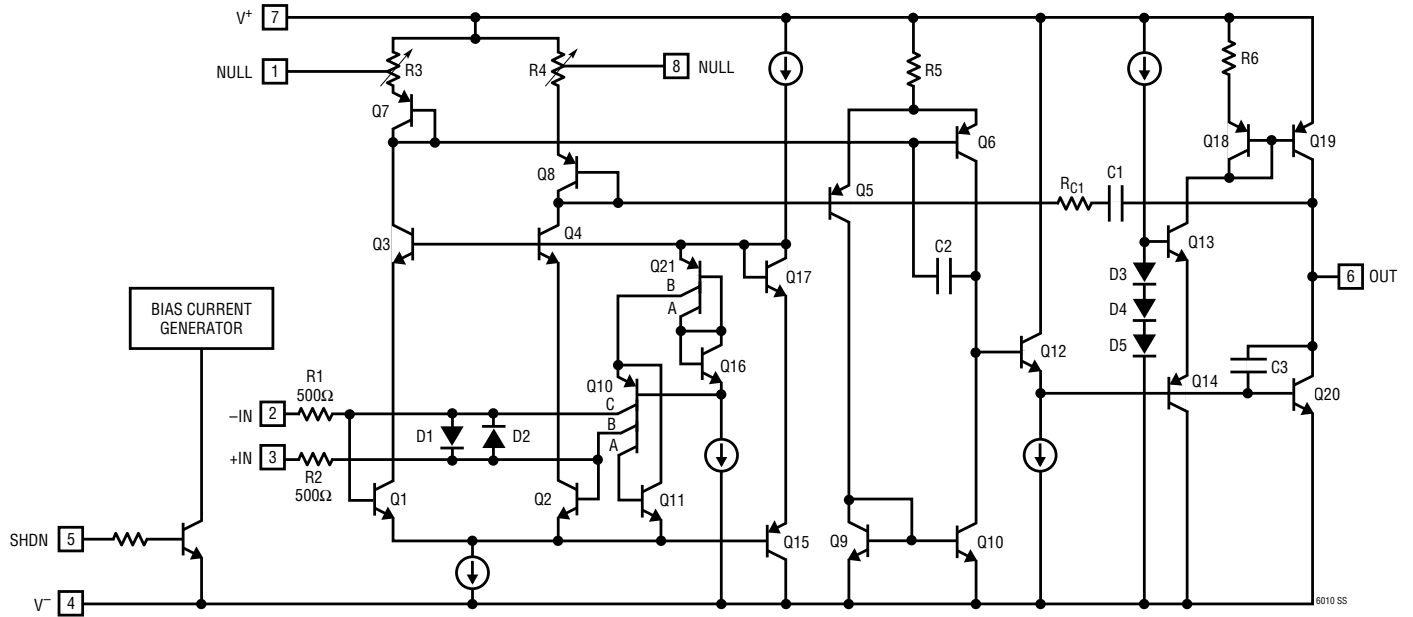


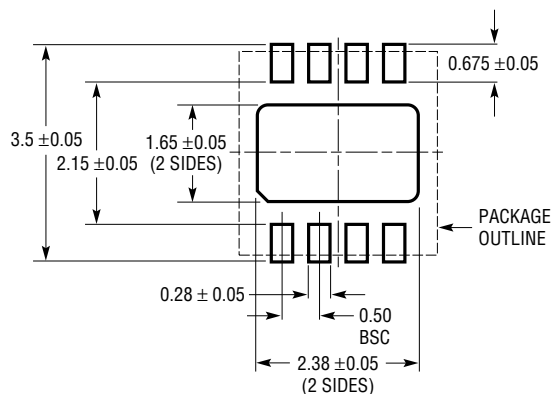
Figure 4. Some Op Amp Configurations Do Not Require Rail-to-Rail Inputs to Achieve Rail-to-Rail Outputs

SIMPLIFIED SCHEMATIC

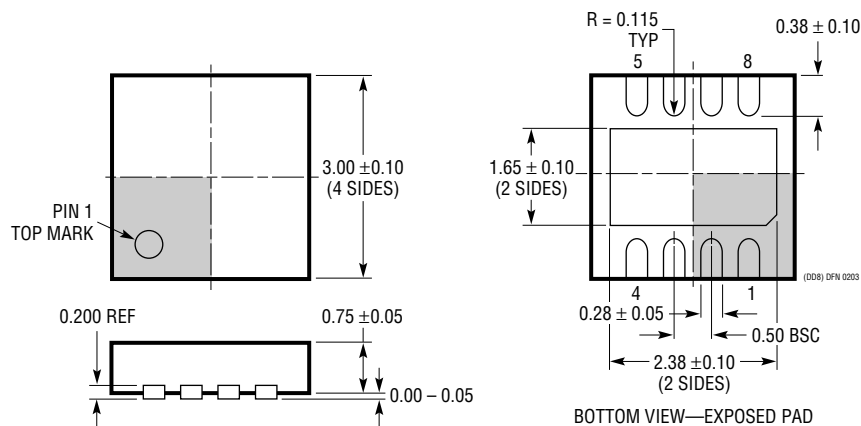


PACKAGE DESCRIPTION

DD Package
8-Lead Plastic DFN (3mm × 3mm)
 (Reference LTC DWG # 05-08-1698)



RECOMMENDED SOLDER PAD PITCH AND DIMENSIONS

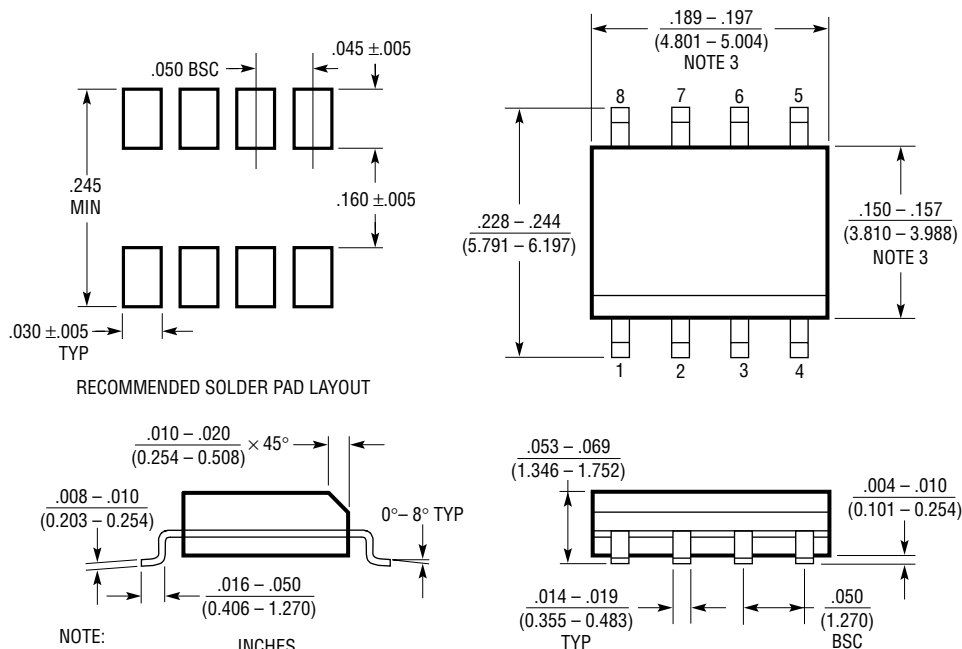


NOTE:

1. DRAWING TO BE MADE A JEDEC PACKAGE OUTLINE MO-229 VARIATION OF (WEED-1)
2. ALL DIMENSIONS ARE IN MILLIMETERS
3. DIMENSIONS OF EXPOSED PAD ON BOTTOM OF PACKAGE DO NOT INCLUDE MOLD FLASH. MOLD FLASH, IF PRESENT, SHALL NOT EXCEED 0.15mm ON ANY SIDE
4. EXPOSED PAD SHALL BE SOLDER PLATED

PACKAGE DESCRIPTION

S8 Package 8-Lead Plastic Small Outline (Narrow .150 Inch) (Reference LTC DWG # 05-08-1610)

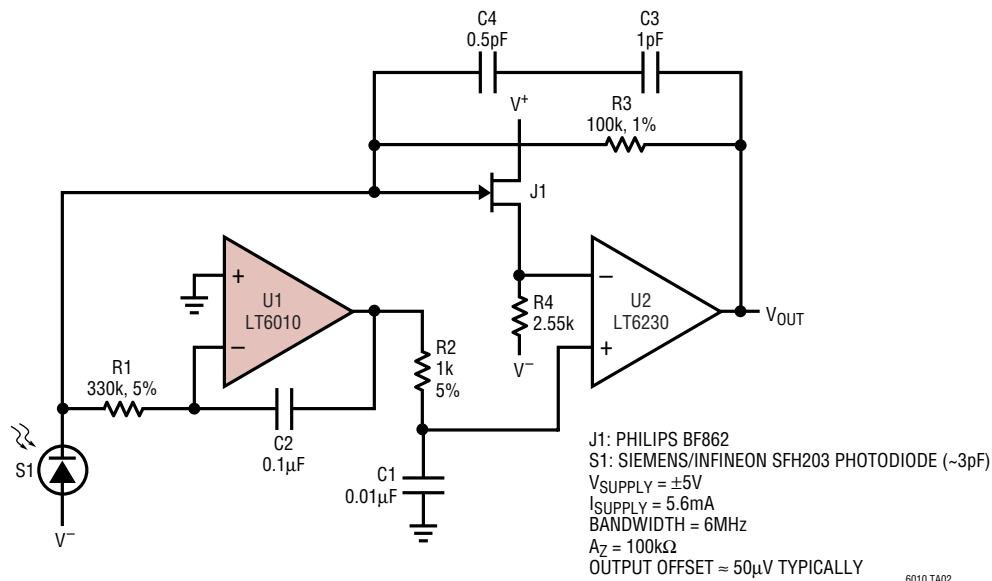


- NOTE:
1. DIMENSIONS IN $\frac{\text{INCHES}}{\text{MILLIMETERS}}$
 2. DRAWING NOT TO SCALE
 3. THESE DIMENSIONS DO NOT INCLUDE MOLD FLASH OR PROTRUSIONS.
MOLD FLASH OR PROTRUSIONS SHALL NOT EXCEED .006" (0.15mm)

S08 0303

TYPICAL APPLICATION

Precision JFET Input Transimpedance Photodiode Amplifier



RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LT6011/6012	Dual/Quad Precision Op Amps	135µA, Rail-to-Rail Output
LT1001	Low Power, Picoamp Input Precision Op Amp	250pA Input Bias Current
LT1880	Rail-to-Rail Output, Picoamp Input Precision Op Amp	C _{LOAD} up to 1000pF

Данный компонент на территории Российской Федерации

Вы можете приобрести в компании MosChip.

Для оперативного оформления запроса Вам необходимо перейти по данной ссылке:

<http://moschip.ru/get-element>

Вы можете разместить у нас заказ для любого Вашего проекта, будь то серийное производство или разработка единичного прибора.

В нашем ассортименте представлены ведущие мировые производители активных и пассивных электронных компонентов.

Нашей специализацией является поставка электронной компонентной базы двойного назначения, продукции таких производителей как XILINX, Intel (ex.ALTERA), Vicor, Microchip, Texas Instruments, Analog Devices, Mini-Circuits, Amphenol, Glenair.

Сотрудничество с глобальными дистрибьюторами электронных компонентов, предоставляет возможность заказывать и получать с международных складов практически любой перечень компонентов в оптимальные для Вас сроки.

На всех этапах разработки и производства наши партнеры могут получить квалифицированную поддержку опытных инженеров.

Система менеджмента качества компании отвечает требованиям в соответствии с ГОСТ Р ИСО 9001, ГОСТ РВ 0015-002 и ЭС РД 009

Офис по работе с юридическими лицами:

105318, г.Москва, ул.Щербаковская д.3, офис 1107, 1118, ДЦ «Щербаковский»

Телефон: +7 495 668-12-70 (многоканальный)

Факс: +7 495 668-12-70 (доб.304)

E-mail: info@moschip.ru

Skype отдела продаж:

moschip.ru

moschip.ru_4

moschip.ru_6

moschip.ru_9