

100mW Stereo Headphone Amplifier

- Operating from **V_{cc}=2V to 5.5V**
- 100mW into 16Ω at 5V
- 38mW into 16Ω at 3.3V
- 11.5mW into 16Ω at 2V
- Switch ON/OFF click reduction circuitry
- High power supply rejection ratio: 85dB at 5V
- High signal-to-noise ratio: 110dB(A) at 5V
- High crosstalk immunity: 100dB (F=1kHz)
- Rail-to-rail input and output
- Unity-gain stable
- Available in **SO-8, MiniSO-8 & DFN8**

Description

The TS482 is a dual audio power amplifier able to drive a 16 or 32Ω stereo headset down to low voltages.

It is delivering up to 100mW per channel (into 16Ω loads) of continuous average power with 0.1% THD+N from a 5V power supply.

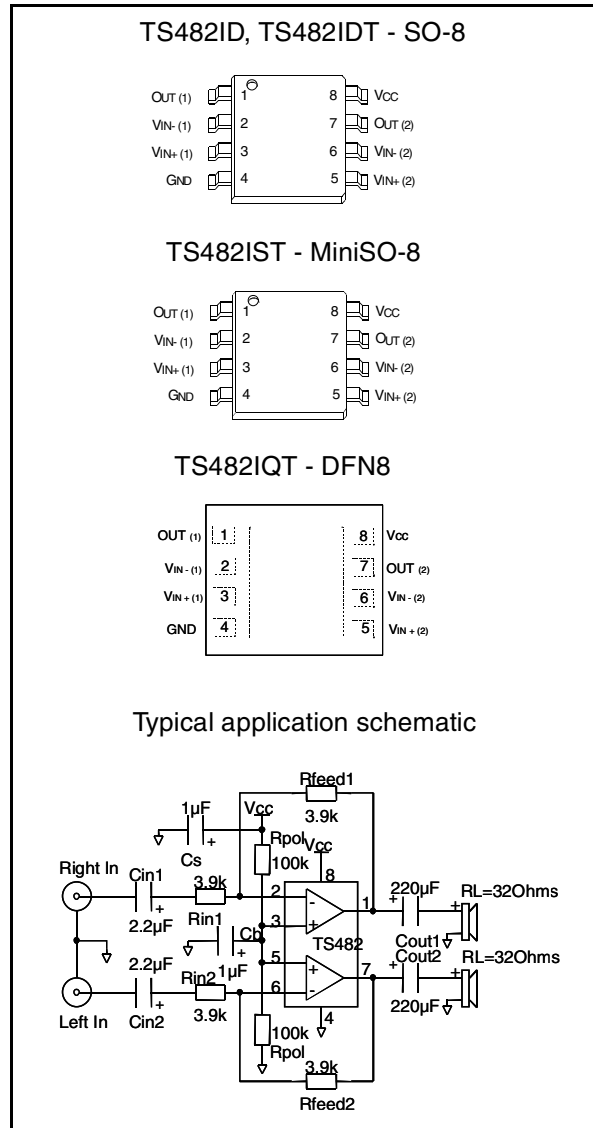
The unity gain stable TS482 can be configured by external gain-setting resistors.

Applications

- Stereo headphone amplifier
- Optical storage
- Computer motherboard
- PDA, organizers & notebook computers
- High-end TV, set-top box, DVD players
- Sound cards

Order Codes

Part Number	Temperature Range	Package	Packing	Marking
TS482ID/IDT	-40, +85°C	SO-8	Tube or Tape & Reel	4821
TS482IST		miniSO-8	Tape & Reel	
TS482IQT		DFN8		



1 Absolute Maximum Ratings

Table 1. Key parameters and their absolute maximum ratings

Symbol	Parameter	Value	Unit
V _{CC}	Supply voltage ⁽¹⁾	6	V
V _i	Input Voltage	-0.3 to V _{CC} +0.3	V
T _{oper}	Operating Free Air Temperature Range	-40 to + 85	°C
T _{stg}	Storage Temperature	-65 to +150	°C
T _j	Maximum Junction Temperature	150	°C
R _{thja}	Thermal Resistance Junction to Ambient SO8 MiniSO8 DFN8	175 215 70	°C/W
Pd	Power Dissipation ⁽²⁾ SO-8 MiniSO-8 DFN8	0.71 0.58 1.79	W
ESD	Human Body Model (pin to pin)	2	kV
ESD	Machine Model - 220pF - 240pF (pin to pin)	200	V
Latch-up	Latch-up Immunity (all pins)	200	mA
	Lead Temperature (soldering, 10sec)	250	°C
	Lead Temperature (soldering, 10sec) for lead-free	260	°C
	Output Short-Circuit Duration	see note ⁽³⁾	

1. All voltages values are measured with respect to the ground pin.
2. Pd has been calculated with Tamb = 25°C, Tjunction = 150°C.
3. Attention must be paid to continuous power dissipation. Exposure of the IC to a short circuit on one or two amplifiers simultaneously can cause excessive heating and the destruction of the device.

Table 2. Operating conditions

Symbol	Parameter	Value	Unit
V _{CC}	Supply Voltage	2 to 5.5	V
R _L	Load Resistor	>= 16	Ω
C _L	Load Capacitor R _L = 16 to 100Ω R _L > 100Ω	400 100	pF
V _{icm}	Common Mode Input Voltage Range	G _{ND} to V _{CC}	V
R _{thja}	Thermal Resistance Junction to Ambient SO-8 MiniSO-8 DFN8 ⁽¹⁾	150 190 41	°C/W

1. When mounted on a 4-layer PCB.

2 Electrical Characteristics

Table 3. Electrical characteristics when $V_{CC} = +5V$, $GND = 0V$, $T_{amb} = 25^{\circ}C$ (unless otherwise specified)

Symbol	Parameter	Min.	Typ.	Max.	Unit
I_{CC}	Supply Current No input signal, no load		5.5	7.2	mA
V_{IO}	Input Offset Voltage ($V_{ICM} = V_{CC}/2$)		1	5	mV
I_{IB}	Input Bias Current ($V_{ICM} = V_{CC}/2$)		200	500	nA
P_O	Output Power THD+N = 0.1% Max, F = 1kHz, $R_L = 32\Omega$ THD+N = 1% Max, F = 1kHz, $R_L = 32\Omega$ THD+N = 0.1% Max, F = 1kHz, $R_L = 16\Omega$ THD+N = 1% Max, F = 1kHz, $R_L = 16\Omega$	60 95	65 67.5 100 107		mW
THD + N	Total Harmonic Distortion + Noise ($A_v=-1$) ⁽¹⁾ $R_L = 32\Omega$, $P_{out} = 60mW$, $20Hz \leq F \leq 20kHz$ $R_L = 16\Omega$, $P_{out} = 90mW$, $20Hz \leq F \leq 20kHz$		0.03 0.03		%
PSRR	Power Supply Rejection Ratio ($A_v=1$), inputs floating F = 100Hz, Vripple = 100mVpp		85		dB
I_O	Max Output Current THD +N < 1%, $R_L = 16\Omega$ connected between out and $V_{CC}/2$	106	120		mA
V_O	Output Swing V_{OL} : $R_L = 32\Omega$ V_{OH} : $R_L = 32\Omega$ V_{OL} : $R_L = 16\Omega$ V_{OH} : $R_L = 16\Omega$	4.45 4.2	0.4 4.6 0.55 4.4	0.48 0.65	V
SNR	Signal-to-Noise Ratio (Filter Type A, $A_v=-1$) $R_L = 32\Omega$, THD +N < 0.2%, $20Hz \leq F \leq 20kHz$	95	110		dB
Crosstalk	Channel Separation, $R_L = 32\Omega$ F = 1kHz F = 20Hz to 20kHz Channel Separation, $R_L = 16\Omega$ F = 1kHz F = 20Hz to 20kHz		100 80 100 80		dB
C_I	Input Capacitance		1		pF
GBP	Gain Bandwidth Product ($R_L = 32\Omega$)	1.35	2.2		MHz
SR	Slew Rate, Unity Gain Inverting ($R_L = 16\Omega$)	0.45	0.7		V/ μ s

1. Fig. 68 to 79 show dispersion of these parameters.

Table 4. Electrical characteristics when $V_{CC} = +3.3V$, $GND = 0V$, $T_{amb} = 25^{\circ}C$ (unless otherwise specified) ⁽¹⁾

Symbol	Parameter	Min.	Typ.	Max.	Unit
I_{CC}	Supply Current No input signal, no load		5.3	7.2	mA
V_{IO}	Input Offset Voltage ($V_{ICM} = V_{CC}/2$)		1	5	mV
I_{IB}	Input Bias Current ($V_{ICM} = V_{CC}/2$)		200	500	nA
P_O	Output Power THD+N = 0.1% Max, F = 1kHz, $R_L = 32\Omega$ THD+N = 1% Max, F = 1kHz, $R_L = 32\Omega$ THD+N = 0.1% Max, F = 1kHz, $R_L = 16\Omega$ THD+N = 1% Max, F = 1kHz, $R_L = 16\Omega$	23	27 28 38 42		mW
THD + N	Total Harmonic Distortion + Noise ($A_v=-1$) ⁽¹⁾ $R_L = 32\Omega$, $P_{out} = 16mW$, $20Hz \leq F \leq 20kHz$ $R_L = 16\Omega$, $P_{out} = 35mW$, $20Hz \leq F \leq 20kHz$		0.03 0.03		%
PSRR	Power Supply Rejection Ratio ($A_v=1$), inputs floating F = 100Hz, Vripple = 100mVpp		80		dB
I_O	Max Output Current THD +N < 1%, $R_L = 16\Omega$ connected between out and $V_{CC}/2$	64	75		mA
V_O	Output Swing V_{OL} : $R_L = 32\Omega$ V_{OH} : $R_L = 32\Omega$ V_{OL} : $R_L = 16\Omega$ V_{OH} : $R_L = 16\Omega$	2.85 2.68	0.3 3 0.45 2.85	0.38 0.52	V
SNR	Signal-to-Noise Ratio (Filter Type A, $A_v=-1$) $R_L = 32\Omega$, THD +N < 0.2%, $20Hz \leq F \leq 20kHz$	92	107		dB
Crosstalk	Channel Separation, $R_L = 32\Omega$ F = 1kHz F = 20Hz to 20kHz Channel Separation, $R_L = 16\Omega$ F = 1kHz F = 20Hz to 20kHz		100 80 100 80		dB
C_I	Input Capacitance		1		pF
GBP	Gain Bandwidth Product ($R_L = 32\Omega$)	1.2	2		MHz
SR	Slew Rate, Unity Gain Inverting ($R_L = 16\Omega$)	0.45	0.7		V/ μ s

1. Fig. 68 to 79 show dispersion of these parameters.

1. All electrical values are guaranteed with correlation measurements at 2V and 5V.

Table 5. Electrical characteristics when $V_{CC} = +2.5V$, $GND = 0V$, $T_{amb} = 25^{\circ}C$ (unless otherwise specified)⁽²⁾

Symbol	Parameter	Min.	Typ.	Max.	Unit
I_{CC}	Supply Current No input signal, no load		5.1	7.2	mA
V_{IO}	Input Offset Voltage ($V_{ICM} = V_{CC}/2$)		1	5	mV
I_{IB}	Input Bias Current ($V_{ICM} = V_{CC}/2$)		200	500	nA
P_O	Output Power THD+N = 0.1% Max, F = 1kHz, $R_L = 32\Omega$ THD+N = 1% Max, F = 1kHz, $R_L = 32\Omega$ THD+N = 0.1% Max, F = 1kHz, $R_L = 16\Omega$ THD+N = 1% Max, F = 1kHz, $R_L = 16\Omega$	12.5 17.5	13.5 14.5 20.5 22		mW
THD + N	Total Harmonic Distortion + Noise ($A_v=-1$) ⁽¹⁾ $R_L = 32\Omega$, $P_{out} = 10mW$, $20Hz \leq F \leq 20kHz$ $R_L = 16\Omega$, $P_{out} = 16mW$, $20Hz \leq F \leq 20kHz$		0.03 0.03		%
PSRR	Power Supply Rejection Ratio ($A_v=1$), inputs floating F = 100Hz, Vripple = 100mVpp		75		dB
I_O	Max Output Current THD +N < 1%, $R_L = 16\Omega$ connected between out and $V_{CC}/2$	45	56		mA
V_O	Output Swing V_{OL} : $R_L = 32\Omega$ V_{OH} : $R_L = 32\Omega$ V_{OL} : $R_L = 16\Omega$ V_{OH} : $R_L = 16\Omega$	2.14 1.97	0.25 2.25 0.35 2.15	0.325 0.45	V
SNR	Signal-to-Noise Ratio (Filter Type A, $A_v=-1$) $R_L = 32\Omega$, THD +N < 0.2%, $20Hz \leq F \leq 20kHz$	89	102		dB
Crosstalk	Channel Separation, $R_L = 32\Omega$ F = 1kHz F = 20Hz to 20kHz Channel Separation, $R_L = 16\Omega$ F = 1kHz F = 20Hz to 20kHz		100 80 100 80		dB
C_I	Input Capacitance		1		pF
GBP	Gain Bandwidth Product ($R_L = 32\Omega$)	1.2	2		MHz
SR	Slew Rate, Unity Gain Inverting ($R_L = 16\Omega$)	0.45	0.7		V/ μ s

1. Fig. 68 to 79 show dispersion of these parameters.

2. All electrical values are guaranteed with correlation measurements at 2V and 5V.

Table 6. Electrical characteristics when $V_{CC} = +2V$, $GND = 0V$, $T_{amb} = 25^{\circ}C$ (unless otherwise specified)

Symbol	Parameter	Min.	Typ.	Max.	Unit
I_{CC}	Supply Current No input signal, no load		5	7.2	mA
V_{IO}	Input Offset Voltage ($V_{ICM} = V_{CC}/2$)		1	5	mV
I_{IB}	Input Bias Current ($V_{ICM} = V_{CC}/2$)		200	500	nA
P_O	Output Power THD+N = 0.1% Max, F = 1kHz, $R_L = 32\Omega$ THD+N = 1% Max, F = 1kHz, $R_L = 32\Omega$ THD+N = 0.1% Max, F = 1kHz, $R_L = 16\Omega$ THD+N = 1% Max, F = 1kHz, $R_L = 16\Omega$	7 9.5	8 9 11.5 13		mW
THD + N	Total Harmonic Distortion + Noise ($A_v=-1$) ⁽¹⁾ $R_L = 32\Omega$, $P_{out} = 6.5mW$, $20Hz \leq F \leq 20kHz$ $R_L = 16\Omega$, $P_{out} = 8mW$, $20Hz \leq F \leq 20kHz$		0.02 0.025		%
PSRR	Power Supply Rejection Ratio ($A_v=1$), inputs floating F = 100Hz, Vripple = 100mVpp		75		dB
I_O	Max Output Current THD +N < 1%, $R_L = 16\Omega$ connected between out and $V_{CC}/2$	33	41.5		mA
V_O	Output Swing V_{OL} : $R_L = 32\Omega$ V_{OH} : $R_L = 32\Omega$ V_{OL} : $R_L = 16\Omega$ V_{OH} : $R_L = 16\Omega$	1.67 1.53	0.24 1.73 0.33 1.63	0.295 0.41	V
SNR	Signal-to-Noise Ratio (Filter Type A, $A_v=-1$) $R_L = 32\Omega$, THD +N < 0.2%, $20Hz \leq F \leq 20kHz$	88	101		dB
Crosstalk	Channel Separation, $R_L = 32\Omega$ F = 1kHz F = 20Hz to 20kHz Channel Separation, $R_L = 16\Omega$ F = 1kHz F = 20Hz to 20kHz		100 80 100 80		dB
C_I	Input Capacitance		1		pF
GBP	Gain Bandwidth Product ($R_L = 32\Omega$)	1.2	2		MHz
SR	Slew Rate, Unity Gain Inverting ($R_L = 16\Omega$)	0.42	0.65		V/ μ s

1. Fig. 68 to 79 show dispersion of these parameters.

Table 7. Components description

Components	Functional Description
Rin	Inverting input resistor which sets the closed loop gain in conjunction with Rfeed. This resistor also forms a high pass filter with Cin ($f_c = 1 / (2 \times \pi \times R_{in} \times C_{in})$)
Cin	Input coupling capacitor which blocks the DC voltage at the amplifier input terminal
Rfeed	Feed back resistor which sets the closed loop gain in conjunction with Rin
Cs	Supply Bypass capacitor which provides power supply filtering
Cb	Bypass capacitor which provides half supply filtering
Cout	Output coupling capacitor which blocks the DC voltage at the load input terminal This capacitor also forms a high pass filter with RL ($f_c = 1 / (2 \times \pi \times R_L \times C_{out})$)
Rpol	These 2 resistors form a voltage divider which provide a DC biasing voltage ($V_{cc}/2$) for the 2 amplifiers.
Av	Closed loop gain = $-R_{feed} / R_{in}$

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Figure 1. Open loop gain and phase vs. frequency response

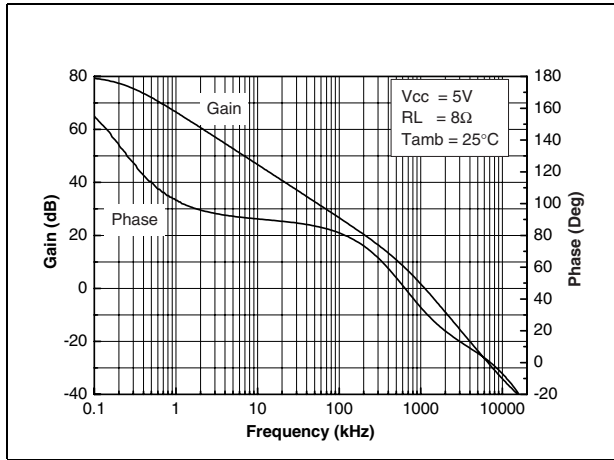


Figure 2. Open loop gain and phase vs. frequency response

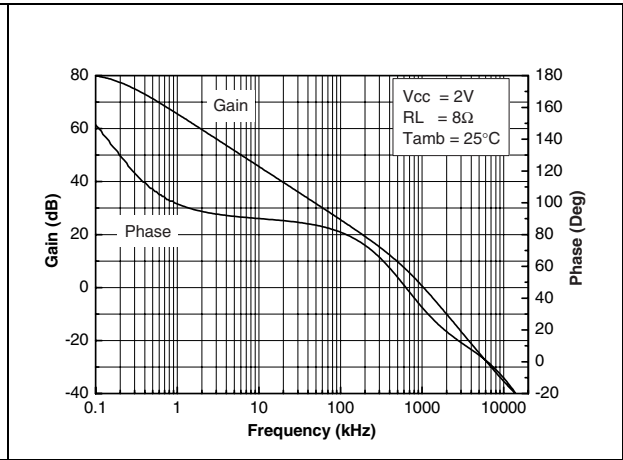


Figure 3. Open loop gain and phase vs. frequency response

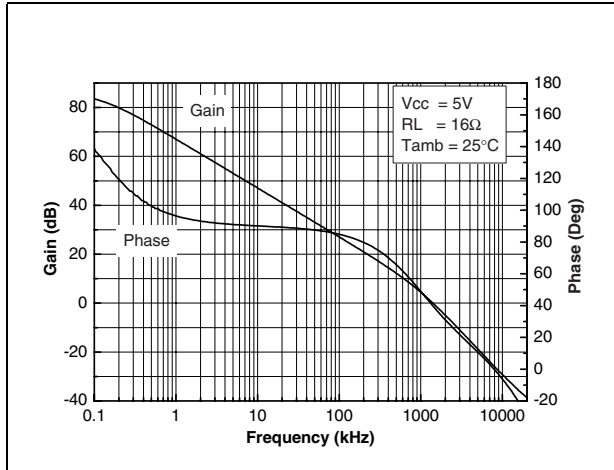


Figure 4. Open loop gain and phase vs. frequency response

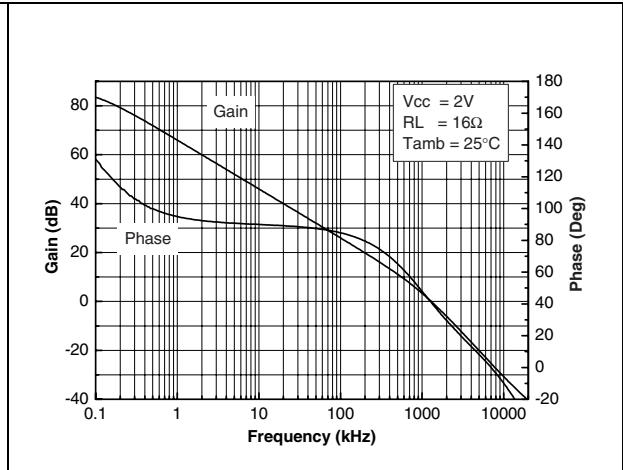


Figure 5. Open loop gain and phase vs. frequency response

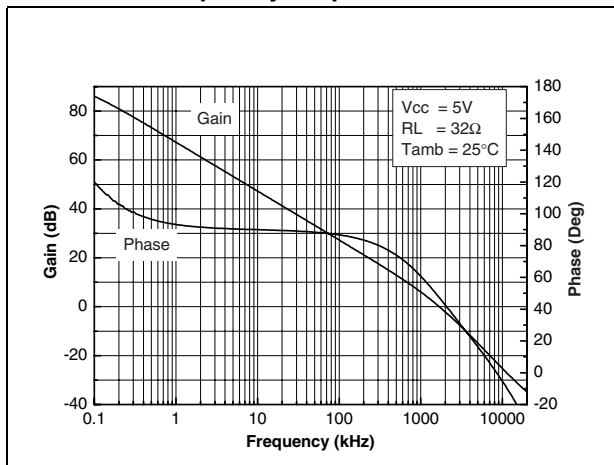


Figure 6. Open loop gain and phase vs. frequency response

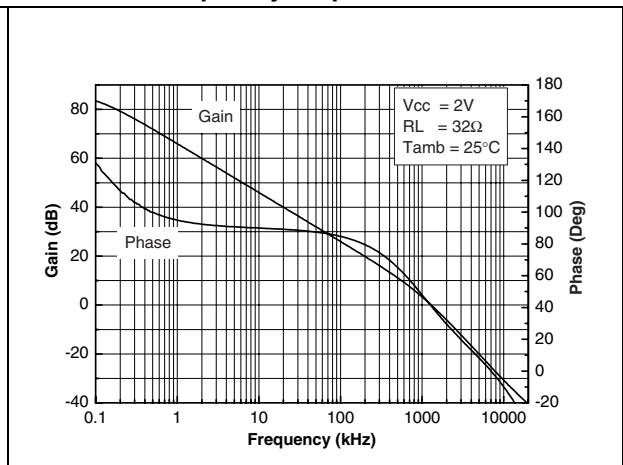


Figure 7. Open loop gain and phase vs. frequency response

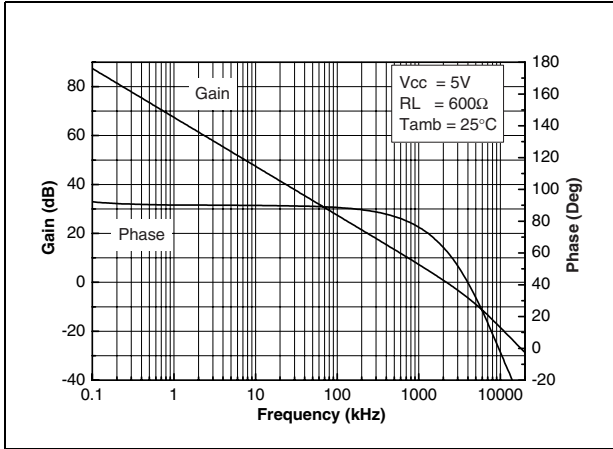


Figure 8. Open loop gain and phase vs. frequency response

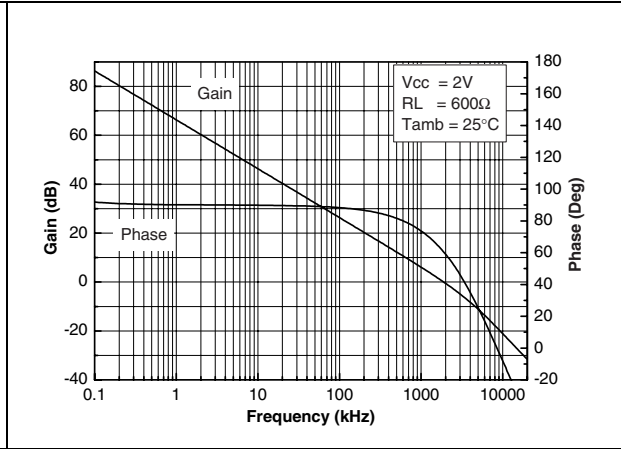


Figure 9. Open loop gain and phase vs. frequency response

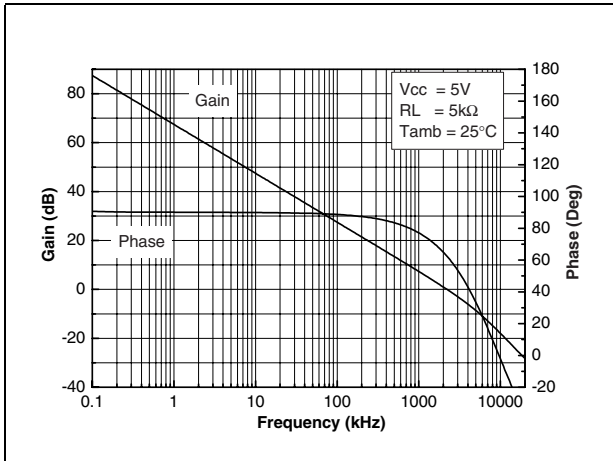


Figure 10. Open loop gain and phase vs. frequency response

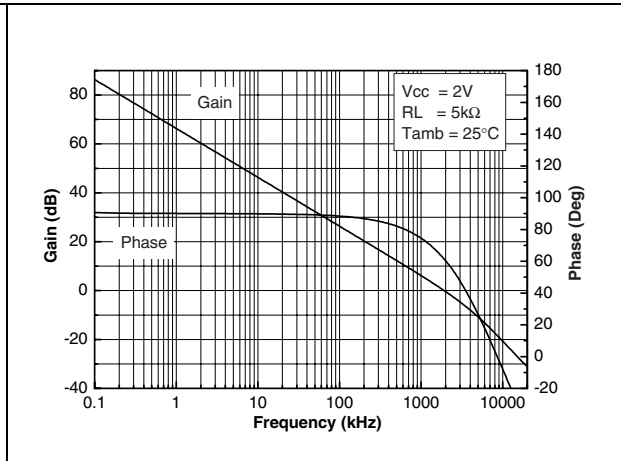


Figure 11. Phase margin vs. power supply voltage

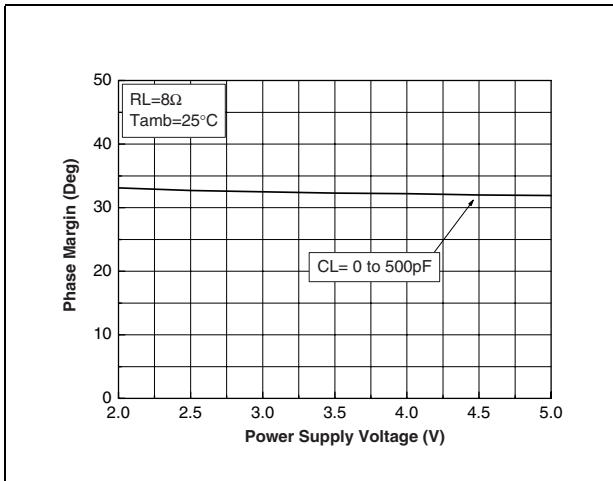


Figure 12. Gain margin vs. power supply voltage

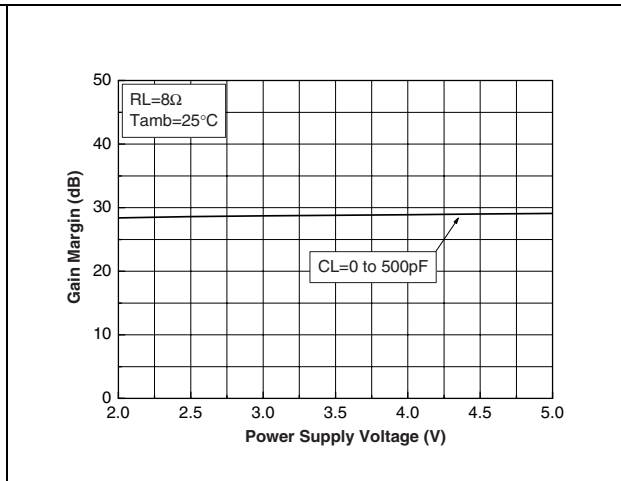


Figure 13. Phase margin vs. power supply voltage Figure 14. Gain margin vs. power supply voltage

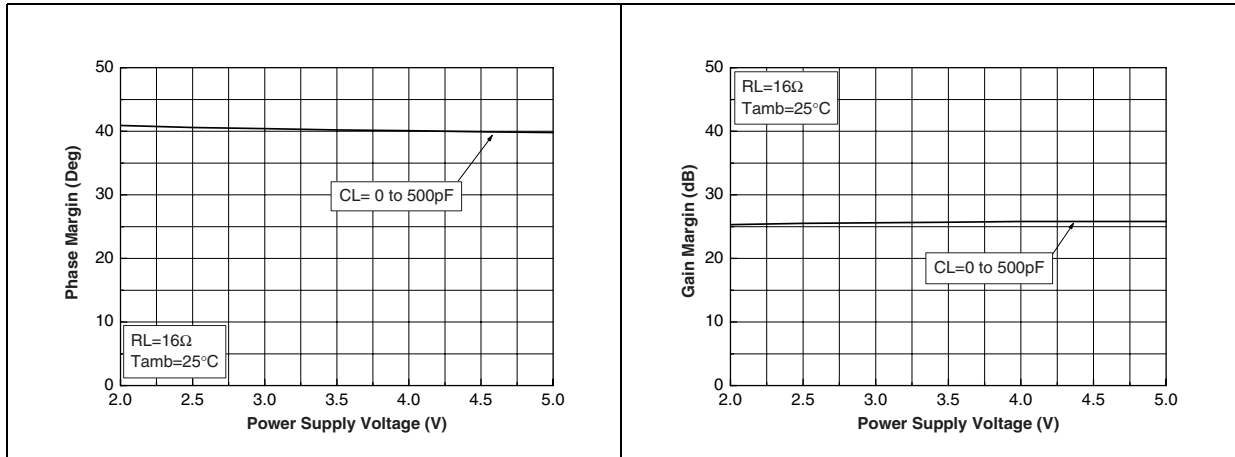


Figure 15. Phase margin vs. power supply voltage Figure 16. Gain margin vs. power supply voltage

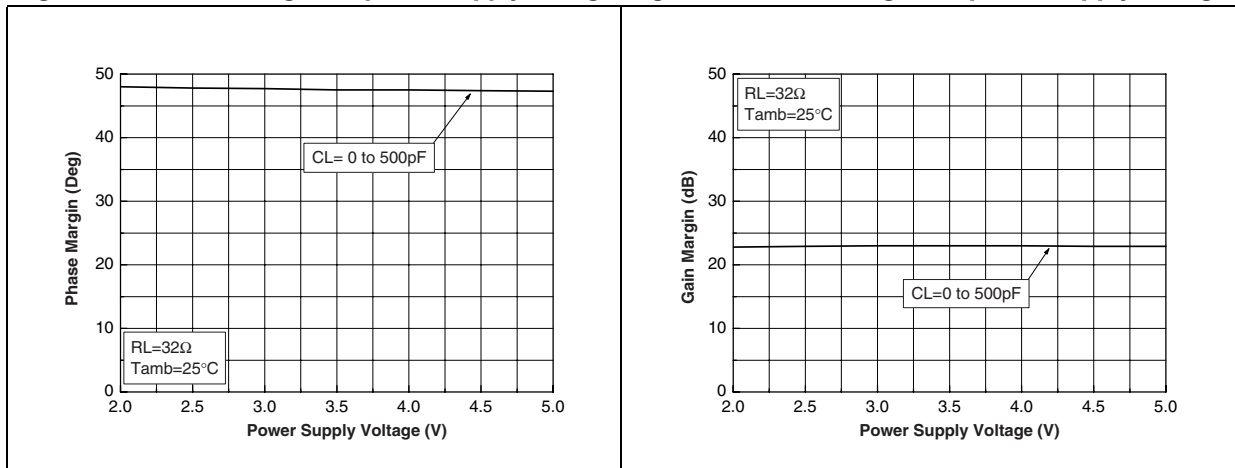


Figure 17. Phase margin vs. power supply voltage Figure 18. Gain margin vs. power supply voltage

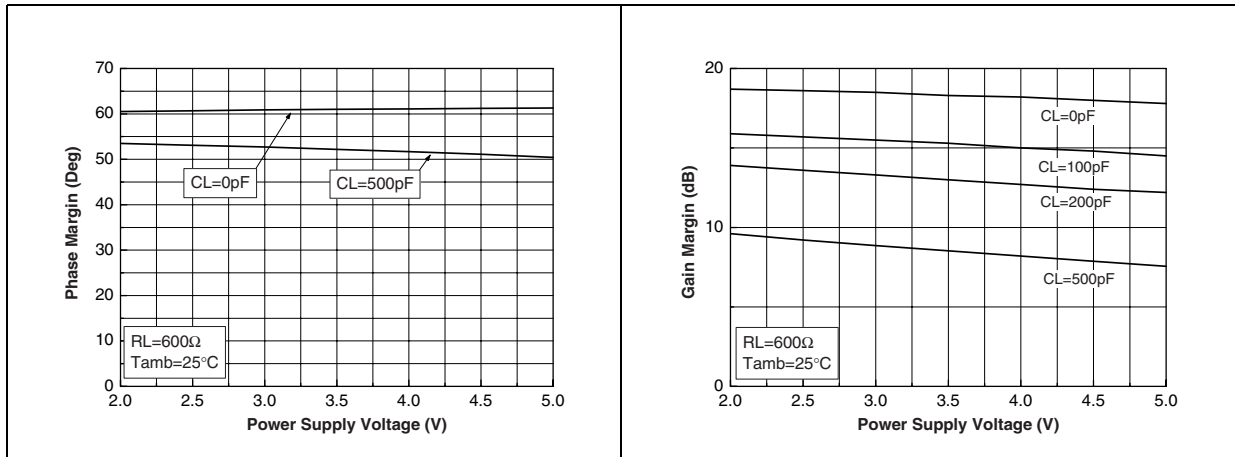


Figure 19. Phase margin vs. power supply voltage Figure 20. Gain margin vs. power supply voltage

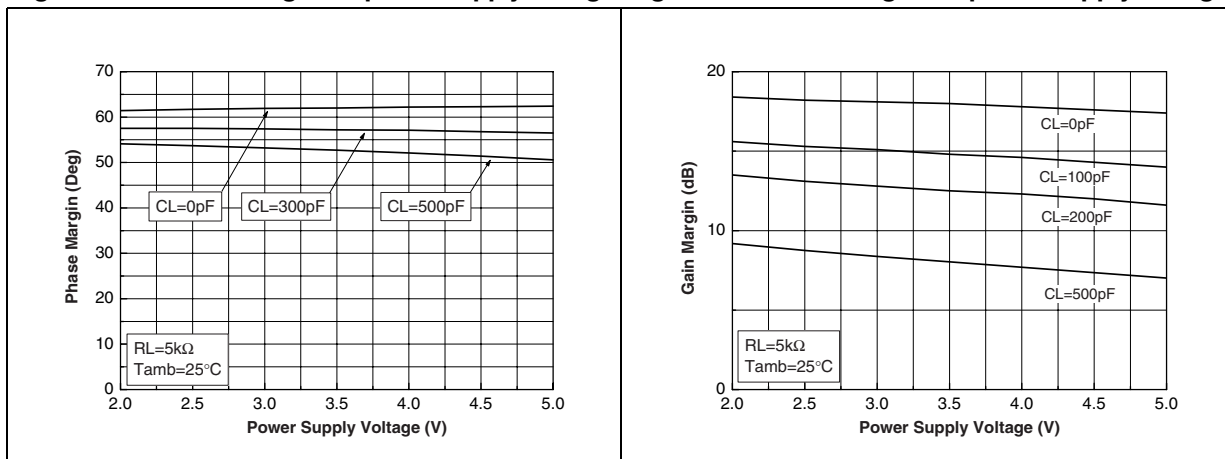


Figure 21. Output power vs. power supply voltage Figure 22. Output power vs. power supply voltage

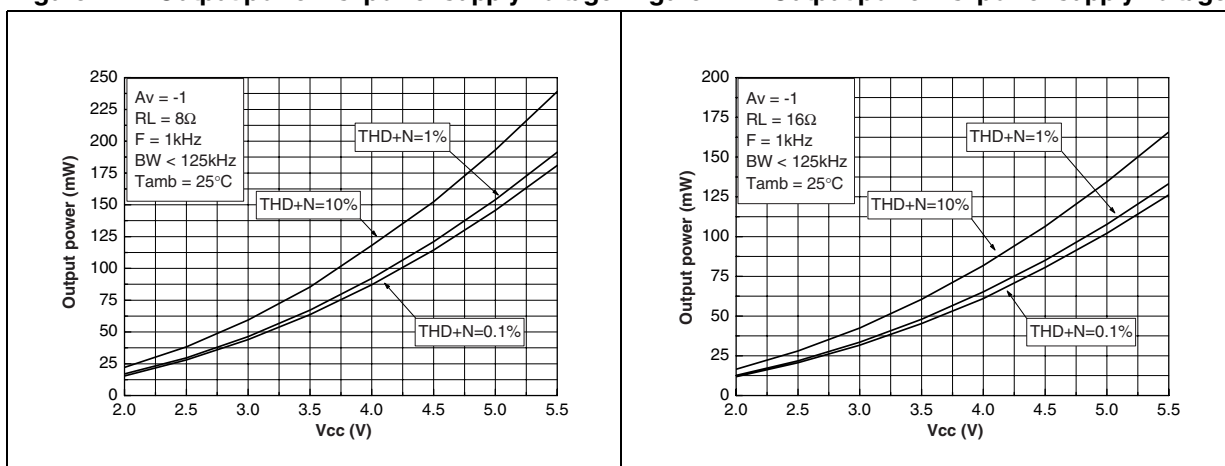


Figure 23. Output power vs. power supply voltage Figure 24. Output power vs. load resistance

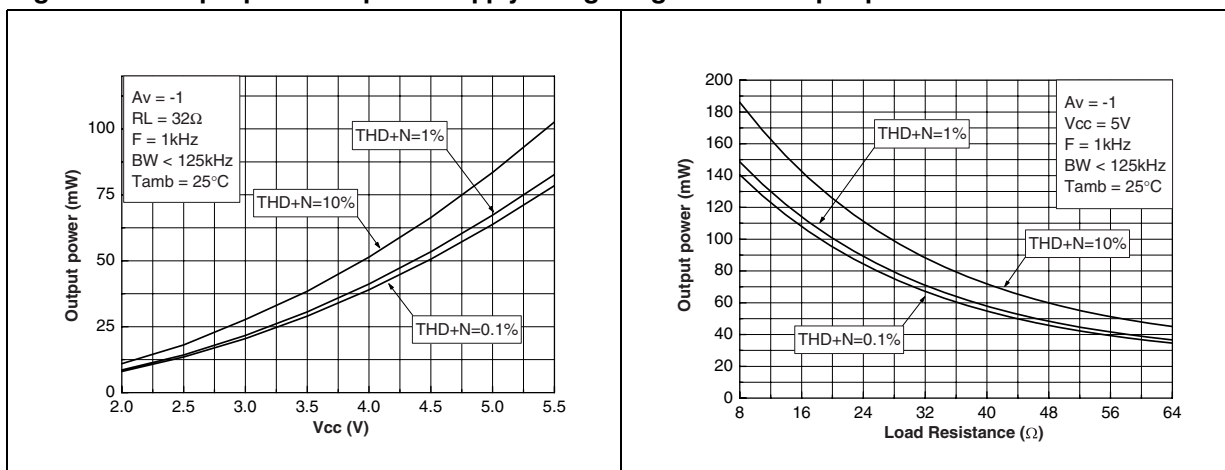


Figure 25. Output power vs. load resistance

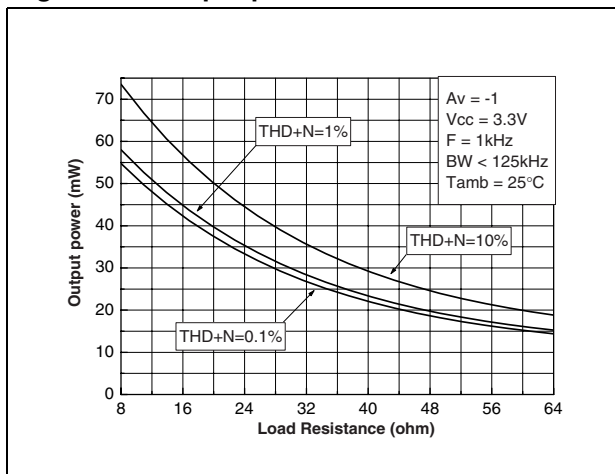


Figure 26. Output power vs. load resistance

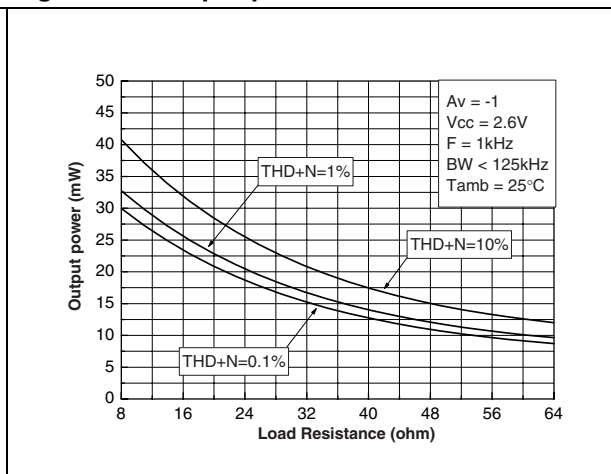


Figure 27. Output power vs. load resistance

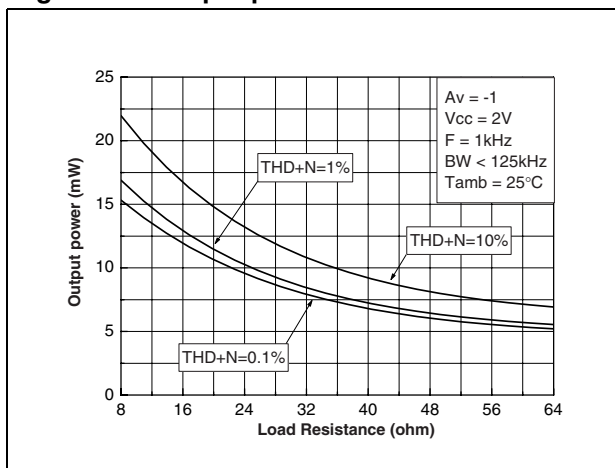


Figure 28. Power dissipation vs. output power

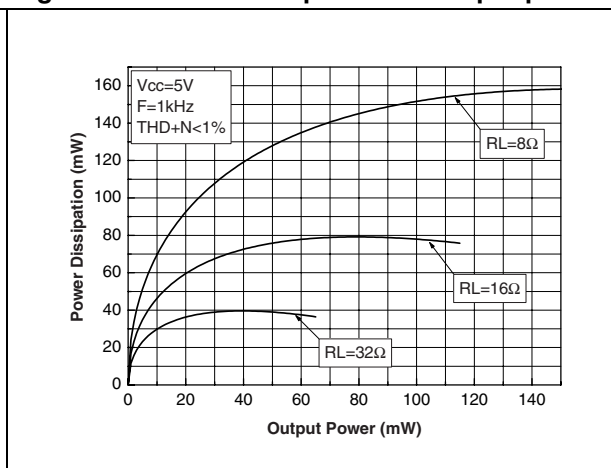


Figure 29. Power dissipation vs. output power

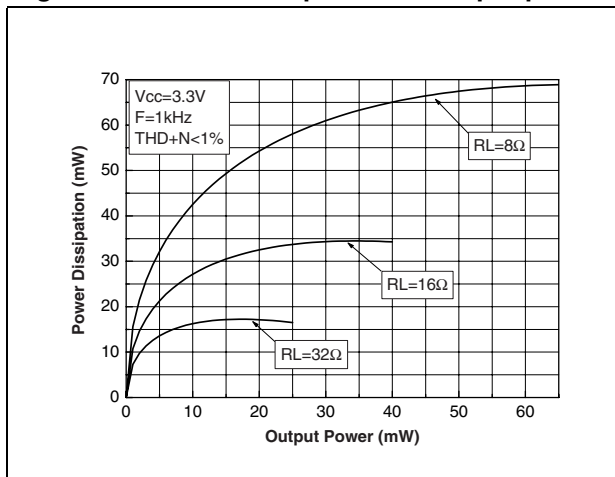


Figure 30. Power dissipation vs. output power

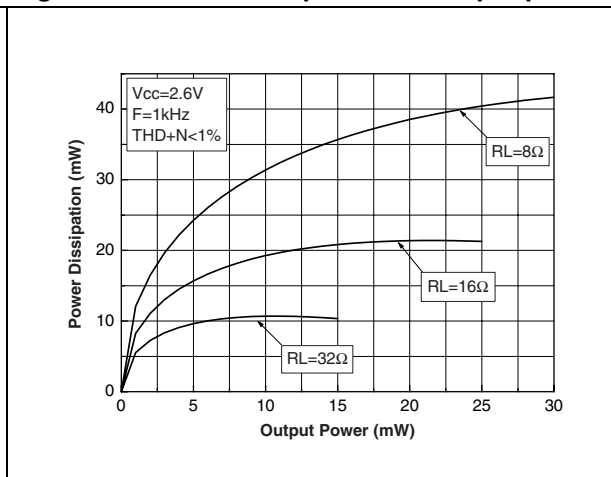


Figure 31. Power dissipation vs. output power Figure 32. Power derating vs. ambient temperature

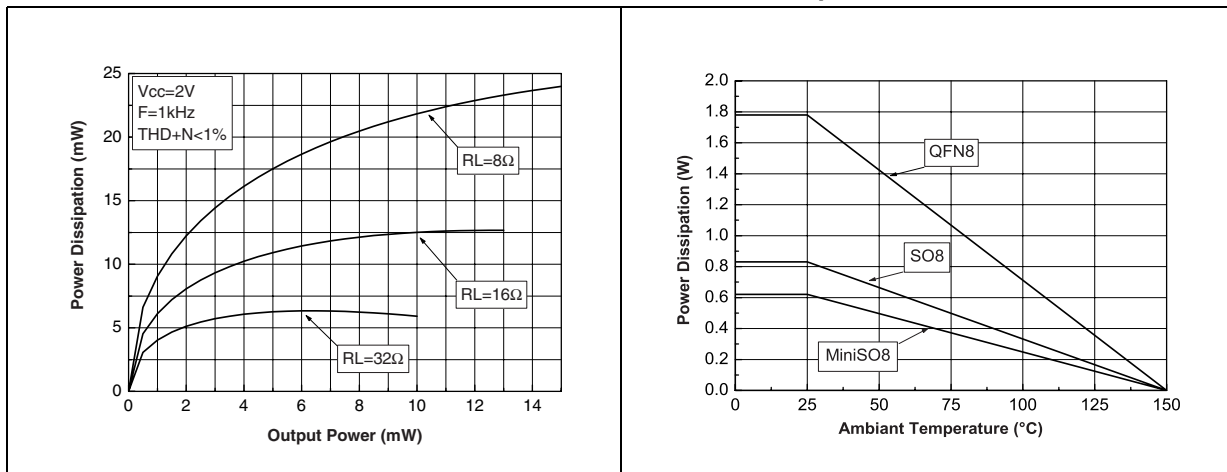


Figure 33. Current consumption vs. power supply voltage Figure 34. Power supply rejection ratio vs. frequency

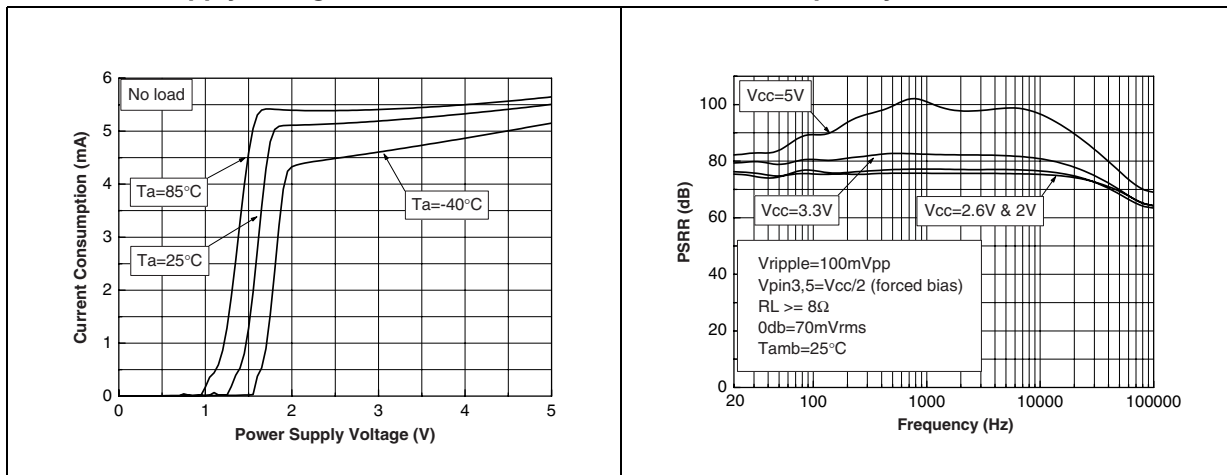


Figure 35. THD + N vs. output power Figure 36. THD + N vs. output power

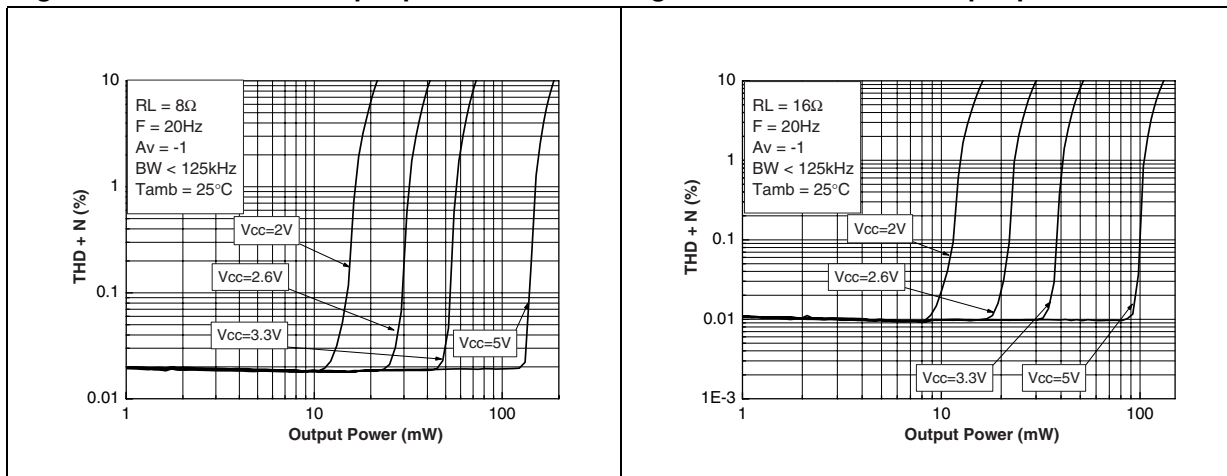


Figure 37. THD + N vs. output power

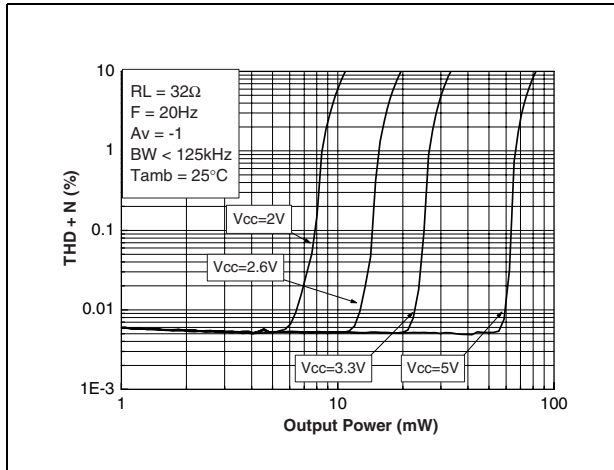


Figure 38. THD + N vs. output power

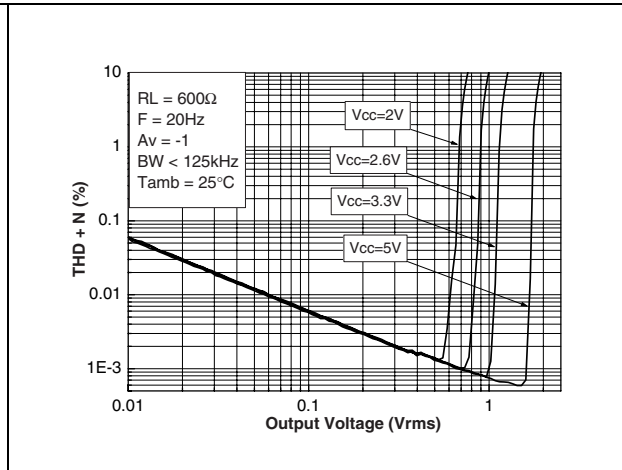


Figure 39. THD + N vs. output power

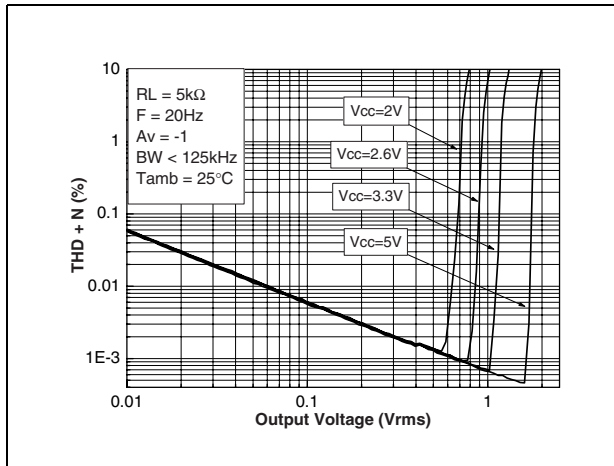


Figure 40. THD + N vs. output power

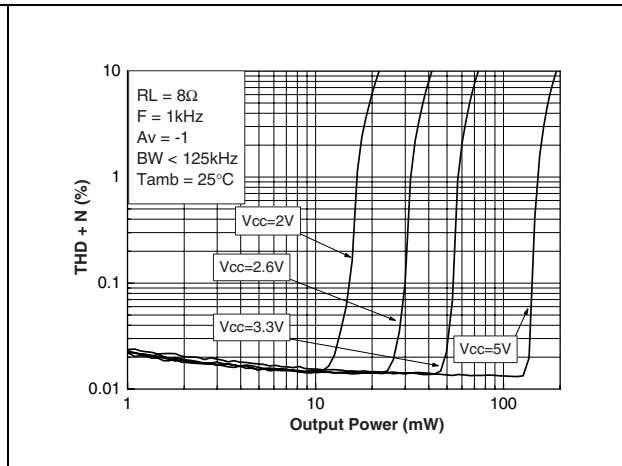


Figure 41. THD + N vs. output power

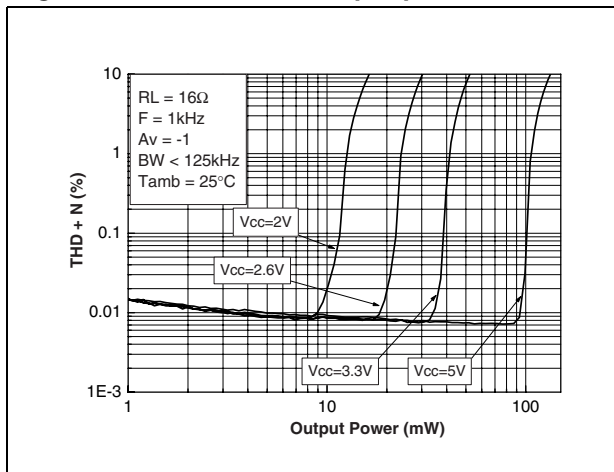


Figure 42. THD + N vs. output power

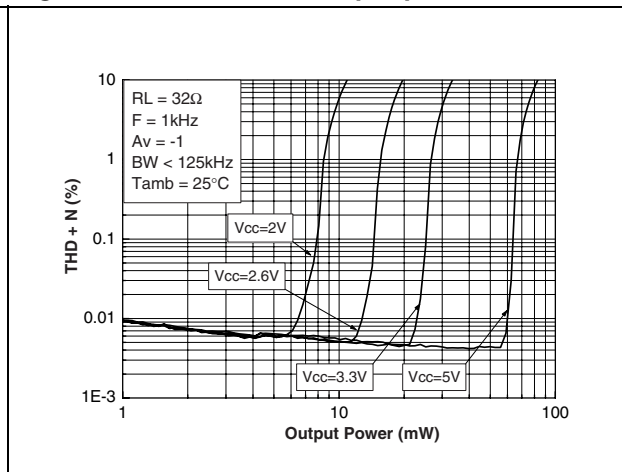


Figure 43. THD + N vs. output power

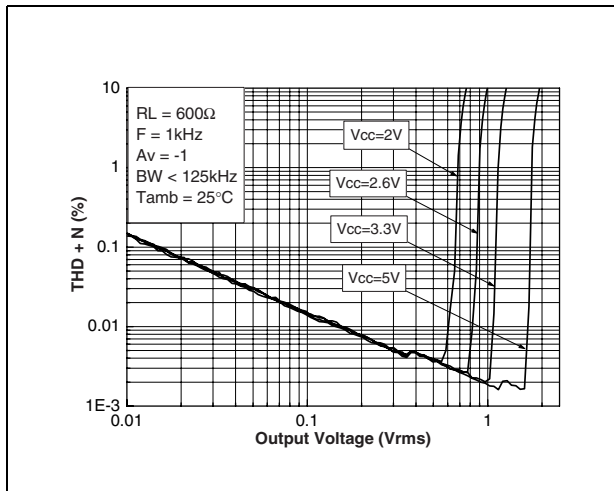


Figure 44. THD + N vs. output power

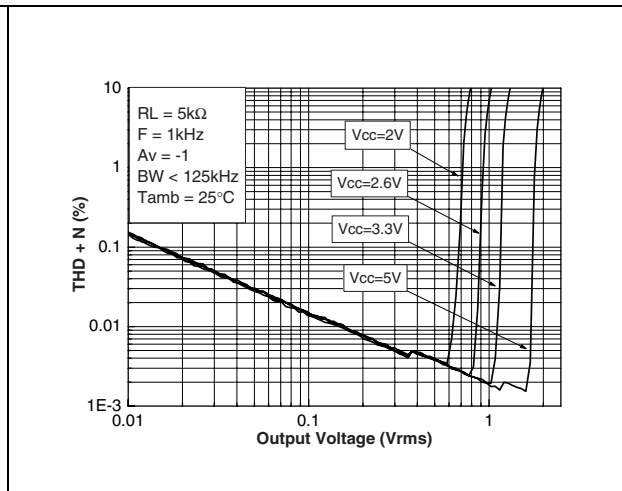


Figure 45. THD + N vs. output power

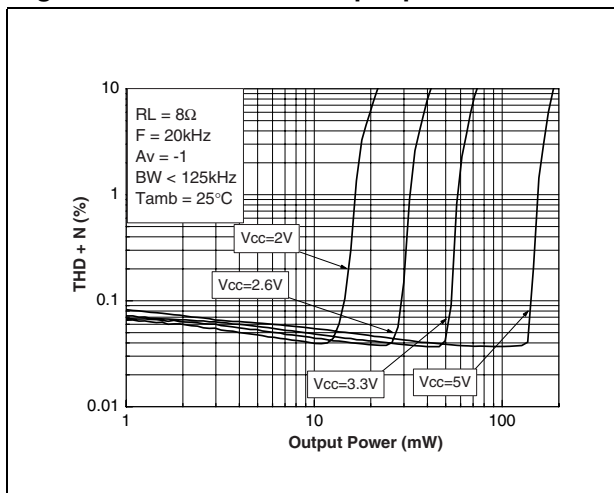


Figure 46. THD + N vs. output power

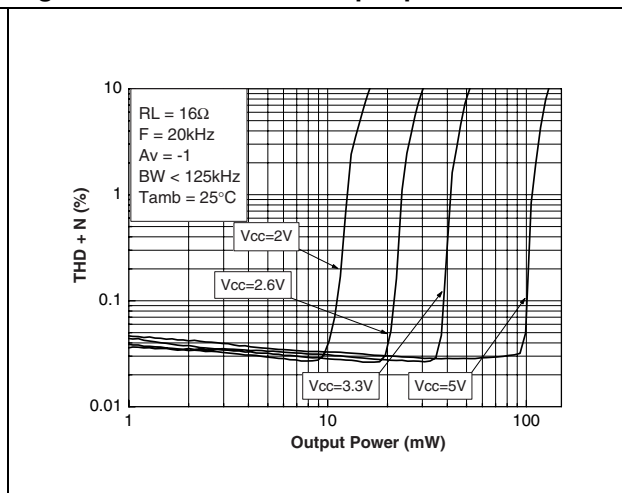


Figure 47. THD + N vs. output power

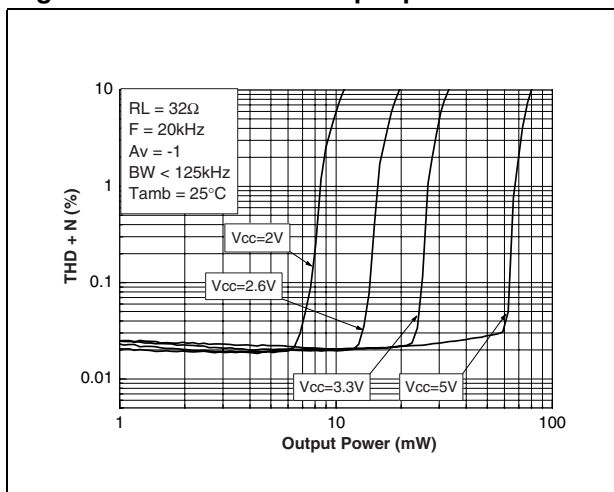


Figure 48. THD + N vs. output power

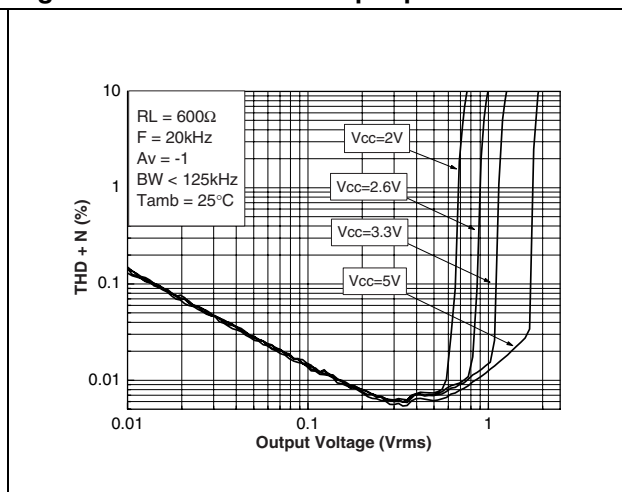


Figure 49. THD + N vs. output power

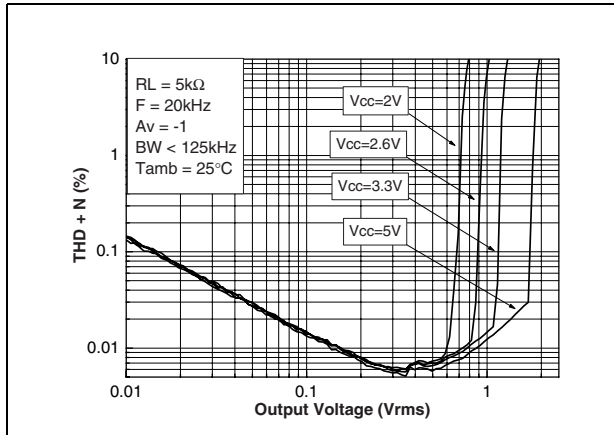


Figure 50. THD + N vs. frequency

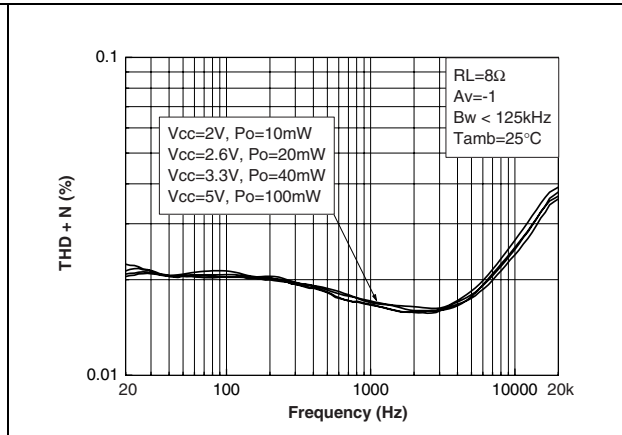


Figure 51. THD + N vs. frequency

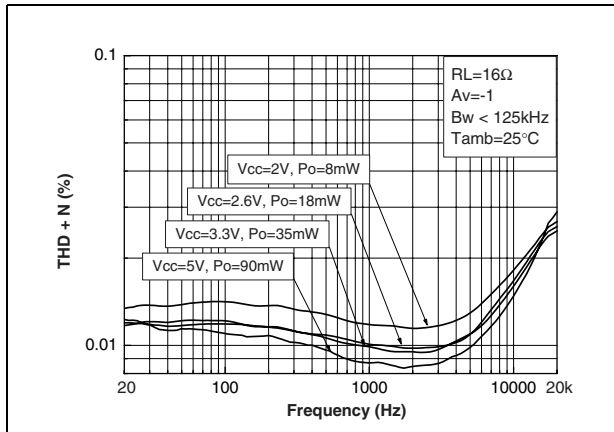


Figure 52. THD + N vs. frequency

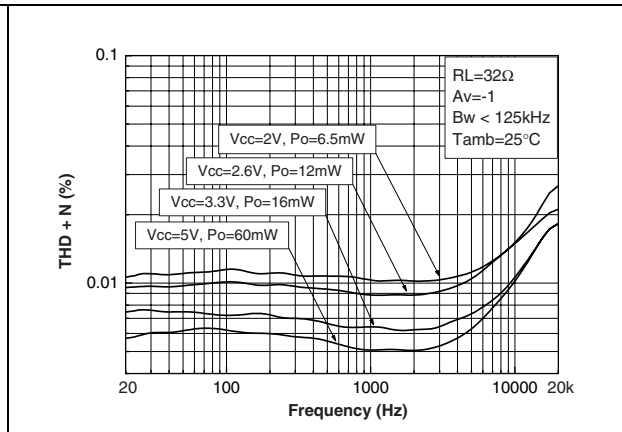


Figure 53. THD + N vs. frequency

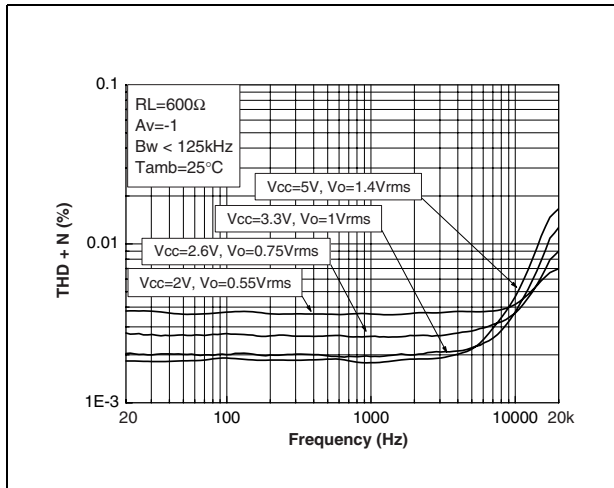


Figure 54. THD + N vs. frequency

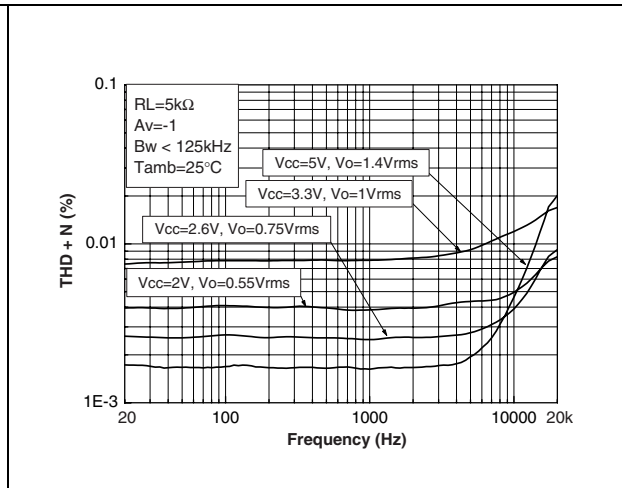


Figure 55. Signal to noise ratio vs. power supply with unweighted filter (20Hz to 20kHz)

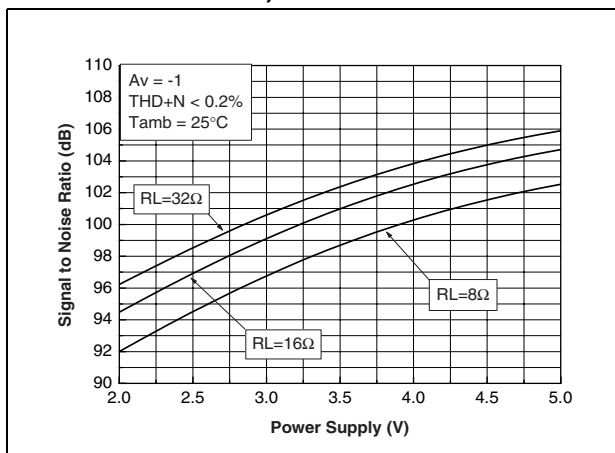


Figure 56. Signal to noise ratio vs. power supply with unweighted filter (20Hz to 20kHz)

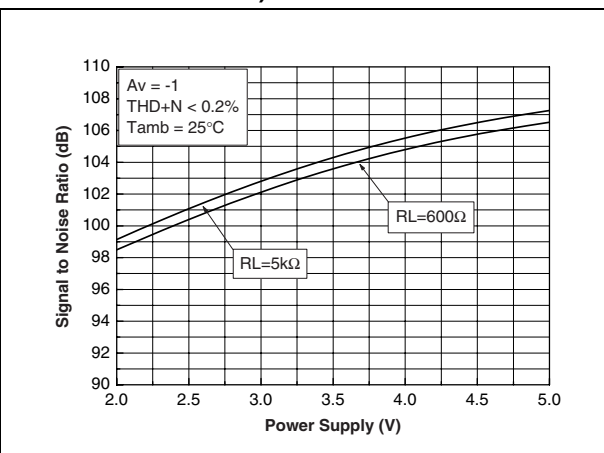


Figure 57. Signal to noise ratio vs. power supply with A weighted filter

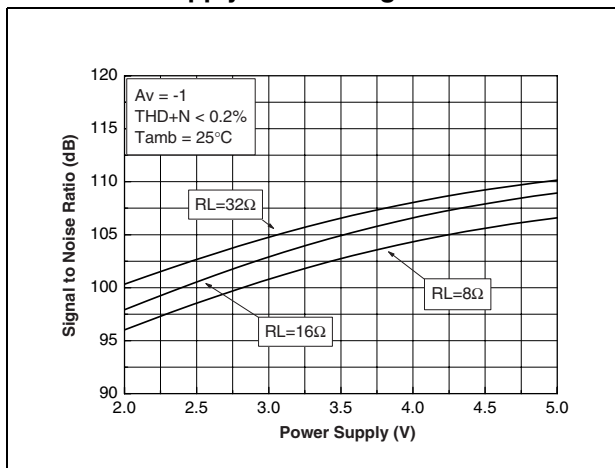


Figure 58. Signal to noise ratio vs. power supply with A weighted filter

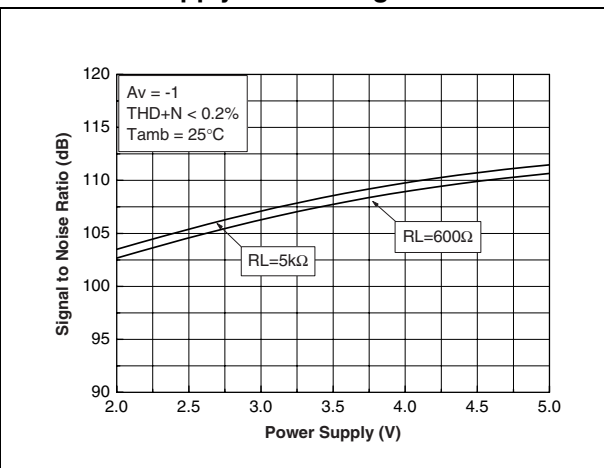


Figure 59. Equivalent input noise voltage vs. frequency

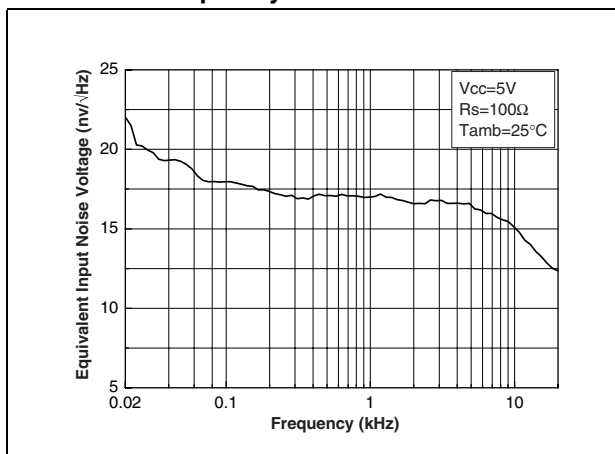


Figure 60. Output voltage swing vs. power supply

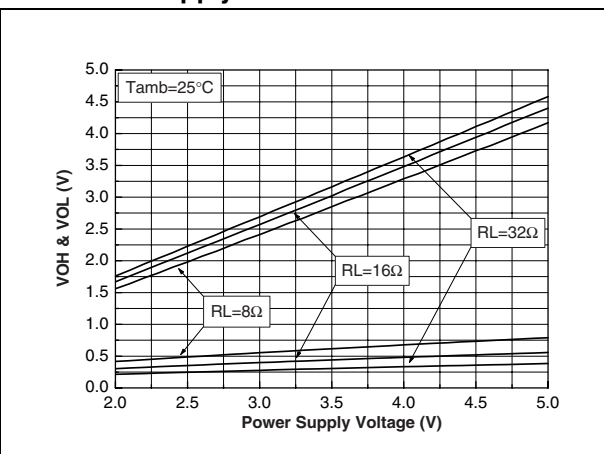


Figure 61. Crosstalk vs. frequency

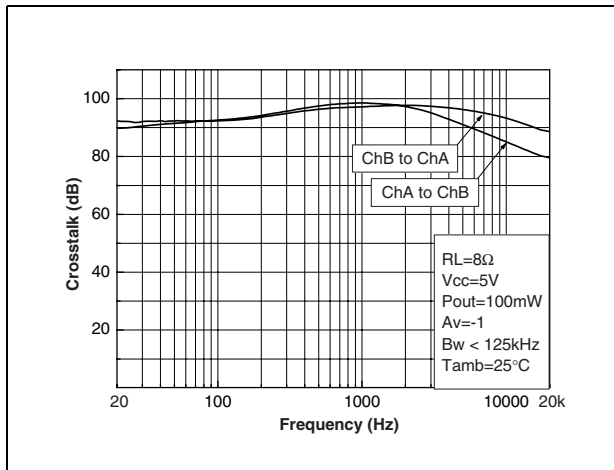


Figure 62. Crosstalk vs. frequency

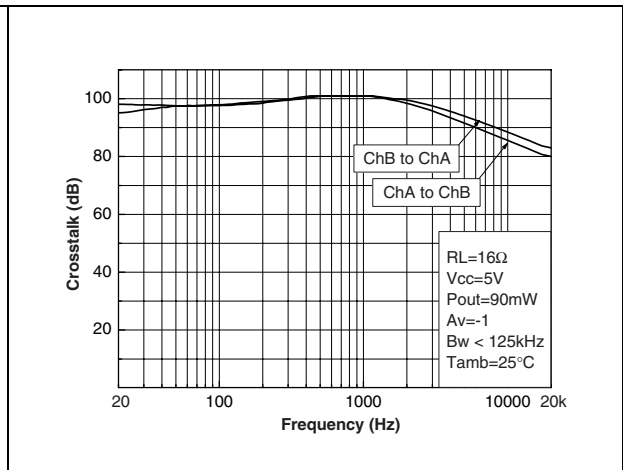


Figure 63. Crosstalk vs. frequency

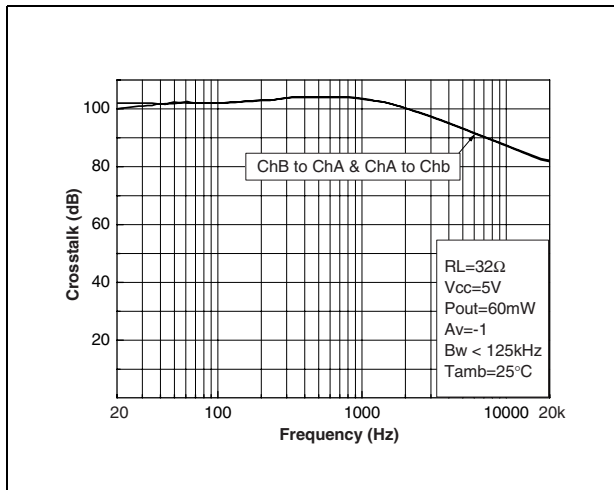


Figure 64. Crosstalk vs. frequency

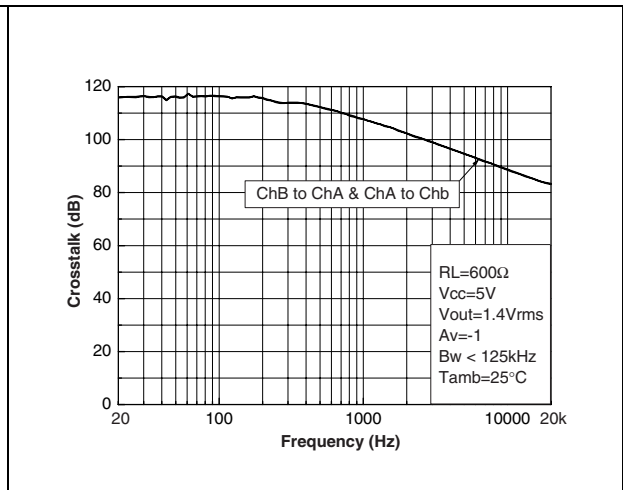


Figure 65. Crosstalk vs. frequency

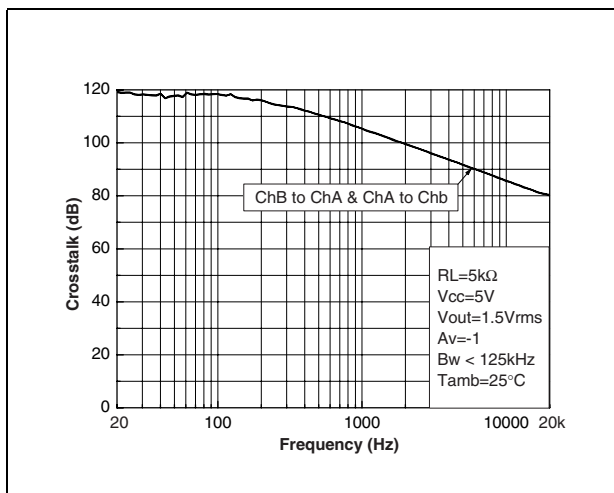


Figure 66. Lower cut off frequency vs. output capacitor

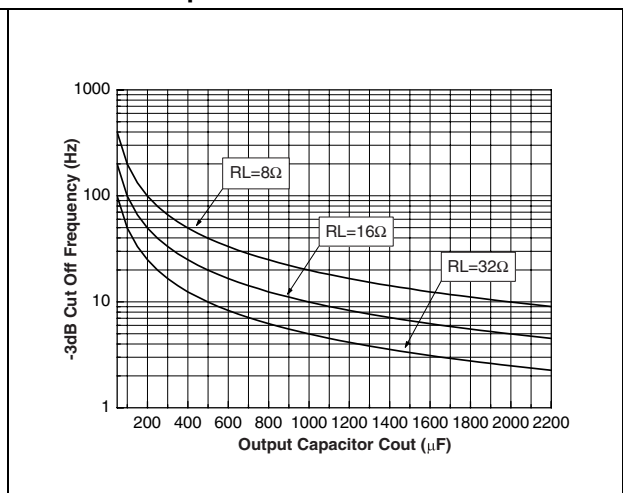


Figure 67. Lower cut off frequency vs. input capacitor

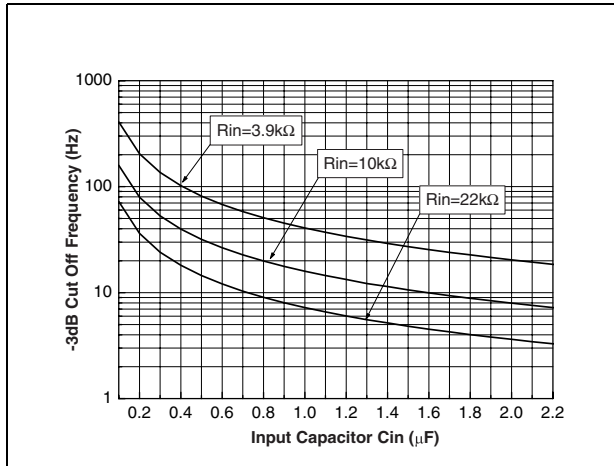


Figure 68. Typical distribution of THD + N

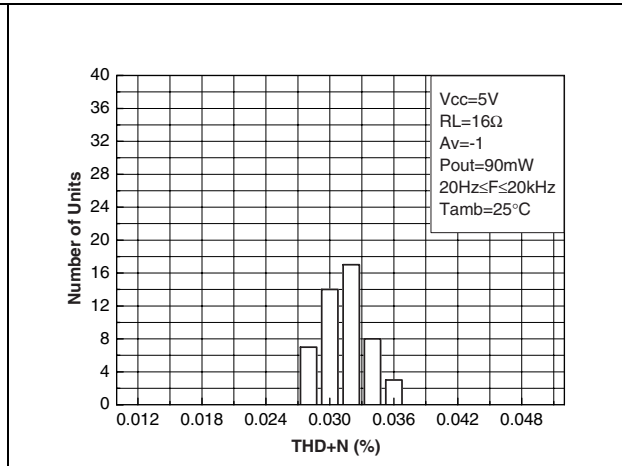


Figure 69. Best case distribution of THD + N

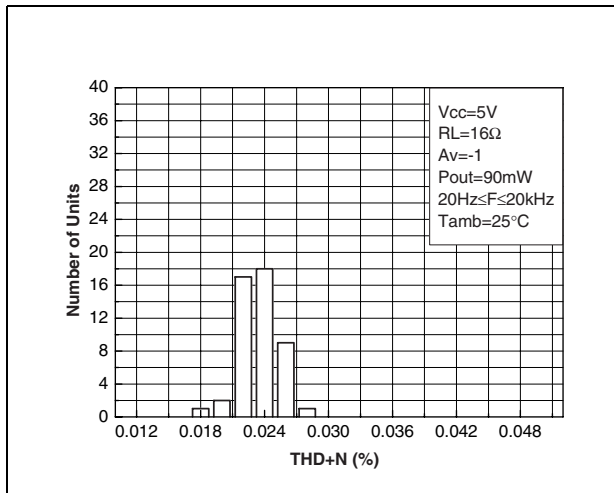


Figure 70. Worst case distribution of THD + N

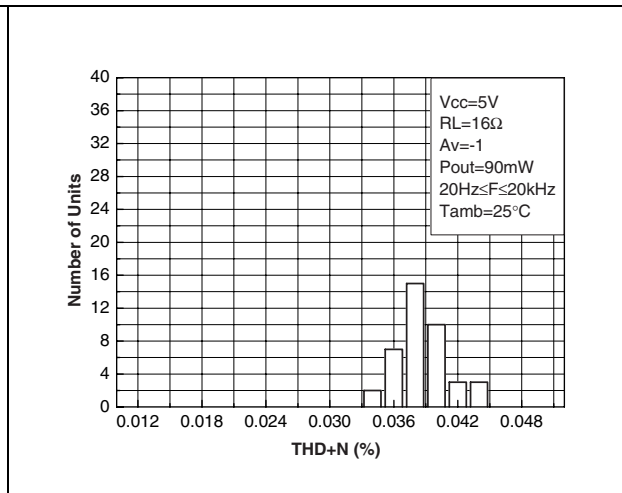


Figure 71. Typical distribution of THD + N

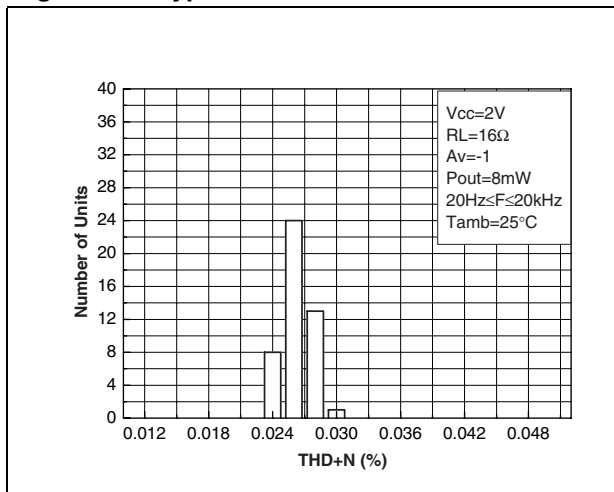


Figure 72. Best case distribution of THD + N

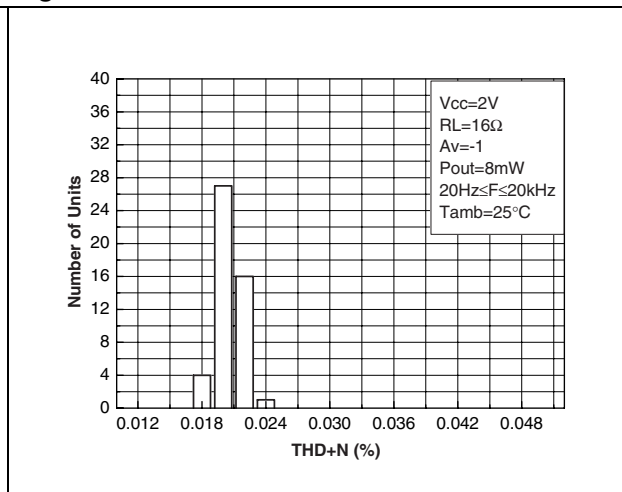


Figure 73. Worst case distribution of THD + N

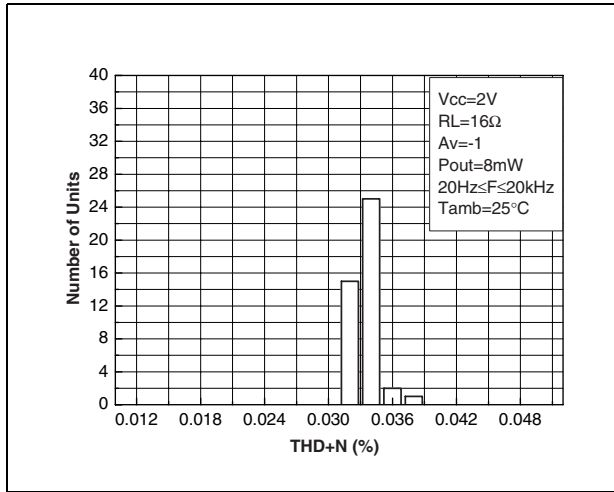


Figure 74. Typical distribution of TDH + N

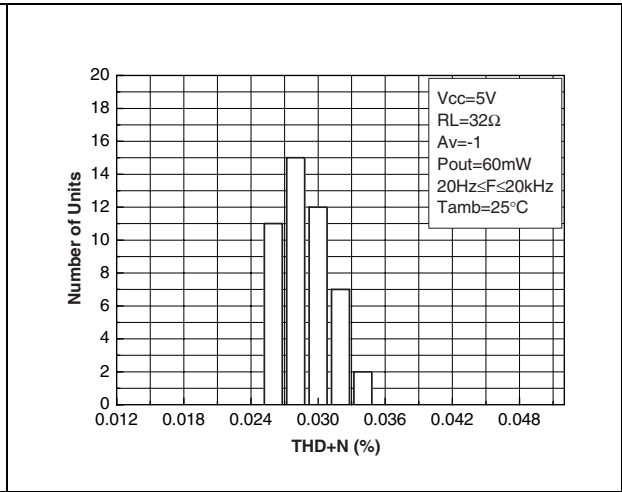


Figure 75. Best case distribution of THD + N

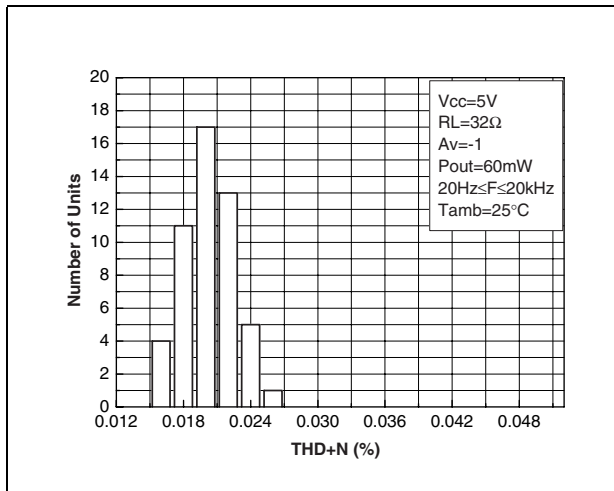


Figure 76. Worst case distribution of THD + N

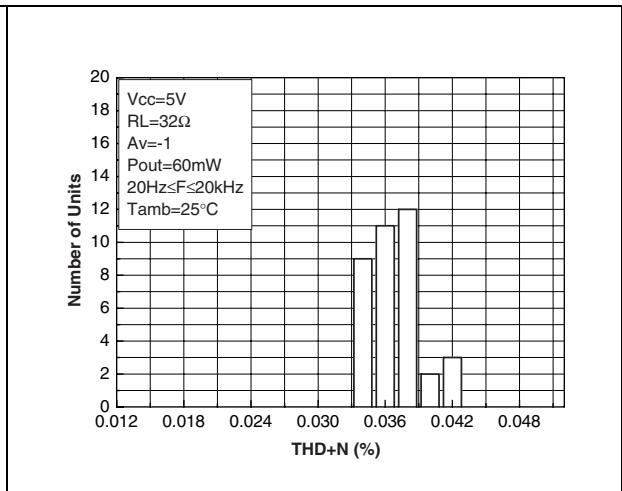


Figure 77. Typical distribution of TDH + N

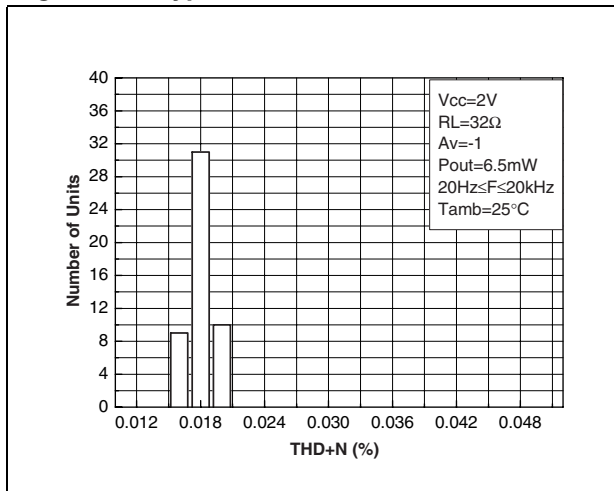


Figure 78. Best case distribution of THD + N

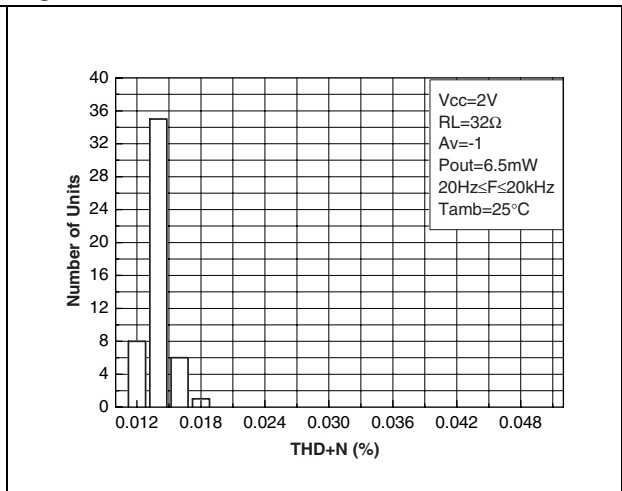
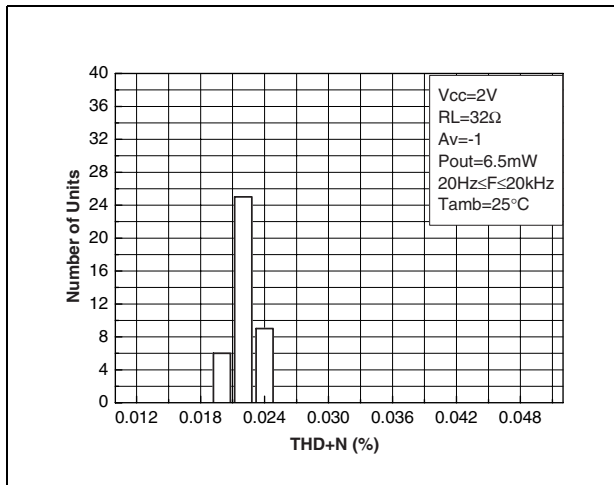


Figure 79. Worst case distribution of THD + N

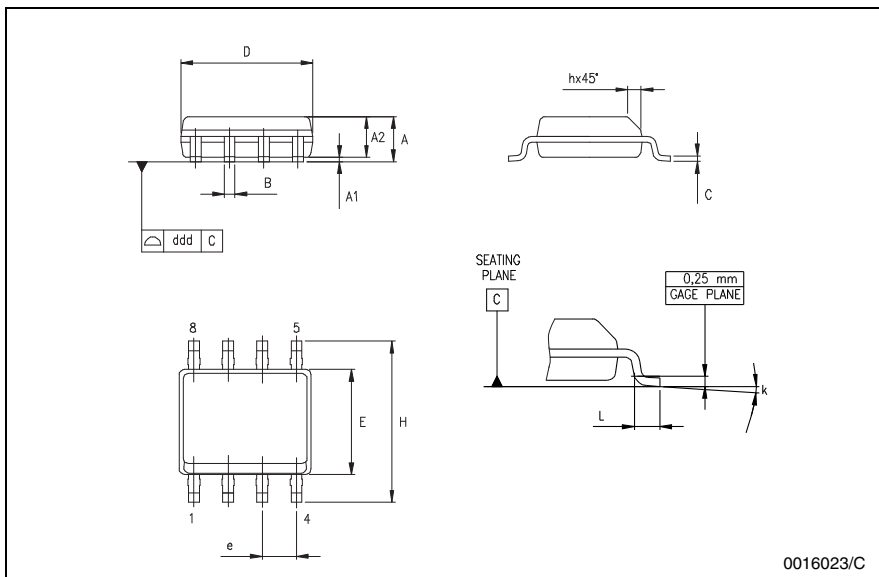


3 Package Mechanical Data

In order to meet environmental requirements, ST offers these devices in ECOPACK® packages. These packages have a Lead-free second level interconnect. The category of second level interconnect is marked on the package and on the inner box label, in compliance with JEDEC Standard JESD97. The maximum ratings related to soldering conditions are also marked on the inner box label. ECOPACK is an ST trademark. ECOPACK specifications are available at: www.st.com.

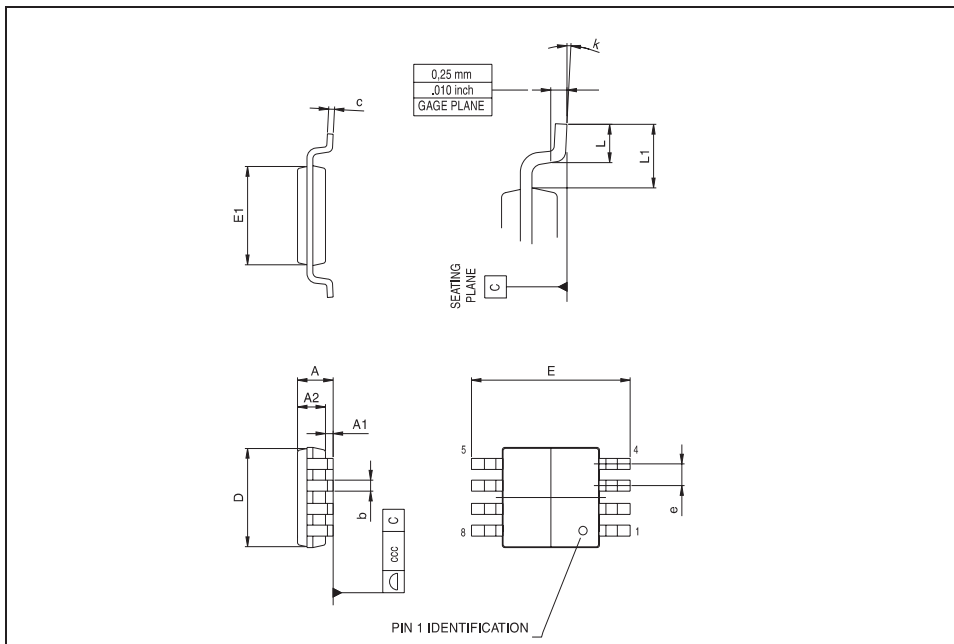
3.1 SO-8 Package

SO-8 MECHANICAL DATA						
DIM.	mm.			inch		
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
A	1.35		1.75	0.053		0.069
A1	0.10		0.25	0.04		0.010
A2	1.10		1.65	0.043		0.065
B	0.33		0.51	0.013		0.020
C	0.19		0.25	0.007		0.010
D	4.80		5.00	0.189		0.197
E	3.80		4.00	0.150		0.157
e		1.27			0.050	
H	5.80		6.20	0.228		0.244
h	0.25		0.50	0.010		0.020
L	0.40		1.27	0.016		0.050
k	8° (max.)					
ddd			0.1			0.04



3.2 MiniSO-8 Package

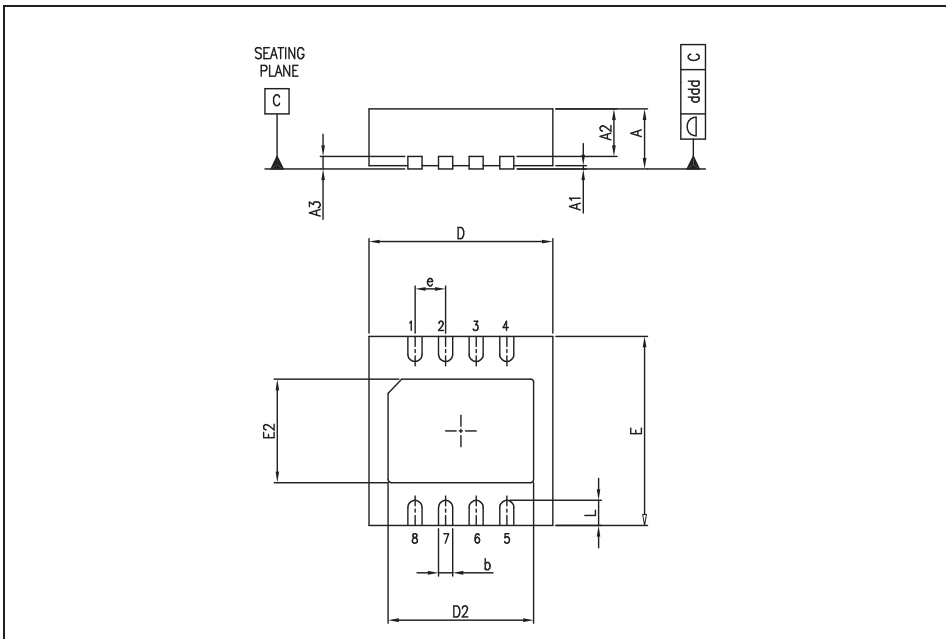
miniSO-8 MECHANICAL DATA						
DIM.	mm.			inch		
	MIN.	TYP	MAX.	MIN.	TYP.	MAX.
A			1.1			0.043
A1	0.05	0.10	0.15	0.002	0.004	0.006
A2	0.78	0.86	0.94	0.031	0.031	0.037
b	0.25	0.33	0.40	0.010	0.13	0.013
c	0.13	0.18	0.23	0.005	0.007	0.009
D	2.90	3.00	3.10	0.114	0.118	0.122
E	4.75	4.90	5.05	0.187	0.193	0.199
E1	2.90	3.00	3.10	.0114	0.118	0.122
e		0.65			0.026	
K	0°		6°	0°		6°
L	0.40	0.55	0.70	0.016	0.022	0.028
L1			0.10			0.004



3.3 DFN8 Package

DFN8 (3x3) MECHANICAL DATA

DIM.	mm.			inch		
	MIN.	TYP	MAX.	MIN.	TYP.	MAX.
A	0.80	0.90	1.00	31.5	35.4	39.4
A1		0.02	0.05		0.8	2.0
A2		0.70			27.6	
A3		0.20			7.9	
b	0.18	0.23	0.30	7.1	9.1	11.8
D	2.875	3.00	3.125		118.1	
D2	2.23	2.38	2.48	87.8	93.7	97.7
E	2.875	3.00	3.125		118.1	
E2	1.49	1.64	1.74	58.7	64.6	68.5
e		0.50			19.7	
L	0.30	0.40	0.50	11.8	15.7	19.7



4 Revision history

Date	Revision	Changes
June 2003	1	Initial release.
Nov. 2005	2	The following changes were made in this revision: <ul style="list-style-type: none">– Lead temperature for lead-free added see Table 1: Key parameters and their absolute maximum ratings on page 2.– Formatting changes throughout.

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