

Reversible Motor Drivers for Brush Motors

0.5A or Less Reversible Motor Driver (Single Motor)

BA6950FS

0.8A Reversible Motor Driver (Single Motor)

No.11008ECT01

BA6951FS

Description These drivers are reversible motor drivers that can directly drive brush motor which require forward and reverse rotations. Four modes of output setting are available by the use of input logic (2 inputs); forward, reverse, stop (idling), and braking. In addition, since voltage applied to motors varies in accord with the control terminal, motor rotating speed can be optionally set and by the built-in current feedback amplifier, the motor can be driven at a constant speed.

Features

- 1) Four-mode outputs of forward, reverse, stop (idling), and braking are enabled in compliance with two inputs
- 2) Motors can be driven at a constant speed by a current feedback amplifier
- 3) Built-in thermal shutdown circuit
 - 4) Built-in current limiting function (BA6951FS)

Applications

Audio-visual equipment; PC peripherals; Car audios; Car navigation systems; OA equipments

●Absolute maximum ratings (Ta=25°C, All voltages are with respect to ground)

Parameter	Symbol	Ratings		Unit
Falameter	Symbol	BA6950FS	BA6951FS	Offic
Supply voltage	VCC	٤	8	
Supply voltage	VB	18		V
Output current	I _{OMAX}	0.4*1 0.8*1		А
Operating temperature	T _{OPR}	-20 ~ 75		°C
Storage temperature	T _{STG}	-55 ~ 150		°C
Power dissipation	Pd	0.813* ²		W
Junction temperature	T _{jmax}	150		°C

*1 Do not, exceed Pd or ASO.

*2 SSOP-A16 package. Mounted on a 70mm x 70mm x 1.6mm FR4 glass-epoxy board with less than 3% copper foil. Derated at 6.4mW/°C above 25°C.

Operating conditions (Ta=25°C)

Parameter	Symbol	Ratings	Unit
Supply voltage	VCC	3 ~ 6	V
Supply voltage	VB	3 ~ 16	V
VTCL voltage	V _{CTL}	0 ~ (VCC-1.8)	V

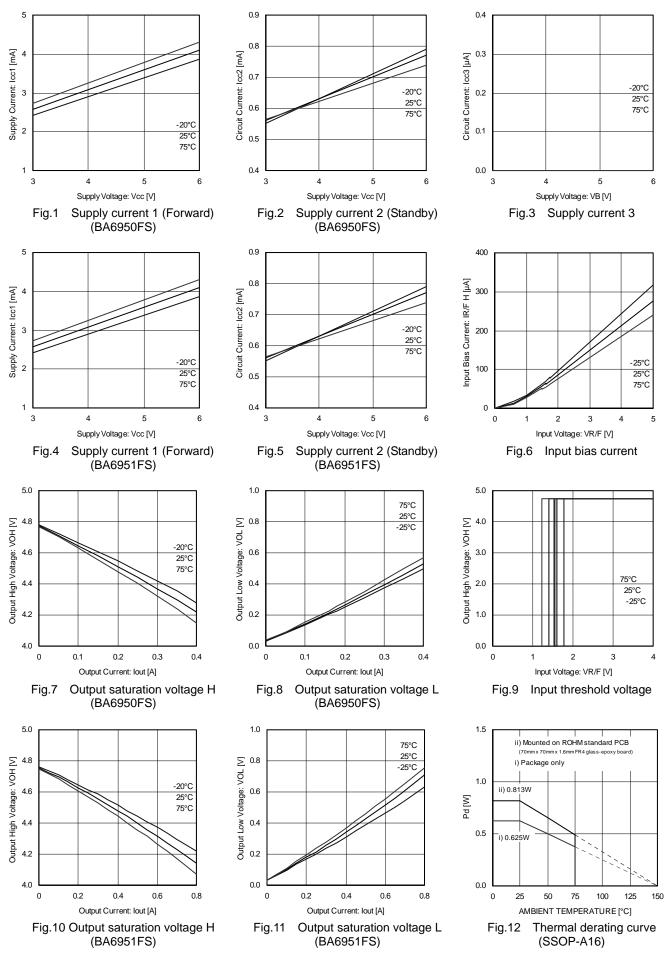
●Electrical characteristics (BA6950FS, unless otherwise specified, Ta=25°C and VCC=4.8V, VB=4.8V)

Parameter	Symbol		Limits		Unit	Conditions
Parameter	Symbol	Min.	Тур.	Max.	Unit	Conditions
Supply current 1	I _{CC1}	-	4.0	6.0	mA	FWD/REV mode, VCTL=0V
Supply current 2	I _{CC2}	-	0.7	1.5	mA	Standby mode, VCTL=0V
Supply current 3	IBOFF	-	0	1	μA	VCC=0V
Input threshold voltage H	V _{R/F H}	2.0	-	VCC	V	
Input threshold voltage L	V _{R/F L}	0	-	0.8	V	
Input bias current	I _{R/F H}	-	80	135	μA	FIN=2V, RIN=2V
CTL amplifier offset voltage	V _{CTLOFS}	-5	0	5	mV	VCTL-RC, VCTL=0V, 1V
CTL amplifier gain	V _{CTLGA}	40	46	52	µA/V	ΔI_{RT1} , VCTL=2V, 1V
CTL output mirror ratio 1	I _{CTLR1}	0.85	1.00	1.15	ratio	I _{RT1} /I _{RC} , I _{RC} =20µA
CTL output mirror ratio 2	I _{CTLR2}	0.90	1.00	1.10	ratio	I _{RT1} /I _{RC} , I _{RC} =200μA
CS amplifier offset voltage	CSOFS	-5	0	5	mV	CS1-CS2, CS1=0V, 0.1V
CS output mirror ratio 1	I _{CSR1}	0.85	1.00	1.15	ratio	I _{RT2} /I _{CS2} , I _{CS} =20µA
CS output mirror ratio 2	I _{CSR2}	0.90	1.00	1.10	ratio	Ι _{RT2} /I _{CS2} , I _{CS} =200μΑ
Output high voltage	V _H	2.0	4.6	-	V	M1, M2, VCTL=0.2V
Output saturation voltage H	V _{OH}	-	0.09	0.3	V	I ₀ =50mA, RT1=VCC
Output saturation voltage L	V _{OL}	-	0.07	0.2	V	I ₀ =50mA, RT1=VCC

●Electrical characteristics (BA6951FS, unless otherwise specified, Ta=25°C and VCC=4.8V, VB=4.8V)

			Limits			Conditions
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Supply current 3	IBOFF	-	0	1	μA	VCC=0V
Input threshold voltage H	V _{R/F H}	2.0	-	VCC	V	
Input threshold voltage L	V _{R/F L}	0	-	0.8	V	
Input bias current	I _{R/F H}	-	80	135	μA	FIN=2V, RIN=2V
CTL amplifier offset voltage	V _{CTLOFS}	-5	0	5	mV	VCTL-RC, VCTL=0V, 1V
CTL amplifier gain	V _{CTLGA}	40	46	52	μA/V	ΔI_{RT1} , VCTL=2V, 1V
CTL output mirror ratio 1	I _{CTLR1}	0.85	1.00	1.15	ratio	I _{RT1} /I _{RC} , I _{RC} =20μΑ
CTL output mirror ratio 2	I _{CTLR2}	0.90	1.00	1.10	ratio	I _{RT1} /I _{RC} , I _{RC} =200μΑ
CS amplifier offset voltage	CS _{OFS}	-5	0	5	mV	ATC-CS, ATC=0V, 0.1V
CS output mirror ratio 1	I _{CSR1}	0.85	1.00	1.15	ratio	I _{RT2} /I _{CS} , I _{CS} =20μΑ
CS output mirror ratio 2	I _{CSR2}	0.90	1.00	1.10	ratio	Ι _{RT2} /I _{CS} , I _{CS} =200μΑ
TL-R _{AOFS} offset voltage	TL-R _{AOFS}	6	18	30	mV	TL=0.3V, R _{ATC} =1.0Ω
Output high voltage	V _H	1.85	2.20	2.55	V	M1, M2, VCTL=1.0V
Output saturation voltage H	V _{OH}	-	0.28	0.56	V	I _O =300mA, RT1=VCC
Output saturation voltage L	V _{OL}	-	0.32	0.64	V	I _O =300mA, RT1=VCC

Electrical characteristic curves (Reference data)



Block diagram and pin configuration

BA6950FS

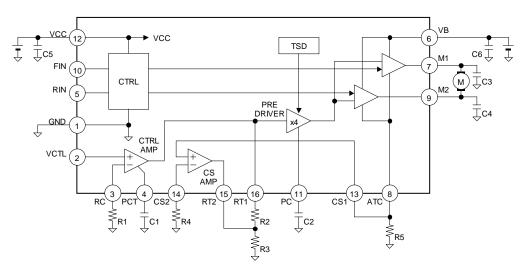


Fig.13 BA6950FS

TADIE I DA0950F5				
Pin	Name	Function		
1	GND	GND		
2	VCTL	Control input		
3	RC	Control gain setting		
4	PCT	CTL amp phase compensation		
5	RIN	Control input (reverse)		
6	VB	Power supply (driver stage)		
7	M1	Driver output		
8	ATC	Current sense pin		
9	M2	Driver output		
10	FIN	Control input (forward)		
11	PC	Phase compensation		
12	VCC	Power supply (small signal)		
13	CS1	CS amp gain setting		
14	CS2	CS amp gain setting		
15	RT2	CTL amp gain setting		
16	RT1	CTL amp gain setting		

Table 1 BA6950FS

		RT1
	Fin	RT2
RC	Fin	CS2
PCT	Fin	CS1
	Fin	VCC
VB	Fin	PC
M1	F	FIN
ATC	F	M2

Fig.14 BA6950FS (SSOP-A16)

Block diagram and pin configuration

BA6951FS

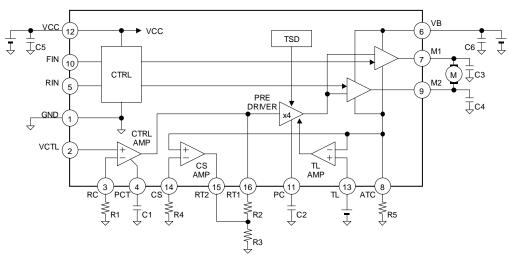


Fig.15 BA6951FS

Pin	Name	Function
1	GND	GND
2	VCTL	Control input
3	RC	Control gain setting
4	PCT	CTL amp phase compensation
5	RIN	Control input (reverse)
6	VB	Power supply (driver stage)
7	M1	Driver output
8	ATC	Current sense pin
9	M2	Driver output
10	FIN	Control input (forward)
11	PC	Phase compensation
12	VCC	Power supply (small signal)
13	TL	Torque limiter setting
14	CS	CS amp gain setting
15	RT2	CTL amp gain setting
16	RT1	CTL amp gain setting

GND	RT1
VCTL	RT2
RC	CS
PCT	TL
RIN	VCC
VB	PC
M1	FIN
ATC	M2

Fig.16 BA6951FS (SSOP-A16)

•External application components

- Resistor for the current sensing, R5
 This is a current sensing resistor, care must be taken to avoid changes in the ground wire pattern in any external connected component.
- Control amplifier gain setting resistor, R1
 VCTL pin voltage is buffered to RC pin, and the control gain V_{CTLGA} can be set by connecting R1. The current decided here is output to RT1 pin.
- Control amplifier phase compensation capacitor, C1
 This phase compensation capacitor for the control amplifier. Please monitor the RT1 pin voltage and confirm no oscillation. About 33pF is recommended.
- 4) Current feedback amplifier gain setting resistor, R4
 CS1 pin voltage (the motor current detection) is buffered to CS2 pin BA6950FS.
 ATC pin voltage (the motor current detection) is buffered to CS pin BA6951FS.
 The current feedback gain can be set by R4 connecting to CS2 or CS pin. The current decided here is output to RT2 pin.
- Pre-amplifier gain setting resistor, R2, R3 These resistors are to add the control amplifier output and the current feedback amplifier output. This amplifier has about fourfold gain.
- Pre-amplifier phase compensation capacitor, C2
 Please connect the capacitor about 0.1µF as the phase compensation of the pre-amplifier, and monitor the driver output no oscillation.
- Stabilization capacitor for the power supply line, C5, C6 Please connect the capacitor of 1µF to 100µF for the stabilization of the power supply line, and confirm the motor operation.
- 8) Phase compensating capacitor, C3, C4 Noise is generated in output pins or oscillation results in accord with the set mounting state such as power supply circuit, motor characteristics, PCB pattern artwork, etc. As noise oscillation measures, connect 0.01µF to 0.1µF capacitors.
- Torque limiter setting, TL pin, BA6951FS only The motor current is limited so that ATC pin voltage should not exceed TL pin voltage.

Functional descriptions

Table 3 Logic table

FIN	RIN	M1	M2	Operation	
L	L	OPEN*	OPEN*	Stop (idling)	
Н	L	L	Н	Forward (M2 > M1)	
L	Н	Н	L	Reverse (M1 > M2)	
Н	Н	L	L	Brake (stop)	

* OPEN is the off state of all output transistors. Please note that this is the state of the connected diodes, which differs from that of the mechanical relay.

•External application components setting procedure

The relation between VCTL and the output high voltage is as follows.

 $\begin{array}{ll} \cdot I_{RT1} = VCTL \, / \, R1 & \cdots (1) & VCTL: \mbox{ Torque control voltage } \\ \cdot I_{RT2} = I_{ACT} \, x \, R5 \, / \, R4 & \cdots (2) & I_{ACT}: \mbox{ Motor current } \\ \cdot V_{RT1} = R3 \, x \, (\, I_{RT1} + I_{RT2} \,) + R2 \, x \, I_{RT1} & \cdots (3) \\ \cdot V_{MX} = 4 \, x \, V_{RT1} & \cdots (4) & V_{M1}, \, V_{M2}: \mbox{ Output high voltage } \end{array}$

$$V_{MX} = \frac{4(R2 + R3)}{R1} \times VTCL + \frac{4R3R5}{R4} \times I_{ACT} \quad \dots (5)$$

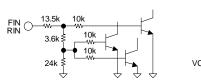
To drive the motor by constant speed as follows.

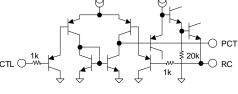
 $R_{L} + R_{ON} + R5 = \frac{4 R3 R5}{R4} \dots (6)$ $R_{L}: Motor coil impedance$ $R_{ON}: On resistance of the driver IC$

R3, R4, and R5 are first set, and then R1 and R2 are set afterwards.

	Та	able 4 External components		
Parts	Default value	Parameter	Recommended condition	
R1	22kΩ	I _{RT1}	I _{RT1} < 1mA	
R2 + R3	1kΩ + 1.5kΩ	V _{RT1}	V _{RT1} x 4 < VB	
R4	560Ω	I _{RT2}	I _{RT2} < 1mA	
R5	5.5Ω	V _{ATC}	V _{ATC} < 1V	
C1	33pF	V _{PCT}		
C2	0.1µF	V _{PC}	Please confirm the motor	
C3, C4	0.1µF	V _{M1} , V _{M2}	operation	
C5, C6	1~100µF	VCC, VB		

Interfaces





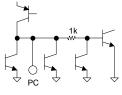


Fig. 17 FIN, RIN

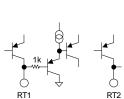


Fig. 20 RT1, RT2

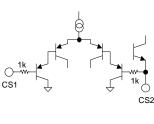


Fig.21 CS1, CS2

(BA6950FS)

Fig.18 VCTL, RC, PCT

Оπ

cs

Fig.22 CS, TL

(BA6951FS)

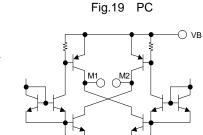


Fig.23 VB, ACT, M1,M2

🔾 АСТ

Notes for use

1) Absolute maximum ratings

Devices may be destroyed when supply voltage or operating temperature exceeds the absolute maximum rating. Because the cause of this damage cannot be identified as, for example, a short circuit or an open circuit, it is important to consider circuit protection measures – such as adding fuses – if any value in excess of absolute maximum ratings is to be implemented.

2) Connecting the power supply connector backward

Connecting the power supply in reverse polarity can damage the IC. Take precautions against reverse polarity when connecting the power supply lines, such as adding an external direction diode.

3) Power supply lines

Return current generated by the motor's Back-EMF requires countermeasures, such as providing a return current path by inserting capacitors across the power supply and GND (10µF, ceramic capacitor is recommended). In this case, it is important to conclusively confirm that none of the negative effects sometimes seen with electrolytic capacitors – including a capacitance drop at low temperatures - occurs. Also, the connected power supply must have sufficient current absorbing capability. Otherwise, the regenerated current will increase voltage on the power supply line, which may in turn cause problems with the product, including peripheral circuits exceeding the absolute maximum rating. To help protect against damage or degradation, physical safety measures should be taken, such as providing a voltage clamping diode across the power supply and GND.

4) Electrical potential at GND

Keep the GND terminal potential to the minimum potential under any operating condition. In addition, check to determine whether there is any terminal that provides voltage below GND, including the voltage during transient phenomena. When both a small signal GND and high current GND are present, single-point grounding (at the set's reference point) is recommended, in order to separate the small signal and high current GND, and to ensure that voltage changes due to the wiring resistance and high current do not affect the voltage at the small signal GND. In the same way, care must be taken to avoid changes in the GND wire pattern in any external connected component.

5) Thermal design

Use a thermal design that allows for a sufficient margin in light of the power dissipation (Pd) under actual operating conditions.

6) Inter-pin shorts and mounting errors

Use caution when positioning the IC for mounting on printed circuit boards. The IC may be damaged if there is any connection error, or if pins are shorted together.

7) Operation in strong electromagnetic fields

Using this product in strong electromagnetic fields may cause IC malfunctions. Use extreme caution with electromagnetic fields.

8) ASO - Area of Safety Operation

When using the IC, set the output transistor so that it does not exceed absolute maximum ratings or ASO.

9) Built-in thermal shutdown (TSD) circuit

The TSD circuit is designed only to shut the IC off to prevent thermal runaway. It is not designed to protect the IC or guarantee its operation in the presence of extreme heat. Do not continue to use the IC after the TSD circuit is activated, and do not operate the IC in an environment where activation of the circuit is assumed.

10) Capacitor between output and GND

In the event a large capacitor is connected between the output and GND, if VCC and VIN are short-circuited with 0V or GND for any reason, the current charged in the capacitor flows into the output and may destroy the IC. Use a capacitor smaller than 0.47µF between output and GND.

11) Testing on application boards

When testing the IC on an application board, connecting a capacitor to a low impedance pin subjects the IC to stress. Therefore, always discharge capacitors after each process or step. Always turn the IC's power supply off before connecting it to or removing it from the test setup during the inspection process. Ground the IC during assembly steps as an antistatic measure. Use similar precaution when transporting or storing the IC.

12) Switching of rotating direction (FWD/REV)

When the rotating direction is changed over by the motor rotating condition, switch the direction after the motor is temporarily brought to the BRAKE condition or OPEN condition. It is recommended to keep the relevant conditions as follows: via BRAKE: Longer than braking time*.

(* the time required for the output voltage to achieve potential below GND when brake is activated.) via OPEN: The time longer than 1 ms is recommended.

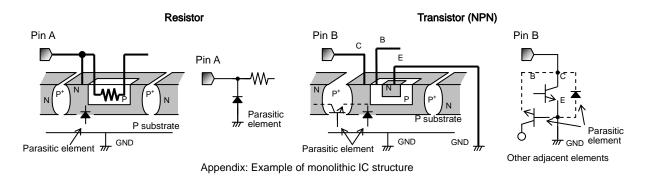
13) Regarding the input pin of the IC

This monolithic IC contains P+ isolation and P substrate layers between adjacent elements, in order to keep them isolated. P-N junctions are formed at the intersection of these P layers with the N layers of other elements, creating a parasitic diode or transistor. For example, the relation between each potential is as follows:

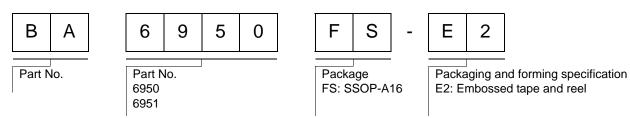
When GND > Pin A and GND > Pin B, the P-N junction operates as a parasitic diode.

When GND > Pin B, the P-N junction operates as a parasitic transistor.

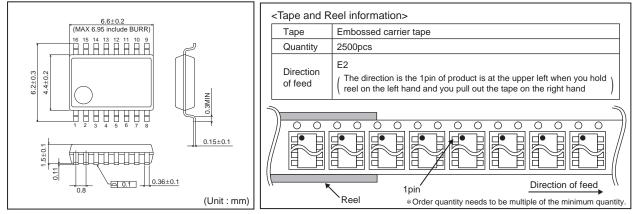
Parasitic diodes inevitably occur in the structure of the IC. The operation of parasitic diodes can result in mutual interference among circuits, as well as operating malfunctions and physical damage. Therefore, do not use methods by which parasitic diodes operate, such as applying a voltage lower than the GND (P substrate) voltage to an input pin.



Ordering part number



SSOP-A16



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 - [f] Sealing or coating our Products with resin or other coating materials
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http://moschip.ru/get-element

Вы можете разместить у нас заказ для любого Вашего проекта, будь то серийное производство или разработка единичного прибора.

В нашем ассортименте представлены ведущие мировые производители активных и пассивных электронных компонентов.

Нашей специализацией является поставка электронной компонентной базы двойного назначения, продукции таких производителей как XILINX, Intel (ex.ALTERA), Vicor, Microchip, Texas Instruments, Analog Devices, Mini-Circuits, Amphenol, Glenair.

Сотрудничество с глобальными дистрибьюторами электронных компонентов, предоставляет возможность заказывать и получать с международных складов практически любой перечень компонентов в оптимальные для Вас сроки.

На всех этапах разработки и производства наши партнеры могут получить квалифицированную поддержку опытных инженеров.

Система менеджмента качества компании отвечает требованиям в соответствии с ГОСТ Р ИСО 9001, ГОСТ РВ 0015-002 и ЭС РД 009

Офис по работе с юридическими лицами:

105318, г.Москва, ул.Щербаковская д.3, офис 1107, 1118, ДЦ «Щербаковский»

Телефон: +7 495 668-12-70 (многоканальный)

Факс: +7 495 668-12-70 (доб.304)

E-mail: info@moschip.ru

Skype отдела продаж: moschip.ru moschip.ru_4

moschip.ru_6 moschip.ru_9