International IOR Rectifier

IR2156(S)PbF BALLAST CONTROL IC

Features

- Ballast control and half bridge driver in one IC
- Programmable preheat frequency
- Programmable preheat time
- Internal ignition ramp
- Programmable over-current threshold
- Programmable run frequency

- Programmable dead time
- DC bus under-voltage reset
- Shutdown pin with hysteresis
- Internal 15.6V zener clamp diode on Vcc
- Micropower startup (150 μA)
- Latch immunity and ESD protection

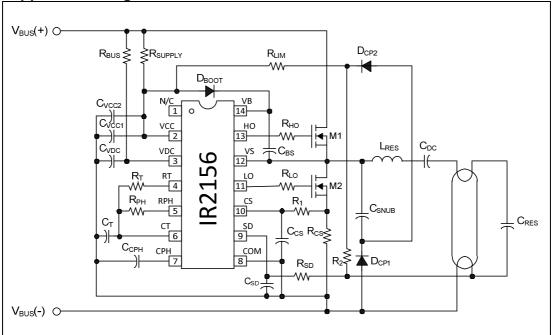
Description

The IR2156 incorporates a high voltage halfbridge gate driver with a programmable oscillator and state diagram to form a complete ballast control IC. The IR2156 features include programmable preheat and run frequencies, programmable preheat time, programmable dead-time, and programmable overcurrent protection. Comprehensive protection features such as protection from failure of a lamp to strike, filament failures, as well as an automatic restart function, have been included in the design.

Packages



Application Diagram



Absolute Maximum Ratings
Absolute Maximum Ratings indicate sustained limits beyond which damage to the device may occur. All voltage parameters are absolute voltages referenced to COM, all currents are defined positive into any lead. The Thermal Resistance and Power Dissipation ratings are measured under board mounted and still air conditions.

Symbol	Definition	Min.	Max.	Units		
VB	High side floating supply voltage	-0.3	625			
Vs	High side floating supply offset voltage	VB - 25	VB + 0.3	V		
VHO	High side floating output voltage	VS - 0.3	VB + 0.3	V		
VLO	Low side output voltage	-0.3	VCC + 0.3			
IOMAX	Maximum allowable output current (HO, LO) due to external power transistor miller effect		-500	500	mA	
VDC	VDC pin voltage		-0.3	VCC+0.3		
VCT	CT pin voltage		-0.3	VCC+0.3	V	
VCPH	CPH pin voltage		-0.3	VCC+0.3		
ICPH	CPH pin current		-5	5		
IRPH	RPH pin current	-5	5	mA		
VRPH	RPH pin voltage	-0.3	VCC+0.3	V		
IRT	RT pin current	-5	5	mA		
VRT	RT pin voltage	-0.3	VCC+0.3	V		
VCS	Current sense pin voltage	-0.3	5.5	V		
ICS	Current sense pin current	-5	5			
ISD	Shutdown pin current	-5	5	mA		
ICC	Supply current (Note 1)		-20	20		
dV/dt	Allowable offset voltage slew rate		-50	50	V/ns	
Do	Package power dissipation @ T _A ≤ +25°C		1.80	107		
PD	PD $PD = (T_{JMAX}-T_A)/R\theta JA$ (14-Pin SOIC)			1.40	W	
R _θ JA	Thermal resistance, junction to ambient	(14-Pin DIP)		70	00044	
			82	°C/W		
TJ	Junction temperature	-55	150			
TS	Storage temperature	-55	150	°C		
TL	Lead temperature (soldering, 10 seconds)		300			

Note 1: This IC contains a zener clamp structure between the chip VCC and COM which has a nominal breakdown voltage of 15.6V. Please note that this supply pin should not be driven by a DC, low impedance power source greater than the VCLAMP specified in the Electrical Characteristics section.

Recommended Operating Conditions For proper operation the device should be used within the recommended conditions.

Symbol	Definition	Min.	Max.	Units		
VBS	High side floating supply voltage	VBSUV+	VCLAMP			
٧s	Steady state high side floating supply offset voltage	-1	600	V		
VCC	Supply voltage	VCCUV+	VCLAMP			
ICC	Supply current	Note 2	10	mA		
СТ	CT lead capacitance	220		pF		
ISD	Shutdown lead current		1	A		
ICS	Current sense pin current	-1	1	mA		
TJ	Junction temperature	-25	125	°C		

Note 2: Enough current should be supplied into the VCC pin to keep the internal 15.6V zener clamp diode on this pin regulating at its voltage, V_{CLAMP}.

Electrical Characteristics

VCC = VBS = VBIAS = 14V +/- 0.25V, VVDC=Open, RT=40K Ω , RPH=100K Ω , CT=470 pF, VCPH=0.0V, VSD=0.0V, VCS=0.0V, CLO=CHO=1000 pF, TA=25C unless otherwise specified.

Symbol	1		Тур	Max	Units	Test Conditions		
Supply C	Supply Characteristics							
VCCUV+	VCC supply undervoltage positive going threshold	10.5	11.5	12.5		VCC rising from 0V		
VCCUV-	VCC supply undervoltage negative going threshold	8.5	9.5	10.5	٧	VCC falling from 14V		
VUVHYS	VCC supply undervoltage lockout hysteresis	1.5	2.0	3.0				
IQCCUV	UVLO mode quiescent current	50	120	200		VCC=11V		
IQCCFLT	Fault-mode quiescent current		200	470	μA	SD = 5.1V, or CS = 1.3V		
IQCC	Quiescent VCC supply current		1.0	1.5	- mA	CT connected to COM, VCC = 14V, RT = $15k\Omega$		
ICC40k	VCC supply current, f = 40kHz	1.3	1.5	1.7	IIIA	VCPH=12V, VVDC=12V		
VCLAMP	VCC zener clamp voltage	14.5	15.6	16.5	V	ICC = 5mA		
Floating Supply Characteristics								
IQBS0	Quiescent VBS supply current	-5	0	5		VHO = VS (CT=0V)		
IQBS1	Quiescent VBS supply current		30	50	μA	VHO = VB (CT=14V)		
lk	Offset supply leakage current			50	μA	VB = VS = 600V		

Electrical Characteristics

VCC = VBS = VBIAS = 14V +/- 0.25V, VVDC=Open, RT=40K Ω , RPH=100K Ω , CT=470 pF, VCPH=0.0V, VSD=0.0V, VCS=0.0V, CLO=CHO=1000 pF, TA=25C unless otherwise specified.

Symbol	Definition	Min	Тур	Max	Units	Test Conditions	
Oscillator, Ballast Control, I/O Characteristics							
foscrun	Oscillator frequency during RUN mode	36.0	40.0	44.0	- kHz	V _{VDC} =14V, VCPH=Open	
foscph	Oscillator frequency during PH mode	49.0	55.0	60.0	KIIZ	V _{VDC} =14V, VCPH=COM	
d	Oscillator duty cycle		50		%		
VCT+	Upper CT ramp voltage threshold		8.3			V _{CC} =14V	
VCT-	Lower CT ramp voltage threshold		4.8		V		
VCTFLT	Fault-mode CT pin voltage		0			SD>5.1V or CS>1.3V	
tDLO	LO output deadtime		2.0				
tDHO	HO output deadtime		2.0		µsec		
RDT	Internal deadtime resistor		3		kΩ		
Preheat C	haracteristics						
ICPH	CPH pin charging current	3.6	4.3	5.2	μΑ	CT=10V, VDC=5V, VCPH=0V	
VCPHFLT	Fault-mode CPH pin voltage		0		mV	SD>5.1V or CS>1.3V	
RPH Char	acteristics						
IRPHLK	Open circuit RPH pin leakage current		0.1		μΑ	CT=10V	
VRPHFLT	Fault-mode RPH pin voltage		0		mV	SD>5.1V or CS>1.3V	
RT Charac	cteristics						
IRTLK Open circuit RT pin leakage current			0.1		μΑ	CT=10V	
VRTFLT	Fault-mode RT pin voltage 0 m\		mV	SD>5.1V or CS>1.3V			
Protection Circuitry Characteristics							
VSDTH+	Rising shutdown pin threshold voltage		5.1		V		
VSDHYS	SD pin Reset threshold voltage		450		mV		
VCSTH+	Over-current sense threshold voltage	1.1	1.25	1.44	V		
tcs	Over-current sense propogation delay		160			Delay from CS to LO	
VCSPW	Over-current sense minimum pulse width		135		nsec	VCS pulse amplitude = VCSTH + 100mV	
RVDC	DC bus sensing resistor	7.5	10.0	14.0	kΩ	VCPH>12V, VDC=7V, CT=0	
VCPH-VDC	CPH to VDC offset voltage	10.3	10.9	11.4	V	VCPH open, VDC=0V	

Electrical Characteristics

VCC = VBS = VBIAS = 14V +/- 0.25V, VVDC=Open, RT=39K Ω , RPH=100K Ω , CT=470 pF, VCPH=0.0V, VSD=0.0V, VCS=0.0V, CLO=CHO=1000 pF, TA=25C unless otherwise specified.

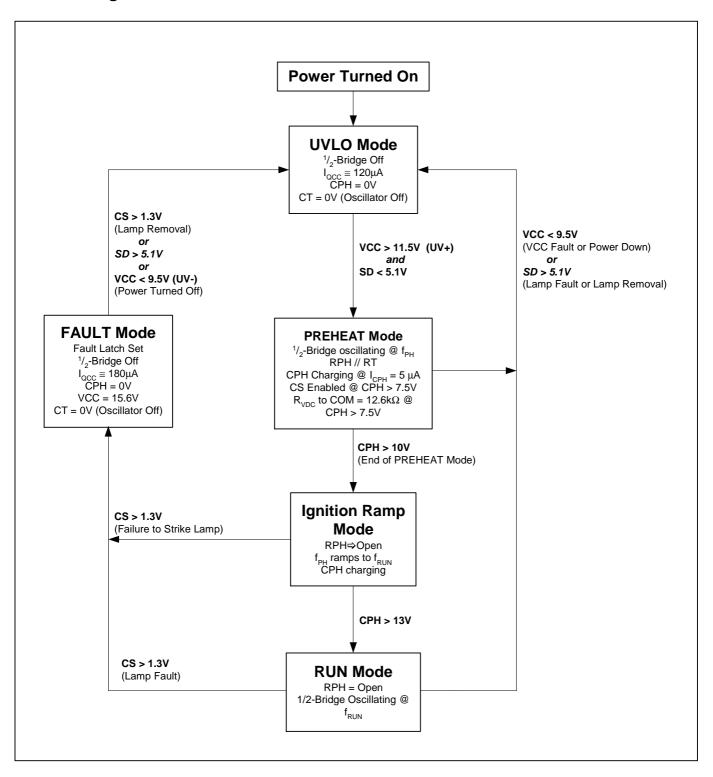
Symbol	Definition		Тур	Max	Units	Test Conditions
Gate Driver Output Characteristics						
VOL	Low-level output voltage		СОМ		V	IO = 0
VOH	High-level output voltage		VCC		V	IO = 0
t _r	Turn-on rise time		110	150		
t _f	Turn-off fall time		55	100	nsec	

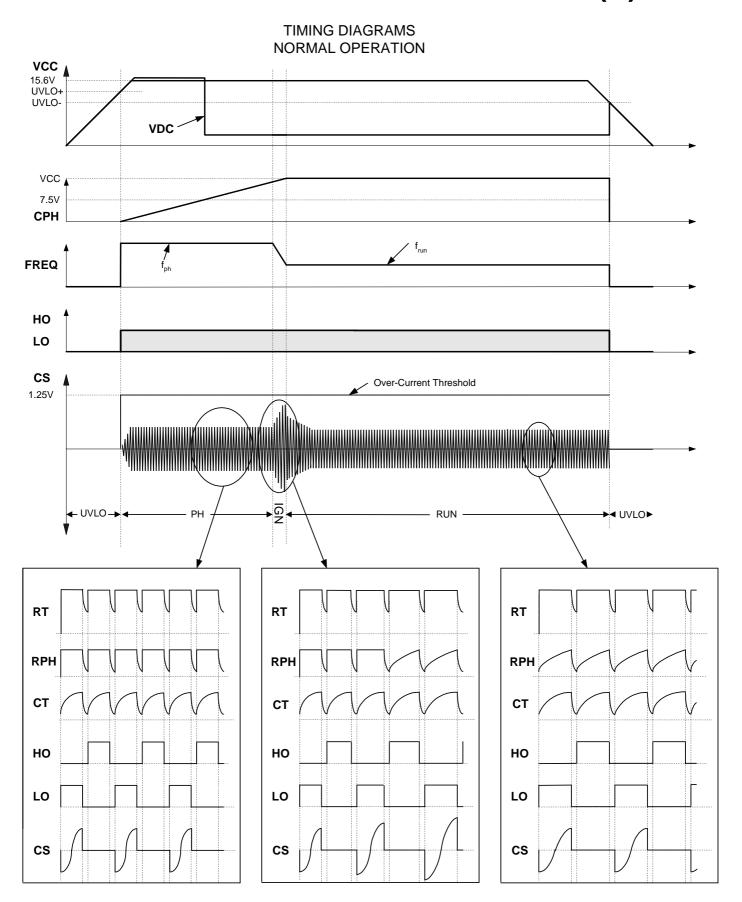
Block Diagram S1 VΒ Driver Logic S2 High-Side Driver R⋛ 40K **≷** ⊠ HO Comp 1 Q RDT Soft Start 2.5K\$ R Q RŞ –⊠ VS S3 S6 RPH ⊠ Low-Side Driver ICPH ↓ Schmitt 1 -⊠ LO CPH ⊠ Fault Logic T) 5.10 5.10 RVDC VDC ⊠ R1 ₹10K R2 Q CS Comp 3 SD 🛭 Comp 2 Voltage Detect COM

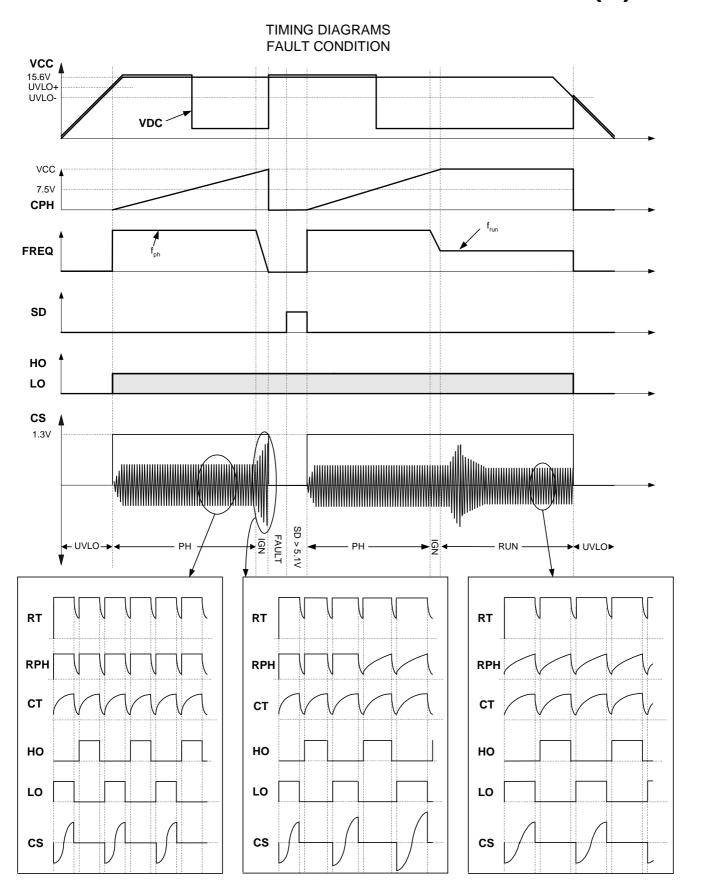
Pin Assignments & Definitions

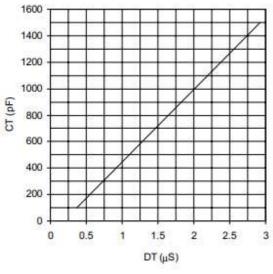
Pin Assignr	nents	Pin #	Symbol	Description
		1	NC	No Connect
NC 1 0	14 VB	2	VCC	Logic & Low-Side Gate Driver Supply
VCC 2	43 110	3	VDC	IC Start-up and DC Bus Sensing Input
VCC 2	13 HO	4	RT	Minimum Frequency Timing Resistor
VDC 3	12 VS	5	RPH	Preheat Frequency Timing Resistor
		6	СТ	Oscillator Timing Capacitor
RT 4	11 LO	7	СРН	Preheat Timing Capacitor
	10 66	8	СОМ	IC Power & Signal Ground
RPH 5	10 CS	9	SD	Shutdown Input
ст 6	9 SD	10	CS	Current Sensing Input
		11	LO	Low-Side Gate Driver Output
CPH 7	8 COM	12	VS	High-Side Floating Return
		13	НО	High-Side Gate Driver Output
		14	VB	High-Side Gate Driver Floating Supply

State Diagram

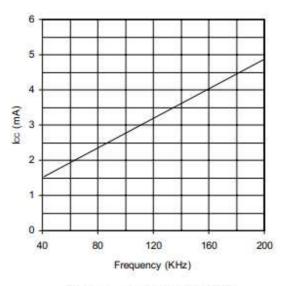




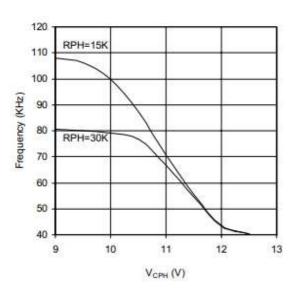




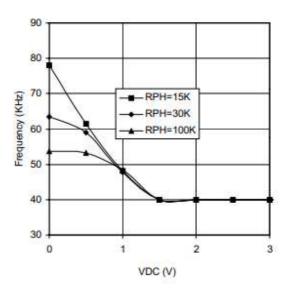
Graph 1. CT vs Dead Time (IR2156)



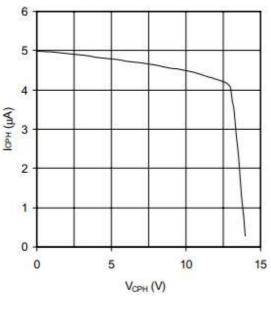
Graph 2. I_{CC} vs Frequency (IR2156)



Graph 3. Frequency vs V_{CPH} (IR2156)



Graph 4. Frequency vs VDC (IR2156)



1000000

CT=220pF

CT=470pF

CT=470pF

CT=470pF

CT=3300pF

CT=5300pF

CT=6800pF

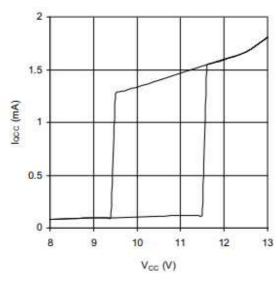
CT=6800pF

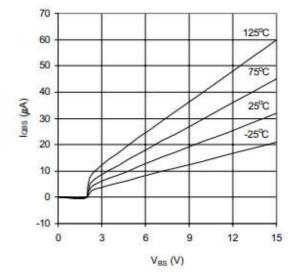
CT=6800pF

CT=6800pF

Graph 5. ICPH vs VCPH (IR2156)

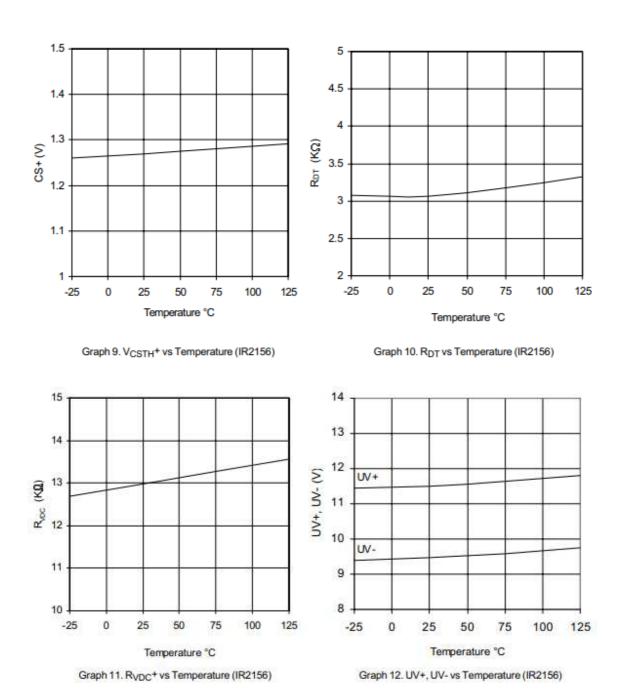
Graph 6. Frequency vs RT (IR2156)

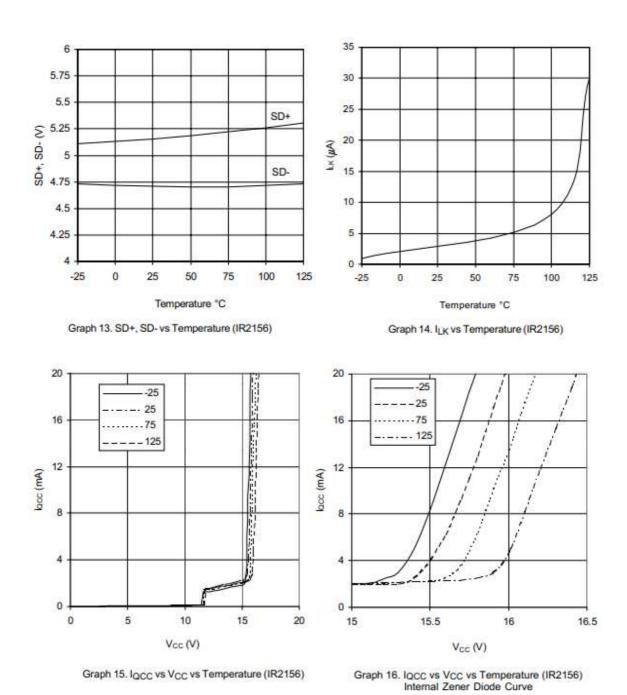


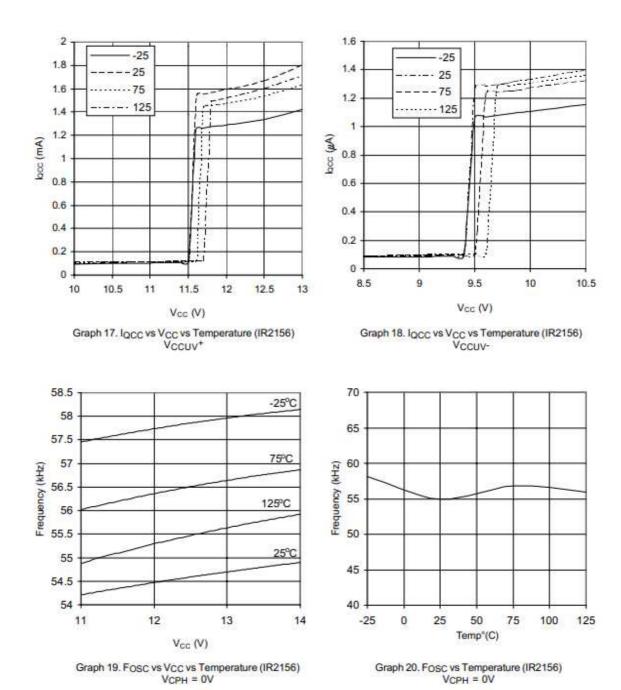


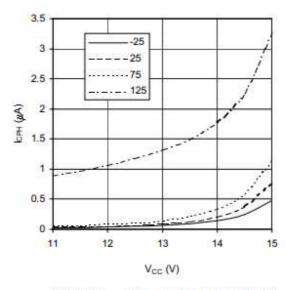
Graph 7. I_{QCC} vs V_{CC} (IR2156) UVLO Hysteresis

Graph 8. IQBS vs VCC vs Temp(IR2156)

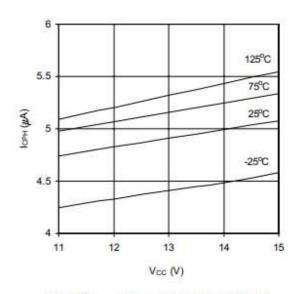




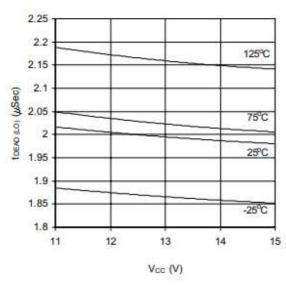




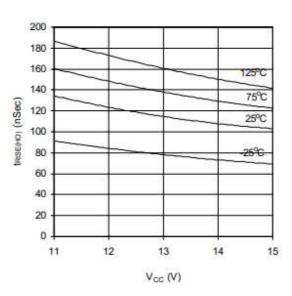
Graph 21. I_{CPH} vs V_{CC} vs Temperature (IR2156) V_{CPH} = V_{CC}



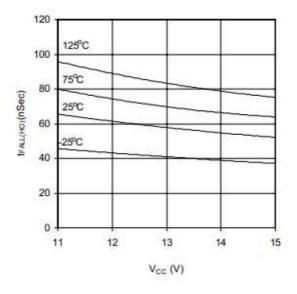
Graph 22. I_{CPH} vs V_{CC} vs Temperature (IR2156) $V_{CPH} = 0V$

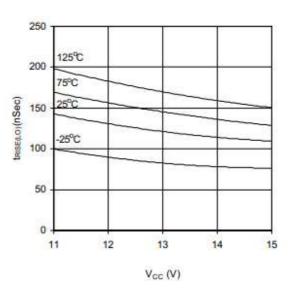


Graph 23. t_{DEAD} vs V_{CC} vs Temperature (IR2156) CT = 1nF



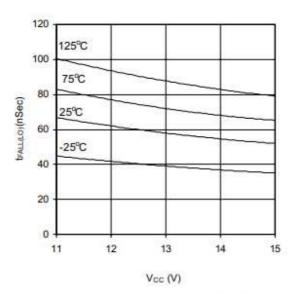
Graph 24. tRISE(HO) vs VCC vs Temperature (IR2156)





Graph 25. t_{FALL(HO)} vs V_{CC} vs Temperature (IR2156)

Graph 26. t_{RISE(LO)} vs V_{CC} vs Temperature (IR2156)



Graph 27. tFALL(LO) vs VCC vs Temperature (IR2156)

Functional Description

Under-voltage Lock-out Mode (UVLO)

The under-voltage lock-out mode (UVLO) is defined as the state the IC is in when VCC is below the turn-on threshold of the IC. To identify the different modes of the IC, refer to the State Diagram shown on page 2 of this document. The IR2156 undervoltage lock-out is designed to maintain an ultra low supply current of less than 200uA, and to guarantee the IC is fully functional before the high and low side output drivers are activated. Figure 1 shows an efficient supply voltage using the start-up current of the IRS2156 together with a charge pump from the ballast output stage (R_{SUPPLY} , C_{VCC} , D_{CP1} and D_{CP2}).

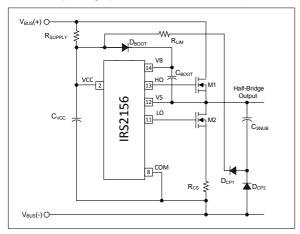


Figure 1, Start-up and supply circuitry.

The start-up capacitor (C_{VCC}) is charged by current through supply resistor (R_{SUPPLY}) minus the start-up current drawn by the IC. This resistor is chosen to provide 2X the maximum start-up current to guarantee ballast start-up at low line input voltage. Once the capacitor voltage on VCC reaches the start-up threshold VCCUV+, and the SD pin is below VSDTH-, the IC turns on and HO and LO begin to oscillate. The capacitor begins to discharge due to the increase in IC operating current (Figure 2).

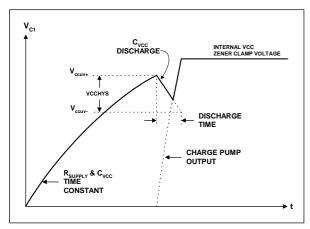


Figure 2, Supply capacitor (C_{VCC}) voltage.

During the discharge cycle, the rectified current from the charge pump charges the capacitor above the IC turn-off threshold. The charge pump and the internal 15.6V zener clamp of the IC take over as the supply voltage. The start-up capacitor and snubber capacitor must be selected such that enough supply

current is available over all ballast operating conditions. An external bootstrap diode (D_{BOOT}) and the supply capacitor (C_{BOOT}) comprise the supply voltage for the high side driver circuitry. To guarantee that the high-side supply is charged up before the first pulse on pin HO, the first pulse from the output drivers comes from the LO pin. During undervoltage lock-out mode, the high-and low-side driver outputs HO and LO are both low, pin CT is connected internally to COM to disable the oscillator, and pin CPH is connected internally to COM for resetting the preheat time.

Preheat Mode (PH)

The preheat mode is defined as the state the IC is in when the lamp filaments are being heated to their correct emission temperature. This is necessary for maximizing lamp life and reducing the required ignition voltage. The IR2156 enters preheat mode when VCC exceeds the VCCUV+ positive-going threshold. HO and LO begin to oscillate at the preheat frequency with 50% duty cycle and with a dead-time which is set by the value of the external timing capacitor, CT, and internal deadtime resistor, RDT. Pin CPH is disconnected from COM and an internal 4uA current source (Figure 3) charges the external preheat timing capacitor on CPH linearly. The over-current protection on pin CS is disabled during preheat.

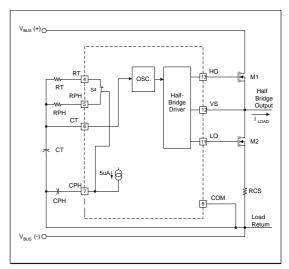


Figure 3, Preheat circuitry

The preheat frequency is determined by the parallel combination of resistors RT and RPH, together with timing capacitor CT. CT charges and discharges between 1/3 and 3/5 of VCC (see Timing Diagram, page 9). CT is charged exponentially through the parallel combination of RT and RPH connected internally to VCC through MOSFET S1. The charge time of CT from 1/3 to 3/5 VCC is the on-time of the respective output gate driver, HO or LO. Once CT exceeds 3/5 VCC, MOSFET S1 is turned off, disconnecting RT and RPH from VCC. CT is then discharged exponentially through an internal resistor, RDT, through MOSFET S3 to COM. The discharge time of CT from 3/5 to 1/3 VCC is the dead-time (both off) of the output gate drivers, HO and LO. The selected value of CT together with RDT therefore program the desired dead-time (see Design Equations, page 12, Equations 1 and 2). Once CT discharges below 1/3 VCC, MOSFET S3 is turned off, disconnecting RDT from COM, and MOSFET S1 is turned on, connecting RT and RPH again to VCC. The frequency remains at the preheat frequency until the voltage on pin CPH exceeds 13V and the IC enters Ignition Mode. During the preheat mode, both the over-current protection and the

DC bus under-voltage reset are enabled when pin CPH exceeds 7.5V.

Ignition Mode (IGN)

The ignition mode is defined as the state the IC is in when a high voltage is being established across the lamp necessary for igniting the lamp. The IR2156 enters ignition mode when the voltage on pin CPH exceeds 13V.

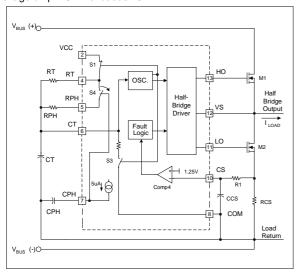


Figure 4, Ignition circuitry.

Pin CPH is connected internally to the gate of a p-channel MOSFET (S4) (see Figure 4) that connects pin RPH with pin RT. As pin CPH exceeds 13V, the gate-to-source voltage of MOSFET S4 begins to fall below the turn-on threshold of S4. As pin CPH continues to ramp towards VCC, switch S4 turns off slowly. This results in resistor RPH being disconnected smoothly from resistor RT, which causes the operating frequency to ramp smoothly from the preheat frequency, through the ignition frequency, to the final run frequency. The over-current threshold on pin CS will protect the ballast against a non-strike or open-filament lamp fault condition. The voltage on pin CS is defined by the lower halfbridge MOSFET current flowing through the external current sensing resistor RCS. The resistor RCS therefore programs the maximum allowable peak ignition current (and therefore peak ignition voltage) of the ballast output stage. The peak ignition current must not exceed the maximum allowable current ratings of the output stage MOSFETs. Should this voltage exceed the internal threshold of 1.3V, the IC will enter FAULT mode and both gate driver outputs HO and LO will be latched low.

Run Mode (RUN)

Once the lamp has successfully ignited, the ballast enters run mode. The run mode is defined as the state the IC is in when the lamp arc is established and the lamp is being driven to a given power level. The run mode oscillating frequency is determined by the timing resistor RT and timing capacitor CT (see Design Equations, page 12, Equations 3 and 4). Should hard-switching occur at the half-bridge at any time due to an open-filament or lamp removal, the voltage across the current sensing resistor, RCS, will exceed the internal threshold of 1.3 volts and the IC will enter FAULT mode. Both gate driver outputs, HO and LO, will be latched low.

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DC Bus Under-voltage Reset

Should the DC bus decrease too low during a brown-out line condition or over-load condition, the resonant output stage to the lamp can shift near or below resonance. This can produce hard-switching at the half-bridge which can damage the halfbridge switches. To protect against this, pin VDC measures the DC bus voltage and pulls down on pin CPH linearly as the voltage on pin VDC decreases 10.9V below VCC. This causes the pchannel MOSFET S4 (Figure 4) to close as the DC bus decreases and the frequency to shift higher to a safe operating point above resonance. The DC bus level at which the frequency shifting occurs is set by the external RBUS resistor and internal RVDC resistor. By pulling down on pin CPH, the ignition ramp is also reset. Therefore, should the lamp extinguish due to very low DC bus levels, the lamp will be automatically ignited as the DC bus increases again. The internal RVDC resistor is connected between pin VDC and COM when CPH exceeds 7.5V (during preheat mode).

Fault Mode (FAULT)

Should the voltage at the current sensing pin, CS, exceed 1.3V at any time after the preheat mode, the IC enters fault mode and both gate driver outputs, HO and LO, are latched in the 'low' state. CPH is discharged to COM for resetting the preheat time, and CT is discharged to COM for disabling the oscillator. To exit fault mode, VCC must be recycled back below the UVLO negative-going turn-off threshold, or, the shutdown pin, SD, must be pulled above VSDTH+. Either of these will force the IC to enter UVLO mode (see State Diagram, page 2). Once VCC is above VCCUV+ and SD is below 4.5V, the IC will begin oscillating again in the preheat mode.

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Design Equations

Note: The results from the following design equations can differ slightly from experimental measurements due to IC tolerances, component tolerances, and oscillator over- and under-shoot due to internal comparator response time.

For additional design support for different lamp types and AC line input configurations, including component calculations, schematics, bill of materials and inductor specifications, please download IR's Ballast Design Assistant (BDA) software at www.irf.com.

Step 1: Program Dead-time

The dead-time between the gate driver outputs HO and LO is programmed with timing capacitor CT and an internal dead-time resistor RDT. The dead-time is the discharge time of capacitor CT from 3/5VCC to 1/3VCC and is given as:

$$t_{DT} = C_T \cdot 2000$$
 [Seconds] (1)

Or,

$$C_T = \frac{t_{DT}}{2000}$$
 [Farads] (2)

Step 2: Program Run Frequency

The final run frequency is programmed with timing resistor RT and timing capacitor CT. The charge time of capacitor CT from 1/3VCC to 3/5VCC determines the on-time of HO and LO gate driver outputs. The run frequency is therefore given as:

$$f_{RUN} = \frac{1}{2 \cdot C_T (0.6 \cdot R_T + 2000)}$$
 [Hertz] (3)

Or,

$$R_T = \frac{1}{1.12 \cdot C_T \cdot f_{RUN}} - 3333$$
 [Ohms] (4)

Step 3: Program Preheat Frequency

The preheat frequency is programmed with timing resistors RT and RPH, and timing capacitor CT. The timing resistors are

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connected in parallel internally for the duration of the preheat time. The preheat frequency is therefore given as:

$$f_{PH} = \frac{1}{2 \cdot C_T \cdot \left(\frac{0.51 \cdot R_T \cdot R_{PH}}{R_T + R_{PH}} + 2000\right)}$$
 [Hertz] (5)

Or,

$$R_{PH} = \frac{\left(\frac{1}{1.12 \cdot C_T \cdot f_{PH}} - 3333\right) \cdot R_T}{R_T - \left(\frac{1}{1.12 \cdot C_T \cdot f_{PH}} - 3333\right)} \quad \text{[Ohms]} \quad (6)$$

Step 4: Program Preheat Time

The preheat time is defined by the time it takes for the capacitor on pin CPH to charge up to 13 volts. An internal current source of 5uA flows out of pin CPH. The preheat time is therefore given as:

$$t_{PH} = C_{PH} \cdot 3.02e6$$
 [Seconds] (7)

Or,

$$C_{PH} = t_{PH} \cdot 0.33e - 6$$
 [Farads] (8)

Step 5: Program Maximum Ignition Current

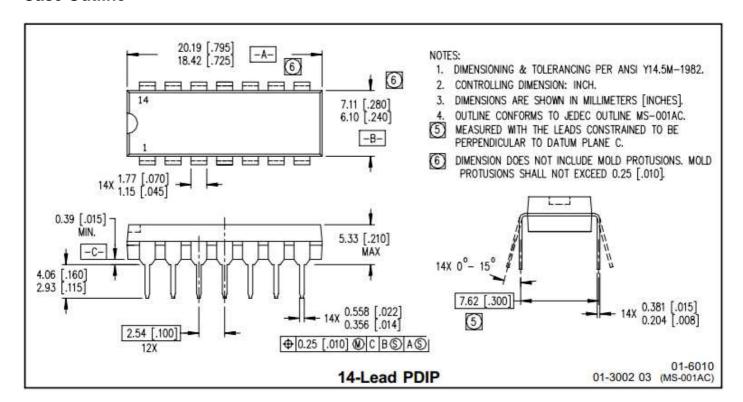
The maximum ignition current is programmed with the external resistor RCS and an internal threshold of 1.25 volts (VCSTH+). This threshold determines the over-current limit of the ballast, which can be exceeded when the frequency ramps down towards resonance during ignition and the lamp does not ignite. The maximum ignition current is given as:

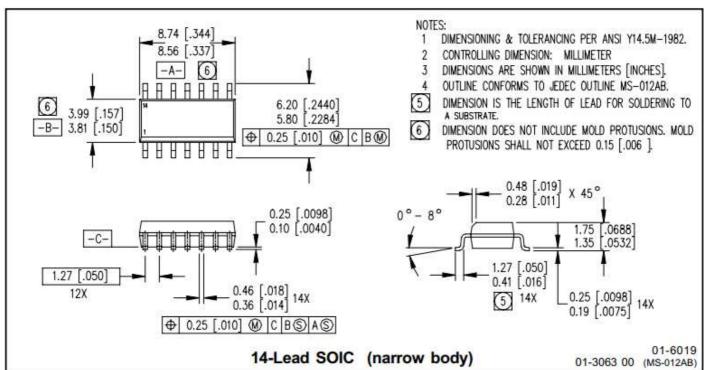
$$I_{IGN} = \frac{1.25}{R_{CS}}$$
 [Amps Peak] (9)

Or

$$R_{CS} = \frac{1.25}{I_{IGN}}$$
 [Ohms]

Case Outline





International IOR Rectifier

IR2156(S)PbF

Qualification: Lead-free MSL3, industrial



http://www.irf.com

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ПОСТАВКА ЭЛЕКТРОННЫХ КОМПОНЕНТОВ

многоканальный

Общество с ограниченной ответственностью «МосЧип» ИНН 7719860671 / КПП 771901001 Адрес: 105318, г.Москва, ул.Щербаковская д.3, офис 1107

Данный компонент на территории Российской Федерации Вы можете приобрести в компании MosChip.

Для оперативного оформления запроса Вам необходимо перейти по данной ссылке:

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Вы можете разместить у нас заказ для любого Вашего проекта, будь то серийное производство или разработка единичного прибора.

В нашем ассортименте представлены ведущие мировые производители активных и пассивных электронных компонентов.

Нашей специализацией является поставка электронной компонентной базы двойного назначения, продукции таких производителей как XILINX, Intel (ex.ALTERA), Vicor, Microchip, Texas Instruments, Analog Devices, Mini-Circuits, Amphenol, Glenair.

Сотрудничество с глобальными дистрибьюторами электронных компонентов, предоставляет возможность заказывать и получать с международных складов практически любой перечень компонентов в оптимальные для Вас сроки.

На всех этапах разработки и производства наши партнеры могут получить квалифицированную поддержку опытных инженеров.

Система менеджмента качества компании отвечает требованиям в соответствии с ГОСТ Р ИСО 9001, ГОСТ РВ 0015-002 и ЭС РД 009

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