# NCP690, NCP691, NCP692

# 1 A, Low $I_{GND}$, Very Low Dropout Regulator (VLDO) with/without Enable

The NCP690, NCP691, NCP692 CMOS LDO family provides 1 A of output current with enhanced ESD in either fixed voltage options or an adjustable output voltage from 5.0 V down to 1.25 V. This device is designed for space constrained and portable battery powered applications and offer additional features such as high PSRR, low Quiescent and Ground current consumption, low noise operation, short circuit and thermal protection. The device is designed to be used with low cost ceramic capacitors and is packaged in the 6−Lead DFN3x3 package.

### Features

- Output Voltage Options: Adjustable, 1.5 V, 1.8 V, 2.5 V, 3.3 V, 5.0 V – Other Options Possible
- Adjustable Output by External Resistors from 5.0 V down to 1.25 V
- Guaranteed 1 A Output Current
- ±1.5% Output Voltage Tolerance over All Operating Conditions (Adjustable)
- ±2% Output Voltage Tolerance over All Operating Conditions (Fixed)
- Typical Noise Voltage of 50 μVrms without a Bypass Capacitor
- Typical Dropout Voltage of 190 mV at 1 A ($V_{out}$ = 2.5 V, $T_J$ = 25°C)
- Active Output Discharge
- Active Low Enable Pin (NCP691 Device)
- Active High Enable Pin (NCP692 device)
- Enhanced ESD: 4 kV and 200 V
- These are Pb−Free Devices

### Applications

- Laptops and PCI Cards
- Modem Banks and Telecom Boards
- DSP, FPGA, Microprocessor Boards
- Portable, Battery−Power Applications
- Hard Disk Drives

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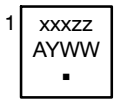
**DFN6 3x3**
**MN SUFFIX**
**CASE 506AH**

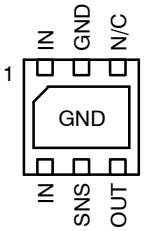
### MARKING DIAGRAM



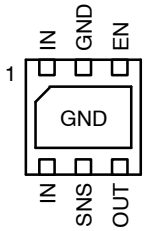

**Pb−Free**

| | |
|---|---|
| xxx | = 690, 691, 692 |
| zz | = 15, 18, 25, 33, 50, AD |
| A | = Assembly Location |
| Y | = Year |
| WW | = Work Week |
| ▪ | = Pb−Free Package |

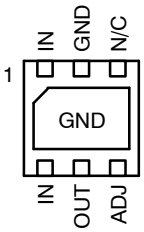
### PIN ASSIGNMENT



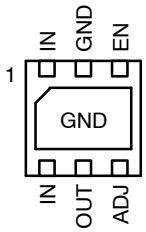

**NCP690 – 6 PIN DFN**
Fixed Version
(Bottom View)




**NCP691, NCP692 – 6 PIN DFN**
Fixed Version
(Bottom View)




**NCP690 – 6 PIN DFN**
Adjustable Version
(Bottom View)




**NCP691, NCP692 – 6 PIN DFN**
Adjustable Version
(Bottom View)

### ORDERING INFORMATION

See detailed ordering and shipping information in the package dimensions section on page 13 of this data sheet.

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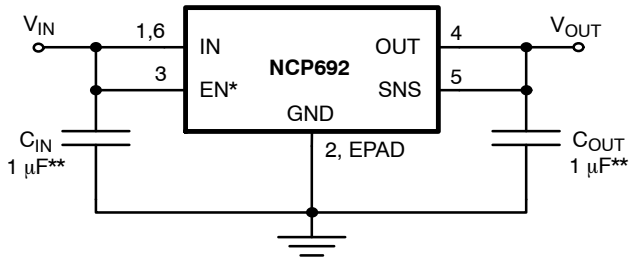
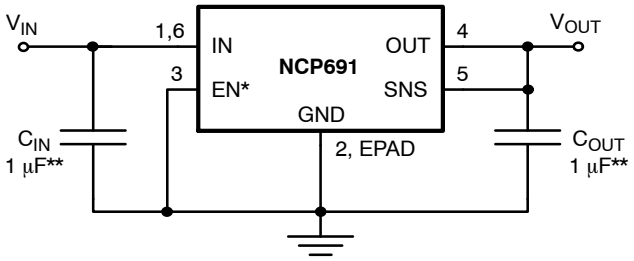


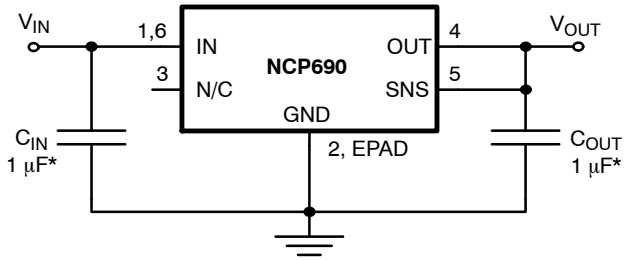

Note: * NCP692 device has EN active high
** Minimum value required for stability

**Figure 1. NCP692 Typical Application Circuit for Fixed Version**
(Output voltage versions: 1.5 V, 1.8 V, 2.5 V, 5.0 V)



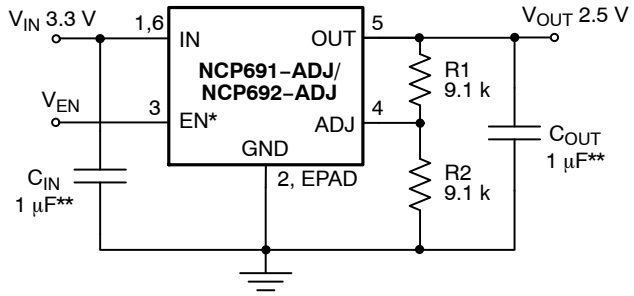

Note: * NCP691 device has EN active low
** Minimum value required for stability

**Figure 2. NCP691 Typical Application Circuit for Fixed Version**
(Output voltage versions: 1.5 V, 1.8 V, 2.5 V, 5.0 V)




Note: * Minimum value required for stability

**Figure 3. NCP690 Typical Application Circuit for Fixed Version**
(Output voltage versions: 1.5 V, 1.8 V, 2.5 V, 5.0 V)




Note: * NCP691−ADJ device has EN active low and
NCP692−ADJ device has EN active high
** Minimum value required for stability

**Figure 4. NCP692 Typical Application Circuit for Adjustable Version**
(Adjustable version for 1.25 V < $V_{OUT}$ ≤ 5.0 V)

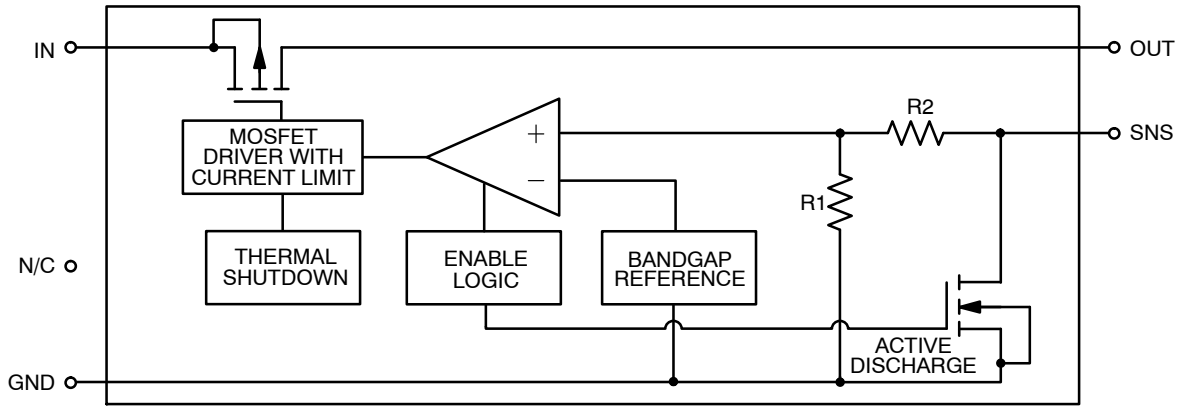



**Figure 5. NCP690 Block Diagram (Fixed Version)**

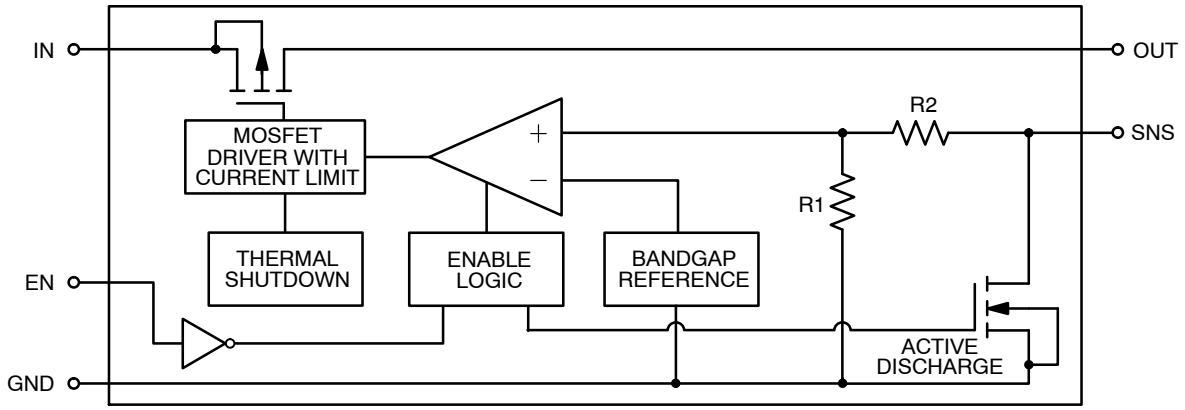



**Figure 6. NCP691 Block Diagram (Fixed Version)**

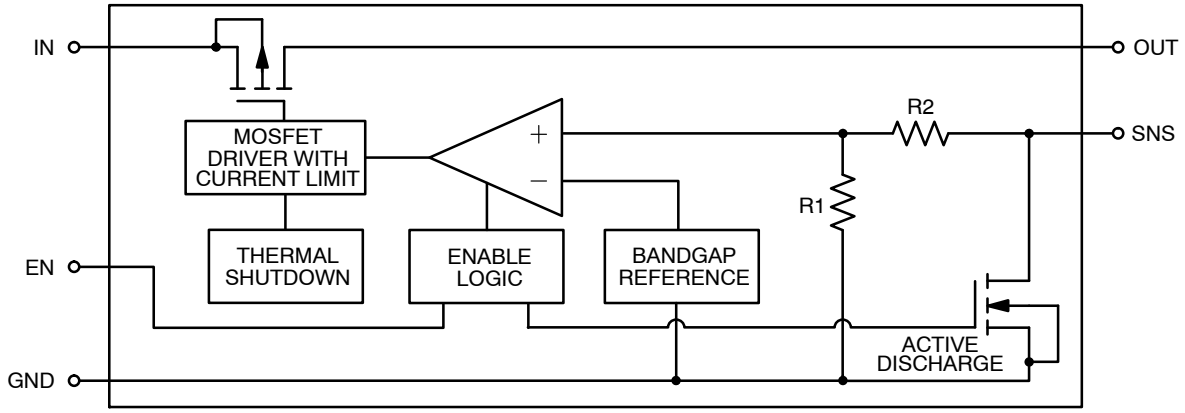



**Figure 7. NCP692 Block Diagram (Fixed Version)**

**Table 1. PIN FUNCTION DESCRIPTION FOR FIXED VERSION**

| Pin No. | Pin Name | Description |
|---|---|---|
| 1, 6 | IN | Voltage inputs which supplies the current to the regulator. Both of these pins should be connected together for full output current capability |
| 2 | GND | Power supply ground of the regulator. Connected to the die through the lead frame. Soldered to the copper plane allows for effective heat removal. |
| 3 | EN | For NCP691 and NCP692 this pin functions as Enable Active Low and Enable Active High respectively. For NCP690 this pin has no special meaning and should be left disconnected. |
| 4 | OUT | Regulated output voltage |
| 5 | SNS | Sense input. This pin should be connected directly to OUT pin. |

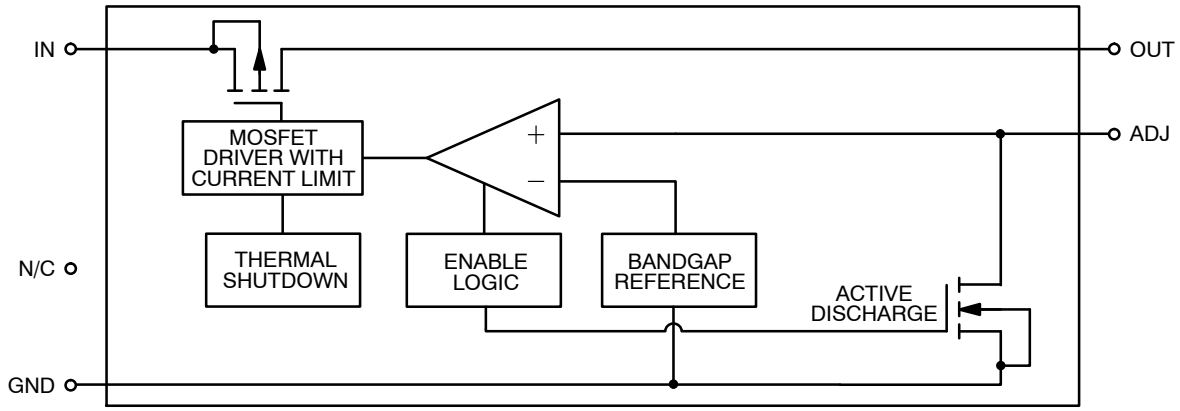



**Figure 8. NCP690 Block Diagram (Adjustable Version)**

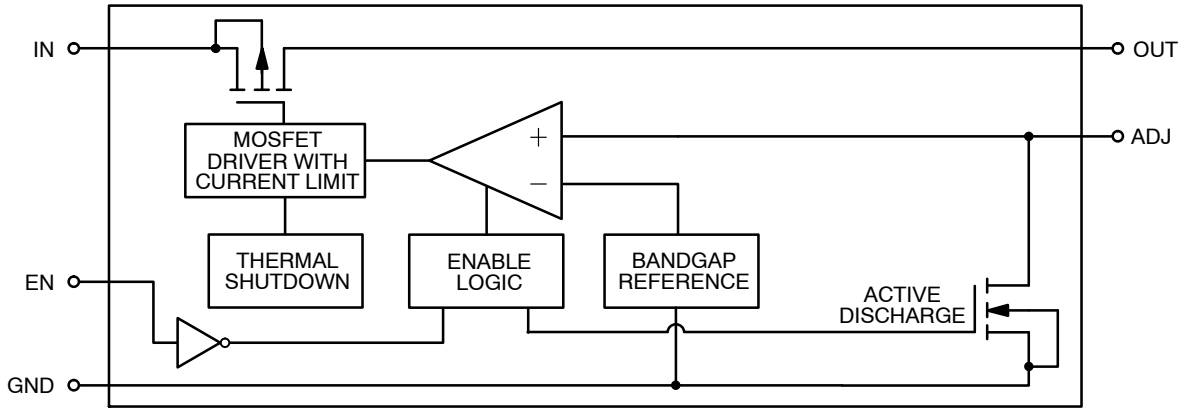



**Figure 9. NCP691 Block Diagram (Adjustable Version)**

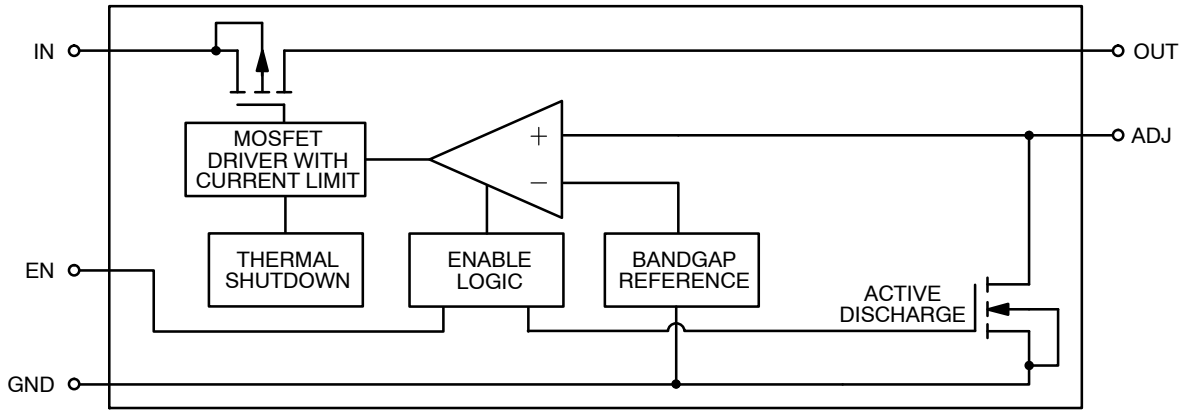



**Figure 10. NCP692 Block Diagram (Adjustable Version)**

**Table 2. PIN FUNCTION DESCRIPTION FOR ADJUSTABLE VERSION**

| Pin No. | Pin Name | Description |
|---|---|---|
| 1, 6 | IN | Voltage inputs which supplies the current to the regulator. Both of these pins should be connected together for full output current capability |
| 2 | GND | Power supply ground of the regulator. Connected to the die through the lead frame. Soldered to the copper plane allows for effective heat removal. |
| 3 | EN | For NCP691 and NCP692 this pin functions as Enable Active Low and Enable Active High respectively. For NCP690 this pin has no special meaning and should be left disconnected. |
| 4 | ADJ | Feedback input. Connect to middle point of resistor divider for Adjustable version. |
| 5 | OUT | Regulated output voltage |

**Table 3. ABSOLUTE MAXIMUM RATINGS**

| Rating | | Symbol | Value | Unit |
|---|---|---|---|---|
| Input Voltage (Note 1) | | $V_{IN}$ | −0.3 to 6.5 | V |
| Chip Enable Voltage | | $V_{EN}$ | −0.3 to 6.5 | V |
| Output Voltage | | $V_{OUT}$ | −0.3 to 6.5 | V |
| Output Voltage / Sense Input, (SNS pin) | | $V_{SNS}$ | −0.3 to 6.5 | V |
| Electrostatic Discharge | Human Body Model | ESD | 4000 | V |
| | Machine Model | | 200 | |
| Maximum Junction Temperature | | $T_{J_MAX}$ | 150 | °C |
| Storage Temperature Range | | $T_{STG}$ | −65 to 150 | °C |

Stresses exceeding Maximum Ratings may damage the device. Maximum Ratings are stress ratings only. Functional operation above the Recommended Operating Conditions is not implied. Extended exposure to stresses above the Recommended Operating Conditions may affect device reliability.

NOTE: This device series contains ESD protection and exceeds the following tests:
ESD HBM tested per JEDEC standard: JESD22−A114
ESD MM tested per JEDEC standard: JESD22−A115
Latch−up Current Maximum Rating: ≤ 150 mA per JEDEC standard: JESD78

**Table 4. PACKAGE THERMAL CHARACTERISTICS**

| Rating | | Condition | Symbol | Value | Unit |
|---|---|---|---|---|---|
| Thermal Resistance, Junction−to−Ambient (Note 2) | DFN6 3x3, 1 oz Cu | 64 $mm^2$ Cu<br>645 $mm^2$ Cu | $R_{\theta JA}$ | 169<br>70 | °C/W |
| | DFN6 3x3, 2 oz Cu | 64 $mm^2$ Cu<br>645 $mm^2$ Cu | $R_{\theta JA}$ | 151<br>62 | °C/W |
| Thermal Resistance, Junction−to−Pin | | | $R_{\theta JL}$ | 15 | °C/W |

**Table 5. OPERATING RANGES**

| Rating | Symbol | Value | Unit |
|---|---|---|---|
| Operating Input Voltage (Notes 3 and 4) | $V_{IN}$ | 1.5 to 6.0 | V |
| Operating Junction Temperature Range | $T_J$ | −40 to 125 | °C |
| Operating Ambient Temperature Range | $T_A$ | −40 to 85 | °C |

1. Minimum $V_{IN}$ = ($V_{OUT}$ + $V_{DO}$) or 1.5 V, whichever is higher.
2. Soldered on FR4 copper area, please refer to Applications Section for Safe Operating Area.
3. Minimum $V_{IN}$ = ($V_{OUT}$ + $V_{DO}$) or 1.5 V, whichever is higher.
4. Refer to Electrical Characteristics and Application Information for Safe Operating Area.

**Table 6. ELECTRICAL CHARACTERISTICS** $V_{IN}$ = ($V_{OUT}$ + 1 V), $V_{EN}$ = $V_{IN}$, $I_{OUT}$ = 1 mA, $C_{IN}$ = 10 μF, $C_{OUT}$ = 10 μF, for typical values $T_J$ = 25°C, for Min/Max values $T_J$ = −40°C to 125°C; unless otherwise noted. (Note 5)

| Parameter | Test Conditions | | Symbol | Min | Typ | Max | Unit |
|---|---|---|---|---|---|---|---|
| Output voltage (Adjustable Version) | $V_{IN}$ = 1.75 V to 6.0 V<br>$I_{OUT}$ = 100 μA to 1 A | | $V_{OUT}$ | 1.231<br>(−1.5%) | 1.250 | 1.269<br>(+1.5%) | V |
| Output voltage (Fixed Version) | $V_{IN}$ = ($V_{OUT}$ + 1 V) to 6.0 V<br>$I_{OUT}$ = 100 μA to 1 A | $V_{OUT}$ = 1.5 V<br>$V_{OUT}$ = 1.8 V<br>$V_{OUT}$ = 2.5 V<br>$V_{OUT}$ = 3.3 V<br>$V_{OUT}$ = 5.0 V | $V_{OUT}$ | 1.470<br>1.764<br>2.450<br>3.234<br>4.900<br>(−2%) | 1.5<br>1.8<br>2.5<br>3.3<br>5.0 | 1.530<br>1.836<br>2.550<br>3.366<br>5.100<br>(+2%) | V |
| Line regulation | $V_{IN}$ = ($V_{OUT}$ + 1 V) to 6.0 V | | $Reg_{LINE}$ | – | 3.2 | 8 | mV |
| Load regulation | $I_{OUT}$ = 100 μA to 1 A | $V_{OUT}$ = 1.5 V<br>$V_{OUT}$ = 1.8 V<br>$V_{OUT}$ = 2.5 V<br>$V_{OUT}$ = 3.3 V<br>$V_{OUT}$ = 5.0 V | $Reg_{LOAD}$ | –<br>–<br>–<br>–<br>– | 10<br>10<br>10<br>10<br>10 | 30<br>30<br>35<br>35<br>40 | mV |
| Dropout voltage (Adjustable Version, Note 6) | $V_{DO}$ = $V_{IN}$ − $V_{OUT}$<br>$V_{OUT}$ = 1.25 V<br>$I_{OUT}$ = 1 A | | $V_{DO}$ | – | 450 | 470 | mV |
| Dropout voltage (Fixed Version, Note 9) | $I_{OUT}$ = 1 A | $V_{OUT}$ = 1.5 V<br>$V_{OUT}$ = 1.8 V<br>$V_{OUT}$ = 2.5 V<br>$V_{OUT}$ = 3.3 V<br>$V_{OUT}$ = 5.0 V | $V_{DO}$ | –<br>–<br>–<br>–<br>– | 290<br>240<br>190<br>180<br>120 | 410<br>380<br>300<br>250<br>210 | mV |
| Ground current | $V_{IN}$ = $V_{OUT}$ + 1 V,<br>$V_{OUT}$ = 1.5 V, 1.8 V,<br>2.5 V, 3.3 V | $I_{OUT}$ = 1 A<br>$I_{OUT}$ = 10 μA<br>$I_{OUT}$ = 100 μA | $I_{GND}$ | –<br>–<br>– | 145<br>145<br>145 | 200<br>200<br>200 | μA |
| | $V_{IN}$ = $V_{OUT}$ + 1 V,<br>$V_{OUT}$ = 5.0 V | $I_{OUT}$ = 1 A<br>$I_{OUT}$ = 10 μA<br>$I_{OUT}$ = 100 μA | | –<br>–<br>– | 145<br>145<br>145 | 240<br>240<br>240 | μA |
| Disable current (NCP692 only, Notes 5 and 7) | $V_{EN}$ < 0.4 V | | $I_{DIS}$ | – | 0.1 | 1 | μA |
| Output Current Limit | $V_{IN}$ = $V_{OUT}$ + 1 V, $V_{OUT}$ = 85% $V_{OUT_NOM}$ | | $I_{LIM}$ | 1.1 | 1.6 | 2.4 | A |
| Short Circuit Current | $V_{OUT}$ = 0 V | | $I_{SC}$ | 1.2 | – | – | A |
| Enable High Level Threshold<br>Enable Low Level Threshold<br>(NCP691 and NCP692) | $V_{EN}$ increasing from low to high logic level<br>$V_{EN}$ decreasing from high to low logic level | | $V_{EN_HI}$<br>$V_{EN_LO}$ | 0.9<br>– | –<br>– | –<br>0.4 | V |
| Enable Input Current (Enable Active Low) (NCP691 only, Note 8) | $V_{EN}$ = 0.9 V to $V_{IN}$ | | $I_{EN_HI}$ | – | 0.01 | 250 | nA |
| Enable Input Current (NCP692 only, Note 8) | $V_{EN}$ = 0 V | | $I_{EN_LO}$ | – | 0.01 | 100 | nA |
| Feedback Current | $V_{FB}$ = 1.25 (Adjustable version only) | | $I_{IFB}$ | – | 210 | 320 | nA |
| Turn−on Time (Note 8) | $V_{IN}$ = 0 V to ($V_{OUT}$ +1 V) or 1.75 V<br>$V_{OUT}$ = 0 V to 90% $V_{OUT_NOM}$ | | $t_{ON}$ | – | 50 | – | μs |

5. Performance guaranteed over the indicated operating temperature range by design and/or characterization tested at $T_J$ = $T_A$ = 25°C. Low duty cycle pulse techniques are used during testing to maintain the junction temperature as close to ambient as possible.
6. Maximum dropout voltage is limited by minimum input voltage. $V_{IN}$ = 1.7 V recommended for guaranteed operation at maximum output current.
7. Refer to the Applications Information Section.
8. Values based on design and/or characterization.
9. Dropout voltage is defined as the differential voltage between $V_{OUT}$ and $V_{IN}$, when $V_{OUT}$ drops 100 mV below its nominal value.

**Table 6. ELECTRICAL CHARACTERISTICS** $V_{IN}$ = ($V_{OUT}$ + 1 V), $V_{EN}$ = $V_{IN}$, $I_{OUT}$ = 1 mA, $C_{IN}$ = 10 μF, $C_{OUT}$ = 10 μF, for typical values $T_J$ = 25°C, for Min/Max values $T_J$ = −40°C to 125°C; unless otherwise noted. (Note 5) (continued)

| Parameter | Test Conditions | | Symbol | Min | Typ | Max | Unit |
|---|---|---|---|---|---|---|---|
| Power supply ripple rejection (Note 8) | $V_{OUT}$ = 1.25 V<br>$V_{IN}$ = $V_{OUT}$ + 1 V, with<br>$V_{PP}$ = 0.5 V, $C_{OUT}$ = 1 μF | f = 120 Hz<br>f = 1 kHz<br>f = 10 kHz | PSRR | –<br>–<br>– | 62<br>55<br>40 | –<br>–<br>– | dB |
| Output noise voltage (Note 8) | BW = 200 Hz to 100 kHz, $C_{IN}$ = 1 μF, $C_{OUT}$ = 10 μF, $T_A$ = 25°C | | $V_N$ | – | 50 | – | $\mu V_{rms}$ |
| Thermal Shutdown Temperature (Note 8) | | | $T_{SD}$ | – | 175 | – | °C |
| Thermal Shutdown Hysteresis (Note 8) | | | $T_{SH}$ | – | 10 | – | °C |

5. Performance guaranteed over the indicated operating temperature range by design and/or characterization tested at $T_J$ = $T_A$ = 25°C. Low duty cycle pulse techniques are used during testing to maintain the junction temperature as close to ambient as possible.
6. Maximum dropout voltage is limited by minimum input voltage. $V_{IN}$ = 1.7 V recommended for guaranteed operation at maximum output current.
7. Refer to the Applications Information Section.
8. Values based on design and/or characterization.
9. Dropout voltage is defined as the differential voltage between $V_{OUT}$ and $V_{IN}$, when $V_{OUT}$ drops 100 mV below its nominal value.

## TYPICAL CHARACTERISTICS

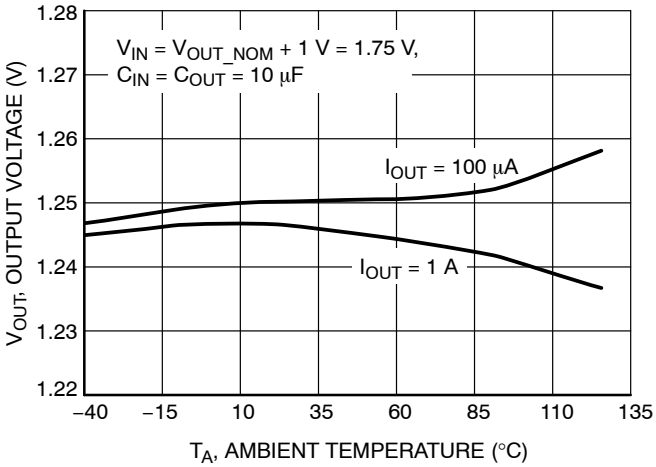



**Figure 11. Output Voltage vs. Temperature ($V_{out}$ = 1.25 V)**

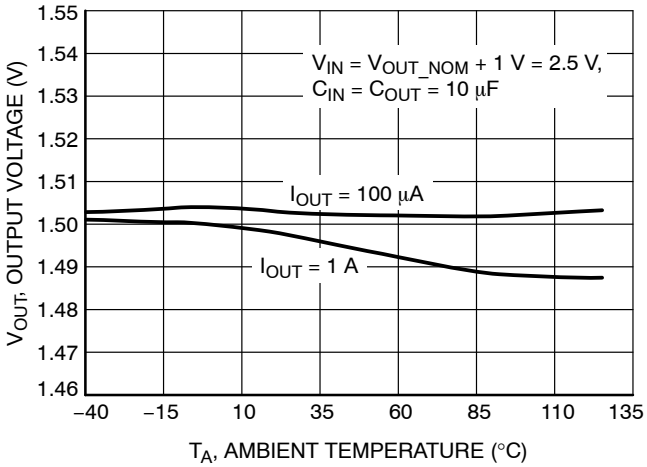



**Figure 12. Output Voltage vs. Temperature ($V_{out}$ = 1.5 V)**

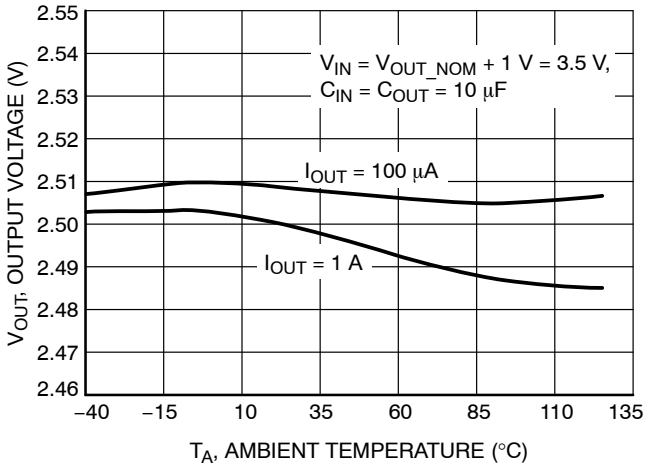



**Figure 13. Output Voltage vs. Temperature ($V_{out}$ = 2.5 V)**

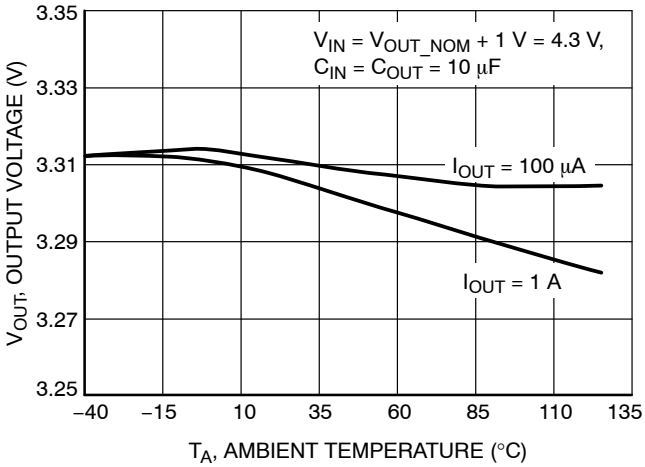



**Figure 14. Output Voltage vs. Temperature ($V_{out}$ = 3.3 V)**

## TYPICAL CHARACTERISTICS

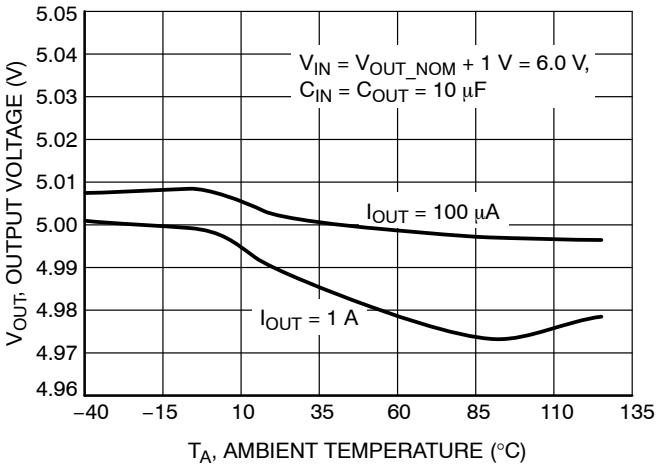



**Figure 15. Output Voltage vs. Temperature ($V_{out}$ = 5.0 V)**

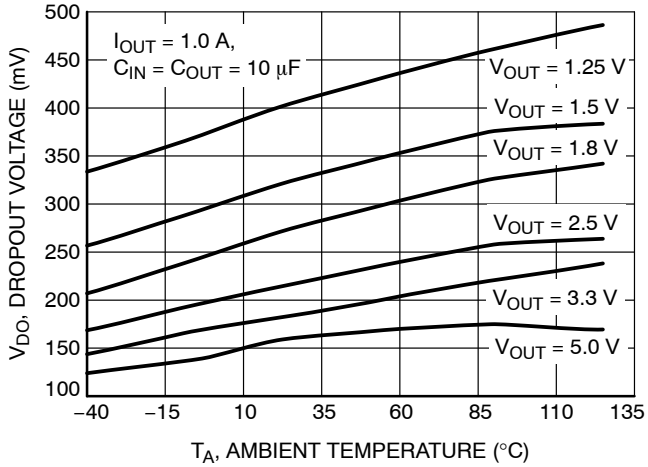



**Figure 16. Dropout Voltage vs. Temperature ($V_{out}$ = 1.25 V, 1.5 V, 1.8 V, 2.5 V, 3.3 V, 5.0 V)**

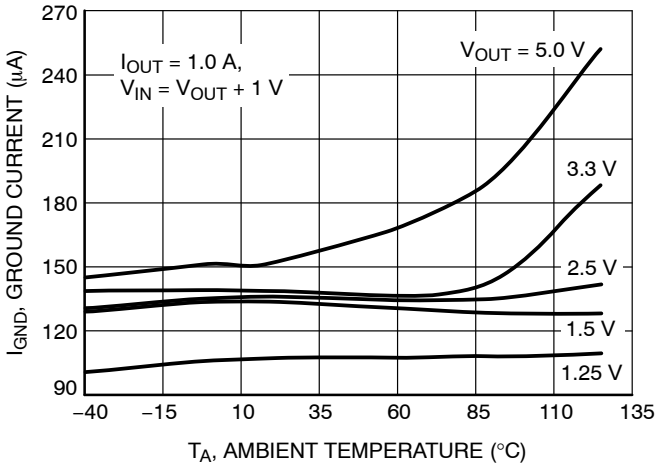



**Figure 17. Ground Current vs. Temperature ($V_{out}$ = 1.25 V, 1.5 V, 2.5 V, 3.3 V, 5.0 V)**

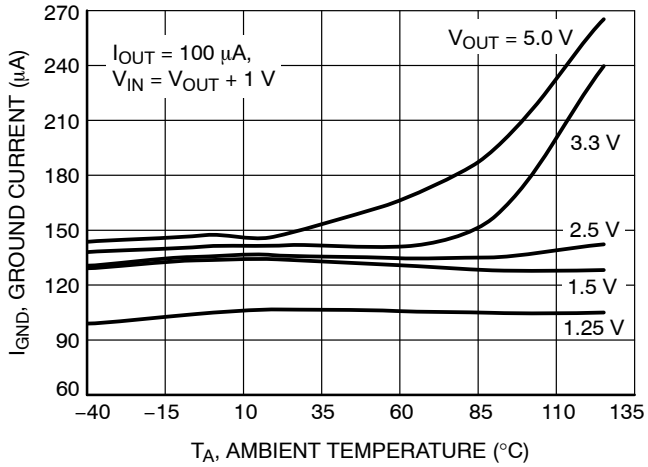



**Figure 18. Ground Current vs. Temperature ($V_{out}$ = 1.25 V, 1.5 V, 2.5 V, 3.3 V, 5.0 V)**

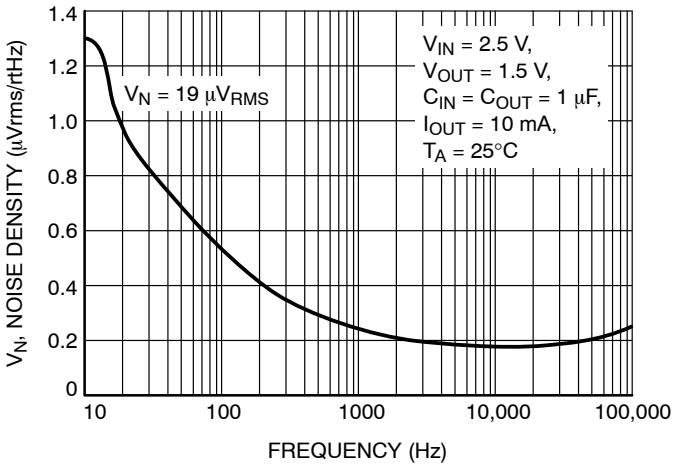



**Figure 19. Noise Density vs. Frequency ($V_{out}$ = 1.5 V)**

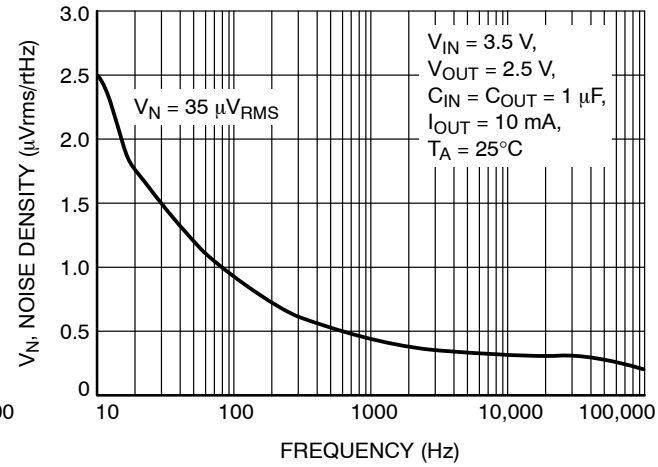



**Figure 20. Noise Density vs. Frequency ($V_{out}$ = 2.5 V)**

## TYPICAL CHARACTERISTICS

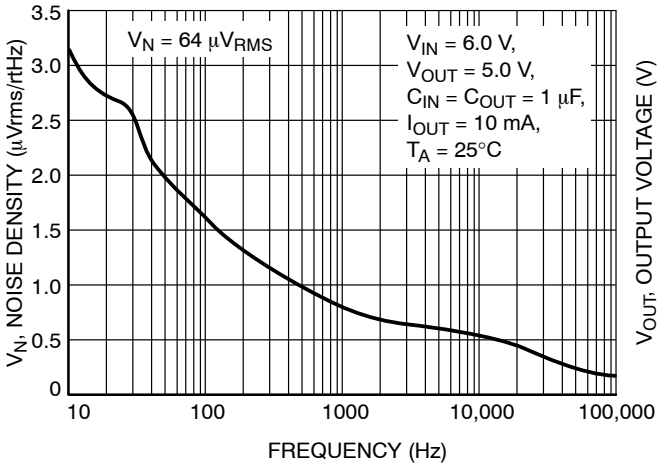



**Figure 21. Noise Density vs. Frequency ($V_{out}$ = 5.0 V)**

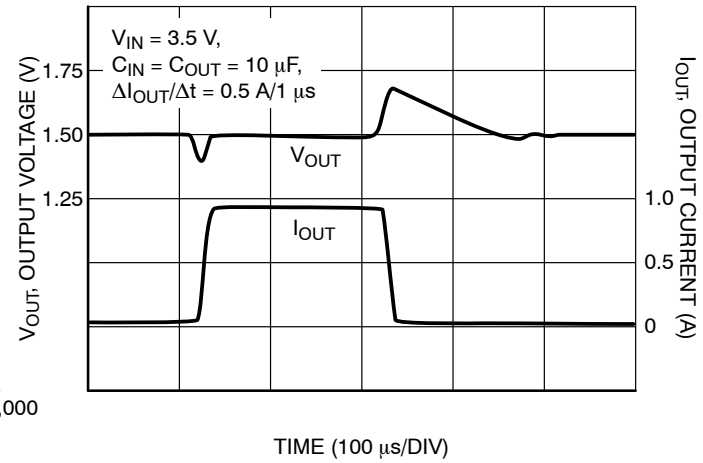



**Figure 22. Load Transient ($V_{out}$ = 1.5 V)**

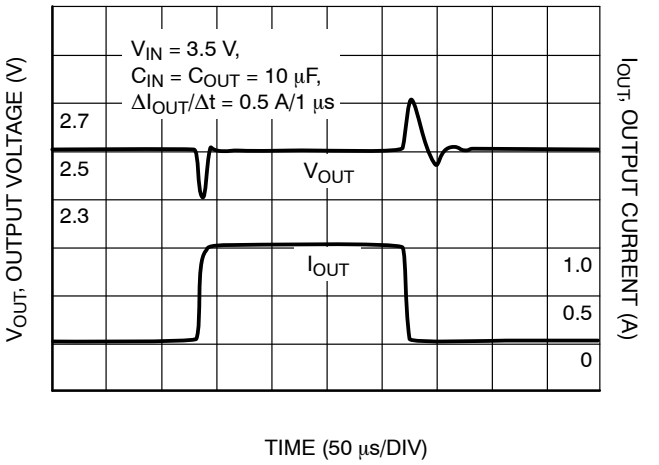



**Figure 23. Load Transient ($V_{out}$ = 2.5 V)**

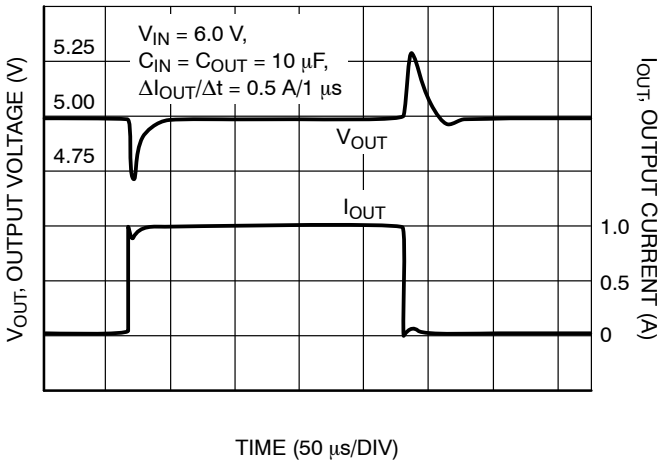



**Figure 24. Load Transient ($V_{out}$ = 5.0 V)**

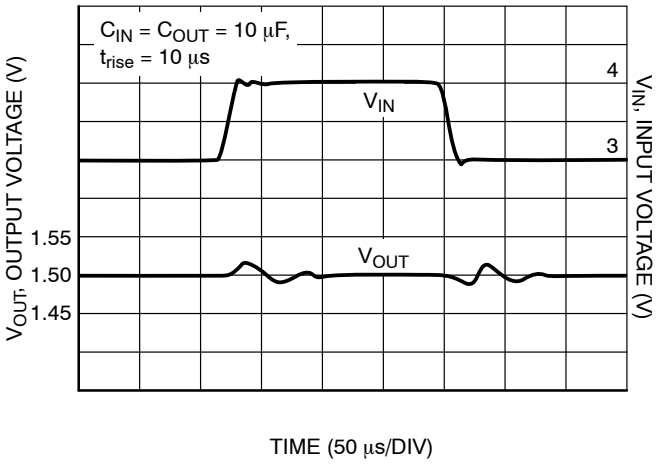



**Figure 25. Line Transient ($V_{out}$ = 1.5 V)**

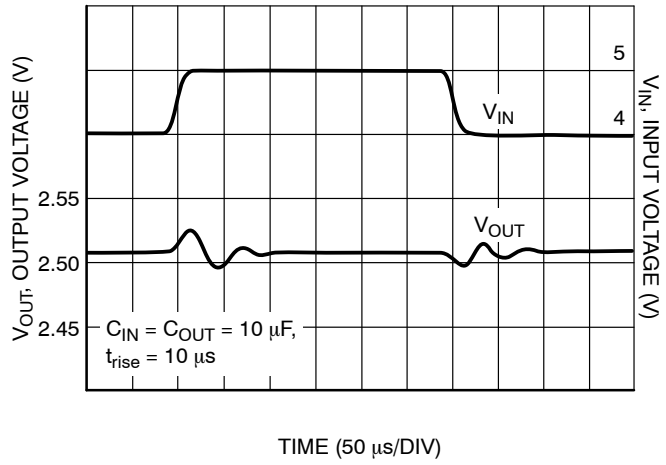



**Figure 26. Line Transient ($V_{out}$ = 2.5 V)**

## TYPICAL CHARACTERISTICS

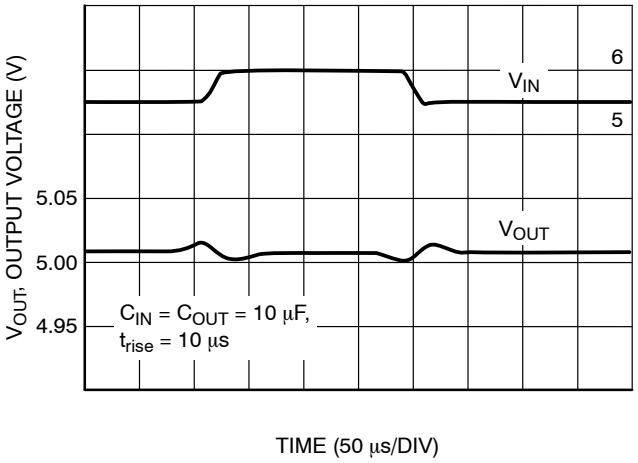



**Figure 27. Line Transient ($V_{out}$ = 5.0 V)**

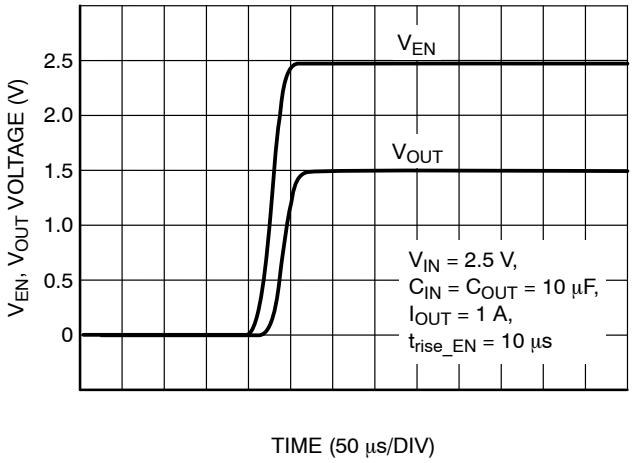



**Figure 28. Start−Up Transient ($V_{out}$ = 1.5 V)**

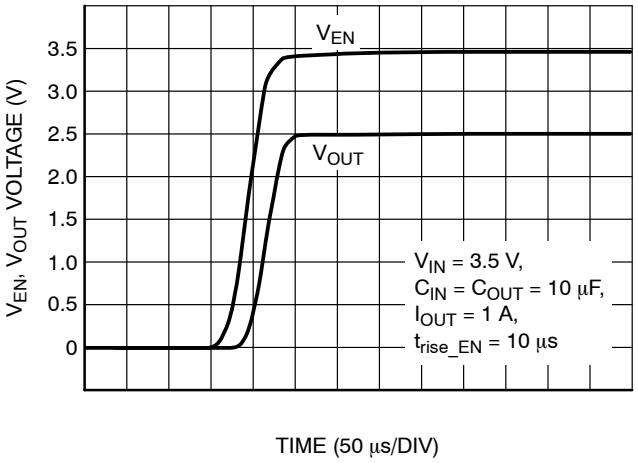



**Figure 29. Start−Up Transient ($V_{out}$ = 2.5 V)**

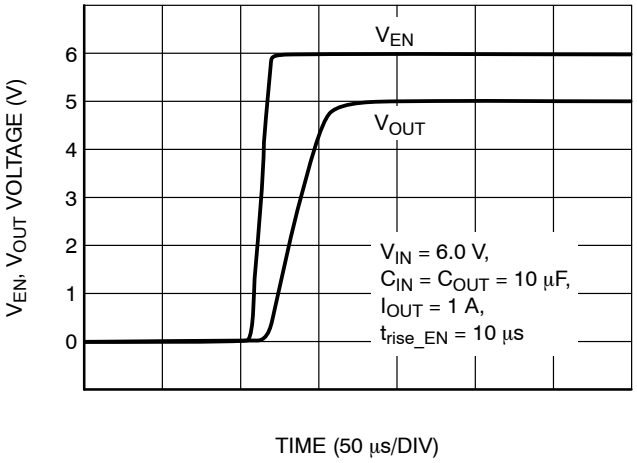



**Figure 30. Start−Up Transient ($V_{out}$ = 5.0 V)**

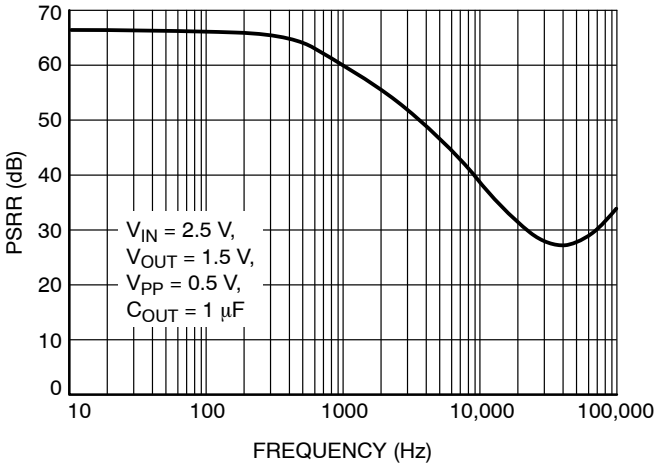



**Figure 31. PSRR vs. Frequency ($V_{out}$ = 1.5 V)**

## DEFINITIONS

### Load Regulation

The change in output voltage for a change in output load current at a constant temperature.

### Dropout Voltage

The input/output differential at which the regulator output no longer maintains regulation against further reductions in input voltage. Measured, when the output drops 100 mV below its nominal value. The junction temperature, load current, and minimum input supply requirements affect the dropout level.

### Output Noise Voltage

This is the integrated value of the output noise over a specified frequency range. Input voltage and output load current are kept constant during the measurement. Results are expressed in $\mu V_{rms}$ or nV $\sqrt{Hz}$.

### Ground Current

Ground Current is the current that flows through the ground pin when the regulator operates without a load on its output ($I_{GND}$). This consists of internal IC operation, bias, etc. It is actually the difference between the input current (measured through the LDO input pin) and the output load current. If the regulator has an input pin that reduces its internal bias and shuts off the output (enable/disable function), this term is called the disable current ($I_{DIS}$.)

### Line Regulation

The change in output voltage for a change in input voltage. The measurement is made under conditions of low dissipation or by using pulse techniques such that the average junction temperature is not significantly affected.

### Line Transient Response

Typical output voltage overshoot and undershoot response when the input voltage is excited with a given slope.

### Load Transient Response

Typical output voltage overshoot and undershoot response when the output current is excited with a given slope between no−load and full−load conditions.

### Thermal Protection

Internal thermal shutdown circuitry is provided to protect the integrated circuit in the event that the maximum junction temperature is exceeded. When activated at typically 175°C, the regulator turns off. This feature is provided to prevent failures from accidental overheating.

### Maximum Package Power Dissipation

The power dissipation level at which the junction temperature reaches its maximum operating value.

## APPLICATIONS INFORMATION

The NCP690 regulator is self−protected with internal thermal shutdown and internal current limit. Typical application circuit is shown in Figure 1.

### Input Decoupling ($C_{IN}$)

A ceramic 10 μF capacitor is recommended and should be connected close to the NCP690 package. Higher capacitance and lower ESR will improve the overall line transient response.

### Output Decoupling ($C_{OUT}$)

The NCP690 does not require a minimum Equivalent Series Resistance (ESR) for the output capacitor. The minimum output decoupling capacitor required for stability is 1 μF. In order to improve the load transient response and start up performance 10 μF capacitor is recommended. The regulator is stable with ceramic chip as well as tantalum capacitors. Larger values improve noise rejection and load transient response.

### No−Load Regulation Considerations

The required minimum 100 μA load current is assured by the internal resistor divider network.

The NCP690 contain an overshoot clamp circuit to improve transient response during a load current step release. When output voltage exceeds the nominal by approximately 20 mV, this circuit becomes active and clamps the output from further voltage increase. Tying the ENABLE pin to $V_{IN}$ will ensure that the part is active whenever the supply voltage is present,

### Noise Decoupling

The NCP690 is a low noise regulator and needs no external noise reduction capacitor. Unlike other low noise regulators which require an external capacitor and have slow startup times, the NCP690 operates without a noise reduction capacitor, has a typical 50 μs turn−on time and achieves a 50 $\mu V_{rms}$ overall noise level between 10 Hz and 100 kHz.

### Enable Operation

The enable pin will turn the regulator on or off. The threshold limits are covered in the electrical characteristics table in this data sheet. The turn−on/turn−off transient voltage being supplied to the enable pin should exceed a slew rate of 10 mV/μs to ensure correct operation. If the enable function is not to be used then the pin should be connected to $V_{IN}$.

### Adjustable Operation

The output voltage can be adjusted from 1 to 4 times the typical 1.250 V regulation voltage by the use of resistor

divider network as shown on Figure 4. The output voltage and resistors should be chosen using Equations 1 and 2.

$$V_{OUT} = 1.250\left(1 + \frac{R_1}{R_2}\right) + (I_{ADJ} \cdot R_1) \quad \text{(eq. 1)}$$

$$R_2 \cong R_1 \frac{1}{\frac{V_{OUT}}{1.25} - 1} \quad \text{(eq. 2)}$$

Input bias current $I_{ADJ}$ is typically less than 210 nA. Choose $R_1$ arbitrarily to minimize errors due to the bias current and to minimize noise contribution to the output voltage. Use Equation 2 to find the required value for $R_2$. If an output voltage of 1.25 V is desired, the adjustable pin should be connected directly to the output pin.

### Thermal Characteristics

As power dissipated in the NCP690 increases, it might become necessary to provide some thermal relief. The maximum power dissipation supported by the device is dependent upon board design and layout. Mounting pad configuration on the PCB, the board material, and the ambient temperature affect the rate of junction temperature rise for the part. When the NCP690 has good thermal conductivity through the PCB, the junction temperature will be relatively low with high power applications. The maximum dissipation the NCP690 can handle is given by:

$$P_{D(MAX)} = \frac{[T_{J(MAX)} - T_A]}{R_{\theta JA}} \quad \text{(eq. 3)}$$

Since $T_J$ is not recommended to exceed 125°C ($T_{J(MAX)}$), then the NCP690 can dissipate up to 1 W when the ambient temperature ($T_A$) is 25°C.

The power dissipated by the NCP690 can be calculated from the following equations:

$$P_D \approx V_{IN}(I_{GND}@I_{OUT}) + I_{OUT}(V_{IN} - V_{OUT}) \quad \text{(eq. 4)}$$

or

$$V_{IN(MAX)} \approx \frac{P_{D(MAX)} + (V_{OUT} \times I_{OUT})}{I_{OUT} + I_{GND}} \quad \text{(eq. 5)}$$

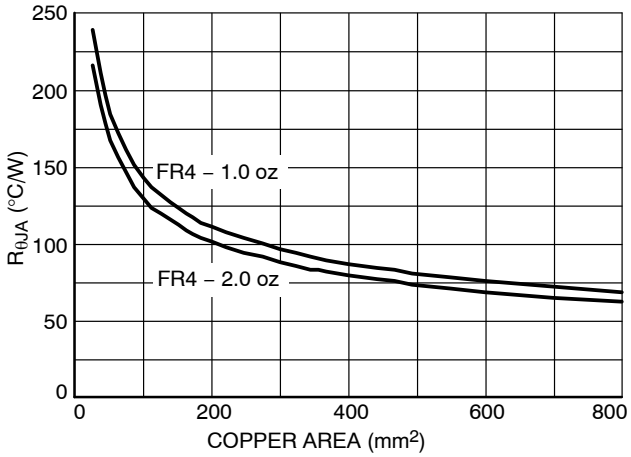



**Figure 32. Thermal Resistance vs. Copper Area**

### Hints

VIN and GND printed circuit board traces should be as wide as possible. When the impedance of these traces is high, there is a chance to pick up noise or cause the regulator to malfunction. Place external components, especially the output capacitor, as close as possible to the NCP690, and make traces as short as possible.

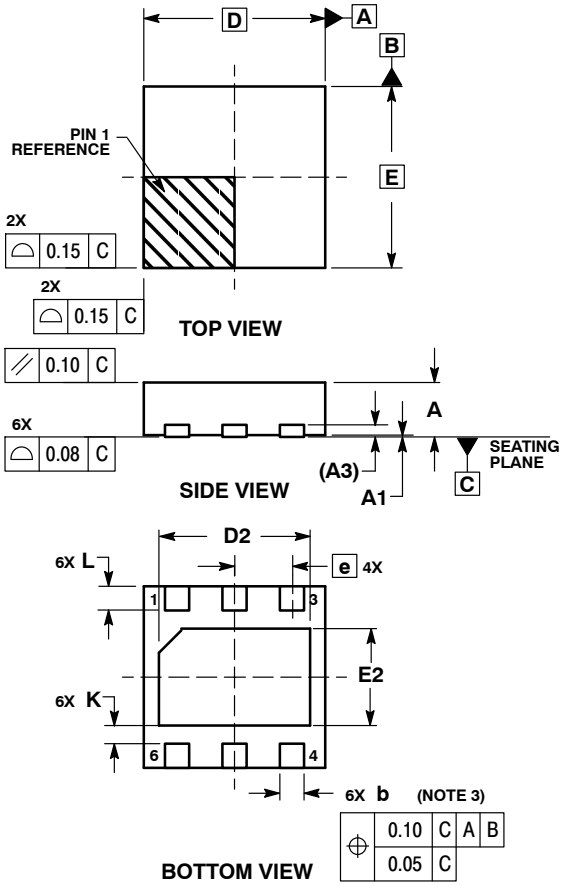
## DEVICE ORDERING INFORMATION

| Device | Nominal Output Voltage | Marking | Package | Shipping† |
|---|---|---|---|---|
| NCP690MN15T2G | 1.5 V | 69015 | DFN6 (Pb-Free) | 3000 / Tape & Reel |
| NCP690MN18T2G | 1.8 V | 69018 | DFN6 (Pb-Free) | 3000 / Tape & Reel |
| NCP690MN25T2G | 2.5 V | 69025 | DFN6 (Pb-Free) | 3000 / Tape & Reel |
| NCP690MN33T2G | 3.3 V | 69033 | DFN6 (Pb-Free) | 3000 / Tape & Reel |
| NCP690MN50T2G | 5.0 V | 69050 | DFN6 (Pb-Free) | 3000 / Tape & Reel |
| NCP690MNADJT2G | ADJ | 690AD | DFN6 (Pb-Free) | 3000 / Tape & Reel |
| NCP691MN15T2G | 1.5 V | 69115 | DFN6 (Pb-Free) | 3000 / Tape & Reel |
| NCP691MN18T2G | 1.8 V | 69118 | DFN6 (Pb-Free) | 3000 / Tape & Reel |
| NCP691MN25T2G | 2.5 V | 69125 | DFN6 (Pb-Free) | 3000 / Tape & Reel |
| NCP691MN33T2G | 3.3 V | 69133 | DFN6 (Pb-Free) | 3000 / Tape & Reel |
| NCP691MN50T2G | 5.0 V | 69150 | DFN6 (Pb-Free) | 3000 / Tape & Reel |
| NCP691MNADJT2G | ADJ | 691AD | DFN6 (Pb-Free) | 3000 / Tape & Reel |
| NCP692MN15T2G | 1.5 V | 69215 | DFN6 (Pb-Free) | 3000 / Tape & Reel |
| NCP692MN18T2G | 1.8 V | 69218 | DFN6 (Pb-Free) | 3000 / Tape & Reel |
| NCP692MN25T2G | 2.5 V | 69225 | DFN6 (Pb-Free) | 3000 / Tape & Reel |
| NCP692MN33T2G | 3.3 V | 69233 | DFN6 (Pb-Free) | 3000 / Tape & Reel |
| NCP692MN50T2G | 5.0 V | 69250 | DFN6 (Pb-Free) | 3000 / Tape & Reel |
| NCP692MNADJT2G | ADJ | 692AD | DFN6 (Pb-Free) | 3000 / Tape & Reel |

†For information on tape and reel specifications, including part orientation and tape sizes, please refer to our Tape and Reel Packaging Specifications Brochure, BRD8011/D.

## PACKAGE DIMENSIONS

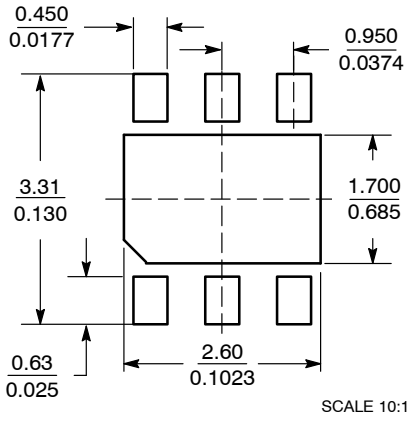
**DFN6 3*3 MM, 0.95 PITCH**
CASE 506AH−01
ISSUE O




NOTES:
1. DIMENSIONS AND TOLERANCING PER ASME Y14.5M, 1994.
2. CONTROLLING DIMENSION: MILLIMETERS.
3. DIMESNION b APPLIES TO PLATED TERMINAL AND IS MEASURED BETWEEN 0.25 AND 0.30 MM FROM TERMINAL.
4. COPLANARITY APPLIES TO THE EXPOSED PAD AS WELL AS THE TERMINALS.

| | MILLIMETERS | | |
|---|---|---|---|
| DIM | MIN | NOM | MAX |
| A | 0.80 | 0.90 | 1.00 |
| A1 | 0.00 | 0.03 | 0.05 |
| A3 | 0.20 REF | | |
| b | 0.35 | 0.40 | 0.45 |
| D | 3.00 BSC | | |
| D2 | 2.40 | 2.50 | 2.60 |
| E | 3.00 BSC | | |
| E2 | 1.50 | 1.60 | 1.70 |
| e | 0.95 BSC | | |
| K | 0.21 | −−− | −−− |
| L | 0.30 | 0.40 | 0.50 |

### SOLDERING FOOTPRINT*




*For additional information on our Pb−Free strategy and soldering details, please download the ON Semiconductor Soldering and Mounting Techniques Reference Manual, SOLDERRM/D.

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