



MELCHER™
The Power Partners.

M Series

50 Watt DC-DC and AC-DC Converters

Features

- Extremely wide operating input voltage ranges from 8 to 385 VDC and 85 to 264 VAC, 47 to 440 Hz
- RoHS lead-free-solder and lead-solder-exempted products available
- Class I equipment
- Input over- and undervoltage lockout
- 1, 2, or 3 individually isolated outputs up to 72 V
- Outputs: SELV, no load, overload, short-circuit proof, rectangular current limiting characteristic
- Adjustable output voltages with remote on/off
- Immunity according to IEC/EN 61000-4-2, -3, -4, -5, -6
- Emissions according to EN 55011/55022
- According to EN 45545 and NF-F-16 (Version V107 or later)
- All PCBs boards coated by protective lacquer
- Very high reliability
- Battery charger models available

Safety-approved to the latest edition of IEC/EN 60950-1 and UL/CSA 60950-1

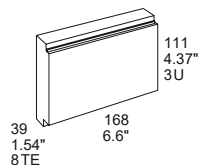


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Description

The M Series of DC-DC and AC-DC converters represents a broad and flexible range of power supplies for use in advanced industrial electronic systems. Features include high efficiency, reliability, low output voltage noise and excellent dynamic response to load/line changes due to individual regulation of each output.

The converter inputs are protected against surges and transients occurring at the source lines. An input over- and undervoltage lockout circuit disables the outputs, if the input voltage is outside the specified range. An inrush current limitation prevents circuit breakers and fuses from tripping at switch-on.

All outputs are open- and short-circuit proof, and are protected against overvoltages by means of built-in suppressor diodes. The outputs can be inhibited by a logic signal applied to the connector (pin 2). If the inhibit function is not used, pin 2 should be connected to pin 23 to enable the outputs.

LED indicators display the status of the converter and allow visual monitoring of the system at any time.

Full input to output, input to case, output to case, and output to output isolation is provided.

The case design allows operation at nominal load up to 71 °C in a free-air ambient temperature. If forced cooling is provided, the ambient temperature may exceed 71 °C but the case temperature should remain below 95 °C under all conditions.

A temperature sensor generates an inhibit signal, which disables the outputs, when the case temperature TC exceeds the limit. The outputs automatically recover, when the temperature drops below the limit.

Various options are available to adapt the converters to individual applications.

The converters may either be plugged into a 19" rack system according to IEC 60927-3 or be mounted onto a chassis or a plate.

Model Selection

Non-standard input/output configurations or special custom adaptations are available on request. Table 1 provides an overview of the basic input and output configurations. More than 1000 different model types have been manufactured with different input/output configurations and customized specialties. Please consult the company for additional model types.

Table 1a: Standard models AM, BM, FM

Output 1		Output 2		Output 3		Operating Input Voltage Range and Efficiency ¹					Options	
$V_{o\ nom}$ [VDC]	$I_{o\ nom}$ [A]	$V_{o\ nom}$ [VDC]	$I_{o\ nom}$ [A]	$V_{o\ nom}$ [VDC]	$I_{o\ nom}$ [A]	$V_{i\ min} - V_{i\ max}$ 8 - 35 VDC	η_{min} [%]	$V_{i\ min} - V_{i\ max}$ 14 - 70 VDC	η_{min} [%]	$V_{i\ min} - V_{i\ max}$ 20 - 100 VDC	η_{min} [%]	
5.1	8.0	-	-	-	-	AM1001-9RG	72	BM1001-9RG	74	FM1001-9RG	74	-7, P, D0 - D9, V0 - V3 ² , F, A, K, H, non-G
12	4.0	-	-	-	-	AM1301-9RG	79	BM1301-9RG	80	FM1301-9RG	80	
15	3.4	-	-	-	-	AM1501-9RG	79	BM1501-9RG	81	FM1501-9RG	81	
24	2.0	-	-	-	-	AM1601-9RG	81	BM1601-9RG	83	FM1601-9RG	82	
48	1.0	-	-	-	-	AM1901-9RG	81	BM1901-9RG	83	FM1901-9RG	83	
5.1	4.0	5.1	4.0	-	-	AM2001-9G	-	BM2001-9G	-	-	-	-7, P, D0 - D9, A, K, H, non-G
5.1	4.0	24	1.0	-	-	AM2060-9G	-	BM2060-9G	-	-	-	
12	2.0	12	2.0	-	-	AM2320-9G	77	BM2320-9G	79	FM2320-9G	80	
15	1.7	15	1.7	-	-	AM2540-9G	78	BM2540-9G	80	FM2540-9G	79	
24	1.0	24	1.0	-	-	AM2660-9G	-	BM2660-9G	-	-	-	
5.1	5.0	12	0.7	12	0.7	AM3020-9G	75	BM3020-9G	76	FM3020-9G	76	-7, P, D0 - D9, A, K, H, non-G
5.1	5.0	15	0.6	15	0.6	AM3040-9G	75	BM3040-9G	76	FM3040-9G	76	
5.1	5.0	24	0.35	24	0.35	AM3060-9G	-	BM3060-9G	-	-	-	

¹ Min. efficiency at $V_{i\ nom}$ and $I_{o\ nom}$. Typ. values are approx. 2% better.

² Option V0, V2, V3 available only for output 1 = 5.1 V (excludes option D)

NFND: Not for new designs.

Preferred: Preferred for new designs

Table 1b: Models CM, DM, LM

Output 1		Output 2		Output 3		Operating Input Voltage Range and Efficiency ¹					Options	
$V_{o\ nom}$ [VDC]	$I_{o\ nom}$ [A]	$V_{o\ nom}$ [VDC]	$I_{o\ nom}$ [A]	$V_{o\ nom}$ [VDC]	$I_{o\ nom}$ [A]	$V_{i\ min} - V_{i\ max}$ 28 - 140 VDC	η_{min} [%]	$V_{i\ min} - V_{i\ max}$ 44 - 220 VDC	η_{min} [%]	$V_{i\ min} - V_{i\ max}$ 88 - 372 VDC 85 - 264 VAC ³	η_{min} [%]	
5.1	8.0	-	-	-	-	CM1001-9RG	74	DM1001-9RG	74	LM1001-9RG	74	-7, E ⁴ , P, D0 - D9, V0 - V3 ² , A, K, H, non-G
12	4.0	-	-	-	-	CM1301-9RG	80	DM1301-9RG	81	LM1301-9RG	80	
15	3.4	-	-	-	-	CM1501-9RG	82	DM1501-9RG	82	LM1501-9RG	79	
24	2.0	-	-	-	-	CM1601-9RG	82	DM1601-9RG	-	LM1601-9RG	82	
48	1.0	-	-	-	-	CM1901-9RG	82	DM1901-9RG	83	LM1901-9RG	82	
5.1	4.0	5.1	4.0	-	-	CM2001-9G	-	DM2001-9G	-	LM2001-9G	-	-7, E ⁴ , P, D0 - D9, A, K, H, non-G
5.1	4.0	24	1.0	-	-	CM2060-9G	-	DM2060-9G	-	LM2060-9G	-	
12	2.0	12	2.0	-	-	CM2320-9G	79	DM2320-9G	80	LM2320-9G	80	
15	1.7	15	1.7	-	-	CM2540-9G	80	DM2540-9G	80	LM2540-9G	79	
24	1.0	24	1.0	-	-	CM2660-9G	-	DM2660-9G	-	LM2660-9G	-	
5.1	5.0	12	0.7	12	0.7	CM3020-9G	76	DM3020-9G	77	LM3020-9G	74	-7, E ⁴ , P, D0 - D9, A, K, H, non-G
5.1	5.0	15	0.6	15	0.6	CM3040-9G	76	DM3040-9G	76	LM3040-9G	73	
5.1	5.0	24	0.35	24	0.35	CM3060-9G	76	DM3060-9G	76	LM3060-9G	-	

Table 1c: EM and battery charger models

Output 1				Output 2		Output 3		Operating Input Voltage Range & Efficiency ¹				Options
$V_{o\ nom}^5$ [VDC]	$I_{o\ nom}$ [A]	$V_{o\ safe}^6$ [VDC]	$V_{o\ max}$ [VDC]	$V_{o\ nom}$ [VDC]	$I_{o\ nom}$ [A]	$V_{o\ nom}$ [VDC]	$I_{o\ nom}$ [A]	$V_{i\ min} - V_{i\ max}$ 67 - 385 VDC	η_{min} [%]	$V_{i\ min} - V_{i\ max}$ 88 - 372 VDC 85 - 264 VAC ³	η_{min} [%]	
Same $V_{o\ nom}$ and $I_{o\ nom}$ as DM models	-	-	-	-	-	-	-	EM1xxx-9RG	-	-	-	-7, E, D, A,
	-	-	-	Same as DM models	-	-	-	EM2xxx-9RG	-	-	-	non-G
	-	-	-	Same as DM models	Same as DM models	-	-	EM3xxx-9RG	-	-	-	
12	3.6	12.84	14.15 – 14.60	-	-	-	-	-	-	LM1781-9RD5G	79	-7, E, A, non-G
24	1.8	25.68	28.30 – 29.15	-	-	-	-	-	-	LM1782-9RD5G	81	
36	1.2	38.52	42.45 – 43.72	-	-	-	-	-	-	LM1783-9RD5G	82	
48	0.9	51.36	56.60 – 58.30	-	-	-	-	-	-	LM1784-9RD5G	81	
60	0.72	64.20	70.75 – 72.87	-	-	-	-	-	-	LM1785-9RD5G	81	

¹ Min. efficiency at $V_{i\ nom}$ and $I_{o\ nom}$. Typ. values are approx. 2% better.

² Option V0, V2, V3 available only for output 1 = 5.1 V (excludes option D)

³ Operating frequency range: 47 – 440 Hz; see *Safety and Installation Instructions* for > 60 Hz !

⁴ Option E only available for CM and LM models (not for DM)

⁵ $V_{o\ nom}$ for EM models

⁶ Setting voltage with open R-input (battery chargers)

NFND: Not for new designs.

Preferred for new designs

Part Number Description

			C	M	2	5	40	-9	E	P	D3	A	H	G
Operating input range V_i :	8 – 35 VDC	A												
	14 – 70 VDC	B												
	20 – 100 VDC	F												
	28 – 140 VDC	C												
	44 – 220 VDC	D												
	67 – 385 VDC	E												
	85 – 264 VAC, 88 – 372 VDC	L												
Series		M												
Number of outputs ⁴	1, 2, 3 ⁴													
Output 1, $V_{o1 \text{ nom}}$:	5.1 V	0, 1, 2												
	12 V	3												
	15 V	4, 5												
	24 V	6												
	other voltages	7, 8												
	48 V	9												
Single-output models (different specs.)		01 – 99												
Outputs 2, 3: $V_{o2 \text{ nom}}$, $V_{o3 \text{ nom}}$:	5.1 V	01 – 19												
	12 V	20 – 39												
	15 V	40 – 59												
	24 V	60 – 69												
	other voltages and specs.	21 – 99												
Ambient temperature range T_A :	–25 to 71 °C	-7												
	–40 to 71 °C	-9												
	customer-specific	-0, -5, -6, -8												
Auxiliary functions and options:														
Inrush current limitation (CM, EM, LM)		E												
Output voltage control input (single-output models)		R ²												
Potentiometers for adjustment of output voltages		P ²												
Save data signal (D0 – D9, to be specified)		D ¹												
ACFAIL signal (V0, V2, V3, to be specified)		V ¹												
Output voltage test sockets		A												
Increased electric strength test voltage		H												
Input fuse built-in (not accessible)		F ³												
Coding strip at the connector		K												
RoHS-compliant for all 6 substances		G												

¹ Option D excludes option V and vice versa

² Feature R is fitted to single-output models only. Option P excludes option R (and vice versa).

³ Only for FM1000

⁴ Models with 220 mm case length. Just add 6000 to the standard model number, e.g., DM3020-9AG → DM9020-9AG.

Note: The sequence of options must follow the order above. The part number description is descriptive only; it is not intended for creating part numbers.

Example: CM2540-9EPD3AHG: DC-DC converter, operating input voltage range 28 – 140 VDC, providing output 1 with 15 V/1.7 A and output 2 with 15 V/1.7 A; temperature range –40 to +71 °C, inrush current limitation, equipped with potentiometers, undervoltage monitor D3, test sockets, tested with higher voltage output to case, RoHS-compliant for all 6 substances.

Product Marking

Basic type designation, applicable approval marks, CE mark, warnings, pin designation, patents and company logo, identification of LEDs, test sockets, and potentiometer.

Specific type designation, input voltage range, nominal output voltages and currents, degree of protection, batch no., serial no., and data code including production site, modification status (version), and date of production.

Functional Description

The input voltage is fed via an input fuse, an input filter, a bridge rectifier, and an inrush current limiter to the input capacitor. This capacitor sources a single-transistor forward converter. Each output is powered by a separate secondary winding of the main transformer. The resultant voltages are rectified and their ripple smoothed by a power choke and an output filter. The main control circuit senses the main output voltage V_{o1} and generates, with respect to the maximum admissible output currents, the control signal for the primary switching transistor. This signal is transferred to the primary side by a coupling transformer.

The auxiliary output voltages V_{o2} and V_{o3} are individually regulated by means of secondary switching transistors. Each auxiliary output's current is sensed using a current transformer. If one of the outputs is driven into current limit, the other outputs will reduce their output voltages as well, because all output currents are controlled by the same main control circuit.

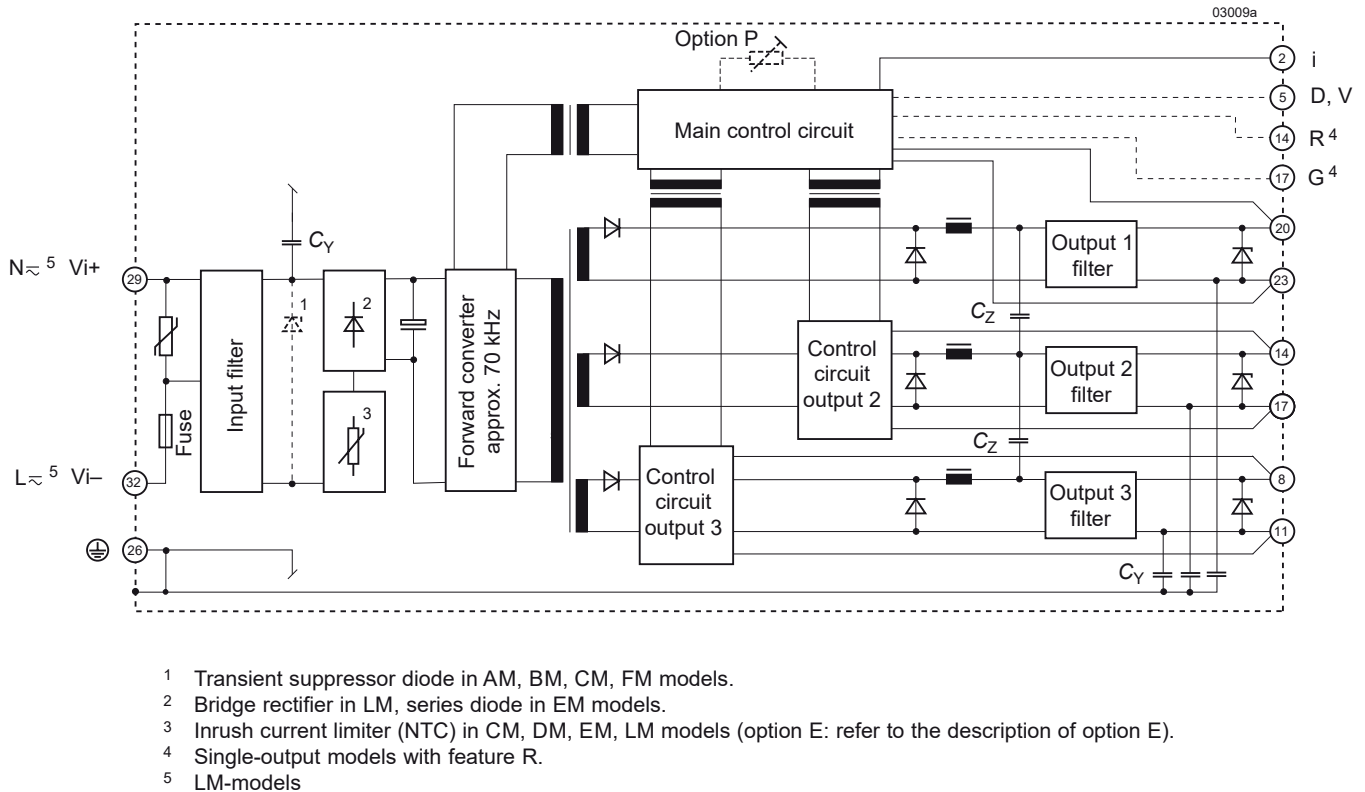


Fig. 1
Block diagram, triple-output models

Electrical Input Data

General conditions:

- $T_A = 25\text{ °C}$, unless T_C is specified.
- Connector pins 2 and 23 interconnected, R input not connected; with option P: $V_o = V_{o\text{ nom}}$

Table 2a: Input data

Model			AM			BM			FM			CM			Unit
Characteristics	Conditions		min	typ	max	min	typ	max	min	typ	max	min	typ	max	
V_i	Operating input voltage	$I_o = 0 - I_{o\text{ nom}}$	8		35	14		70	20		100	28		154 ⁷	VDC
$V_{i\text{ nom}}$	Nominal input voltage	$T_{C\text{ min}} - T_{C\text{ max}}$		15			30			50			60		
I_i	Input current	$V_{i\text{ nom}}, I_{o\text{ nom}}^2$		4.0			2.0			1.2			1.0		A
P_{i0}	No-load input power: – single-output models – double-output models – triple-output models	$V_{i\text{ nom}}$ $I_{o1,2,3} = 0$		1 7 6	1.5 9 9		1 7 6	1.5 9 9		1 7 6	1.5 9 9		1 7 6	1.5 9 9	W
P_{inh}	Idle input power	Inhibited $V_{i\text{ nom}}$		1	1.5		1	1.5		1	1.5		1	1.5	
I_{inrp}^6	Peak inrush current	$V_i = V_{i\text{ max}}$			400			500			400			170 ⁴	
t_{inrr}	Rise time	$R_s = 0\text{ }\Omega^3$		60			50			40			60		
t_{inrh}	Time to half-value	$T_C = 25\text{ °C}$		170			100			60			280		
R_i	Input resistance	$T_C = 25\text{ °C}$	87.5			140			250			824 ⁴			m Ω
C_i	Input capacitance		2600	4000		670	1100		370	600		370	600		μF
$V_{i\text{ abs}}$	Input voltage limits without any damage		0	40		0	80		0	120		0	160		VDC

Table 2b: Input data

Model			DM			EM			LM			Unit
Characteristics	Conditions		min	typ	max	min	typ	max	min	typ	max	
V_i	Operating input voltage	$I_o = 0 - I_{o\text{ nom}}$	-			-			85		264	VDC
$V_{i\text{ nom}}$	Nominal input voltage	$T_{C\text{ min}} - T_{C\text{ max}}$	44		220	67		385	88		372	
I_i	Input current	$V_{i\text{ nom}}, I_{o\text{ nom}}^2$		0.55			0.275			0.2		A
P_{i0}	No-load input power: – single-output models – double-output models – triple-output models	$V_{i\text{ nom}}$ $I_{o1,2,3} = 0$		1 7 6	1.5 9 9		1 7 6	1.5 9 9		1 7 6	1.5 9 9	W
P_{inh}	Idle input power	Inhibited $V_{i\text{ nom}}$		1	1.5		1	1.5		1	1.5	
I_{inrp}^6	Peak inrush current	$V_i = V_{i\text{ max}}$			110 ⁴			160 ⁴			60 ⁴	
t_{inrr}	Rise time	$R_s = 0\text{ }\Omega^3$		40			40			300		
t_{inrh}	Time to half-value	$T_C = 25\text{ °C}$		250			240			900		
R_i	Input resistance	$T_C = 25\text{ °C}$	2000 ⁴			2400 ⁴			6200 ⁴			m Ω
C_i	Input capacitance		140		270	140		270	140		270	μF
$V_{i\text{ abs}}$	Input voltage limits without any damage		0		400 ⁵	- 400		400	- 400		400	VDC
			-		-	-		-	0		284	

¹ In AC powered mode (LM models): Nominal input voltage range: 100 – 240 VAC, operating input frequency range: 47 – 440 Hz

² With multiple-output models, the same condition for each output applies.

³ R_s = source resistance.

⁴ Value for initial switch-on cycle.

⁵ 1 s max., duty cycle 1% max.

⁶ $I_{inrp} = V_i / (R_s + R_i)$; see *Inrush Current*.

⁷ 140 V continuously. CM models with version V106 or greater (or with suffix /131) withstand 154 V for 2 s.

Input Fuse

A fuse holder containing a slow-blow type fuse (size: 5 × 20 mm) is mounted in the back plate of the converter. The fuse protects the converter against severe defects. It may not fully protect it at input voltages exceeding 200 VDC. In applications, where the converters operate at DC source voltages above 200 VDC, an external fuse or a circuit breaker at system level should be installed.

The fuse and a VDR form together with the input filter an effective protection against high input transients.

Note: For applications, where the fuse should not be accessible; see *Option F*.

Table 3: Fuse types (slow-blow)

Series	Schurter type	Part number
AM1000 – 3000	SPT 10 A / 250 V	0001.2514
BM1000 – 3000	SPT 8 A / 250 V	0001.2513
FM1000 – 3000	SPT 5 A / 250 V	0001.2511
CM1000 – 3000	SPT 3.15 A / 250 V	0001.2509
DM1000 – 3000 EM1000 – 3000 LM1000 – 3000	SPT 2.5 A / 250 V	0001.2508

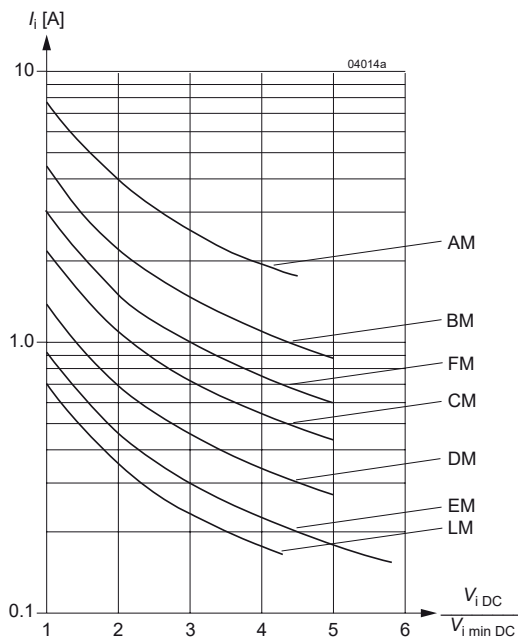


Fig. 2
Typical input current versus relative input voltage at nominal output load

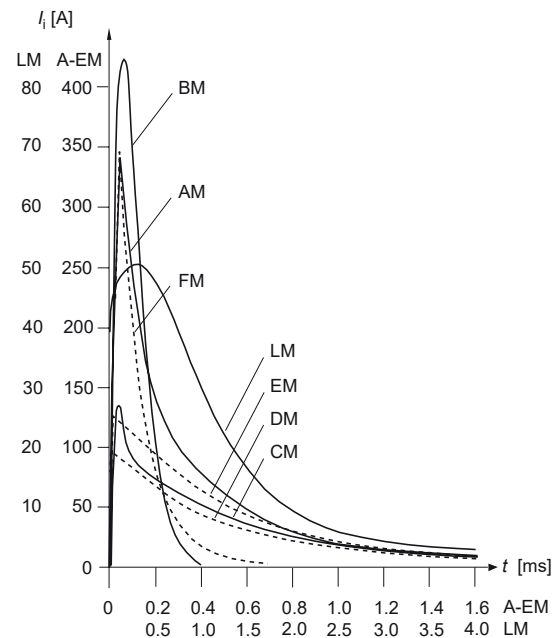


Fig. 3
Typical inrush current at initial switch-on. $V_{i\max}$ (DC) and nominal output load

Inrush Current

The CM, DM, EM, and LM (excluding FM) models incorporate an NTC resistor in the input line, which (during the initial switch-on cycle) limits the peak inrush current in order to prevent the connectors and external switching devices from damage. Subsequent switch-on cycles within a short interval will cause an increase of the peak inrush current due to the warming-up of the NTC resistor. Refer to *Option E* (only available for CM, EM, and LM).

Input Under-/Overvoltage Lockout

If the input voltage remains below $0.8 V_{i\min}$ or exceeds approx. $1.1 V_{i\max}$, an internally generated inhibit signal disables the output(s). When checking this function the absolute maximum input voltage rating $V_{i\text{abs}}$ must be carefully considered (see table *Input data*).

Note: When V_i is between $V_{i\min}$ and the undervoltage lockout level, the output voltage may be below the value defined in table *Output data*.

Reverse Polarity

Reverse polarity at the input of AM, BM, CM, DM, and FM models will cause the fuse to blow. In EM and LM models a series diode will protect the converter. A series diode is not incorporated in AM, BM, CM, DM and FM types to avoid unwanted power losses.

Electrical Output Data

General Conditions:

- $T_A = 25^\circ\text{C}$, unless T_C is specified.
- Connector pins 2 and 23 interconnected, R input not connected; with option P: $V_o = V_{o\text{ nom}}$

Table 4: Output data

Output voltage				5.1 V			12 V			15 V			24 V			48 V			Unit
Characteristics		Conditions	min	typ	max	min	typ	max	min	typ	max	min	typ	max	min	typ	max		
V _o	Output voltage	V _{I nom} ¹ I _{o nom} ¹	5.07		5.13	11.93		12.07	14.91		15.09	23.86		24.14	47.72		48.28	V	
V _{op}	Output overvoltage protection ⁵		7.5			21			25			41			85				
I _{o nom}	Output current	V _{I min} – V _{I max}	see Table 1: Model Selection																
I _{oL}	Output current limitation	T _{C min} – T _{C max}	see Fig. 4: Typical voltage V _o Vo versus output currents I _o																
V _o	Output voltage noise	Switching frequ.	V _{I nom} ¹ I _{o nom} ¹ IEC/EN 61204	15	30	25	50	35	70	40	80	50	100	mV _{pp}					
		Total	BW = 20 MHz	60	120	40	80	40	80	40	80	-							
ΔV _{oV}	Static line regulation	V _{I min} – V _{I nom} V _{I nom} – V _{I max} I _{o nom} ¹	±10	±30	±12	±50	±15	±60	±15	±60	±15	±60	mV						
ΔV _{oI}	Static load regulation	V _{I nom} I _{o nom} – 0 ²	6	25	13	50	17	60	30	80	60	150							
ΔV _{oIc}	Static cross load regulation ⁴	V _{I nom} I _{o nom} – 0 ³	0	±15	0	±20	0	±30	0	±40	-								
V _{od}	Dynamic load regulation	Voltage deviation	V _{I nom} I _{o nom} ↔ 1/3 I _{o nom} ²	±220		±110		±150		±130		±150	ms						
t _d		Recovery time	IEC/EN 61204	0.6		0.6		0.5		1		2							
V _{odc}	Dynamic cross load regulation ⁴	Voltage deviation	V _{I nom} I _{o nom} ↔ 1/3 I _{o nom} ³	+10 -100		+10 -75		+10 -140		+20 -200		- -	mV						
t _{dc}		Recovery time	IEC/EN 61204	0.05 0.5		0.2 0.3		0.5 0.7		1 2		- -	ms						
α _{Vo}	Temperature coefficient ΔV _o / ΔT _C	V _{I min} – V _{I nom}	±0.02		±0.02		±0.02		±0.02		±0.02		%/K						
		0 – I _{o nom}	±1.0		±2.4		±3.0		±4.8		±9.6		mV/K						

¹ With multiple-output models, all outputs are loaded with the nominal current.

² Condition for specified output. With multiple-output models, other output(s) loaded with constant current $I_{o\text{ nom}}$. See fig. 5 *Dynamic load regulation*.

³ Condition for non-specified output, individually tested, other output(s) loaded with constant current $I_{o\text{ nom}}$. See fig. 5 *Dynamic load regulation*.

⁴ Multiple-output models.

⁵ By suppressor diode.

Output Characteristic and Protection

Each output is protected by a suppressor diode, which under worst case conditions may become a short circuit. The suppressor diodes are not designed to withstand externally applied overvoltages. Overload at any of the outputs will cause a shutdown of all outputs. A red LED indicates an overload condition at the respective output.

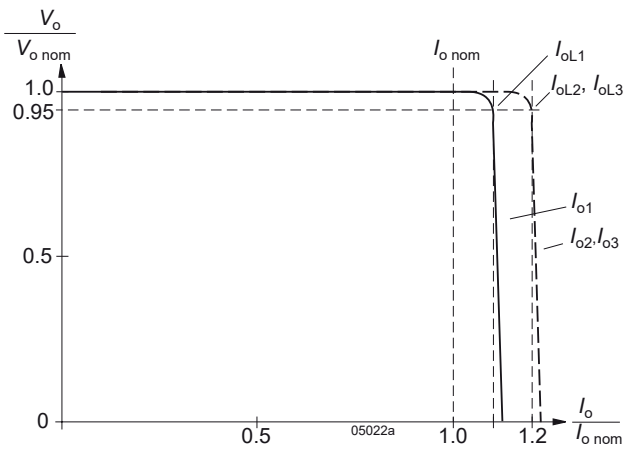


Fig. 4
Typical voltage V_o versus output currents I_o .

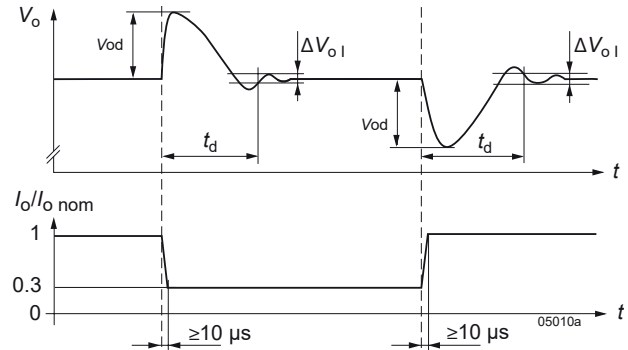


Fig. 5
Dynamic load regulation $V_{o,d}$ versus load change.

Thermal Considerations and Protection

If a converter is located in free, quasi-stationary air (convection cooling) at the indicated maximum ambient temperature $T_{A,max}$ (see table *Temperature specifications*) and is operated at its nominal input voltage and output power, the temperature measured at the measuring point of case temperature T_C (see *Mechanical Data*) will approach the indicated value $T_{C,max}$ after the warm-up phase. However, the relationship between T_A and T_C depends heavily on the conditions of operation and integration into a system. The thermal conditions are influenced by input voltage, output current, airflow, and temperature of surrounding components and surfaces. $T_{A,max}$ is therefore, contrary to $T_{C,max}$, an indicative value only.

Caution: The installer must ensure that under all operating conditions T_C remains within the limits stated in the table *Temperature specifications*.

Notes: Sufficient forced cooling or an additional heat sink allow T_A to pass over 71 °C, if $T_{C,max}$ is not exceeded.

For -7 or -9 models at an ambient temperature T_A of 85 °C with only convection cooling, the maximum permissible current for each output is approx. 50% of its nominal value; see figure 6.

A temperature sensor generates an internal inhibit signal disabling the outputs, when the case temperature exceeds $T_{C,max}$. The outputs automatically recover, when the temperature drops below this limit.

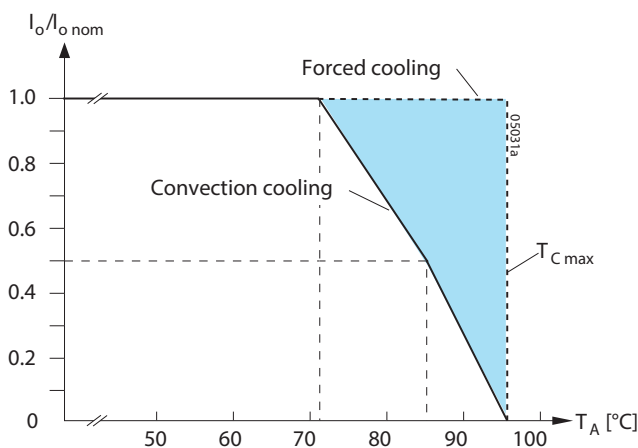


Fig. 6
Output current derating versus temperature

Parallel and Series Connection

Main outputs of equal nominal voltage can be connected in parallel. It is important to assure that the main output of a multiple-output converter is forced to supply a minimum current of 0.1 A to enable correct operation of its own auxiliary outputs.

In parallel operation, one or more of the main outputs may operate continuously in current limitation, causing an increase of the case temperature T_c . Consequently, a reduction of the max. ambient temperature by 10 K is recommended.

Main or auxiliary outputs can be connected in series with any other output of the same or another converter. In series connection, the maximum output current is limited by the lowest current limit. Output ripple and regulation values are added. Connection wiring should be kept as short as possible.

If output terminals are connected together in order to establish multi-voltage configurations, e.g., +5.1 V, ±12 V etc., the common-ground connecting point should be as close as possible to the connectors of the converter in order to avoid excessive output ripple voltages.

Note: Auxiliary outputs should never be connected in parallel!

Output Current Allocation for Special Models

Output currents differing from those given for standard models (see *Model Selection*) can be provided on request. A maximum output power of 50 W should be considered, if an ambient temperature range of –40 to 71 °C is required. The maximum permissible output currents are indicated in the table below. If the output voltages are different from standard values, the relevant output currents have to be adapted accordingly.

With reduced maximum ambient temperature or with forced cooling, the total output power may exceed 50 W. Customized configurations always need to be checked by a feasibility study first. Please ask the Company for more information.

Table 5: Current allocation with special models

Output voltage all types $V_{o1/2/3 \text{ nom}}$ [V]		Output 1 all types $I_{o1 \text{ max}}$ [A]	Output 2 AM – LM2000 $I_{o2 \text{ max}}$ [A]	Output 2 AM – LM3000 $I_{o2 \text{ max}}$ [A]	Output 3 AM – LM3000 $I_{o3 \text{ max}}$ [A]	Temperature	
						T_A [°C]	T_c [°C]
5.1		8.0	4.0	1.8 (2.5 ¹)	1.5	–40 to 71	–25 to 95
12		4.0	2.0	1.5	1.2		
15		3.4	1.7	1.2	1.0		
24		2.0	1.0	0.7	0.5		
²	5.1	10.0	4.5	2.1 (2.8 ¹)	1.8	–25 to 60	–25 to 90
	12	5.0	2.5	1.7	1.5		
	15	4.0	2.0	1.5	1.3		
	24	2.5	1.3	0.9	0.7		
	5.1	11.0	5.0	2.4 (3.0 ¹)	2.0	–25 to 50	–25 to 85
	12	6.0	3.0	2.0	1.7		
	15	4.6	2.3	1.7	1.5		
	24	3.0	1.5	1.0	0.8		

¹ Special high-current components required.

² $V_{i \text{ min}}$ has to be increased.

Hold-up Time and Output Response

When the input voltage is switched off, the output voltage will remain high for a certain hold-up time t_h (see fig. 7) before the output voltage falls below $0.95 V_{o\text{ nom}}$. To achieve the hold-up times indicated in fig. 8, AM, BM, CM, DM, and FM models require an external series diode in the input line. This is necessary to prevent the discharge of the input capacitor through the source impedance or other circuits connected to the same source. EM and LM models have a built-in series diode. In AM, BM, CM, DM, and FM models, no series diode is built-in, since it would generate additional power losses inside the converter.

Note: For hold-up time with option V, refer to *Option V*.

The behavior of the outputs is similar with either the input voltage applied or the inhibit switched low.

No output voltage overshoot occurs, when the converter is turned on or off.

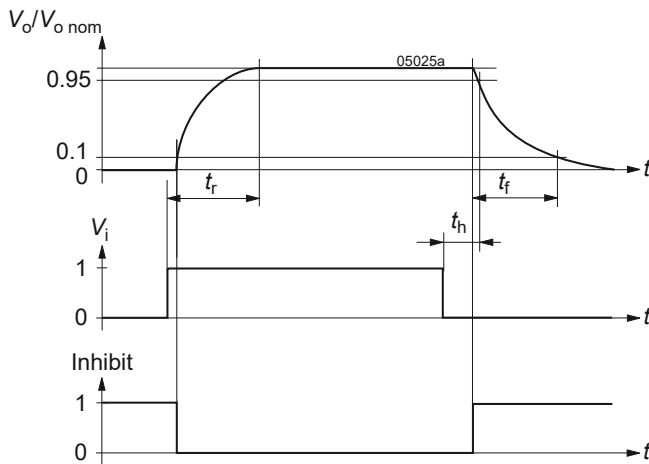


Fig. 7
Output response times versus V_i or inhibit control

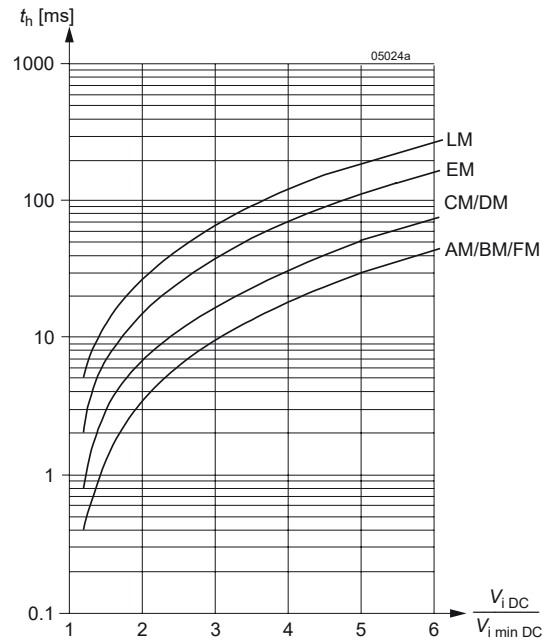


Fig. 8
Typical hold-up time t_h versus input voltage at $I_{o\text{ nom}}$

Table 6: Output response time t_r and t_f (see fig. 7). Values not applicable for models equipped with option E.

Type of converter	t_r at $P_o = 0$ and t_f at $P_o = P_{o\text{ nom}}$		t_r and t_f at $P_o = \frac{3}{4} P_{o\text{ nom}}$		t_r and $P_o = P_{o\text{ nom}}$		Unit
	typ	max	typ	max	typ	max	
AM – LM1001-9R	5	10	5	10	10	20	ms
AM – LM1301-9R	10	20	15	30	20	30	
AM – LM1501-9R	5	10	10	20	30	40	
AM – LM1601-9R	15	30	25	50	40	60	
AM – LM1901-9R	65	130	100	200	165	330	
AM – LM2320-9	20	40	30	60	50	100	
AM – LM2540-9	15	30	20	40	35	70	
AM – LM3020-9	55	110	85	170	145	290	
AM – LM3040-9	40	80	60	120	100	200	

Conditions:

R input not connected. For multiple-output models the figures indicated in the table relate to the output, which reacts slowest. All outputs are resistively loaded. Variation of the input voltage within $V_{i\text{ min}} - V_{i\text{ max}}$ does not influence the values considerably.

Auxiliary Functions

Inhibit

The outputs of the converters may be enabled or disabled by means of a logic signal (TTL, CMOS, etc.) applied between the inhibit input i and the negative pin of output 1 ($Vo1-$). In systems with several converters, this feature can be used, for example, to control the activation sequence of the converters. If the inhibit function is not required, connect the inhibit pin 2 to pin 23 to enable the outputs (active low logic, fail safe). The response times are specified in table 6.

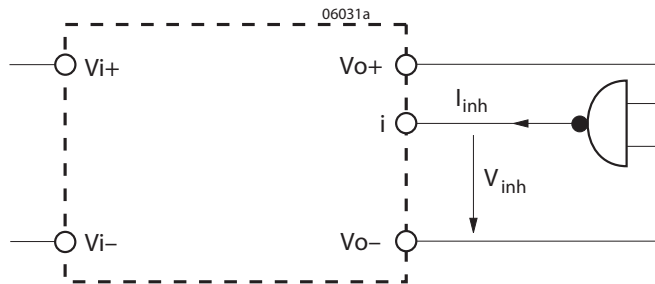


Fig. 9
Definition of V_{inh} and I_{inh} .

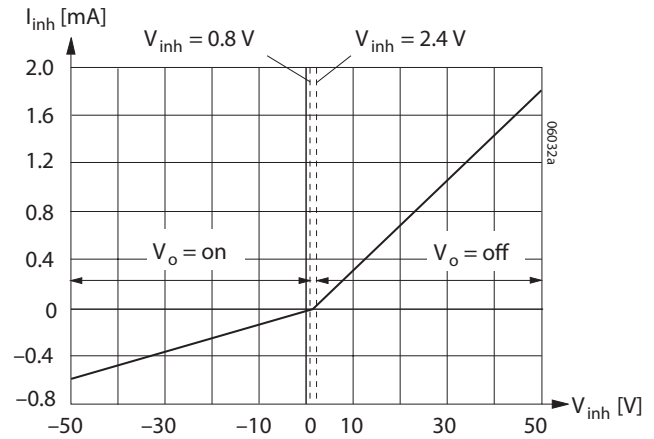


Fig. 9
Definition of V_{inh} and I_{inh} .

Table 7: Inhibit data

Characteristics		Conditions	min	typ	max	Unit
V_{inh}	Inhibit input voltage to keep output voltage	$V_o = \text{on}$	$V_{i \min} - V_{i \max}$	-50	0.8	V
		$V_o = \text{off}$	$T_{C \min} - T_{C \max}$	2.4	50	
I_{inh}	Inhibit current	$V_{inh} = 0$	-60	-100	-220	μA

R-Control for Output Voltage Adjustment

As a standard feature, single-output models without option P offer an adjustable output voltage identified by letter R in the type designation.

Note: With open R input, $V_o = V_{o \text{ nom}}$.

The output voltage V_o can be adjusted by an external voltage (V_{ext}) or by an external resistor (R_{ext1} or R_{ext2}). The adjustment range is approximative 0 – 110% of $V_{o \text{ nom}}$. For output voltages $V_o > V_{o \text{ nom}}$, the minimum input voltage $V_{i \min}$ specified in *Electrical Input Data* increases proportionally to $V_o/V_{o \text{ nom}}$.

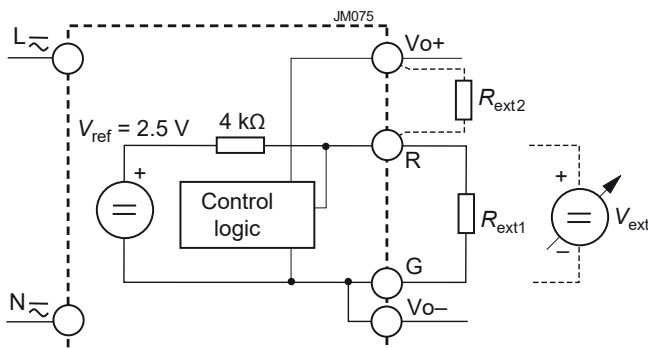


Fig. 11
Output voltage adjustment

a) Adjustment by means of an external resistor R_{ext}

Depending upon the value of the required output voltage, the resistor shall be connected:

either: Between the R and G pin to achieve an output voltage adjustment range of $V_o \approx 0$ to 100 % of $V_{o\ nom}$.

$$R_{ext1} \approx 4\ k\Omega \cdot \frac{V_o}{V_{o\ nom} - V_o}$$

or: Between the R pin and Vo+ to achieve an output voltage range of $V_o \approx 100$ to 110% of $V_{o\ nom}$.
($V_o - 2.5\ V$)

$$R_{ext2} \approx 4\ k\Omega \cdot \frac{2.5\ V \cdot (V_o / V_{o\ nom} - 1)}{V_o - 2.5\ V}$$

Caution: To prevent damage, R_{ext2} should never be less than 47 k Ω .

Note: R inputs of n converters with paralleled outputs may be paralleled too, but if only one external resistor is used, its value should be R_{ext1}/n or R_{ext2}/n respectively.

b) Adjustment by means of an external control voltage V_{ext} between G and R pin.

The control voltage range is 0 to 2.75 V and allows for adjustment in the range of $V_o \approx 0$ to 110% of $V_{o\ nom}$.

$$V_{ext} \approx \frac{V_o \cdot 2.5\ V}{V_{o\ nom}}$$

Caution: The external control voltage should be in the range 0 to +3V to prevent the converter from damage.

Table 8a: R_{ext1} for $V_o < V_{o\ nom}$ (Conditions: $V_{i\ nom}$, $I_{o\ nom}$, rounded up to resistor values E 96, R_{ext2} is not fitted.)

$V_{o\ nom} = 5.1\ V$		$V_{o\ nom} = 12\ V$		$V_{o\ nom} = 15\ V$		$V_{o\ nom} = 24\ V$		$V_{o\ nom} = 48\ V$	
$V_o\ [V]$	$R_{ext1}\ [k\Omega]$	$V_o\ [V]$	$R_{ext1}\ [k\Omega]$	$V_o\ [V]$	$R_{ext1}\ [k\Omega]$	$V_o\ [V]$	$R_{ext1}\ [k\Omega]$	$V_o\ [V]$	$R_{ext1}\ [k\Omega]$
0.5	0.432	2	0.806	2	0.619	4	0.806	8	0.806
1.0	0.976	3	1.33	4	1.47	6	1.33	12	1.33
1.5	1.65	4	2	6	2.67	8	2	16	2
2.0	2.61	5	2.87	8	4.53	10	2.87	20	2.87
2.5	3.83	6	4.02	9	6.04	12	4.02	24	4.02
3.0	5.76	7	5.62	10	8.06	14	5.62	28	5.62
3.5	8.66	8	8.06	11	11	16	8.06	32	8.06
4.0	14.7	9	12.1	12	16.2	18	12.1	36	12.1
4.5	30.1	10	20	13	26.1	20	20	40	20
5.0	200	11	44.2	14	56.2	22	44.2	44	44.2

Table 8b: R_{ext2} for $V_o > V_{o\ nom}$ (Conditions: $V_{i\ nom}$, $I_{o\ nom}$, rounded up to resistor values E 96, R_{ext1} is not fitted.)

$V_{o\ nom} = 5.1\ V$		$V_{o\ nom} = 12\ V$		$V_{o\ nom} = 15\ V$		$V_{o\ nom} = 24\ V$		$V_{o\ nom} = 48\ V$	
$V_o\ [V]$	$R_{ext2}\ [k\Omega]$	$V_o\ [V]$	$R_{ext2}\ [k\Omega]$	$V_o\ [V]$	$R_{ext2}\ [k\Omega]$	$V_o\ [V]$	$R_{ext2}\ [k\Omega]$	$V_o\ [V]$	$R_{ext2}\ [k\Omega]$
5.15	464	12.1	1780	15.2	1470	24.25	3160	48.5	6810
5.20	215	12.2	909	15.4	750	24.50	1620	49.0	3480
5.25	147	12.3	619	15.6	511	24.75	1100	49.5	2370
5.30	110	12.4	464	15.8	383	25.00	825	50.0	1780
5.35	90.9	12.5	383	16.0	332	25.25	715	50.5	1470
5.40	78.7	12.6	316	16.2	274	25.50	590	51.0	1270
5.45	68.1	12.7	274	16.4	237	25.75	511	51.5	1100
5.50	61.9	12.8	249	16.5	226	26.00	453	52.0	953
		13.0	200			26.25	402	52.5	845
		13.2	169			26.40	383	52.8	806

Display Status of LEDs

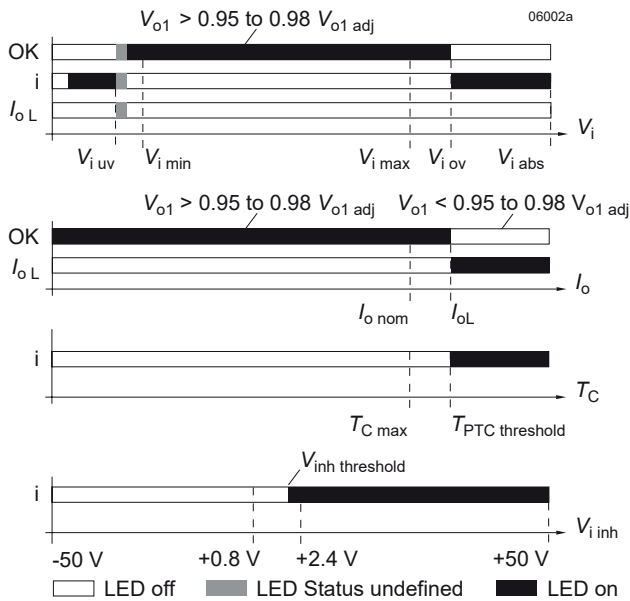


Fig. 12
LED indicators

Electromagnetic Compatibility (EMC)

A suppressor diode or a metal oxide VDR (depending upon converter model) together with an input fuse and an input filter form an effective protection against high input transient voltages, which typically occur in most installations, but especially in battery-driven mobile applications. The M Series has been successfully tested to the following specifications:

Electromagnetic Immunity

Table 9: Immunity type tests

Phenomenon	Standard	Level	Coupling mode ¹	Value applied	Waveform	Source impeded.	Test procedure	In oper.	Perf. crit. ²			
Supply related surge	RIA 12 ³	A ⁴	+i/−i	3.5 • V _{Batt}	2/20/2 ms	0.2 Ω	1 positive surge	yes	A			
		B		1.5 • V _{Batt}	0.1/1/0.1 s							
Direct transients		C	+i/−i, −i/c	960 V _p	10/100 μs	5 Ω	5 pos. & 5 neg. impulses	yes	A			
		D ³		1800 V _p	5/50 μs							
		E		3600 V _p	0.5/5 μs							
		F		4800 V _p	0.1/1 μs							
		G		8400 V _p	0.05/0.1 μs							
		Indirect couples transients		H	−o/c, +o/−o, −o/−i	1800 V _p				5/50 μs	100 Ω	
J			3600 V _p	0.5/5 μs								
K			4800 V _p	0.1/1 μs								
L			8400 V _p	0.05/0.1 μs								
Electrostatic discharge (to case)		IEC/EN 61000-4-2	4 ⁵	contact discharge	±8000 V _p	1/50 ns				330 Ω 150 pF	10 pos. & 10 neg. discharges	yes
	air discharge			±15000 V _p								
Electromagnetic field	IEC/EN 61000-4-3	x ⁶	antenna	20 V/m	AM 80% / 1 kHz	N/A			80 – 1000 MHz	yes	A ¹¹	
Electromagnetic field, pulse modulated	ENV 50204	4 ⁷		10 V/m	50% duty cycle, 200 Hz repetition frequency		900 ±5 MHz	yes	A			
Electrical fast transients / burst	IEC/EN 61000-4-4	3 ⁸	capacitive, o/c	±2000 V _p	bursts of 5/50 ns; 2.5 / 5 kHz over 15 ms; burst period: 300 ms	50 Ω	60 s positive 60 s negative transients per coupling mode	yes	A ¹¹			
		3 ⁸	direct, i/c, +i/−i	±2000 V _p					A ¹¹			
		4		±4000 V _p					B			
Surges	IEC/EN 61000-4-5	3 ⁹	i/c	2000 V _p	1.2 / 50 μs	12 Ω	5 pos. & 5 neg. surges per coupling mode	yes	A			
			+i/−i	1000 V _p		2 Ω						
Conducted disturbances	IEC/EN 61000-4-6	3 ¹⁰	i, o, signal wires	10 VAC (140 dBμV)	AM 80% / 1 kHz	150 Ω	0.15 – 80 MHz	yes	A			
Power frequency magnetic field	IEC/EN 61000-4-8	3 ¹¹	-	300 A/m			60 s in all 3 axes	yes	A			

¹ i = input, o = output, c = case

² A = normal operation, no deviation from specs.; B = normal operation, temporary loss of function or deviation from specs possible

³ RIA 12 covers or exceeds IEC 60571-1 and EN 50155:1995. Surge D corresponds to EN 50155:2001, waveform A; surge G corresponds to EN 50155:2001, waveform B.

⁴ Only met with EM (110 V battery) and extended input range models (customer-specific) of BM (24 V battery) and CM (48 V battery). Standard DK models (72 V battery) are not damaged, but overvoltage lockout will occur during the surge.

⁵ Exceeds EN 50121-3-2:2015 table 6.3 and EN 50121-4:2006 table 1.4.

⁶ Corresponds to EN 50121-3-2:2015 table 6.1 and exceeds EN 50121-4:2006 table 1.1. Valid for version V104 or higher.

⁷ Compliance with digital mobile phones.

⁸ Corresponds to EN 50121-3-2:2015 table 5.2 and EN 50121-4:2006 table 2.2.

⁹ Covers or exceeds EN 50121-3-2:2015 table 4.3 and EN 50121-4:2006 table 2.3.

¹⁰ Corresponds to EN 50121-3-2:2015 table 5.1 and EN 50121-4:2006 table 3.1 (radio frequency common mode).

¹¹ Perf. criterion B for triple-output models.

¹² Corresponds to EN 50121-4:2006 table 1.3 for AC systems

Electromagnetic Emissions

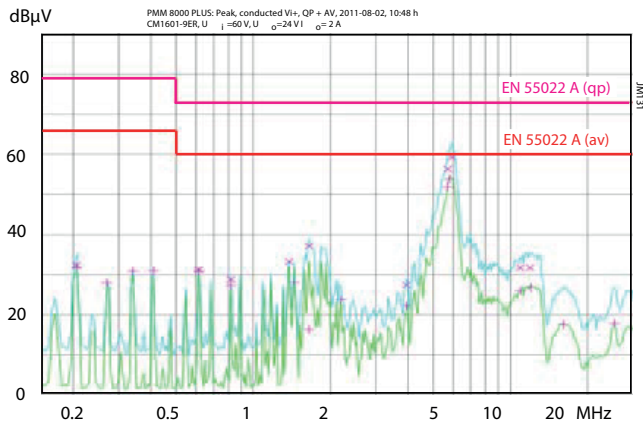


Fig. 13a
Typ. conducted disturbances at the input (quasi-peak and average) of CM1601-9ER according to IEC/EN 55011/22, measured at $V_i = 60$ VDC and $I_{o\ nom}$.

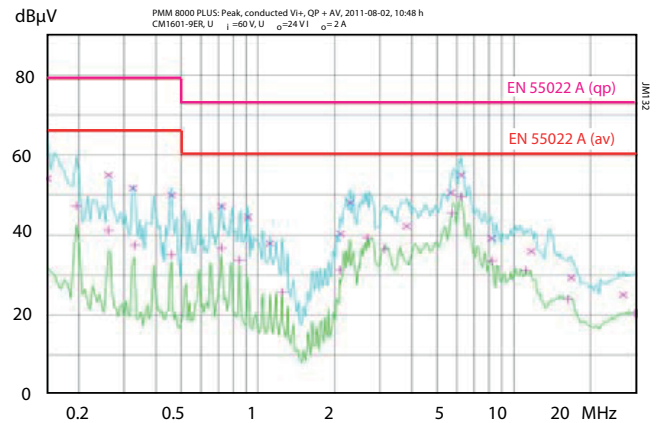


Fig. 13b
Typ. conducted disturbances at the input (quasi-peak and average) of LM1601-9R according to IEC/EN 55011/22, measured at $V_i = 230$ VAC and $I_{o\ nom}$.

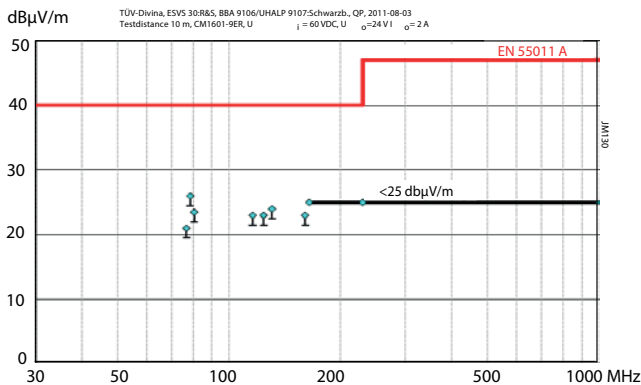


Fig. 14a
Typical radiated emissions of CM1601-9ER according to IEC/EN 55011/22, normalized to a distance of 10 m, measured at $V_i = 60$ VDC and $I_{o\ nom}$.

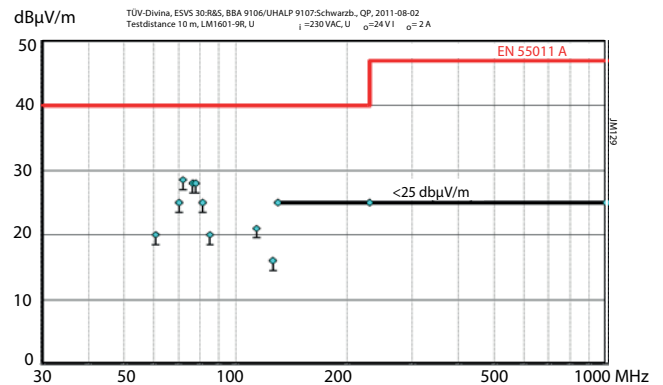


Fig. 14b
Typical radiated emissions of LM1601-9R according to IEC/EN 55011/22, normalized to a distance of 10 m, measured at $V_i = 230$ VAC and $I_{o\ nom}$.

Immunity to Environmental Conditions

Table 10: Mechanical and climatic stress

Test method		Standard	Test Conditions	Status
Db	Damp heat test, cyclic	EN 50155:2007, clause 12.2.5 IEC/EN 60068-2-30	Temperature: 55 °C and 25 °C Cycles (respiration effect): 2 Duration: 2x 24 h	Converter not operating
Bd	Dry heat test, steady state	EN 50155:2007, clause 12.2.4 IEC/EN 60068-2-2	Temperature: 70 °C Duration: 6 h	Converter operating
Ad	Cooling test, steady state	EN 50155:2007, clause 12.2.3 IEC/EN 60068-2-1	Temperature, duration: - 40 °C, 2 h Performance test: +25 °C	Converter not operating
Kb	Salt mist, cyclic sodium chloride (NaCl) solution	IEC/EN 60068-2-52	Concentration: 5% (30 °C) Duration: 2 h per cycle Storage: 40 °C, 93% rel. humidity Storage duration: 22 h per cycle, 3 cycles	Converter not operating
Fc	Vibration (sinusoidal)	IEC/EN 60068-2-6 MIL-STD-810D section 514.3	Acceleration amplitude: 0.35 mm (10 – 60 Hz) 5 g _n = 49 m/s ² (60 - 2000 Hz) Frequency (1 Oct/min): 10 – 2000 Hz Test duration: 7.5 h (2.5 h in each axis)	Converter operating
Fda	Random vibration wide band Reproducibility high	IEC 60068-2-35 DIN 40046 part 23	Acceleration spectral density: 0.05 g _n ² /Hz Frequency band: 20 – 500 Hz Acceleration magnitude: 4.9 g _{n rms} Test duration: 3 h (1 h in each axis)	Converter operating
Eb	Bump (half-sinusoidal)	IEC/EN 60068-2-29 MIL-STD-810D section 516.3	Acceleration amplitude: 40 g _n = 392 m/s ² Bump duration: 6 ms Number of bumps: 6000 (1000 in each direction)	Converter operating
Ea	Shock (half-sinusoidal)	IEC/EN 60068-2-27 MIL-STD-810D section 516.3	Acceleration amplitude: 100 g _n = 981 m/s ² Bump duration: 6 ms Number of bumps: 18 (3 in each direction)	Converter operating
-	Shock	EN 50155:2007 clause 12.2.11, EN 61373 sect. 10, class B, body mounted ¹	Acceleration amplitude: 5.1 g _n Bump duration: 30 ms Number of bumps: 18 (3 in each direction)	Converter operating
-	Simulated long life testing at increased random vibration levels	EN 50155:2007 clause 12.2.11, EN 61373 sect. 8 and 9, class B, body mounted ¹	Acceleration spectral density: 0.02 g _n ² /Hz Frequency band: 5 – 150 Hz Acceleration magnitude: 0.8 g _{n rms} Test duration: 15 h (5 h in each axis)	Converter operating

¹ Body mounted = chassis of a railway coach

Temperatures

Table 11: Temperature specifications, valid for an air pressure of 800 – 1200 hPa (800 – 1200 mbar)

Model			-7 (option)		-9 (standard)		Unit
Characteristics	Conditions		min	max	min	max	
T_A	Ambient temperature	Converter operating	- 25	71	- 40	71	° C
T_C	Case temperature		- 25	95	- 40	95	
T_S	Storage temperature	Not operating	- 40	85	- 55	85	

Reliability

Table 12: MTBF

Ratings at specified case temperature	Converter model	Ground benign 40 °C	Ground fixed		Ground mobile 50 °C	Device hours ²	Unit
			40 °C	70 °C			
MTBF ¹	AM – LM1000	320 000	130 000	40 000	35 000	880 000	h
	AM – LM2000	225 000	105 000	32 000	28 000	720 000	
	AM – LM3000	225 000	80 000	28 000	25 000	740 000	

¹ Calculated in accordance with MIL-HDBK-217E

² Statistical values, based on an average of 4300 working hours per year in general field use over 3 years

Mechanical Data

Dimensions in mm.

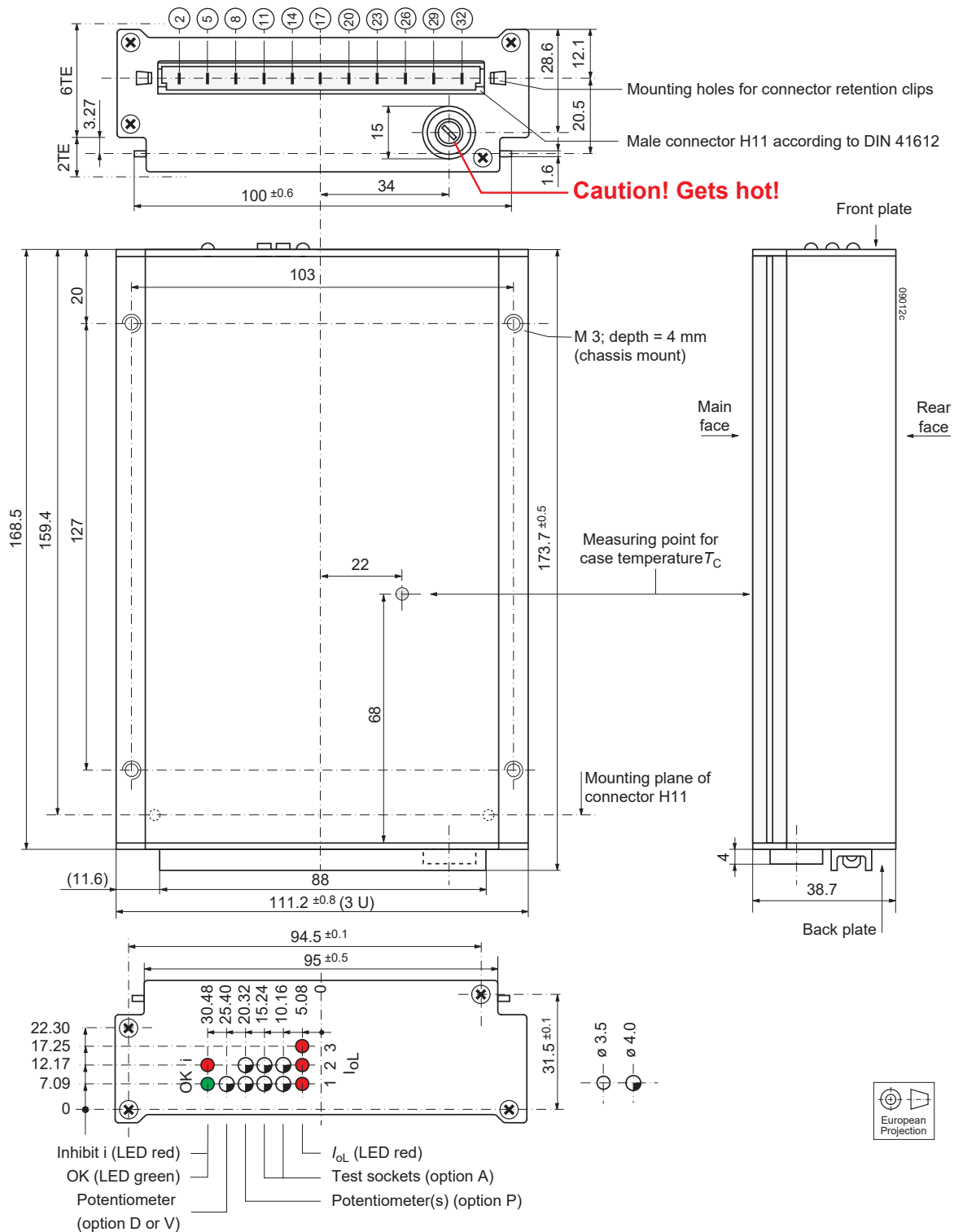


Fig. 15
Case M02, weight 770 g (approx.).
Case aluminum, black finish and self cooling.

Note: Long case, elongated by 60 mm for 220 mm rack depth, is available on request.

Safety and Installation Instructions

Connector Pin Allocation

Pin no. 26 (protective earth) is a leading pin, ensuring that it makes contact with the female connector first.

Table 13: Pin allocation

Electrical determination	AM - LM1000		AM - LM2000		AM - LM3000	
	Pin	Ident	Pin	Ident	Pin	Ident
Inhibit	2	i	2	i	2	i
Safe data or ACFAIL	5 ⁵	D or V	5 ⁵	D or V	5 ⁵	D or V
Output voltage (positive)	8	Vo+	8	n.c.	8	Vo3+
Output voltage (negative)	11	Vo-	11	n.c.	11	Vo3-
Voltage adjust	14	R ¹				
Adjust return	17	G ¹				
Output voltage (positive)			14	Vo2+	14	Vo2+
Output voltage (negative)			17	Vo2-	17	Vo2-
Output voltage (positive)	20	Vo+	20	Vo1+	20	Vo1+
Output voltage (negative)	23	Vo-	23	Vo1-	23	Vo1-
Protective earth PE ²	26	⊕	26	⊕	26	⊕
DC input voltage ³	29	Vi+	29	Vi+	29	Vi+
DC input voltage	32	Vi-	32	Vi-	32	Vi-
AC input voltage ⁴	29	N≈	29	N≈	29	N≈
AC input voltage	32	L≈	32	L≈	32	L≈

¹ Not connected if option P is fitted

² Leading pin

³ AM, BM, CM, DM, EM, and FM models

⁴ LM models

⁵ Not connected if option neither option D or V is fitted

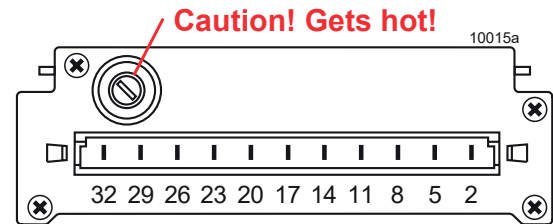


Fig. 16
View of male H11 connector.

Installation Instructions

All M Series converters are components, intended exclusively for inclusion within other equipment by professional installers. Installation must strictly follow the national safety regulations in compliance with the enclosure, mounting, creepage, clearance, casualty, markings and segregation requirements of the end-use application.

Connection to the system shall be made via the female connector H11. Other installation methods may not meet the safety requirements.

The converters are provided with the leading pin 26 (⊕), which is reliably connected with the case. For safety reasons, it is essential to connect pin 26 with the protective earth of the supply system.

An input fuse is connected in the line to pin 32 (Vi- or L≈). Since this fuse is designed to protect the converter in case of an overcurrent and does not necessarily cover all customer needs, an external fuse suitable for the application and in compliance with the local requirements may be necessary in the wiring to one or both input pins (no. 29 and/or no. 32), particularly if the phase or neutral line cannot be assigned to the corresponding terminals (LM models operated with AC).

Important: Whenever the inhibit function is not in use, pin 2 (i) should be connected to pin 23 (Vo-) to enable the output(s).

Caution: Do not open the converters, or warranty will be invalidated.

Make sure that there is sufficient air flow possible for convection cooling. This should be verified by measuring the case temperature $T_{C,}$ when the converter is installed and operated in the end-use application. The maximum specified case temperature $T_{C, \max}$ shall not be exceeded. See also *Thermal Considerations*.

Operation of LM Models at Greater than 63 Hz

In such a case, the converters may exceed the leakage current of 3.5 mA imposed in the safety standards. A warning marking is required in the end-use product.

Protection Degree and Cleaning Liquids

Condition: Female connector fitted to the converter.

IP 40: All models, except those with options P or A, and except those with option D/V with potentiometer.

IP 30: All models fitted with options A or option D/V without potentiometer.

IP 20: All models fitted with option P or with option D/V with potentiometer.

In order to avoid possible damage, any penetration of liquids (e.g., cleaning fluids) has to be avoided.

Railway Applications

The M Series converters have been designed observing the railway standards EN 50155 and EN 50121. All boards are coated with a protection lacquer.

Standards and Approvals

The converters correspond to class I equipment and have been approved according to the standards IEC/EN 60950-1 and UL/CSA 60950-1 2nd Ed.

The converters have been evaluated for:

- Class I equipment
- Building in
- Basic insulation between input and case and double or reinforced insulation between input and output, based on the input voltage of 250 VAC or 400 VDC
- Functional insulation between output(s) and case
- Functional insulation between the outputs
- Pollution degree 2 environment
- Overvoltage category II
- Altitude up to 2000 m

The converters are subject to manufacturing surveillance in accordance with the above mentioned standards and with ISO 9001:2015.

Isolation

The electric strength test is performed in the factory as routine test in accordance with EN 50514 and IEC/EN 60950.

The company will not honor any warranty claims resulting from incorrectly executed electric strength field tests.

Table 14: Isolation

Characteristics		Input to Case + Output(s)	Output(s) to Case (standard)	Output(s) to Case (option H)	Output to Output	Unit
Electric strength test	Factory test >1 s	2.8 ¹	1.4	2.8	0.3	kVDC
	AC test voltage equivalent to factory test	2.0 ¹	1.0	2.0	0.2	kVAC
Insulation resistance at 500 VDC		>300	>300	>300	>100 ²	MΩ
Creepage distances		≥ 3.2 ³	---	---	---	mm

¹ According to IEC/EN 60950, sub-assemblies connecting input to output are pre-tested with 5.6 kVDC or 4 kVAC.

² Tested at 300 VDC

³ Input to outputs: ≥ 6.4 mm

Safety of Operator-Accessible Output Circuits

If the output circuit of a DC-DC converter is operator-accessible, it shall be an SELV circuit according to the IEC/EN 60950 safety standards.

Since the M Series converters provide double or reinforced insulation between input and output based upon a rated primary input voltage of 250 VAC or 400 VDC, only functional insulation is needed between the AC mains and the input of the converter.

Only voltage adaption and rectification to the specified input voltage range of a DC/DC converter is needed.

Table 15 shows a possible installation configuration, compliance with which causes the output circuit of the DC-DC converter to be an SELV circuit according to IEC/EN 60950 up to a configured output voltage (sum of nominal voltages if in series or +/- configuration) of 48 V. However, it is the sole responsibility of the installer to assure the compliance with the relevant and applicable safety regulations.

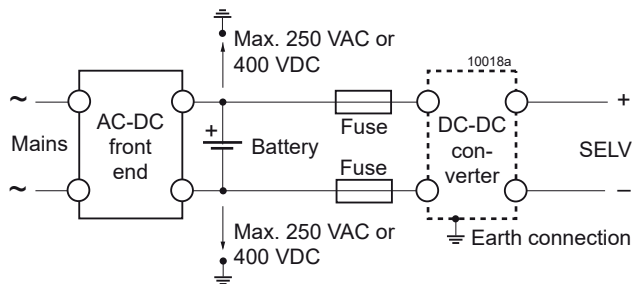


Fig. 17
Schematic safety concept

Table 15: Safety concept leading to an SELV output circuit

Conditions	Front end			DC-DC converter		Result
Nominal supply voltage	Minimum required grade of insulation, to be provided by the AC-DC front end, including mains supplied battery charger	Maximum rated DC output voltage from the front end	Minimum required safety status of the front end output circuit	Equipment	Measures to achieve the specified safety status of the output circuit	Safety status of the DC-DC converter output circuit
Mains 250 VAC	Operational (i.e. there is no need for electrical isolation between the mains supply voltage and the DC-DC converter input voltage)	400 VDC ¹ (The rated voltage between any input pin and earth can be up to 250 VAC or 400 VDC)	Primary circuit	A-LM	Double or reinforced insulation, based on 250 VAC and 400 VDC (provided by the DC-DC converter) and earthed case ²	SELV circuit

¹ The front end output voltage should match the specified operating input voltage range of the DC-DC converter.

² The earth connection has to be provided by the installer according to the safety standard IEC/EN 60950.

Description of Options

Table 16: Survey of options

Option	Function of option	Characteristics
- 7	Former standard operational ambient temperature range	$T_A = -25$ to 71 °C
A	Test sockets at front panel for check of output voltage	V_o internally measured at the connector terminals
E	Electronic inrush current limitation circuitry	Active inrush current limitation, only for CM, EM, LM models
P ¹	Potentiometer for fine adjustment of output voltage	Adjustment range $\pm 5\%$ of $V_{o\text{ nom}}$, excludes R input
F	Input fuse built-in	Fuse not externally accessible, only for FM1000
H	Enhanced output to case electric strength test voltage	See table <i>Isolation</i>
D ²	Input and/or output undervoltage monitoring circuitry	Safe data signal output (D0 – D9)
V ^{2,3}	Input and/or output undervoltage monitoring circuitry	ACFAIL signal according to VME specifications (V0, V2, V3)
K	Coding strip at the connector	Ensuring correct population of DIN-racks
G	RoHS	RoHS-compatible for all six substances

¹ Models equipped with option P do not provide the R function; pins 14 and 17 are not connected.

² Option D excludes option V and vice versa.

³ Only available if main output voltage $V_{o1} = 5.1$ V

-7 Former Standard Temperature Range

Option -7 stays for the operational ambient temperature range from -25 to 71 °C, which may be preferred by some customers for reasons of documentation or approvals.

A Test Sockets

Test sockets (pin $\varnothing = 2$ mm, distance $d = 5.08$ mm) are located at the front of the converter. The output voltage is sensed at the connector pins inside of the converter. Outputs 2 and 3 of triple-output models are not sensed.

P Potentiometer

Built-in multi-turn potentiometers provide an output voltage adjustment range of minimum $\pm 5\%$ of $V_{o\text{ nom}}$ and are accessible through holes in the front cover. Compensation of voltage drop across connector and wiring becomes easily achievable. For output voltages $V_o > V_{o\text{ nom}}$, the minimum input voltage according to *Electrical Input Data* increases proportionally to $V_o/V_{o\text{ nom}}$. Triple-output models allow only the adjustment of V_{o1} .

Note: Potentiometers are not recommended for mobile applications.

E Electronic Inrush Current Limitation

Available for CM, EM and LM models.

The standard version of the models CM, DM, EM and LM include a passive inrush current limitation with an NTC resistor. For applications requiring an improved inrush current limitation, an active electronic circuit as shown in fig.18 has been developed. Typical inrush current waveforms of units equipped with this option are shown below.

CM models meet the CEPT/ETSI standards for 48V supply voltage according to ETS 300132-2, if fitted with option E combined with option D6 (input voltage monitoring). Option D6, externally adjustable via potentiometer, is necessary to disable the converter at input voltages below the actual service ranges, avoiding an excessive input current when the input voltage is raised slowly according to ETS 300132-2. Option D6 threshold level $V_{ti} + V_{hi}$ (refer to description of option D) should be adjusted to $36 - 40.5$ V for 48 V nominal supply voltage (for 60 V systems, threshold should be set to $44 - 50$ V). The D output (pin 5) should be connected to the inhibit (pin 2). For applications, where potentiometers are not allowed, refer to option D9.

Table 18: Inrush current characteristics with option E

Characteristics		CM at $V_i = 110$ VDC		EM, LM at $V_i = 110$ VDC		EM, LM at $V_i = 372$ VDC		Unit
		typ	max	typ	max	typ	max	
$I_{\text{inr p}}$	Peak inrush current	6.5	8	2.2	4	7.3	10	A
t_{inr}	Inrush current duration	22	30	10	20	20	40	ms

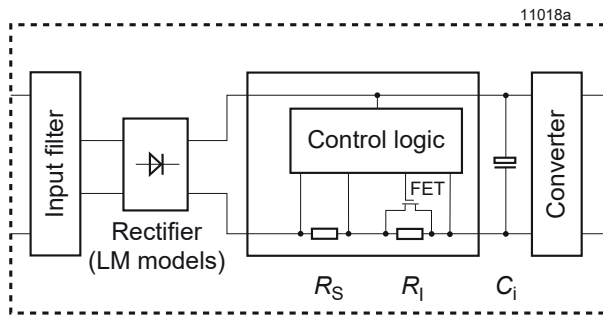


Fig. 18
Option E block diagram

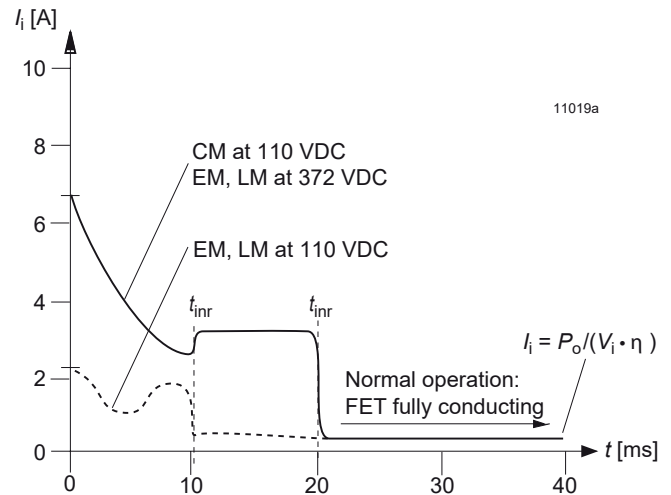


Fig. 19
Typical inrush current waveforms of CM, EM, and LM converters with option E

Precautions:

In order to avoid overload of the series resistor R_s , the on/off switching cycle should be limited to 12 s, if switched on/off continuously. There should not be more than 10 start-up cycles within 20 s at a case temperature of 25 °C.

If CM models are driven by input voltages below 35 VDC or LM models below 100 VAC, the maximum case temperature should be derated by 10 °C, or the total output power should be derated by 20%. EM and LM models driven by DC input voltages do not need to be derated within the full specified input voltage range.

F Fuse Not Accessible

Standard M converters have a fuseholder containing a 5 × 20 mm fuse, which is externally accessible and located in the back plate near to the connector. Some applications require an inaccessible fuse. Option F provides a fuse mounted directly onto the main PCB inside the case (only FM1000).

The full self-protecting functions of the converter do normally not lead to a broken fuse, except as a result of inverse polarity at the input of an AM, BM, CM, DM, or FM models, or if a power component inside fails. In such cases the defective converter must be returned to the Company for repair.

H Enhanced Electric Strength Test

Electric strength test output to case; see table *Isolation*.

D Undervoltage Monitor

The input and/or output undervoltage monitor operates independently of the built-in input undervoltage lock-out circuit. A logic “low” (JFET output) or “high” signal (NPN output) is generated at pin 5, when one of the monitored voltages drops below the preselected threshold level V_t . The return for this signal is V_{o1} (pin 23). The D output recovers, when the monitored voltage(s) exceed(s) $V_t + V_h$. The threshold level V_t is either adjustable by a potentiometer accessible through a hole in the front cover, or adjusted in the factory to a fixed value specified by the customer.

Option D exists in various versions D0 – D9, as shown in the Table 19.

JFET output (D0 – D4):

Connector pin D is internally connected via the drain-source path of a JFET (self-conducting type) to the negative potential of output 1. $V_D \leq 0.4$ V (logic low) corresponds to a monitored voltage level (V_i and/or V_{o1}) $< V_t$. The current I_D through the JFET should not exceed 2.5 mA. The JFET is protected by a 0.5 W Zener diode of 8.2 V against external overvoltages.

NPN output (D5 – D9):

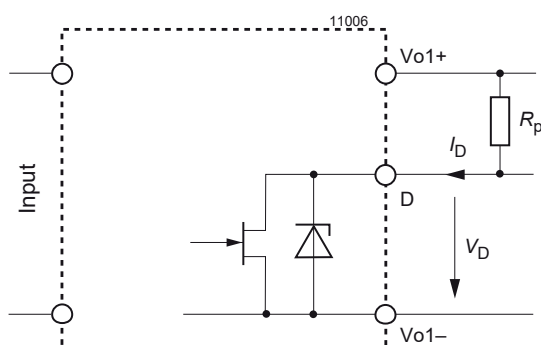
Connector pin D is internally connected via the collector-emitter path of a NPN transistor to the negative potential of output 1. $V_D \leq 0.4$ V (logic low) corresponds to a monitored voltage level (V_i and/or V_{o1}) $> V_t + V_h$. The current I_D through the open collector should not exceed 20 mA. The NPN output is not protected against external overvoltages. V_D should not exceed 40 V.

Output type		Monitoring		Minimum adjustment range of threshold level V_t		Typ. hysteresis V_h [% of V_t] for $V_{t\min} - V_{t\max}$	
JFET	NPN	V_i	V_{o1}	V_{ti}	V_{to}	V_{hi}	V_{ho}
D1	D5	no	yes	---	$3.5\text{ V} - 48\text{ V}^1$	---	$2.3 - 1\text{ V}$
D2	D6	yes	no	$V_{i\min} - V_{i\max}^1$	---	$3.0 - 0.5\text{ V}$	---
D3	D7	yes	yes	$V_{i\min} - V_{i\max}^1$	$0.95 - 0.98\text{ V}_{o1}^2$	$3.0 - 0.5\text{ V}$	"0"
D4	D8	no	yes	---	$0.95 - 0.98\text{ V}_{o1}^2$	---	"0"
D0	D9	no	yes	---	$3.5\text{ V} - 48\text{ V}^3$	---	$1.8 - 1\text{ V}$
		yes	no	$V_{i\min} - V_{i\max}^{3,4}$	---	$2.2 - 0.4\text{ V}$	---
		yes	yes	$V_{i\min} - V_{i\max}^{3,4}$	$0.95 - 0.98\text{ V}_{o1}^2$	$2.2 - 0.4\text{ V}$	"0"

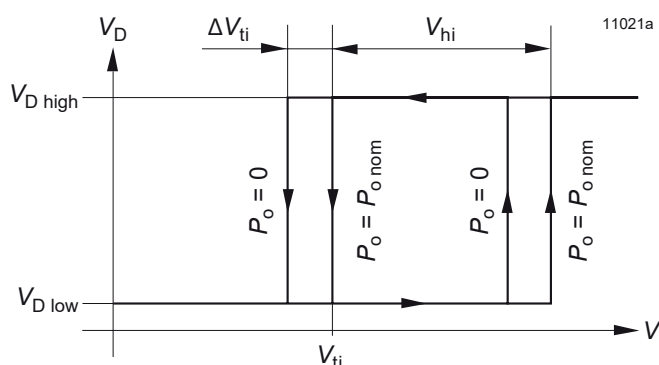
⁴ Adjusted at $I_{0 \text{ nom}}$

V_i, V_{o1} status	D output, V_D
V_i or $V_{o1} < V_t$	low, L, $V_D \leq 0.4$ V at $I_D = 2.5$ mA
V_i and $V_{o1} > V_i + V_n$	high, H, $I_D \leq 25$ μ A at $V_D = 5.25$ V

V_i, V_{o1} status	D output, V_D
V_i or $V_{o1} < V_t$	high, H, $I_D \leq 25 \mu A$ at $V_D = 40 V$
V_i and $V_{o1} > V_i + V_o$	low, L, $V_D \leq 0.4 V$ at $I_D = 20 mA$



If V_i is monitored, the internal input voltage after the input filter and rectifier (EM and LM types) is measured. Consequently, this voltage differs from the voltage at the connector pins by the voltage drop ΔV_{ti} across input filter and rectifier. The threshold level of the D0 and D9 options is adjusted in the factory at nominal output current $I_{o, nom}$ and $T_A = 25^\circ\text{C}$. The value of ΔV_{ti} depends upon input voltage range (AM, BM, etc.), threshold level V_i , temperature, and input current.



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The figure displays timing diagrams for three types of MOSFETs: NPN, JFET, and 11008a. The diagrams are organized into four columns, each representing a different input voltage failure or switch-on cycle scenario. The rows show the drain voltage (V_D), drain current (I_D), gate voltage (V_G), and input voltage (V_i) over time (t).

- Row 1 (V_D):** Shows the drain voltage. For NPN and JFET, it has levels for $V_{D\text{ high}}$ and $V_{D\text{ low}}$. For 11008a, it shows a high level and a low level. The number '3' indicates a specific region or condition.
- Row 2 (I_D):** Shows the drain current. For NPN and JFET, it has levels for $I_{D\text{ high}}$ and $I_{D\text{ low}}$. For 11008a, it shows a high level and a low level.
- Row 3 (V_G):** Shows the gate voltage. For NPN and JFET, it has levels for $V_{G\text{ high}}$ and $V_{G\text{ low}}$. For 11008a, it shows a high level and a low level.
- Row 4 (V_i):** Shows the input voltage. For NPN and JFET, it has levels for $V_{i\text{ high}}$ and $V_{i\text{ low}}$. For 11008a, it shows a high level and a low level.

The four columns represent different input voltage failure or switch-on cycle scenarios:

- Input voltage failure:** Shows a transition from high to low input voltage, with a delay t_h^1 before the drain voltage drops.
- Switch-on cycle:** Shows a transition from low to high input voltage, with a delay $t_{\text{low min}}^4$ before the drain voltage rises.
- Input voltage sag:** Shows a temporary drop in input voltage, with a delay $t_{\text{low min}}^4$ before the drain voltage drops.
- Switch-on cycle and subsequent input voltage failure:** Shows a transition from low to high input voltage, followed by a failure, with delays $t_{\text{low min}}^4$ and $t_{\text{high min}}$ indicated.

The timing diagram for an NPN common-emitter amplifier shows three waveforms over time t . The first waveform is the drain voltage V_D , which is a square wave switching between $V_{D \text{ high}}$ and $V_{D \text{ low}}$. The second waveform is the drain current I_D , which is zero when V_D is high and rises to a peak (labeled 2) and then decays (labeled 3) when V_D transitions to low. The third waveform is the output voltage V_{o1} , which is a square wave between $V_{o1 \text{ nom}}$ and $V_{to} + V_{ho}$ until V_D transitions to high, then decays (labeled 3) and then recovers (labeled 4) to $V_{o1 \text{ nom}}$ when V_D transitions to low. The recovery time is labeled $t_{\text{low min}}$.

- 1 See *Electrical Output Data* for hold-up time.
- 2 With output voltage monitoring the hold-up time $t_h = 0$
- 3 The D signal remains high, if the D output is connected to an external source.

Fig. 23
Relationship between V_i , V_o , V_D , $V_o/V_{o\text{nom}}$ versus time

V ACFAIL signal (VME)

Available for converters with $V_{o1} = 5.1$ V. This option defines an undervoltage monitoring circuit for the input or the input and main output voltage equivalent to option D and generates the ACFAIL signal (V signal), which conforms to the VME standard. The low state level of the ACFAIL signal is specified at a sink current of $I_V = 48$ mA to $V_V \leq 0.6$ V (open-collector output). The pull-up resistor feeding the open-collector output should be placed on the VME backplane.

After the ACFAIL signal has gone low, the VME standard requires a hold-up time t_h of at least 4 ms before the 5.1 V output drops to 4.875 V, when the 5.1 V output is fully loaded. This hold-up time t_h is provided by the internal input capacitance. Consequently the working input voltage and the threshold level V_{ti} should be adequately above the minimum input voltage V_{imin} of the converter, so that enough energy is remaining in the input capacitance. If the input voltage is below the required level, an external hold-up capacitor (C_{iext}) should be added.

Formula for threshold level for desired value of t_h :

$$V_{ti} = \sqrt{\frac{2 \cdot P_o \cdot (t_h + 0.3 \text{ ms}) \cdot 100}{C_{imin} \cdot \eta}} + V_{imin}^2$$

Formula for additional external input capacitor

$$C_{iext} = \frac{2 \cdot P_o \cdot (t_h + 0.3 \text{ ms}) \cdot 100}{\eta \cdot (V_{ti}^2 - V_{imin}^2)} - C_{imin}$$

where as:

C_{imin} = internal input capacitance [mF],
according to table below
 C_{iext} = external input capacitance [mF]
 P_o = output power [W]
 η = efficiency [%]
 t_h = hold-up time [ms]
 V_{imin} = minimum input voltage [V]¹
 V_{ti} = threshold level [V]

Notes: The threshold level V_{ti} of option V2 and V3 is adjusted in the factory to a value according to the table below. A decoupling diode should be connected in series with the input of AM, BM, CM, DM, and FM converters to avoid the input capacitance discharging through other loads connected to the same source voltage. If LM models are powered by AC, an external input capacitor cannot be applied unless an additional rectifier is provided.

Table 20: Available internal input capacitance and factory potentiometer setting of U_{ti} with resulting hold-up time

Types	AM	BM	CM	DM	EM	FM	LM	Unit
C_{imin}	2.6	0.67	0.37	0.14	0.14	0.37	0.14	mF
V_{ti}	9.5	19.5	39	61	104	39	120	VDC
t_h	0.34	0.69	1.92	1.73	6.69	2.92	8.18	ms

Option V operates independently of the built-in input undervoltage lockout circuit. A logic "low" signal is generated at pin 5 as soon as one of the monitored voltages drops below the preselected threshold level V_t . The return for this signal is Vo1– (pin 23). The V output recovers, when the monitored voltage exceeds $V_t + V_h$. The threshold level V_t is either adjustable by a potentiometer, accessible through a hole in the front cover, or adjusted in the factory to a determined customer-specific value. Versions V0, V2 and V3 are available as shown below.

Table 21: Undervoltage monitor functions

V output (VME compatible)	Monitoring		Minimum adjustment range of threshold level V_t		Typical hysteresis U_h [% of V_t] for $V_{tmin} - V_{tmax}$	
	V_i	V_{o1}	V_{ti}	V_{to}	V_{hi}	V_{ho}
V2	yes	no	$V_{imin} - V_{imax}^1$	---	3.0 – 0.5 V	---
V3	yes	yes	$V_{imin} - V_{imax}^1$	$0.95 - 0.98 V_{o1}^2$	3.0 – 0.5 V	"0"
V0	yes	no	$V_{imin} - V_{imax}^{3,4}$	---	2.2 – 0.4 V	---
	yes	yes	$V_{imin} - V_{imax}^{3,4}$	$0.95 - 0.98 V_{o1}^2$	2.2 – 0.4 V	"0"

¹ Threshold level adjustable by potentiometer (not recommended for mobile applications).

² Fixed value between 95% and 98% of V_{o1} (tracking), output undervoltage monitoring is not a requirement of VME standard.

³ Adjusted at I_{onorm} .

⁴ Fixed value, resistor-adjusted ($\pm 2\%$) acc. to customer's specifications; individual type designation is determined by the company.

V output (V0, V2, V3):

Connector pin V is internally connected to the open collector of a NPN transistor. The emitter is connected to the negative potential of output 1. $V_V = 0.6$ V (logic low) corresponds to a monitored voltage level (V_i and/or V_{o1}) $< U_t$. The current I_V through the open collector should not exceed 50 mA. The NPN output is not protected against external overvoltages. V_V should not exceed 80 V.

V_i, V_{o1} status	V output, V_V
V_i or $V_{o1} < V_t$	low, L, $V_V \leq 0.6$ V at $I_V = 50$ mA
V_i and $V_{o1} > V_t + V_h$	high, H, $I_V \leq 25$ μ A at $V_V = 5.1$ V

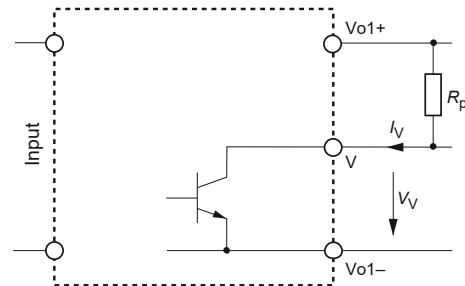


Fig. 24
Output configuration of options V0, V2, V3

Threshold tolerances and hysteresis

V_i is monitored after the input filter and rectifier (EM and LM models). Consequently, this voltage differs from the voltage at the connector pins by the voltage drop ΔV_{ti} across input filter and rectifier. The threshold level of option V0 is factory-adjusted at $I_{o, nom}$ and $T_A = 25$ °C. ΔV_{ti} depends upon the input voltage range (AM, BM, ...), threshold level V_{ti} , temperature, and input current.

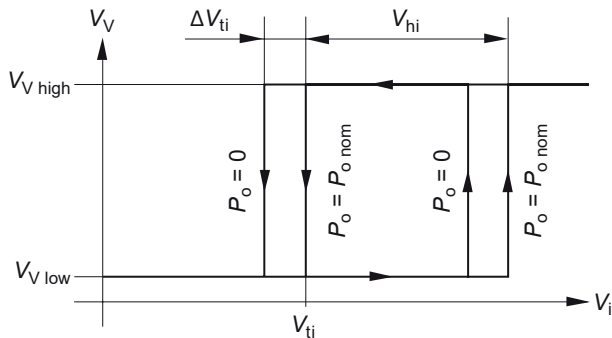


Fig. 25
Definition of V_{ti} , ΔV_{ti} and V_{hi}

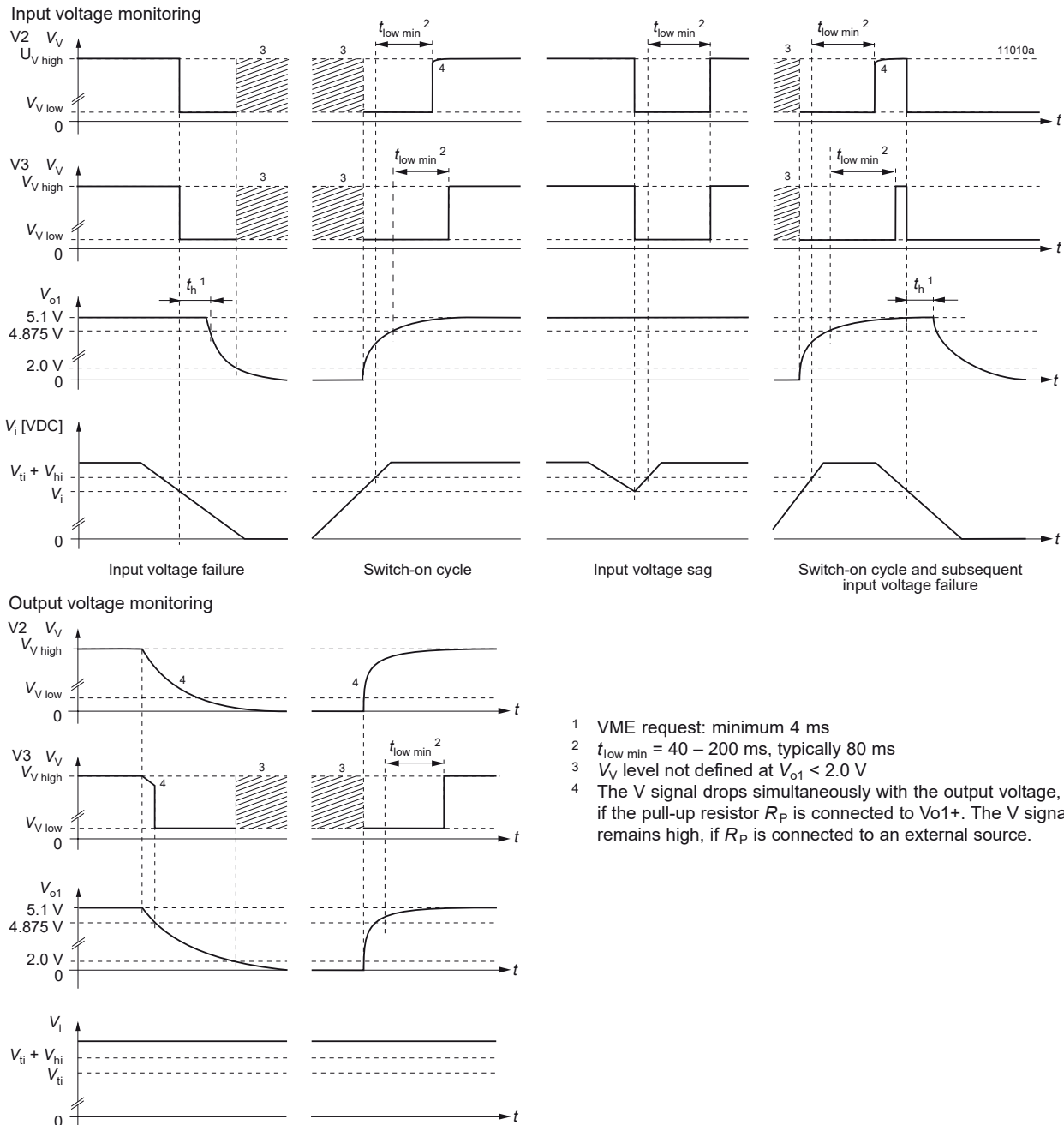


Fig. 26
Relationship between V_i , V_{o1} , V_V , I_V , and $V_{o1}/V_{o\text{nom}}$ versus time.

K Coding Strip

A plastic part across the connector ensures correct population of the DIN-rack.

G RoHS

RoHS-compatible for all six substances. For the dimensions of the cooling plates, see *Mechanical Data*. Option B2 is for custom-specific models with elongated case (for 220 mm DIN-rack depth).

Accessories

A great variety of electrical and mechanical accessories are available including:

- Various mating H11 connectors including solder, fast-on, or press-fit terminals
- Pair of connector retention clips HZZ01209-G
- Code key system: 5 coding wedges HZZ00202-G
- Various front panels for 19" rack mounting
- Flexible H11 PCB board HZZ01208-G for connecting with a mother board
- Universal mounting bracket UMB-LHMQ (HZZ00610-G) for chassis or DIN-rail mounting in upright position.
- DIN-rail mounting brackets DMB-MHQ (HZZ00619-G)
- Mounting plate M (HZZ01208) for chassis or a wall mounting, where only frontal access is given
- Battery sensor [S-KSMH...] for using the converter as battery charger (different cell characteristics).

For additional accessory product information, see the accessory data sheets listed with each product series or individual model at our web site.



Fig. 27
Different front panels

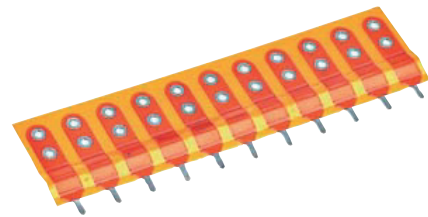


Fig. 28
Flexible H11 PCB (HZZ01208-G)

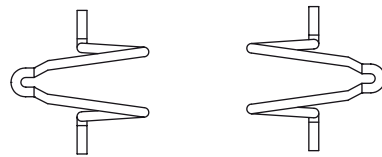


Fig. 29
A pair of connector retention clips (HZZ01209-G)



Fig. 30
Mounting plate M (HZZ01210), connector with fast-on terminals (HZZ00101-G), secured with retention clips (HZZ01209-G)



Fig. 31
Universal mounting bracket for DIN-rail mounting (HZZ00610-G)

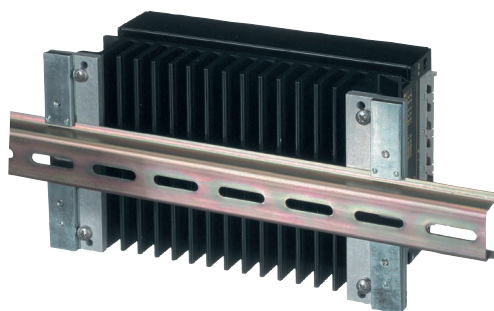


Fig. 32
DIN-rail mounting brackets DMB-MHQ (HZZ00619-G)

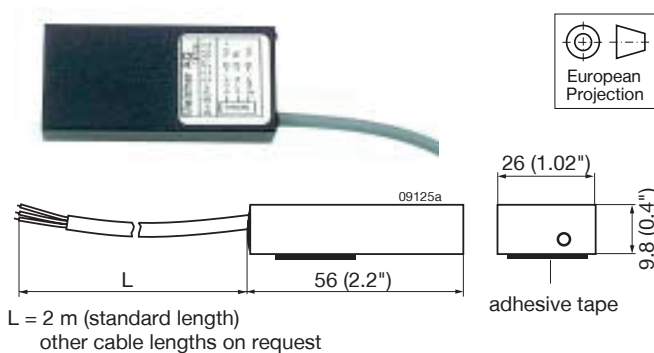


Fig.33
Battery temperature sensor S-KSMH

NUCLEAR AND MEDICAL APPLICATIONS - These products are not designed or intended for use as critical components in life support systems, equipment used in hazardous environments, or nuclear control systems.

TECHNICAL REVISIONS - The appearance of products, including safety agency certifications pictured on labels, may change depending on the date manufactured. Specifications are subject to change without notice.

Данный компонент на территории Российской Федерации

Вы можете приобрести в компании MosChip.

Для оперативного оформления запроса Вам необходимо перейти по данной ссылке:

<http://moschip.ru/get-element>

Вы можете разместить у нас заказ для любого Вашего проекта, будь то серийное производство или разработка единичного прибора.

В нашем ассортименте представлены ведущие мировые производители активных и пассивных электронных компонентов.

Нашей специализацией является поставка электронной компонентной базы двойного назначения, продукции таких производителей как XILINX, Intel (ex.ALTERA), Vicor, Microchip, Texas Instruments, Analog Devices, Mini-Circuits, Amphenol, Glenair.

Сотрудничество с глобальными дистрибьюторами электронных компонентов, предоставляет возможность заказывать и получать с международных складов практически любой перечень компонентов в оптимальные для Вас сроки.

На всех этапах разработки и производства наши партнеры могут получить квалифицированную поддержку опытных инженеров.

Система менеджмента качества компании отвечает требованиям в соответствии с ГОСТ Р ИСО 9001, ГОСТ РВ 0015-002 и ЭС РД 009

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