Toshiba BiCD Integrated Circuit Silicon Monolithic

TB67S101ANG

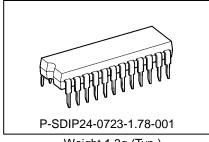
PHASE-in controlled Bipolar Stepping Motor Driver

TB67S101ANG is a two-phase bipolar stepping motor driver using a PWM chopper. The interface is Phase control. Fabricated with the BiCD process, the rating is $50\ V/4.0\ A$

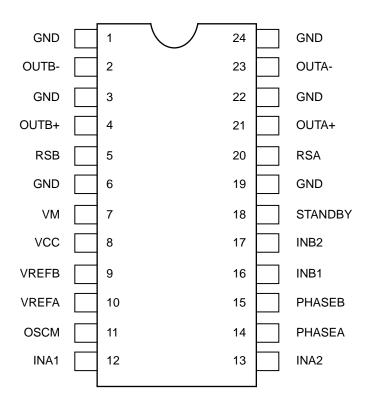
Features

- · BiCD process integrated monolithic IC.
- PWM controlled constant-current drive.
- · Allows full, half, quarter step operation.
- Low on-resistance (High + Low side = 0.49 Ω (typ.)) MOSFET output stage.
- High efficiency motor current control mechanism (Advanced Dynamic Mixed Decay)
- High voltage and current (For specification, please refer to absolute maximum ratings and operation ranges)
- Built-in error detection circuits (Thermal shutdown (TSD), over-current shutdown (ISD), and power-on reset (POR))
- · Built-in VCC regulator for internal circuit use.
- Chopping frequency of a motor can be customized by external resistor and capacitor.
- · Package: P-SDIP24-0723-1.78-001

Note: Please be careful about thermal conditions during use.

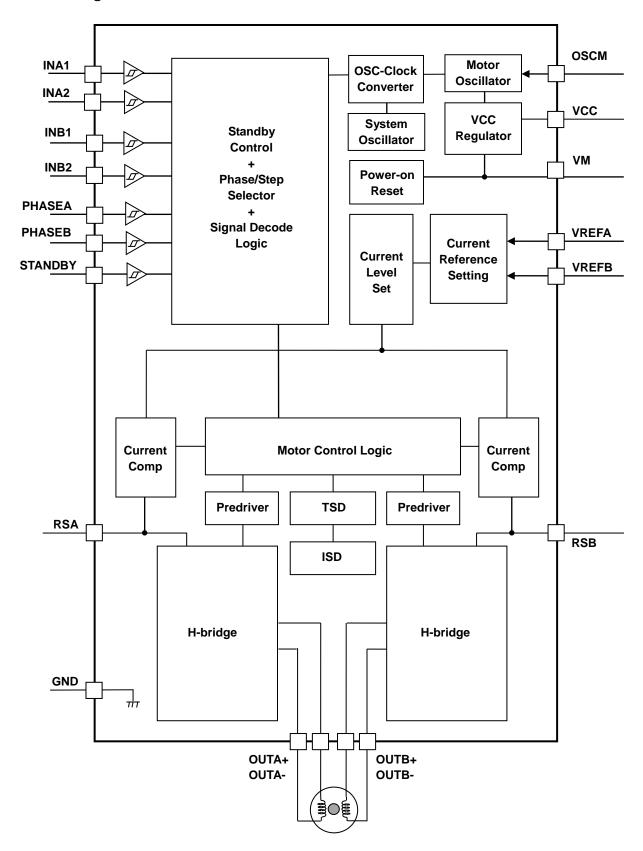


1. Pin assignment (Top View)



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2. Block diagram



Functional blocks/circuits/constants in the block chart etc. may be omitted or simplified for explanatory purposes.

Notes

All the grounding wires of the TB67S101ANG must run on the solder mask on the PCB and be externally terminated at only one point. Also, a grounding method should be considered for efficient heat dissipation.

Careful attention should be paid to the layout of the output, VCC(VM) and GND traces, to avoid short circuits across output pins or to the power supply or ground. If such a short circuit occurs, the device may be permanently damaged.

Also, the utmost care should be taken for pattern designing and implementation of the device since it has power supply pins (VM, RS, OUT, GND) through which a particularly large current may run. If these pins are wired incorrectly, an operation error may occur or the device may be destroyed.

The logic input pins must also be wired correctly. Otherwise, the device may be damaged owing to a current running through the IC that is larger than the specified current.

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3. Pin explanations

Pin No.1 – 24

| Pin No. 1 – 24 | Pin Name | Function |
|----------------|----------|--|
| 1 | GND | Ground pin |
| 2 | OUTB- | Motor Bch (-) output pin |
| 3 | GND | Ground pin |
| 4 | OUTB+ | Motor Bch (+) output pin |
| 5 | RSB | Motor Bch current sense pin |
| 6 | GND | Ground pin |
| 7 | VM | Motor power supply pin |
| 8 | VCC | Internal VCC regulator monitor pin |
| 9 | VREFB | Motor Bch output set pin |
| 10 | VREFA | Motor Ach output set pin |
| 11 | OSCM | Oscillating circuit frequency for chopping set pin |
| 12 | INA1 | Motor Ach excitation control input 1 |
| 13 | INA2 | Motor Ach excitation control input 2 |
| 14 | PHASEA | Current direction signal input for motor Ach |
| 15 | PHASEB | Current direction signal input for motor Bch |
| 16 | INB1 | Motor Bch excitation control input 1 |
| 17 | INB2 | Motor Bch excitation control input 2 |
| 18 | STANDBY | All-function-initializing and Low power dissipation mode |
| 19 | GND | Ground pin |
| 20 | RSA | Motor Ach current sense pin |
| 21 | OUTA+ | Motor Ach (+) output pin |
| 22 | GND | Ground pin |
| 23 | OUTA- | Motor Ach (-) output pin |
| 24 | GND | Ground pin |

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4. Input/Output equivalent circuit

| Pin name | IN/OUT signal | Equivalent circuit |
|---|--|--|
| INA1 INA2 PHASEA INB1 INB2 PHASEB STANDBY | Digital input (VIH/VIL) VIH: 2.0 V (min) to 5.5 V (max) VIL: 0 V (min) to 0.8 V (max) | Logic Input Pin Cy 00 Cy |
| VCC VREFA VREFB | VCC voltage range 4.75 V (min) to 5.0 V (typ.) to 5.25 V (max) VREF voltage range 0 V to 3.6 V | VCC Δ VREF Δ VREF Δ M M M M M M M M M M M M M |
| OSCM | OSCM frequency setting range 0.64 MHz (min) to 1.12 MHz (typ.) to 2.4 MHz (max) | OSCM \square |
| OUTA+ OUTA- OUTB+ OUTB- RSA RSB | VM power supply voltage range 10 V (min) to 47 V (max) OUT pin voltage 10 V (min) to 47 V (max) | RS OUT- |

The equivalent circuit diagrams may be simplified or some parts of them may be omitted for explanatory purposes.

5. Function explanation (Stepping motor)

Motor output current (lout) : The flow from OUT+ to OUT- is plus current. The flow from OUT+ to OUT+ is minus current.

<Full step resolution>

| Ach | | | | Bch | | | |
|--------|------|------|---------|--------|--------|------|---------|
| Input | | | Output | | Output | | |
| PHASEA | INA1 | INA2 | Iout(A) | PHASEB | INB1 | INB2 | lout(B) |
| Н | Н | Н | +100% | Н | Н | Н | +100% |
| L | Н | Н | -100% | Н | Н | Н | +100% |
| L | Н | Н | -100% | L | Н | Н | -100% |
| Н | Н | Н | +100% | L | Н | Н | -100% |

Please set INA1, INA2, INB1, and INB2 to Low until VM power supply reaches the proper operating range.

<Half step resolution>

| Ach | | | | Bch | | | |
|--------|-------|------|---------|--------|-------|------|---------|
| | Input | | Output | | Input | | Output |
| PHASEA | INA1 | INA2 | Iout(A) | PHASEB | INB1 | INB2 | lout(B) |
| Н | Н | Н | +100% | Н | Н | Н | +100% |
| - | L | L | 0% | Н | Н | Н | +100% |
| L | Н | Н | -100% | Н | Н | Н | +100% |
| L | Н | Н | -100% | - | L | L | 0% |
| L | Н | Н | -100% | L | Н | Н | -100% |
| - | L | L | 0% | L | Н | Н | -100% |
| Н | Н | Н | +100% | L | Н | Н | -100% |
| Н | Н | Н | +100% | - | L | L | 0% |

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-: Don't care

<Quarter step resolution>

| Ach | | | | Bch | | | | |
|--------|------|------|---------|--------|-------|------|---------|--|
| Input | | | Output | | Input | | | |
| PHASEA | INA1 | INA2 | Iout(A) | PHASEB | INB1 | INB2 | lout(B) | |
| Н | Н | L | +71% | Н | Н | L | +71% | |
| Н | L | Н | +38% | Н | Н | Н | +100% | |
| Х | L | L | 0% | Н | Н | Н | +100% | |
| L | L | Н | -38% | Н | Н | Н | +100% | |
| L | Н | L | -71% | Н | Н | L | +71% | |
| L | Н | Н | -100% | Н | L | Н | +38% | |
| L | Н | Н | -100% | Х | L | L | 0% | |
| L | Н | Н | -100% | L | L | Н | -38% | |
| L | Н | L | -71% | L | Н | L | -71% | |
| L | L | Н | -38% | L | Н | Н | -100% | |
| Х | L | L | 0% | L | Н | Н | -100% | |
| Н | L | Н | +38% | L | Н | Н | -100% | |
| Н | Н | L | +71% | L | Н | L | -71% | |
| Н | Н | Н | +100% | L | L | Н | -38% | |
| Н | Н | Н | +100% | Х | L | L | 0% | |
| Н | Н | Н | +100% | Н | L | Н | +38% | |

X : Don't care

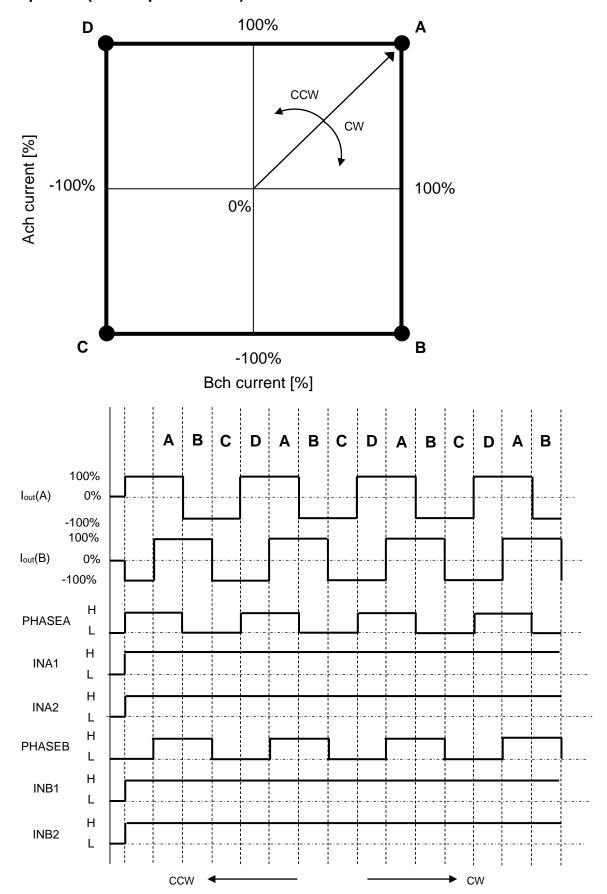
Others

| Pin Name | Н | L | Notes |
|--------------------------|------------------------|--------------------------------|---|
| INA1, INA2 INB1, INB2 | The current value of e | ach ch is set up with 2 value. | Please refer to the above-mentioned current value setting table. |
| PHASEA PHASEB | OUT+: H OUT-: L | OUT+: L OUT-: H | In PHASE=H, Charge current flows in the direction of OUT- from OUT+. |
| STANDBY | Standby release | Standby mode | In STANDBY= L, an internal oscillating circuit and a motor output part are stopped. (The drive of a motor cannot be performed.) |

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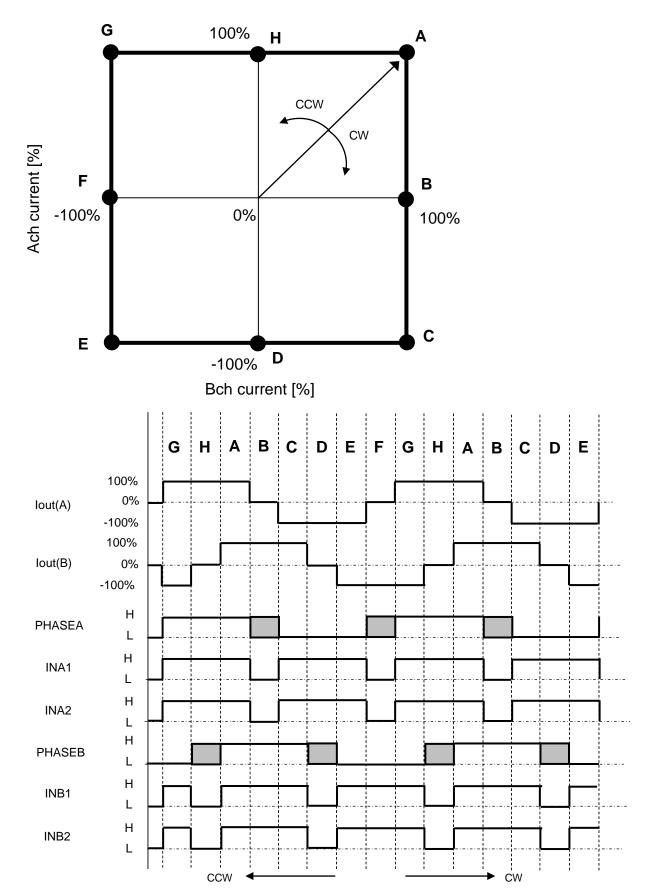
Current phasor (Full step resolution)



Timing charts may be simplified for explanatory purpose.

 $Please\ set\ INA1,\ INA2,\ INB1,\ and\ INB2\ to\ Low\ until\ VM\ power\ supply\ reaches\ the\ proper\ operating\ range.$

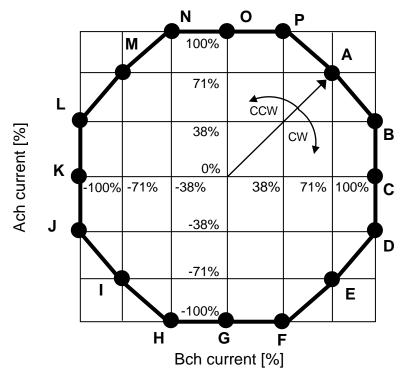
Current phasor (Half step resolution)

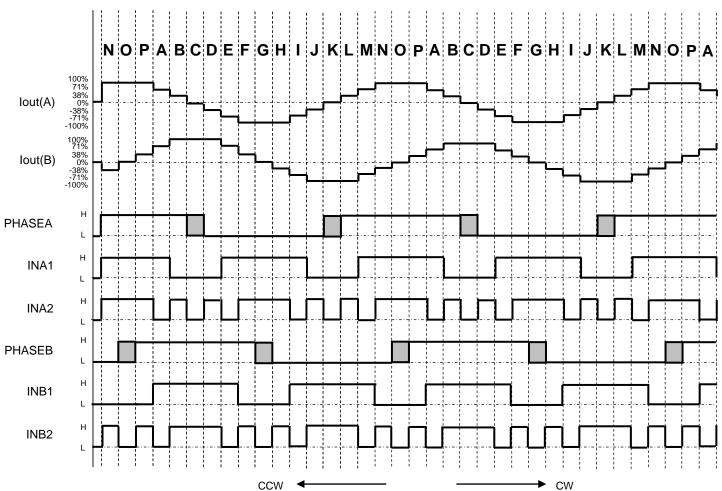


Timing charts may be simplified for explanatory purpose.

 $Please\ set\ INA1,\ INA2,\ INB1,\ and\ INB2\ to\ Low\ until\ VM\ power\ supply\ reaches\ the\ proper\ operating\ range.$

Current phasor (Quarter step resolution)





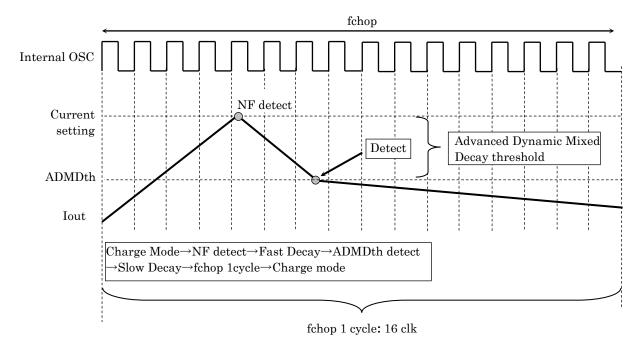
Timing charts may be simplified for explanatory purpose.

 $Please\ set\ INA1,\ INA2,\ INB1,\ and\ INB2\ to\ Low\ until\ VM\ power\ supply\ reaches\ the\ proper\ operating\ range.$

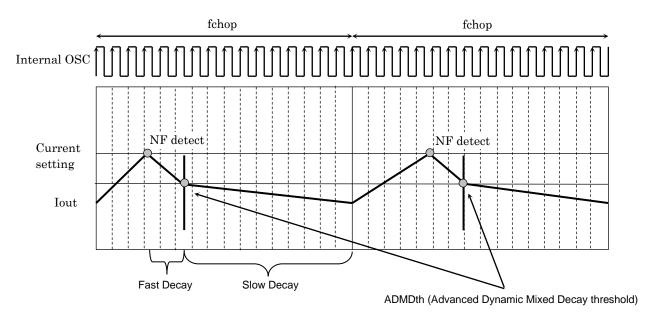
6. Decay function

ADMD(Advanced Dynamic Mixed Decay) constant current control

The Advanced Dynamic Mixed Decay threshold, which determines the current ripple level during current feedback control, is a unique value.



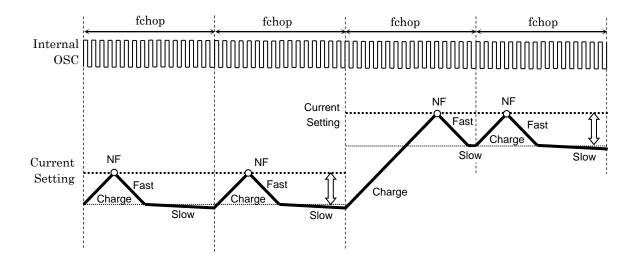
Auto Decay Mode current waveform



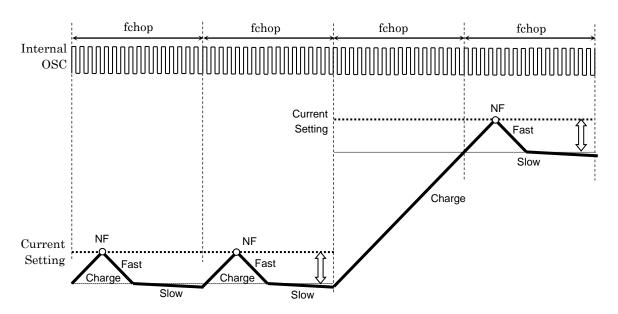
Timing charts may be simplified for explanatory purpose.

ADMD current waveform

•When the next current step is higher :

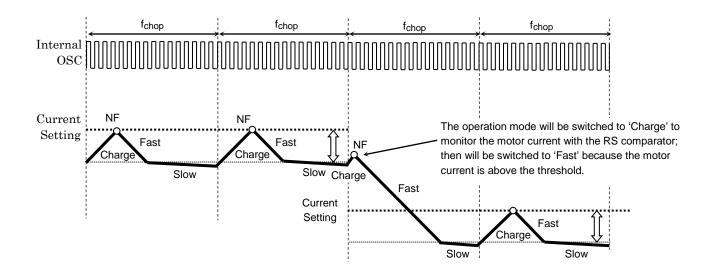


•When Charge period is more than 1 fchop cycle :

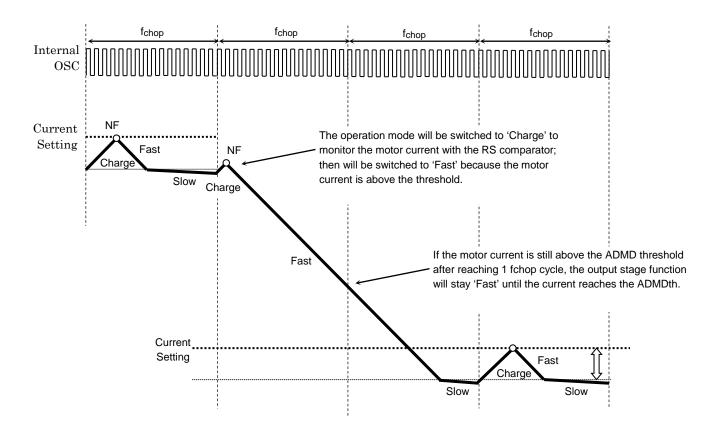


When the Charge period is longer than fchop cycle, the Charge period will be extended until the motor current reaches the NF threshold. Once the current reaches the next current step, then the sequence will go on to decay mode.

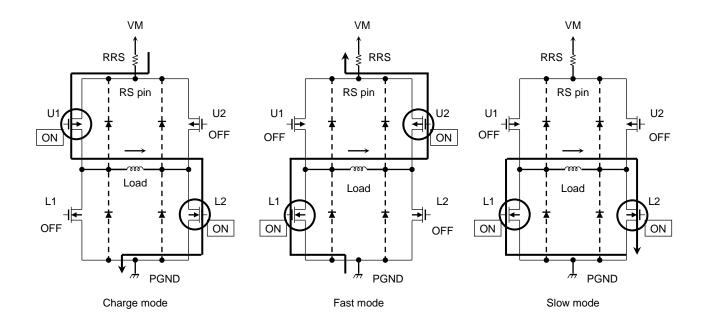
•When the next current step is lower:



•When the Fast continues past 1 fchop cycle (the motor current not reaching the ADMD threshold during 1 fchop cycle)



7. Output transistor function mode



Output MOSFET function

| CLK | U1 | U2 | L1 | L2 |
|--------|-----|-----|-----|-----|
| CHARGE | ON | OFF | OFF | ON |
| SLOW | OFF | OFF | ON | ON |
| FAST | OFF | ON | ON | OFF |

Note: This table shows an example of when the current flows as indicated by the arrows in the figures shown above. If the current flows in the opposite direction, refer to the following table.

| CLK | U1 | U2 | L1 | L2 |
|--------|-----|-----|-----|-----|
| CHARGE | OFF | ON | ON | OFF |
| SLOW | OFF | OFF | ON | ON |
| FAST | ON | OFF | OFF | ON |

This IC controls the motor current to be constant by 3 modes listed above.

The equivalent circuit diagrams may be simplified or some parts of them may be omitted for explanatory purposes.

8. Calculation of the Predefined Output Current

For PWM constant-current control, this IC uses a clock generated by the OSCM oscillator.

The peak output current (Setting current value) can be set via the current-sensing resistor (RS) and the reference voltage (Vref), as follows:

$$lout(max) = Vref(gain) \times \frac{Vref(V)}{RRS(\Omega)}$$

Vref(gain): the Vref decay rate is 1 / 5.0 (typ.)

For example: In the case of a 100% setup when Vref = 3.0 V, Torque = 100%, RRS = 0.51 Ω , the motor constant current (Setting current value) will be calculated as:

lout = $3.0 \text{ V} / 5.0 / 0.51 \Omega = 1.18 \text{ A}$

9. Calculation of the OSCM oscillation frequency (chopper reference frequency)

An approximation of the OSCM oscillation frequency (fOSCM) and chopper frequency (fchop) can be calculated by the following expressions.

```
\begin{split} \text{fOSCM} &= 1/[0.56 \text{ x } \{\text{Cx}(\text{R1} + 500)\}] \\ &\qquad \qquad \text{.......C, R1: External components for OSCM } (\text{C} = 270 \text{ pF} \text{ , R1} = 5.1 \text{ k}\Omega \Longrightarrow \text{fOSCM} = \text{About 1.12 MHz } (\text{Typ.})) \end{split} \begin{aligned} \text{fchop} &= \text{fOSCM} / \text{16} \\ &\qquad \qquad \text{.......fOSCM} &= 1.12 \text{ MHz} \Longrightarrow \text{fchop} = \text{Arround 70 kHz} \end{aligned}
```

If chopping frequency is raised, Ripple of current will become small and wave-like reproducibility will improve. However, the gate loss inside IC goes up and generation of heat becomes large.

By lowering chopping frequency, reduction in generation of heat is expectable. However, Ripple of current may become large. It is a standard about about 70 kHz. A setup in the range of 50 to 100 kHz is recommended.



Absolute maximum ratings (Ta = 25°C)

| Characteristics | Symbol | Rating | Unit | Remarks |
|-----------------------------|---------|------------|------|--------------------------|
| Motor power supply | VM | 50 | V | - |
| Motor output voltage | Vout | 50 | V | - |
| Motor output current | lout | 4.0 | Α | - |
| Internal Logic power supply | VCC | 6.0 | V | When externally applied. |
| La mia in must valta na | VIN(H) | 6.0 | V | - |
| Logic input voltage | VIN(L) | -0.4 | V | - |
| Vref input voltage | Vref | 5.0 | V | - |
| Power dissipation | PD | 1.78 | W | Note1 |
| Operating temperature | Topr | -20 to 85 | °C | - |
| Storage temperature | Tstg | -55 to 150 | °C | - |
| Junction temperature | Tj(max) | 150 | °C | - |

Note1: Device alone (Ta = 25°C). If the Ta exceeds above 25°C, derate PD by 14.2 mW/°C.

Ta: Ambient temperature

Topr: Ambient temperature while the IC is active

Tj: Junction temperature while the IC is active. The maximum junction temperature is limited by the thermal shutdown (TSD) circuitry. It is advisable to keep the maximum current below a certain level so that the maximum junction temperature, Tj (MAX), will not exceed 120°C.

Caution: Absolute maximum ratings

The absolute maximum ratings of a semiconductor device are a set of ratings that must not be exceeded, even for a moment. Do not exceed any of these ratings.

Exceeding the rating (s) may cause device breakdown, damage or deterioration, and may result in injury by explosion or combustion.

The value of even one parameter of the absolute maximum ratings should not be exceeded under any circumstances. The TB67S101ANG does not have overvoltage detection circuit. Therefore, the device is damaged if a voltage exceeding its rated maximum is applied.

All voltage ratings, including supply voltages, must always be followed. The other notes and considerations described later should also be referred to.

Operation Ranges (Ta = -20 to 85°C)

| Characteristics | Symbol | Min | Тур. | Max | Unit | Remarks |
|-----------------------|--------------|-----|------|-----|------|---------------------------|
| Motor power supply | VM | 10 | 24 | 47 | V | |
| Motor output current | lout | - | 1.0 | - | Α | Note1 |
| Logio input voltago | VIN(H) | 2.0 | , | 5.5 | ٧ | Logic input High Level |
| Logic input voltage | VIN(L) | 0 | - | 0.8 | V | Logic input Low Level |
| Phase input frequency | fPHASE | - | - | 400 | kHz | |
| Chopper frequency | fchop(range) | 40 | 70 | 150 | kHz | |
| Vref input voltage | Vref | GND | 2.0 | 3.6 | V | |

Note1: Maximum current for actual usage may be limited by the operating circumstances such as operating conditions (exciting mode, operating time, and so on), ambient temperature, and heat conditions (board condition and so on).

Electrical Specifications 1 (Ta = 25°C, VM = 24 V, unless specified otherwise)

| Characteristic | S | Symbol | Test condition | Min | Тур. | Max | Unit |
|----------------------------|--------------|-----------|--|-----|------|-----|------|
| La sia issaut valta sa | HIGH | VIN(H) | Logic input pin (*) | 2.0 | - | 5.5 | V |
| Logic input voltage | LOW | VIN(L) | Logic input pin (*) | 0 | - | 0.8 | V |
| Logic input hysteresis | voltage | VIN(HYS) | Logic input pin (*) | 100 | - | 300 | mV |
| Logic input current | HIGH | IIN(H) | Logic input voltage = 3.3 V | - | 33 | • | μΑ |
| Logic input current | LOW | IIN(L) | Logic input voltage = 0 V | - | - | 1 | μΑ |
| | | IM1 | Output pins = open, STANDBY = L | - | 2 | 3.5 | mA |
| Dower concump | tion | IM2 | Output pins = open, STANDBY = H | - | 3.5 | 5.5 | mA |
| Power consump | lion | IM3 | Output pins=open Full step resolution | - | 5.5 | 7 | mA |
| Output lookage ourrent | High-side | IOH | VRS = VM = 50 V, Vout = 0 V | | - | 1 | μA |
| Output leakage current | Low-side | IOL | VRS = VM = Vout = 50 V | 1 | | μA | |
| Motor current channel | differential | ∆lout1 | Current differential between Ch | -5 | 0 | 5 | % |
| Motor current setting | accuracy | ∆lout2 | lout = 1.5 A | -5 | 0 | 5 | % |
| RS pin curren | t | IRS | VRS = VM = 24 V | 0 | - | 10 | μΑ |
| Motor output ON-resistance | | Ron(S)_PN | Tj = 25°C, Forward direction (High-side + Low-side) | _ | 0.49 | 0.6 | Ω |

^{*:} VIN(H) is defined as the VIN voltage that causes the outputs (OUTA,OUTB) to change when a pin under test is gradually raised from 0 V. VIN(L) is defined as the VIN voltage that causes the outputs (OUTA, OUTB) to change when the pin is then gradually lowered. The difference between VIN(H) and VIN(L) is defined as the input hysteresis.

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^{*:} When the logic signal is applied to the device whilst the VM power supply is not asserted; the device is designed not to function, but for safe usage, please apply the logic signal after the VM power supply is asserted and the VM voltage reaches the proper operating range.

Electrical Specifications 2 (Ta = 25°C, VM = 24 V, unless specified otherwise)

| Characteristics | Symbol | Test condition | Min | Тур. | Max | Unit |
|--|--------------------|----------------|-------|-------|-------|------|
| Vref input current | Iref | Vref = 2.0 V | - | 0 | 1 | μΑ |
| VCC voltage | VCC | ICC = 5.0 mA | 4.75 | 5.0 | 5.25 | V |
| VCC current | ICC | VCC = 5.0 V | - | 2.5 | 5 | mA |
| Vref gain rate | Vref(gain) | Vref = 2.0 V | 1/5.2 | 1/5.0 | 1/4.8 | _ |
| Thermal shutdown(TSD) threshold (Note1) | T _j TSD | _ | 145 | 160 | 175 | °C |
| VM recovery voltage | VMR | _ | 7.0 | 8.0 | 9.0 | V |
| Over-current detection (ISD) threshold (Note2) | ISD | _ | 4.1 | 4.9 | 5.7 | А |

Note1: About TSD

When the junction temperature of the device reached the TSD threshold, the TSD circuit is triggered; the internal reset circuit then turns off the output transistors. Noise rejection blanking time is built-in to avoid misdetection. Once the TSD circuit is triggered, the device will be set to standby mode, and can be cleared by reasserting the VM power source, or setting the DMODE pins to standby mode. The TSD circuit is a backup function to detect a thermal error, therefore is not recommended to be used aggressively.

Note2: About ISD

When the output current reaches the threshold, the ISD circuit is triggered; the internal reset circuit then turns off the output transistors. Once the ISD circuit is triggered, the device keeps the output off until power-on reset (POR), is reasserted or the device is set to standby mode by DMODE pins. For fail-safe, please insert a fuse to avoid secondary trouble.

Back-EMF

While a motor is rotating, there is a timing at which power is fed back to the power supply. At that timing, the motor current recirculates back to the power supply due to the effect of the motor back-EMF.

If the power supply does not have enough sink capability, the power supply and output pins of the device might rise above the rated voltages. The magnitude of the motor back-EMF varies with usage conditions and motor characteristics. It must be fully verified that there is no risk that the TB67S101A or other components will be damaged or fail due to the motor back-EMF.

Cautions on Overcurrent Shutdown (ISD) and Thermal Shutdown (TSD)

The ISD and TSD circuits are only intended to provide temporary protection against irregular conditions such as an output short-circuit; they do not necessarily guarantee the complete IC safety.

If the device is used beyond the specified operating ranges, these circuits may not operate properly: then the device may be damaged due to an output short-circuit.

The ISD circuit is only intended to provide a temporary protection against an output short-circuit. If such a condition persists for a long time, the device may be damaged due to overstress. Overcurrent conditions must be removed immediately by external hardware.

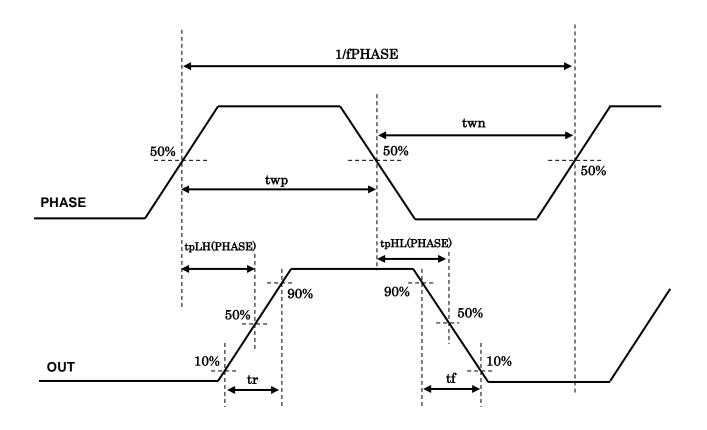
IC Mounting

Do not insert devices incorrectly or in the wrong orientation. Otherwise, it may cause breakdown, damage and/or deterioration of the device.

AC Electrical Specification (Ta = 25°C, VM = 24 V, 6.8 mH/5.7 Ω)

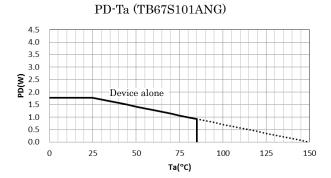
| Characteristics | Symbol | Test condition | Min | Тур. | Max | Unit |
|--------------------------------|-------------|---|-----|------|------|------|
| | fPHASE(min) | _ | 100 | - | 1 | |
| Minimum PHASE pulse width | twp | _ | 50 | - | - | ns |
| | twn | _ | 50 | - | - | |
| | tr | _ | 30 | 80 | 130 | |
| Output transistor | tf | _ | 40 | 90 | 140 | ns |
| switching specific | tpLH(PHASE) | PHASE - Output | 250 | - | 1200 | |
| | tpHL(PHASE) | PHASE - Output | 250 | - | 1200 | |
| Analog noise blanking time | AtBLK | VM = 24 V, lout = 1.5 A Analog tBLK | 250 | 400 | 550 | ns |
| Oscillator frequency accuracy | ⊿fOSCM | $COSC = 270 \text{ pF}, ROSC = 5.1 \text{ k}\Omega$ | -15 | - | +15 | % |
| Oscillator reference frequency | fOSCM | COSC = 270 pF, ROSC = 5.1 k Ω | 952 | 1120 | 1288 | kHz |
| Chopping frequency | fchop | Output: Active(IOUT =1.5 A), fOSCM = 1120 kHz | - | 70 | - | kHz |

AC Electrical Specification Timing chart



Timing charts may be simplified for explanatory purpose.

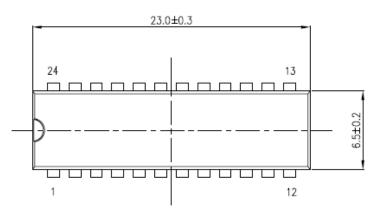
(For reference) Power dissipation and Ambient temperature



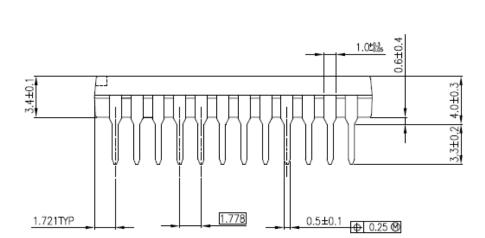
The power dissipation depends on the PCB layout and mounting conditions so please becareful. Also, when the ambient temperature is high, the allowed power disspation will be smaller.

Package Dimensions

P-SDIP24-0723-1.78-001







Weight: 1.3 g (Typ.)

Notes on Contents

Block Diagrams

Some of the functional blocks, circuits, or constants in the block diagram may be omitted or simplified for explanatory purposes.

Equivalent Circuits

The equivalent circuit diagrams may be simplified or some parts of them may be omitted for explanatory purposes.

Timing Charts

Timing charts may be simplified for explanatory purposes.

Application Circuits

The application circuits shown in this document are provided for reference purposes only. Thorough evaluation is required, especially at the mass-production design stage.

Toshiba does not grant any license to any industrial property rights by providing these examples of application circuits.

Test Circuits

Components in the test circuits are used only to obtain and confirm the device characteristics. These components and circuits are not guaranteed to prevent malfunction or failure from occurring in the application equipment.

IC Usage Considerations

Notes on handling of ICs

- (1) The absolute maximum ratings of a semiconductor device are a set of ratings that must not be exceeded, even for a moment. Do not exceed any of these ratings. Exceeding the rating(s) may cause device breakdown, damage or deterioration, and may result in injury by explosion or combustion.
- Use an appropriate power supply fuse to ensure that a large current does not continuously flow in the case of overcurrent and/or IC failure. The IC will fully break down when used under conditions that exceed its absolute maximum ratings, when the wiring is routed improperly or when an abnormal pulse noise occurs from the wiring or load, causing a large current to continuously flow and the breakdown can lead to smoke or ignition. To minimize the effects of the flow of a large current in the case of breakdown, appropriate settings, such as fuse capacity, fusing time and insertion circuit location, are required.
- (3) If your design includes an inductive load such as a motor coil, incorporate a protection circuit into the design to prevent device malfunction or breakdown caused by the current resulting from the inrush current at power ON or the negative current resulting from the back electromotive force at power OFF. IC breakdown may cause injury, smoke or ignition. Use a stable power supply with ICs with built-in protection functions. If the power supply is unstable, the protection function may not operate, causing IC breakdown. IC breakdown may cause injury, smoke or ignition.
- (4) Do not insert devices in the wrong orientation or incorrectly. Make sure that the positive and negative terminals of power supplies are connected properly.
 - Otherwise, the current or power consumption may exceed the absolute maximum rating, and exceeding the rating(s) may cause device breakdown, damage or deterioration, and may result in injury by explosion or combustion.
 - In addition, do not use any device inserted in the wrong orientation or incorrectly to which current is applied even just once.
- (5) Carefully select external components (such as inputs and negative feedback capacitors) and load components (such as speakers), for example, power amp and regulator.

 If there is a large amount of leakage current such as from input or negative feedback capacitor, the IC output DC voltage will increase. If this output voltage is connected to a speaker with low input withstand voltage, overcurrent or IC failure may cause smoke or ignition. (The overcurrent may cause smoke or ignition from the IC itself.) In particular, please pay attention when using a Bridge Tied Load (BTL) connection-type IC that inputs output DC voltage to a speaker directly.

Points to remember on handling of ICs

Overcurrent detection Circuit

Overcurrent detection circuits (referred to as current limiter circuits) do not necessarily protect ICs under all circumstances. If the overcurrent detection circuits operate against the overcurrent, clear the overcurrent status immediately.

Depending on the method of use and usage conditions, exceeding absolute maximum ratings may cause the overcurrent detection circuit to operate improperly or IC breakdown may occur before operation. In addition, depending on the method of use and usage conditions, if overcurrent continues to flow for a long time after operation, the IC may generate heat resulting in breakdown.

Thermal Shutdown Circuit

Thermal shutdown circuits do not necessarily protect ICs under all circumstances. If the thermal shutdown circuits operate against the over-temperature, clear the heat generation status immediately.

Depending on the method of use and usage conditions, exceeding absolute maximum ratings may cause the thermal shutdown circuit to operate improperly or IC breakdown to occur before operation.

Heat Radiation Design

When using an IC with large current flow such as power amp, regulator or driver, design the device so that heat is appropriately radiated, in order not to exceed the specified junction temperature (TJ) at any time or under any condition. These ICs generate heat even during normal use. An inadequate IC heat radiation design can lead to decrease in IC life, deterioration of IC characteristics or IC breakdown. In addition, when designing the device, take into consideration the effect of IC heat radiation with peripheral components.

Back-EMF

When a motor rotates in the reverse direction, stops or slows abruptly, current flows back to the motor's power supply owing to the effect of back-EMF. If the current sink capability of the power supply is small, the device's motor power supply and output pins might be exposed to conditions beyond the absolute maximum ratings. To avoid this problem, take the effect of back-EMF into consideration in system design.

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