

Compact Video Driver Series for DSCs and Portable Devices

Ultra-compact Waferlevel



Chip Size Packeage Output Capacitor-less Single Output Video Drivers

BH76906GU, BH76909GU, BH76912GU, BH76916GU, BH76706GU

No. 09064EAT01

Description

Due to a built-in charge pump circuit, this video driver does not require the large capacity tantalum capacitor at the video output pin that is essential in conventional video drivers. Features such as a built-in LPF that has bands suited to mobile equipment, current consumption of 0 µA at standby, and low voltage operation from as low as 2.5 V make it optimal for digital still cameras, mobile phones, and other equipment in which high density mounting is demanded.

Features

- 1) WLCSP ultra-compact package (1.6 mm x 1.6 mm x 0.75 mm)
- 2) Improved noise characteristics over BH768xxFVM series
- 3) Four video driver amplifier gains in lineup: 6 dB, 9 dB, 12 dB, 16.5 dB
- 4) Large output video driver of maximum output voltage 5.2 Vpp. Ample operation margin for supporting even low voltage operation
- 5) Output coupling capacitor not needed, contributing to compact design
- 6) Built-in standby function and circuit current of 0 µA (typ) at standby
- 7) Clear image playback made possible by built-in 8th-order 4.5 MHz LPF
- 8) Due to use of bias input format, supports not only video signals but also chroma signals and RGB signals
- 9) Due to built-in output pin shunt switch, video output pin can be used as video input pin (BH76706GU)

Mobile phone, digital still camera, digital video camera, hand-held game, portable media player

Line up matrix

Product Name	Product Name Video Driver Amplifier Gain		Video Output Pin Shunt Function
BH76906GU	6dB	1Vpp	
BH76909GU	9dB	0.7Vpp	
BH76912GU	12dB	0.5Vpp	_
BH76916GU	16.5dB	0.3Vpp	
BH76706GU	6dB	1Vpp	0

Absolute Maximum Ratings (Ta = 25 °C)

Parameter	Symbol	Rating	Unit
Supply voltage	Vcc	3.55	V
Power dissipation	Pd	580	mW
Operating temperature range	Topr	-40~+85	°C
Storage temperature range	Tstg	-55~+125	°C

^{*} When mounted on a 50 mm×58 mm×1.6 mm glass epoxy board, reduce by 5.8mW/°C above Ta=+25°C.

Operating Range

Parameter	Symbol	Min.	Тур.	Max.	Unit
Supply voltage	Vcc	2.5	3.0	3.45	V

Electrical Characteristics

[Unless otherwise specified, Typ. : Ta = 25 °C, VCC = 3V]

							e specified, Typ. : Ta = 25 °C, VCC = 3V]		
Parameter	Symbol	BH76906 GU	Ty BH76909 GU	pical Valu BH76912 GU	Jes BH76916 GU	BH76706 GU	Unit	Measurement Conditions	
Circuit current 1-1	I _{CC1-1}			15.0			mΑ	In active mode (No signal)	
Circuit current 1-2	I _{CC1-2}	17.0				mA	In active mode (Outputting NTSC color bar signal)		
Circuit current 2	I_{CC2}			0.0			μA	In standby mode	
Circuit current 3	I _{CC3}		-	_		100	μΑ	In input mode (Applying B3 = 1.5 V)	
Standby switch input current High Level	I _{thH1}		4	! 5			μA	Applying B3 = 3.0 V	
Standby switch switching voltage High Level	V_{thH1}		1.2	√ min		_	V	Active mode	
Standby switch switching voltage Low Level	V_{thL1}		0.45	5V _{max}			V	Standby mode	
Standby switch outflow current High Level	I _{thH2}					0	μA	Applying B3 = 3.0 V	
Standby switch outflow current Middle Level	I _{thM2}					8	μA	Applying B3 = 1.5 V	
Standby switch outflow current Low Level	I _{thL2}					23	μA	Applying B3 = 0 V	
Mode switching voltage High Level	V_{thH2}		-	_		VCC -0.2 (MIN.)	V	Standby mode	
Mode switching voltage Middle Level	V_{thM2}					VCC/2 (TYP.)	V	Input mode	
Mode switching voltage low Level	V_{thL2}					0.2 (MAX.)	V	Active mode	
Voltage gain	G_V	6.0	9.0	12.0	16.5	6.0	dB	Vo=100kHz, 1.0Vpp	
Maximum output level	Vomv			5.2			Vpp	f=10kHz,THD=1%	
Frequency characteristic 1	G _{f1}).2		-0.2	dB	f=4.5MHz/100KHz	
Frequency characteristic 2	G _{f2}			1.5		-1.4	dB	f=8.0MHz/100KHz	
Frequency characteristic 3	G_{f3}			26		-28	dB	f=18MHz/100KHz	
Frequency characteristic 4	G_{f4}		-4	44		-48	dB	f=23.5MHz/100KHz	
Differential gain	D_G			0.5			%	V ₀ =1.0V _p -p Inputting standard staircase Signal	
Differential phase	D _P	1.0		_	deg	V ₀ =1.0V _{p-p} Inputting standard staircase signal			
Y signal to noise ratio	SN_Y	+74	+73	+70	+70	+74	dB	100 kHz~6MHz band Inputting 100% white video signal	
C AM signal to noise ratio	SN _{CA}	+77	+77 +76 +75 +75		+77	dB	100~500 kHz band Inputting 100% chroma video signal		
C PM signal to noise ratio	SN _{CP}	+65			•	dB	100~500 kHz band Inputting 100% chroma video signal		
Current able to flow into output pin	lextin	30				mA	Applying 4.5 V to output pin through 150 Ω		
Output DC offset	Voff	±50max				mV	With no signal Voff = (Vout pin voltage) ÷ 2		
Input impedance	Rin	150				kΩ	Measure inflowing current when applying A3 = 1 V		
Output pin shunt switch on resistance	Ron			_		3	Ω	111 7.19.10	

●Test Circuit Diagram

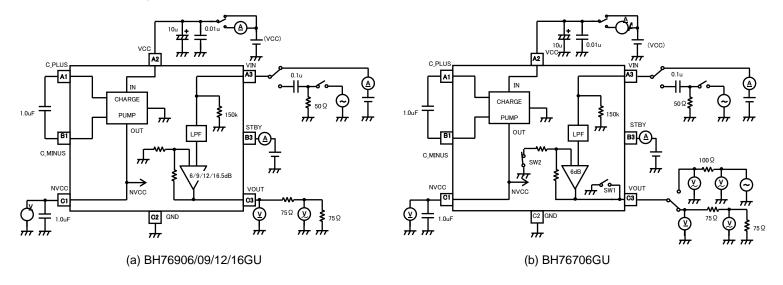
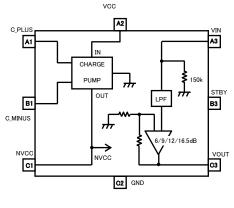


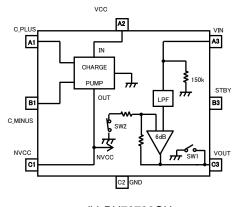
Fig. 1

※ A test circuit is a circuit for shipment inspection and differs from an application circuit example.

●Block Diagram



(a) BH76906/09/12/16GU



(b) BH76706GU

Fig. 2

Operation Logic

BH769xxGU

STBY Pin Logic	Operating Mode	
Н	Active	
L	Standby	
OPEN	Standby	

BH76706GU

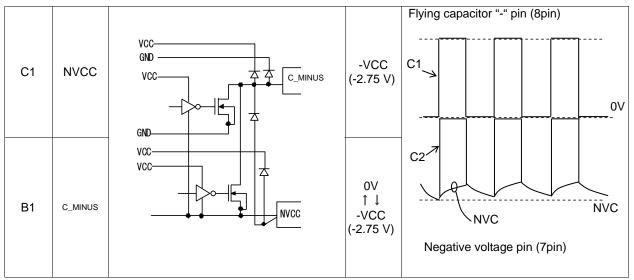
STBY Pin Logic	Operating Mode	SW1	SW2
Н	Standby	OFF	OFF
M	Input (Record)	ON	OFF
L	L Active (Playback)		ON

●Pin Descriptions

Pin Descr	iptions				
Ball	Pin Name	Pin Internal Equivalent Circuit Diagram	DC Voltage	Functional Description	
A1	C_PLUS	VCC VCC VCC GND GND WCC	+VCC ↑ ↓ 0V	Flying capacitor "+" pin See functional descriptions of 7pin, 8pin	
A2	VCC	invo	VCC	VCC pin	
A3	VIN	VCC VIN 100 3.9k 3.9k 150K	0V	Video signal input pin VIN 1 \(\mu \) F Suitable input signals include composite video signals, chroma signals, R.G.B. signals	
В3	STBY	BH769xxGU VCC STBY 250K 200K GND GND STBY CCC STBY STBY STBY ACC STBY S	VCC to 0V	ACTIVE/STANBY switching pin Pin Voltage MODE 1.2 V~VCC (H) ACTIVE 0 V~0.45 V STANBY MODE switching pin Pin Voltage MODE 2.8 V~VCC (H) STANBY 1.3 V~1.7 V (M) GND (Record) 0 V~0.2 V (L) (Playback)	
С3	VOUT	VCC NVCC NVCC BH76706GU only	0V	Video signal output pin VOUT 75 Ω 75 Ω 75 Ω	
C2	GND	VCC GND ,	0V	GND pin	

Note 1) DC voltages in the figure are those when VCC = 3.0 V. Moreover, these values are reference values which are not guaranteed.

Note 2) Numeric values in the figure are settings which do not guarantee ratings.

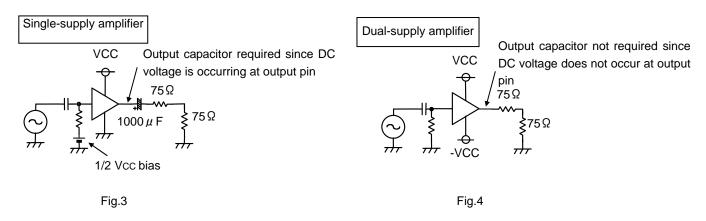


Note 1) DC voltages in the figure are those when VCC = 3.0 V. Moreover, these values are reference values which are not guaranteed.

Note 2) Numeric values in the figure are settings which do not guarantee ratings.

Description of Operation

1) Principles of output coupling capacitorless video drivers



For an amplifier operated from a single power supply (single-supply), since the operating point has a potential of approximately 1/2 Vcc, a coupling capacitor is required for preventing direct current in the output. Moreover, since the load resistance is 150 Ω (75 Ω + 75 Ω) for the video driver, the capacity of the coupling capacitor must be on the order of 1000 μ F if you take into account the low band passband. (Fig.3)

For an amplifier operated from dual power supplies (± supply), since the operating point can be at GND level, a coupling capacitor for preventing output of direct current is not needed.

Moreover, since a coupling capacitor is not needed, in principle, there is no lowering of the low band characteristic at the output stage. (Fig.4)

2) Occurrence of negative voltage due to charge pump circuit

A charge pump, as shown in Fig. 5, consists of a pair of switches (SW1, SW2) and a pair of capacitors (flying capacitor, anchor capacitor). Switching the pair of switches as shown in Fig. 5 causes a negative voltage to occur by shifting the charge in the flying capacitor to the anchor capacitor as in a bucket relay.

In this IC, by applying a voltage of +3 V, a negative voltage of approximately -2.8 V is obtained.

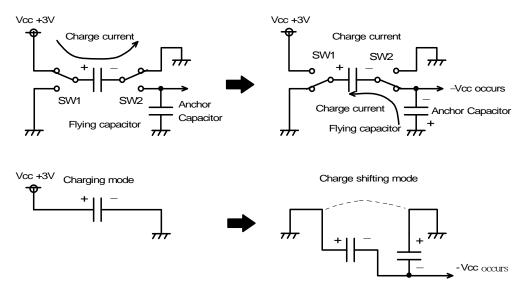


Fig.5 Principles of Charge Pump Circuit

3) Configuration of BH769xxGU and BH76706GU

As shown in Fig. 6, a BH769xxGU or BH76706GU is a dual-supply amplifier and charge pump circuit integrated in one IC. Accordingly, while there is +3 V single-supply operation, since a dual-supply operation amplifier is used, an output coupling capacitor is not needed.

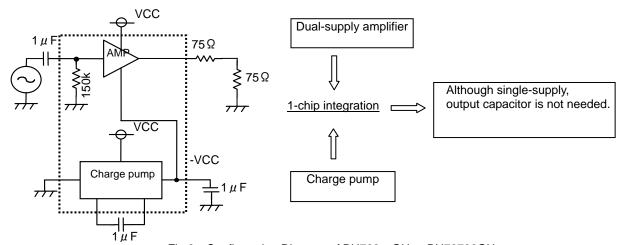


Fig.6 Configuration Diagram of BH769xxGU or BH76706GU

4) Input pin format and sag characteristic

While a BH769xxGU or BH76706GU is a low voltage operation video driver, since it has a large dynamic range of approximately 5.2 Vpp, a resistance termination method that is compatible regardless of signal form (termination by 150 $k\Omega$) is used, and not a clamp method that is an input method exclusively for video signals.

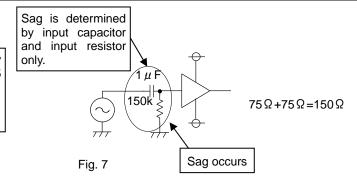
Therefore, since a BH769xxGU or BH76706GU operates normally even if there is no synchronization signal in the input signal, it is compatible with not only normal video signals but also chroma signals and R.G.B. signals and has a wide application range.

Moreover, concerning sag (lowering of low band frequency) that occurs at the input pin and becomes a problem for the resistance termination method, since the input termination resistor is a high 150 k Ω , even if it is combined with a small capacity input capacitor, a sag characteristic that is not a problem in actual use is obtained.

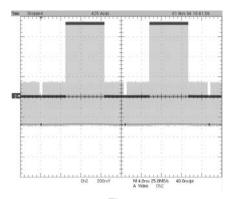
In evaluating the sag characteristic, it is recommended that you use an H-bar signal in which sag readily stands out. (Fig. 8 to Fig. 10)

Input capacitor and input impedance cutoff frequency is the same as when output capacitor in generic 75 Ω driver is made 1000 $\mu \text{F}.$

1 μ F x 150 $k\dot{\Omega}$ = 1000 μ F x 150 Ω (Input pin time constant) (Output pin time constant)



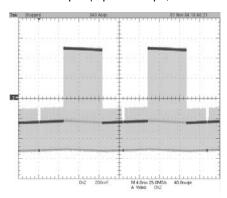
a) Video signal without sag (TG-7/1 output, H-bar)





TV screen output image of H-bar signal

Fig. 8 b) BH769xxGU or BH76706GU output (Input = 1.0 μ F, TG-7/1 output, H-bar)



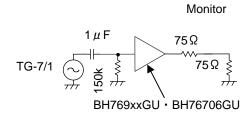


Fig. 9

c) $1000 \mu F + 150 \Omega$ sag waveform (TG-7/1 output, H-bar)

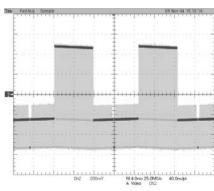
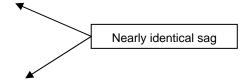
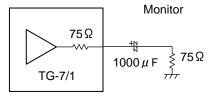


Fig. 10

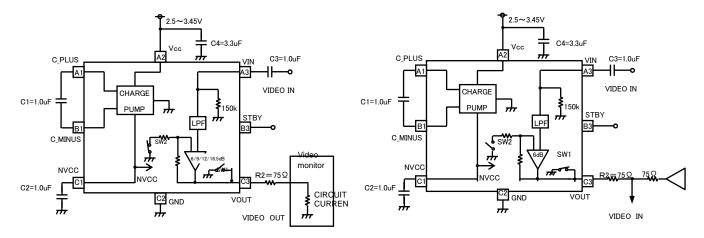




Application Circuit Example

At playback (Active mode)

Recording (Input mode) BH76706GU only



* SW1 and SW2 are built-in BH76706GU only

See page 3/16 for STBY pin logic in each mode

Fig.11

We are confident in recommending the above application circuit example, but we ask that you carefully check not just the static characteristics but also transient characteristics of this circuit before using it.

Caution on use

- Wiring from the decoupling capacitor C4 to the IC should be kept as short as possible.
 Moreover, this capacitor's capacitance value may have ripple effects on the IC, and may affect the S-N ratio for signals, so we recommend using as large a decoupling capacitor as possible. (Recommended C4: 3.3 μF, B characteristics, 6.3 V or higher maximum voltage)
- Make mount board patterns follow the layout example shown on page 10 as closely as possible.
- Capacitors to use In view of the temperature characteristics, etc., we recommend a ceramic capacitor with B characteristics.
- 3. The NVCC (C1 pin) terminal generates a voltage that is used within the IC, so it should never be connected to a load unless absolutely necessary. Moreover, this capacitor (C2) has a large capacitance value but very little negative voltage ripple.

(Recommended C2: 1.0 μF, B characteristic, 6.3 V or higher maximum voltage)

- 4. Capacitors C1 and C4 should be placed as close as possible to the IC. If the wiring to the capacitor is too long, it can lead to intrusion of switching noise. (Recommended C1: 1.0 μF, B characteristics, 6.3 V or higher maximum voltage)
- 5. The HPF consists of input coupling capacitor C3 and 150 kΩ of internal input impedance. Be sure to check for video signal sag before determining the C3 value. The cut-off frequency fc can be calculated using the following formula. fc = 1/(2 π × C3 × 150k Ω) (Recommended C3: 1.0 μF, B characteristic, 6.3 V or higher maximum voltage)
- 6. The output resistor R2 should be placed close to the IC.
- 7. If the IC is mounted in the wrong direction, there is a risk of damage due to problems such as inverting VCC and GND. Be careful when mounting it.
- 8. A large current transition occurs in the power supply pin when the charge pump circuit is switched. If this affects other ICs (via the power supply line), insert a resistor (approximately 10 Ω) in the VCC line to improve the power supply's ripple effects. Although inserting a 10 Ω resistor lowers the voltage by about 0.2 V, this IC has a wide margin for low-voltage operation, so dynamic range problems or other problems should not occur. (See Figures 12 to 14.)

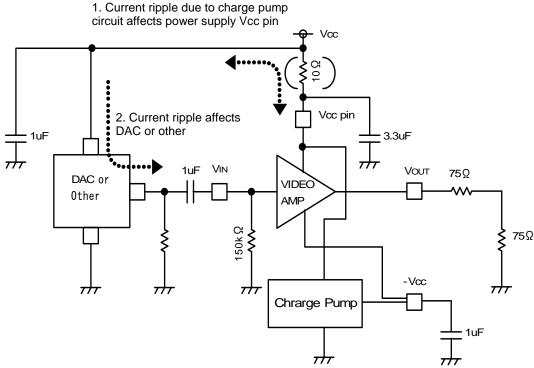


Fig.12 Effects of Charge Pump Circuit Current Ripple on External Circuit

1) Decoupling capacitor only

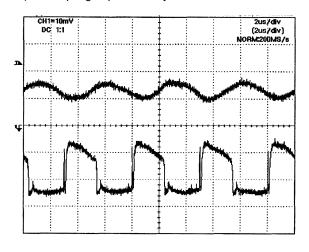


Fig.13

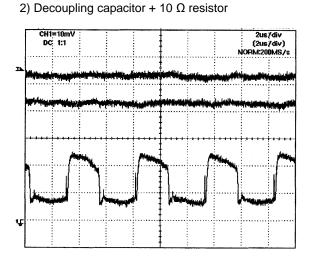
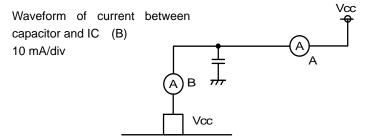


Fig.14

Waveform of current between power supply and capacitor (A) 10 mA/div



Waveform of current between power supply and capacitor (A)

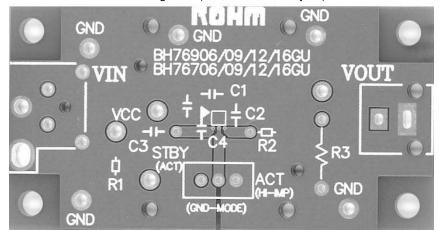
10 mA/div

Waveform of current between resistor and capacitor (B)

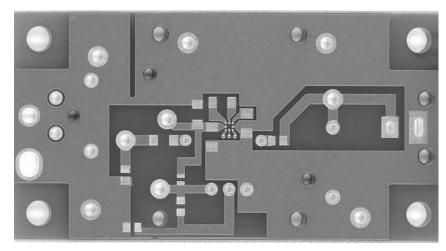
10 mA/div

Waveform of current between capacitor and IC (C) 10 mA/div

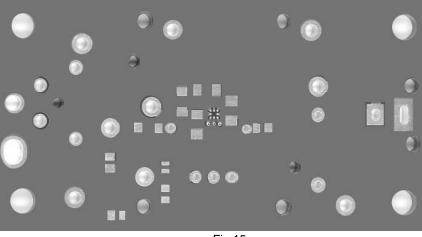
● Evaluation Board Pattern Diagram (Double-sided, 2 layers)



Layer 1 wiring + Silkscreen legend



Layer 2 wiring

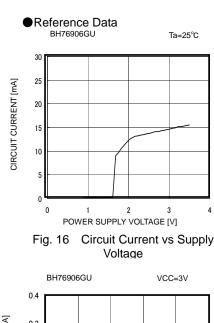


Solder pattern

Parts List

Fig.15

Symbol	Function	Recommended Value	Remarks
C1	Flying capacitor	1 μ F	B characteristic recommended
C2	Tank capacitor	1 μ F	B characteristic recommended
C3	Input coupling capacitor	1μF	B characteristic recommended
C4	Decoupling capacitor	3.3 μ F	B characteristic recommended
R1	Input termination resistor	75Ω	Needed when connected to video signal measurement set
R2	Output resistor	75Ω	_
R3	Output termination resistor	75Ω	Not needed when connected to TV or video signal measurement set
	Input connector	BNC	
	Output connector	RCA (Pin jack)	



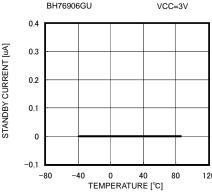


Fig. 19 Standby Circuit Current vs Ambient Temperature

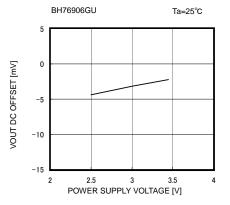
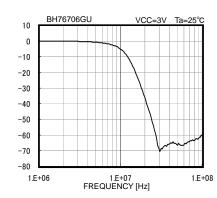


Fig. 22 VOUT Pin Output DC Offset vs Supply Voltage



VOLTAGE GAIN [dB]

Fig. 25 Frequency Characteristic

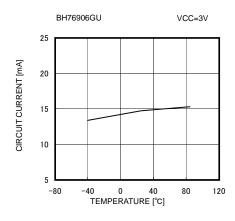


Fig. 17 Circuit Current vs Ambient Temperature

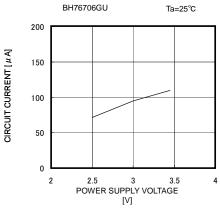


Fig. 20 GND Mode Circuit Current vs Supply Voltage

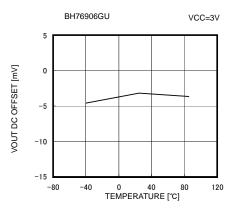


Fig. 23 VOUT Pin Output DC Offset vs Ambient Temperature

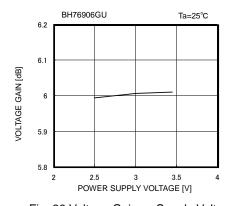


Fig. 26 Voltage Gain vs Supply Voltage

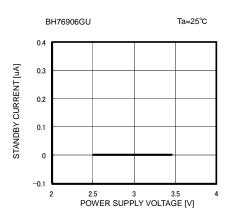
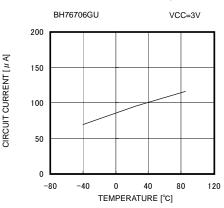


Fig. 18 Standby Circuit Current vs Supply Voltage



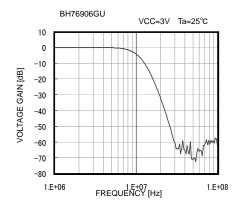


Fig. 24 Frequency Characteristic

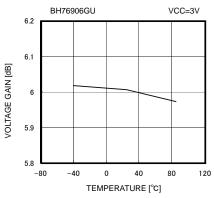


Fig. 27 Voltage Gainvs Ambient Temperature

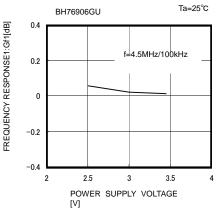


Fig. 28 Frequency Characteristic 1 vs Supply Voltage

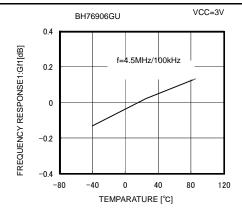


Fig. 29 Frequency Characteristic 1 vs Ambient Temperature

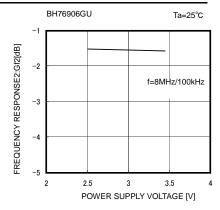


Fig. 30 Frequency Characteristic 2 vs Supply Voltage

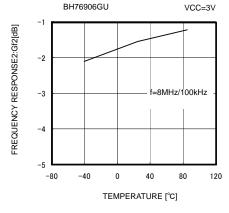


Fig. 31 Frequency Characteristic 2 vs Ambient Temperature

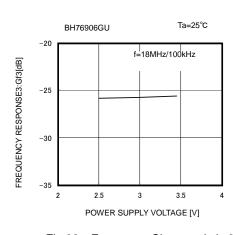
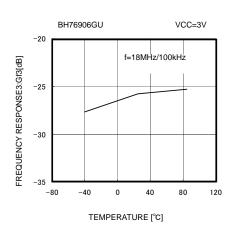


Fig.32 Frequency Characteristic 3 vs Supply Voltage



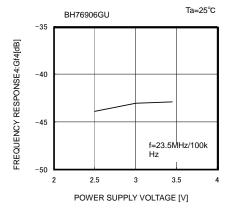


Fig. 34 Frequency Characteristic4 vs Supply Voltage

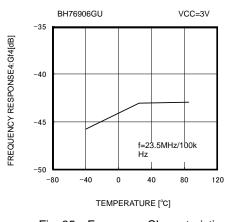
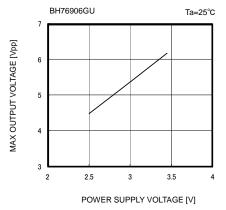


Fig. 35 Frequency Characteristic 4 vs Ambient Temperature



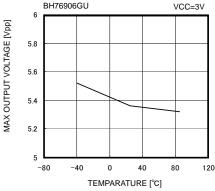


Fig. 37 Max. Output Level vs Ambient Temperature

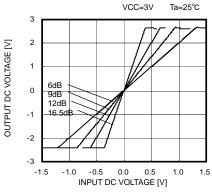


Fig. 38 DC I/O Characteristic

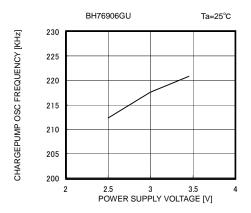


Fig. 39 Charge Pump Oscillation Frequency vs Supply Voltage

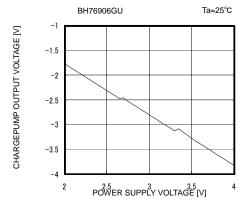


Fig. 41 Charge Pump Output Voltage vs Supply Voltage

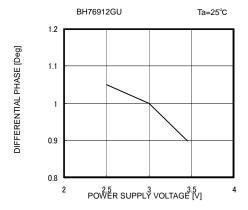


Fig. 43 Differential Phase vs Supply Voltage

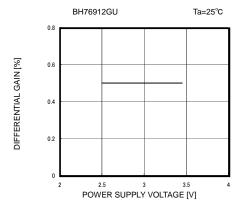


Fig. 45 Differential Gain vs Supply Voltage

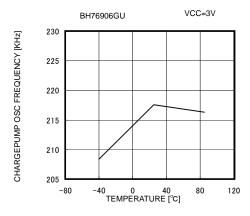


Fig. 40 Charge Pump Oscillation Frequency vs Ambient Temperature

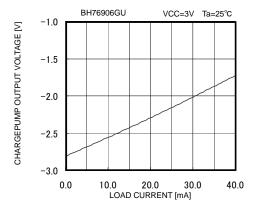
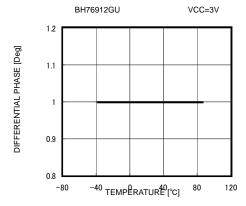


Fig. 42 Charge Pump Load Regulation



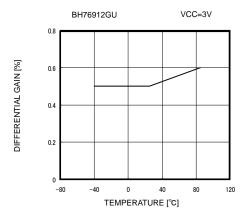


Fig. 46 Differential Gain vs Ambient Temperature

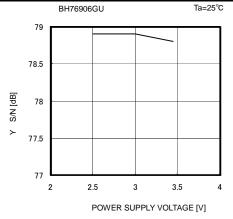


Fig. 47 Y S/N vs Supply Voltage

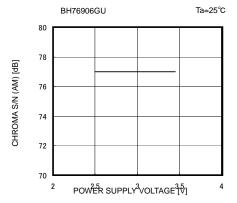


Fig. 49 C AM S/N vs Supply Voltage

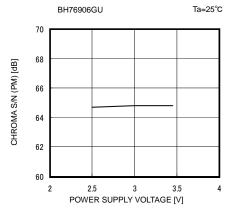


Fig. 51 C PM S/N vs Supply Voltage

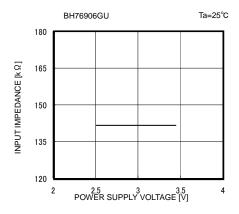


Fig. 53 Input Impedance vs Supply Voltage

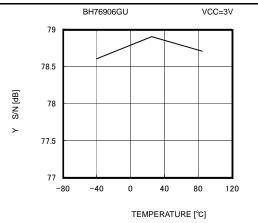


Fig.48 Y S/N vs Ambient Temperature

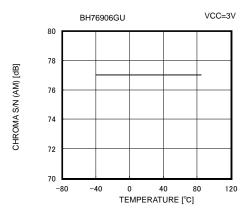


Fig. 50 C AM S/N vs Ambient Temperature

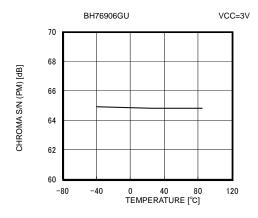


Fig. 52 C PM S/N vs Ambient Temperature

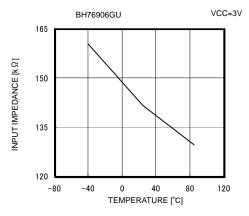


Fig. 54 Input Impedance vs Ambient Temperature

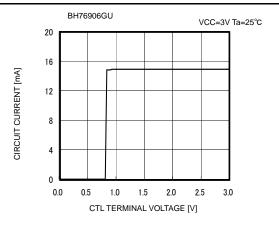


Fig. 55 Control Pin Characteristic

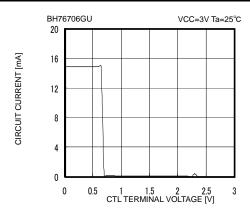


Fig. 56 Control Pin Characteristic

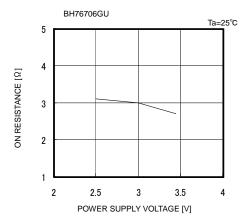


Fig. 57 Output Pin Shunt Switch On Resistance vs Supply Voltage

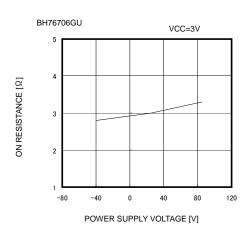


Fig. 58 Output Pin Shunt Switch On Resistance vs Ambient Temperature

Performing separate electrostatic damage countermeasures When adding an externally attached electrostatic countermeasure element to the output pin, connect a varistor in the position shown in Fig. 59 (if connected directly to the output pin, the IC could oscillate depending on the capacity of the varistor). For this IC, since the output waveform is GND-referenced and swings positive and negative, a normal Zener diode cannot be used.

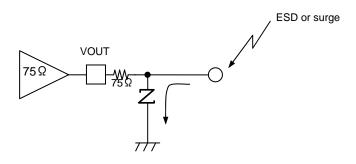
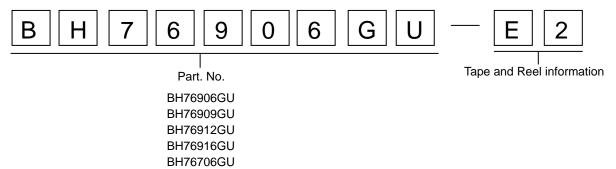
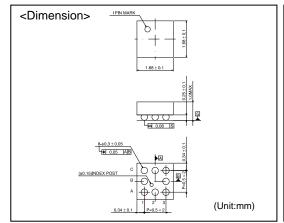


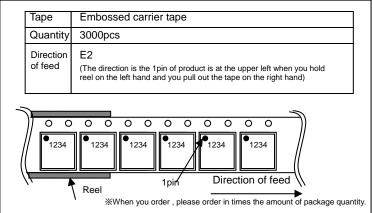
Fig.59 Using Externally Attached Varistor

Selection of order type



VCSP85H1





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(Note1) Medical Equipment Classification of the Specific Applications

JAPAN	USA	EU	CHINA	
CLASSⅢ	CLASSⅢ	CLASS II b	CLASSIII	
CLASSIV	CLASSIII	CLASSⅢ	CLASSIII	

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 - [c] Use of our Products in places where the Products are exposed to sea wind or corrosive gases, including Cl₂, H₂S, NH₃, SO₂, and NO₂
 - [d] Use of our Products in places where the Products are exposed to static electricity or electromagnetic waves
 - [e] Use of our Products in proximity to heat-producing components, plastic cords, or other flammable items
 - [f] Sealing or coating our Products with resin or other coating materials
 - [g] Use of our Products without cleaning residue of flux (even if you use no-clean type fluxes, cleaning residue of flux is recommended); or Washing our Products by using water or water-soluble cleaning agents for cleaning residue after soldering
 - [h] Use of the Products in places subject to dew condensation
- 4. The Products are not subject to radiation-proof design.
- 5. Please verify and confirm characteristics of the final or mounted products in using the Products.
- 6. In particular, if a transient load (a large amount of load applied in a short period of time, such as pulse. is applied, confirmation of performance characteristics after on-board mounting is strongly recommended. Avoid applying power exceeding normal rated power; exceeding the power rating under steady-state loading condition may negatively affect product performance and reliability.
- 7. De-rate Power Dissipation (Pd) depending on Ambient temperature (Ta). When used in sealed area, confirm the actual ambient temperature.
- 8. Confirm that operation temperature is within the specified range described in the product specification.
- 9. ROHM shall not be in any way responsible or liable for failure induced under deviant condition from what is defined in this document.

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- 2. In principle, the reflow soldering method must be used; if flow soldering method is preferred, please consult with the ROHM representative in advance.

For details, please refer to ROHM Mounting specification

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This Product is electrostatic sensitive product, which may be damaged due to electrostatic discharge. Please take proper caution in your manufacturing process and storage so that voltage exceeding the Products maximum rating will not be applied to Products. Please take special care under dry condition (e.g. Grounding of human body / equipment / solder iron, isolation from charged objects, setting of lonizer, friction prevention and temperature / humidity control).

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 - [b] the temperature or humidity exceeds those recommended by ROHM
 - the Products are exposed to direct sunshine or condensation
 - [d] the Products are exposed to high Electrostatic
- 2. Even under ROHM recommended storage condition, solderability of products out of recommended storage time period may be degraded. It is strongly recommended to confirm solderability before using Products of which storage time is exceeding the recommended storage time period.
- 3. Store / transport cartons in the correct direction, which is indicated on a carton with a symbol. Otherwise bent leads may occur due to excessive stress applied when dropping of a carton.
- Use Products within the specified time after opening a humidity barrier bag. Baking is required before using Products of which storage time is exceeding the recommended storage time period.

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