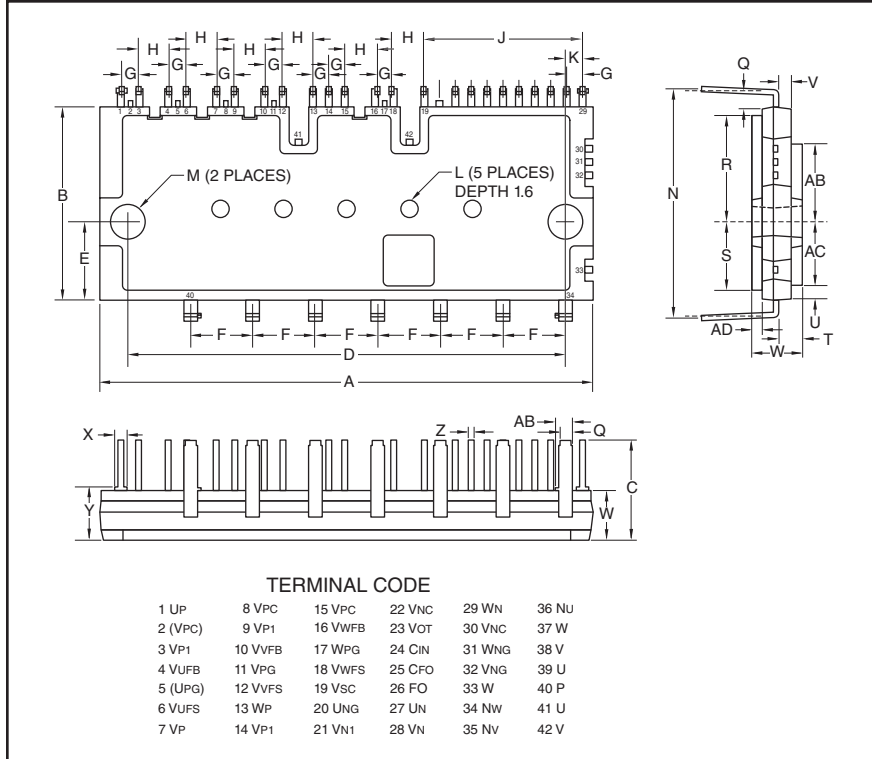


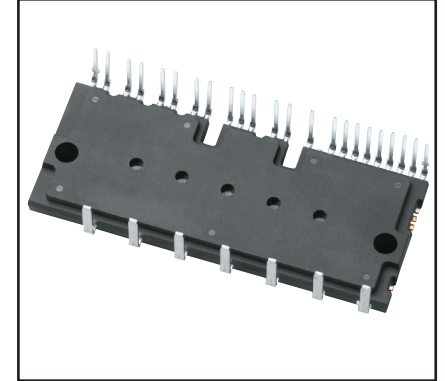
Intellimod™ Module Dual-In-Line Intelligent Power Module 75 Amperes/600 Volts



Outline Drawing and Circuit Diagram

Dimensions	Inches	Millimeters
A	3.11±0.02	79.0±0.5
B	1.22±0.02	31.0±0.5
C	0.63	16.0
D	2.76±0.01	70.0±0.3
E	0.5	12.7
F	0.39±0.01	10.0±0.3
G	0.1±0.01	2.54±0.3
H	0.2±0.01	5.08±0.3
J	1.0	25.4
K	0.11	2.8
L	0.12 Dia.	2.9 Dia.
M	0.18±0.01 Dia.	4.5±0.2 Dia.
N	1.42±0.02	36.2±0.5
P	0.03	0.7

Dimensions	Inches	Millimeters
Q	0.08	2.0
R	0.66	16.73
S	0.44	11.13
T	0.15±0.04	3.8±1.0
U	0.082	2.1
V	0.086	2.2
W	0.31	8.0
X	0.07	1.8
Y	0.34	8.6
Z	0.03	0.8
AA	0.10	2.7
AB	0.48	12.33
AC	0.39	10.12
AD	0.068	1.75



Description:

DIPIMs are intelligent power modules that integrate power devices, drivers, and protection circuitry. Design time is reduced by the use of application-specific HVICs and value-added features such as linear temperature feed-back. Overall efficiency and reliability are increased by the use of full gate CSTBT technology and low thermal impedance.

Features:

- ☐ Low-loss, Full Gate CSTBT IGBTs
- ☐ Single Power Supply
- ☐ Integrated HVICs
- ☐ Direct Connection to CPU

Applications:

- ☐ Variable Speed Pumps
- ☐ Variable Speed Compressors
- ☐ Small Motor Control

Ordering Information:

PS21A7A is a 600V, 75 Ampere DIP Intelligent Power Module.

PS21A7A
Intellimod™ Module
Dual-In-Line Intelligent Power Module
 75 Amperes/600 Volts

Absolute Maximum Ratings, $T_j = 25^\circ\text{C}$ unless otherwise specified

Characteristics	Symbol	PS21A7A	Units
Self-protection Supply Voltage Limit (Short Circuit Protection Capability)*	$V_{CC(\text{prot.})}$	400	Volts
Module Case Operation Temperature (See T_C Measurement Point Below)	T_C	-20 to 100	$^\circ\text{C}$
Storage Temperature	T_{stg}	-40 to 125	$^\circ\text{C}$
Mounting Torque, M4 Mounting Screws	—	13	in-lb
Module Weight (Typical)	—	65	Grams
Isolation Voltage, AC 1 minute, 60Hz Sinusoidal, Connection Pins to Heatsink Plate	V_{ISO}	2500	Volts

IGBT Inverter Sector

Supply Voltage (Applied between P-NU, NV, NW)	V_{CC}	450	Volts
Supply Voltage, Surge (Applied between P-NU, NV, NW)	$V_{CC(\text{surge})}$	500	Volts
Collector-Emitter Voltage ($T_C = 25^\circ\text{C}$)	V_{CES}	600	Volts
Collector Current ($T_C = 25^\circ\text{C}$)	$\pm I_C$	75	Amperes
Peak Collector Current ($T_C = 25^\circ\text{C}$, <1ms)	$\pm I_{CP}$	150	Amperes
Collector Dissipation ($T_C = 25^\circ\text{C}$, per 1 Chip)	P_C	162	Watts
Power Device Junction Temperature**	T_j	-20 to 150	$^\circ\text{C}$

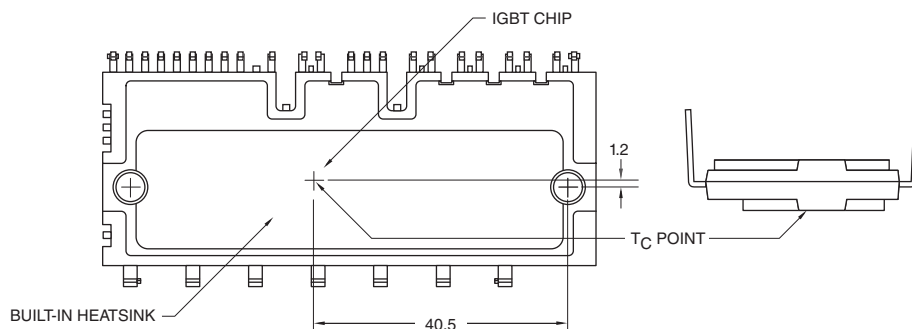
Control Sector

Supply Voltage (Applied between V_{P1} - V_{PC} , V_{N1} - V_{NC})	V_D	20	Volts
Supply Voltage (Applied between V_{UFB} - V_{UFS} , V_{VFB} - V_{VFS} , V_{WFB} - V_{WFS})	V_{DB}	20	Volts
Input Voltage (Applied between U_P , V_P , W_P - V_{PC} , U_N , V_N , W_N - V_{NC})	V_{IN}	-0.5 ~ $V_D+0.5$	Volts
Fault Output Supply Voltage (Applied between F_O - V_{NC})	V_{FO}	-0.5 ~ $V_D+0.5$	Volts
Fault Output Current (Sink Current at F_O Terminal)	I_{FO}	1	mA
Current Sensing Input Voltage (Applied between C_{IN} - V_{NC})	V_{SC}	-0.5 ~ $V_D+0.5$	Volts

* $V_D = 13.5 \sim 16.5\text{V}$, Inverter Part, $T_j = 125^\circ\text{C}$, Non-repetitive, Less than $2\mu\text{s}$

**The maximum junction temperature rating of the power chips integrated within the DIPIM is 150°C ($@T_f \leq 100^\circ\text{C}$). However, to ensure safe operation of the DIPIM, the average junction temperature should be limited to $T_{j(\text{avg})} \leq 125^\circ\text{C}$ ($@T_f \leq 100^\circ\text{C}$).

T_C Measurement Point



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Electrical and Mechanical Characteristics, $T_j = 25^\circ\text{C}$ unless otherwise specified

Characteristics	Symbol	Test Conditions	Min.	Typ.	Max.	Units
IGBT Inverter Sector						
Collector-Emitter Saturation Voltage	$V_{CE(sat)}$	$I_C = 75A, T_j = 25^{\circ}C, V_D = V_{DB} = 15V, V_{IN} = 5V$	—	1.55	2.05	Volts
		$I_C = 75A, T_j = 125^{\circ}C, V_D = V_{DB} = 15V, V_{IN} = 5V$	—	1.65	2.10	Volts
Diode Forward Voltage	V_{EC}	$T_j = 25^{\circ}C, -I_C = 75A, V_{IN} = 0V$	—	1.70	2.20	Volts
Inductive Load Switching Times	t_{on}		1.80	2.40	—	μs
	t_{rr}	$V_{CC} = 300V, V_D = V_{DB} = 15V,$	—	0.30	—	μs
	$t_{C(on)}$	$I_C = 75A, T_j = 125^{\circ}C, V_{IN} = 0 \leftrightarrow 5V,$	—	0.40	—	μs
	t_{off}	Inductive Load (Upper-Lower Arm)	—	3.40	—	μs
	$t_{C(off)}$		—	0.65	—	μs
Collector-Emitter Cutoff Current	I_{CES}	$V_{CE} = V_{CES}, T_j = 25^{\circ}C$	—	—	1.0	mA
		$V_{CE} = V_{CES}, T_j = 125^{\circ}C$	—	—	10	mA
Control Sector						
Circuit Current	I_D	$V_D = V_{DB} = 15V, V_{IN} = 5V,$ Total of $V_{P1}-V_{PC}, V_{N1}-V_{NC}$	—	—	7.00	mA
		$V_D = V_{DB} = 15V, V_{IN} = 5V,$ $V_{UFB}-V_{UFS}, V_{VFB}-V_{VFS}, V_{WFB}-V_{WFS}$	—	—	0.55	mA
		$V_D = V_{DB} = 15V, V_{IN} = 0V,$ Total of $V_{P1}-V_{PC}, V_{N1}-V_{NC}$	—	—	7.00	mA
		$V_D = V_{DB} = 15V, V_{IN} = 0V,$ $V_{UFB}-V_{UFS}, V_{VFB}-V_{VFS}, V_{WFB}-V_{WFS}$	—	—	0.55	mA
Fault Output Voltage	V_{FOH}	$V_{SC} = 0V, F_O$ Terminal Pull-up to 5V by 10k Ω	4.9	—	—	Volts
	V_{FOL}	$V_{SC} = 1V, I_{FO} = 1mA$	—	—	0.95	Volts
Input Current	I_{IN}	$V_{IN} = 5V$	1.0	1.5	2.0	mA
Short-Circuit Trip Level*	I_{SC}	$-20^{\circ}C \leq T_C \leq 100^{\circ}C, V_D = 15V, R_S = 18.7\Omega$	127.5	—	225.0	Amps
Supply Circuit Undervoltage	UV_{DBt}	Trip Level, $T_C \leq 100^{\circ}C$	10.0	—	12.0	Volts
Protection	UV_{DBr}	Reset Level, $T_C \leq 100^{\circ}C$	10.5	—	12.5	Volts
	UV_{Dt}	Trip Level, $T_C \leq 100^{\circ}C$	10.3	—	12.5	Volts
	UV_{Dr}	Reset Level, $T_C \leq 100^{\circ}C$	10.8	—	13.0	Volts
Fault Output Pulse Width**	t_{FO}	$C_{FO} = 22nF$	1.0	1.8	—	ms
ON Threshold Voltage	$V_{th(on)}$	Applied between $U_P, V_P, W_P-V_{PC},$	2.1	2.3	2.6	Volts
OFF Threshold Voltage	$V_{th(off)}$	U_N, V_N, W_N-V_{NC}	0.8	1.4	2.1	Volts
Temperature Output***	V_{OT}	At LVIC Temperature = $85^{\circ}C$	3.50	3.63	3.76	Volts

* Short-Circuit protection is functioning only at the lower arms. Please select the value of the external shunt resistor such that the SC trip level is less than 85A.

**Fault signal is asserted when the lower arm short circuit or control supply under-voltage protective functions operate. The fault output pulse-width t_{FO} depends on the capacitance value of C_{FO} according to the following approximate equation: $C_{FO} = (12.2 \times 10^{-6} \times t_{FO} [\text{F}])$.

***When the temperature rises excessively, the controller (MCU) should stop the DIPIPM.

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Intellimod™ Module
Dual-In-Line Intelligent Power Module
75 Amperes/600 Volts

Thermal Characteristics, $T_j = 25^\circ\text{C}$ unless otherwise specified

Characteristic	Symbol	Condition	Min.	Typ.	Max.	Units
Thermal Resistance Junction to Case	$R_{th(j-c)Q}$	IGBT Part (Per 1/6 Module)	—	—	0.77	$^\circ\text{C/Watt}$
Thermal Resistance Junction to Case	$R_{th(j-c)D}$	FWDi Part (Per 1/6 Module)	—	—	1.25	$^\circ\text{C/Watt}$

Recommended Conditions for Use

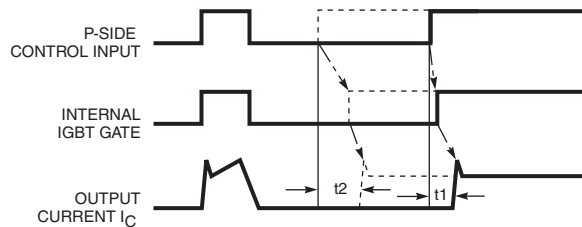
Characteristic	Symbol	Condition	Min.	Typ.	Max.	Units
Supply Voltage	V_{CC}	Applied between P-NU, NV, NW	0	300	400	Volts
Control Supply Voltage	V_D	Applied between $V_{P1-V_{PC}}$, $V_{N1-V_{NC}}$	13.5	15.0	16.5	Volts
	V_{DB}	Applied between $V_{UFB-V_{UFS}}$, $V_{VFB-V_{VFS}}$, $V_{WFB-V_{WFS}}$	13.0	15.0	18.5	Volts
Control Supply Variation	ΔV_D , ΔV_{DB}		-1	—	1	V/ μs
Arm Shoot-through Blocking Time	t_{DEAD}	For Each Input Signal, $T_C \leq 00^\circ\text{C}$	2.0	—	—	μs
PWM Input Frequency	f_{PWM}	$T_C \leq 100^\circ\text{C}$, $T_j \leq 125^\circ\text{C}$	—	—	20	kHz
Allowable rms Current*	I_O	$V_{CC} = 300\text{V}$, $V_D = 15\text{V}$, $f_{PWM} = 5\text{kHz}$, PF = 0.8, Sinusoidal PWM, $T_j \leq 125^\circ\text{C}$, $T_C \leq 100^\circ\text{C}$	—	—	37.0	Arms
		$V_{CC} = 300\text{V}$, $V_D = 15\text{V}$, $f_{PWM} = 15\text{kHz}$, PF = 0.8, Sinusoidal PWM, $T_j \leq 125^\circ\text{C}$, $T_C \leq 100^\circ\text{C}$	—	—	17.0	Arms
Minimum Input Pulse Width	$P_{WIN(on)}^{**}$		0.3	—	—	μs
	$P_{WIN(off)}^{***}$					
	$I_C \leq 75\text{A}$	$200 \leq V_{CC} \leq 350\text{V}$, $13.5 \leq V_D \leq 16.5\text{V}$,	3.0	—	—	μs
	$75 \leq I_C \leq 127.5\text{A}$	$13.5 \leq V_{DB} \leq 18.5\text{V}$, $-20^\circ\text{C} \leq T_C \leq 100^\circ\text{C}$, N-line Wiring Inductance Less Than 10nH	5.0	—	—	μs
V_{NC} Variation	V_{NC}	Between V_{NC-NU} , NV, NW (Including Surge)	-5.0	—	5.0	Volts

* The allowable rms current value depends on the actual application conditions.

**If input signal ON pulse is less than $P_{WIN(on)}$, the device may not respond.

***The IPM may fail to respond to an ON pulse if the preceeding OFF pulse is less than $P_{WIN(off)}$.

Delayed Response Against Shorter Input OFF Signal Than $P_{WIN(off)}$, P-side only

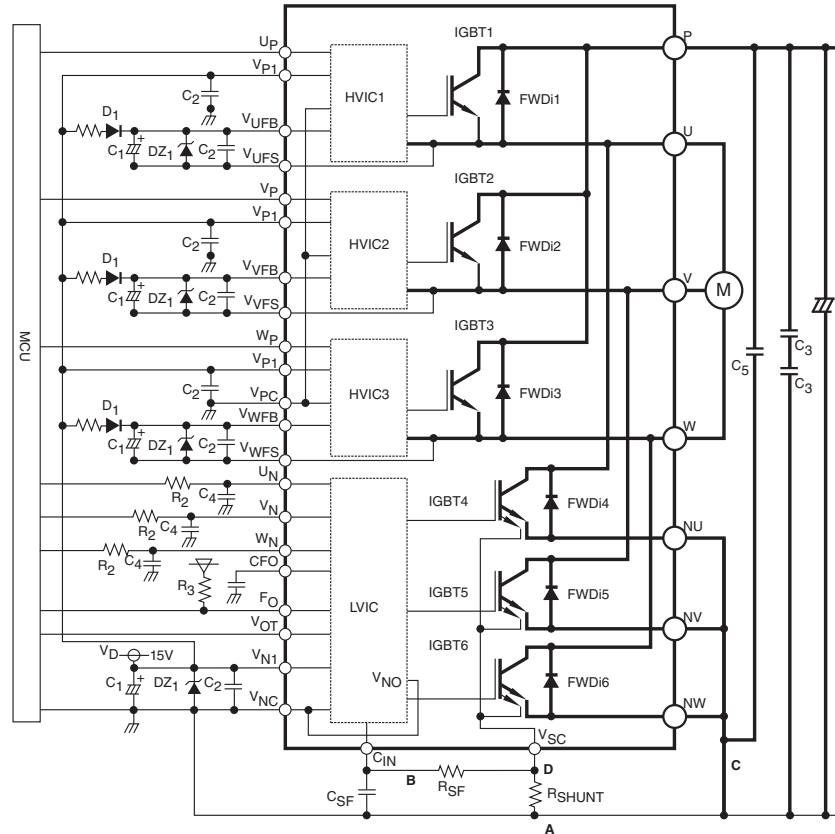


Solid Line – OFF Pulse Width > $P_{WIN(off)}$; Turn ON time $t1$.

Dotted Line – OFF Pulse Width < $P_{WIN(off)}$; Turn ON time $t2$.

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 75 Amperes/600 Volts

Application Circuit



Component Selection:

Dsgn.	Typ. Value	Description
D1	1A, 600V	Control and boot strap supply overvoltage suppression
DZ1	24V, 1W	Control and boot strap supply over voltage suppression
C1	10-100µF, 50V	Boot strap supply reservoir – electrolytic long life/low impedance, 105°C
C2	0.22-2.0µF, 50V	Local decoupling/High frequency noise filters – multilayer ceramic (Note 4)
C3	200 to 2500µF, 450V	Main DC bus filter capacitor – electrolytic, long life, high ripple current, 105°C
C4	100pF, 50V	Optional input signal noise filter – multilayer ceramic (Note 11)
C5	0.1-0.22µF, 1000V	Surge voltage suppression (Note 2)
CFSF	1000pF, 50V	Short circuit detection filter capacitor – multilayer ceramic
RSF	1.8kΩ	Short circuit detection filter resistor
RSHUNT	20ohm-500ohm	Current sensing resistor
R1	1-10Ω	Boot strap supply inrush limiting resistor – non-inductive, temperature stable, tight tolerance (Note 5)
R2	330Ω	Optional input signal noise filter (Note 11)
R3	10kΩ	Fault signal pull-up resistor (Note 9)

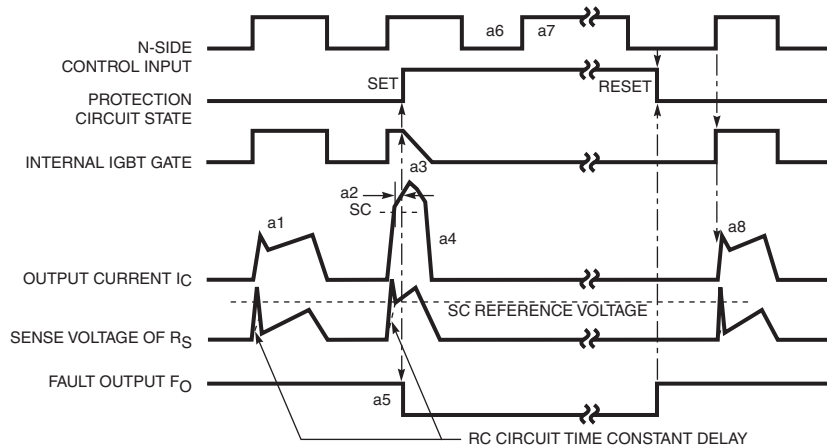
Notes:

- 1) If control GND is connected to power GND by broad pattern, it may cause malfunction by power GND fluctuation. It is recommended to connect control GND at only a point at which NU, NV, NW are connected to power GND line.
- 2) To prevent surge destruction, the wiring between the smoothing capacitor and the P-N1 terminals should be as short as possible. Generally inserting a 0.1µ – 0.22µF snubber capacitor C3 between the P-N1 terminals is recommended.
- 3) The time constant R1, C4 of RC filter for preventing the protection circuit malfunction should be selected in the range of 1.5µ – 2µs. SC interrupting time might vary with the wiring pattern. Tight tolerance, temp-compensated type is recommended for R1, C4.
- 4) All capacitors should be mounted as close to the terminals of the DIPIM as possible. (C1: good temperature, frequency characteristics electrolytic type, and C2: good temperature, frequency and DC bias characteristic ceramic type are recommended.)
- 5) It is recommended to insert a Zener diode DZ1 (24V/1W) between each pair of control supply terminals to prevent surge destruction.
- 6) To prevent erroneous SC protection, the wiring from VSC terminals to CIN filter should be divided at the point D that is close to the terminal of sense resistor and the wiring should be patterned as short as possible.
- 7) For sense resistor, the variation within 1% (including temperature characteristics), low inductance type is recommended. 1/8W is recommended, but an evaluation of your system is recommended.
- 8) To prevent erroneous operation, wiring A, B, and C should be as short as possible.
- 9) FO output is open drain type. It should be pulled up to the positive side of 5V or 15V power supply with a resistor that limits FO sink current (IFO) under 1mA. (Over 5.1kΩ is needed and 10kΩ is recommended for 5V supply.)
- 10) Error signal output width (tFO) can be set by the capacitor connected to the CFO terminal. $t_{FO}(\text{typ}) = C_{FO} / 9.1 \times 10^{-6} \text{ (s)}$.
- 11) Input drive is high-active type. There is a 3.3kΩ pull-down resistor integrated in the IC input circuit. To prevent malfunction, the wiring of each input should be patterned as short as possible. When inserting the RC filter, make sure the input signal level meets the turn-on and turn-off threshold voltage. Thanks to HVIC inside the module, connection to the MCU may be direct or with an opto-coupler.

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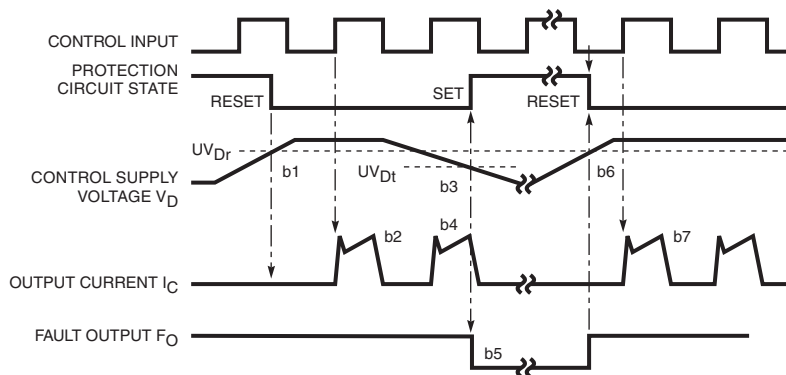
Protection Function Timing Diagrams

Short Circuit Protection (N-side Only with External Shunt Resistor and RC Filter)



- a1: Normal operation – IGBT turns on and carries current.
- a2: Short circuit current is detected (SC trigger).
- a3: All N-side IGBT's gate are hard interrupted.
- a4: All N-side IGBT's turn off.
- a5: F_O output with a fixed pulse width (determined by the external capacitance C_{FO}).
- a6: Input "L" – IGBT off.
- a7: Input "H" – IGBT on, but during the F_O output period the IGBT will not turn on.
- a8: IGBT turns on when L→H signal is input after F_O is reset.

Under-Voltage Protection (N-side , UV_D)

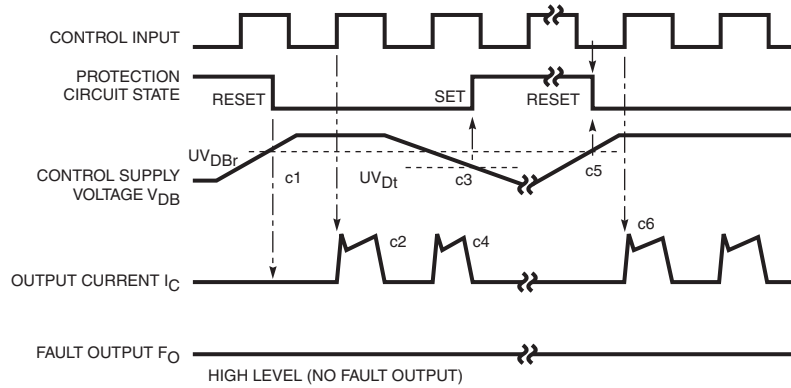


- b1: Control supply voltage V_D rises – After V_D level reaches under voltage reset level (UV_{Dr}), the circuits start to operate when next input is applied.
- b2: Normal operation – IGBT turns on and carries current.
- b3: V_D level dips to under voltage trip level (UV_{Dt}).
- b4: All N-side IGBT's turn off in spite of control input condition.
- b5: F_O is low for a minimum period determined by the capacitance C_{FO} but continuously during UV period.
- b6: V_D level reaches UV_{Dr} .
- b7: Normal operation – IGBT turns on and carries current.

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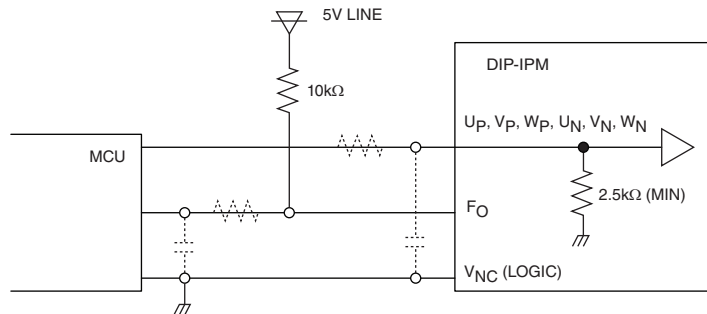
Protection Function Timing Diagrams

Under-Voltage Protection (P-side, UV_{DB})



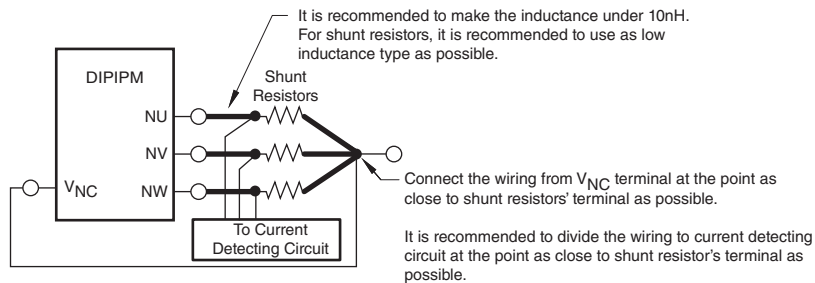
- c1: Control supply voltage V_{DB} rises – After V_{DB} level reaches under voltage reset level (UV_{DBr}), the circuits starts to operate when next input is applied.
- c2: Normal operation – IGBT turns on and carries current.
- c3: V_{DB} level dips to under voltage trip level (UV_{Dt}).
- c4: P-side IGBT turns off in spite of control input signal level, but there is no F_O signal output.
- c5: V_{DB} level reaches UV_{DBr} .
- c6: Normal operation – IGBT on and carries current.

Typical Interface Circuit



NOTE: RC coupling at each input (parts shown dotted) may change depending on the PWM control scheme used in the application and the wiring impedance of the printed circuit board. The DIP-IPM input signal section integrates a 2.5kΩ (min) pull-down resistor. Therefore, when using an external filtering resistor, care must be taken to satisfy the turn-on threshold voltage requirement.

Wiring Method Around Shunt Resistor



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Сотрудничество с глобальными дистрибьюторами электронных компонентов, предоставляет возможность заказывать и получать с международных складов практически любой перечень компонентов в оптимальные для Вас сроки.

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Система менеджмента качества компании отвечает требованиям в соответствии с ГОСТ Р ИСО 9001, ГОСТ РВ 0015-002 и ЭС РД 009

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