

LED Driver for High Power LEDs

ILD4001

Step down LED Controller for high power LEDs

Data Sheet

Revision 2.0, 2011-06-09

Industrial and Multimarket

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Revision History

Page or Item	Subjects (major changes since previous revision)
Revision 2.0, 2011-06-09	
All	Preliminary status removed
Table 4	DC characteristics updated
Table 5	Switching characteristics updated
Chapter 6.4	LED current vs T_S added
Revision 1.5, 2011-05-30	
Table 2	ESD capability updated
Table 5	AC characteristics updated
Table 7	Analog dimming updated
Chapter 6.4	All figures updated

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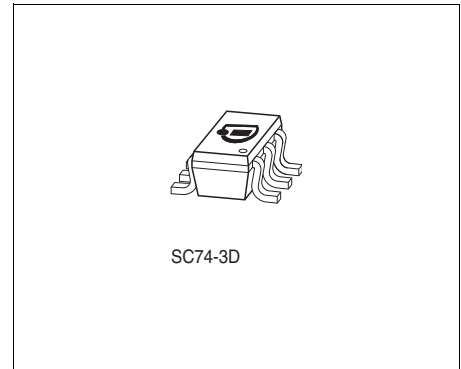
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Step down LED Controller for high power LEDs

1 Features

- Wide input voltage range: 4.5 V ... 42 V
- Capable to drive N-channel MOSFETs that provide up to 3 A output current and up to 98% efficiency
- Temperature shut down mechanism
- Switching frequency up to 500 kHz
- Analog and PWM dimming possible
- Typical 3 % output current accuracy
- Very low LED current drift over temperature
- Minimum external component required
- Small package: SC-74



Applications

- LED controller for indoor and outdoor illumination
- LED replacement lamps, e.g. MR16 halogen replacement
- Retail, office and residential high power luminaires
- Architectural lighting
- Downlights and light engines
- Appliances, e.g. fridge / freezer

Product Name	Package	Pin Configuration						Marking
ILD4001	SC74-6-4	1 = V_S	2 = GND	3 = EN	4 = V_{drive}	5 = GND	6 = V_{sense}	01

2 Product Brief

The ILD4001 is a hysteretic buck LED controller IC for driving high power LEDs in indoor and outdoor lighting applications.

The LED controller is capable of driving an external MOSFET power transistor with the internal push-pull output stage to achieve LED currents of 350 mA up to 3 A and more depending on the dimensioning of the MOSFET, the thermal budget of the circuit board and the current sense resistor.

The ILD4001 is widely suitable for LED applications with supply voltages up to 42 V. A multifunctional enable pin allows dimming of the LEDs with DC voltage or a PWM signal. Furthermore the enable pin can be used to switch the LED controller on and off to minimize power consumption in standby.

The ILD4001 incorporates an integrated thermal shutdown function pulling low the V_{drive} output signal once the junction temperature exceeds the threshold temperature. Once the junction temperature drops below the threshold temperature the V_{drive} output is activated again.

To provide maximum design flexibility, the ILD4001 is housed in a small SC-74 package.

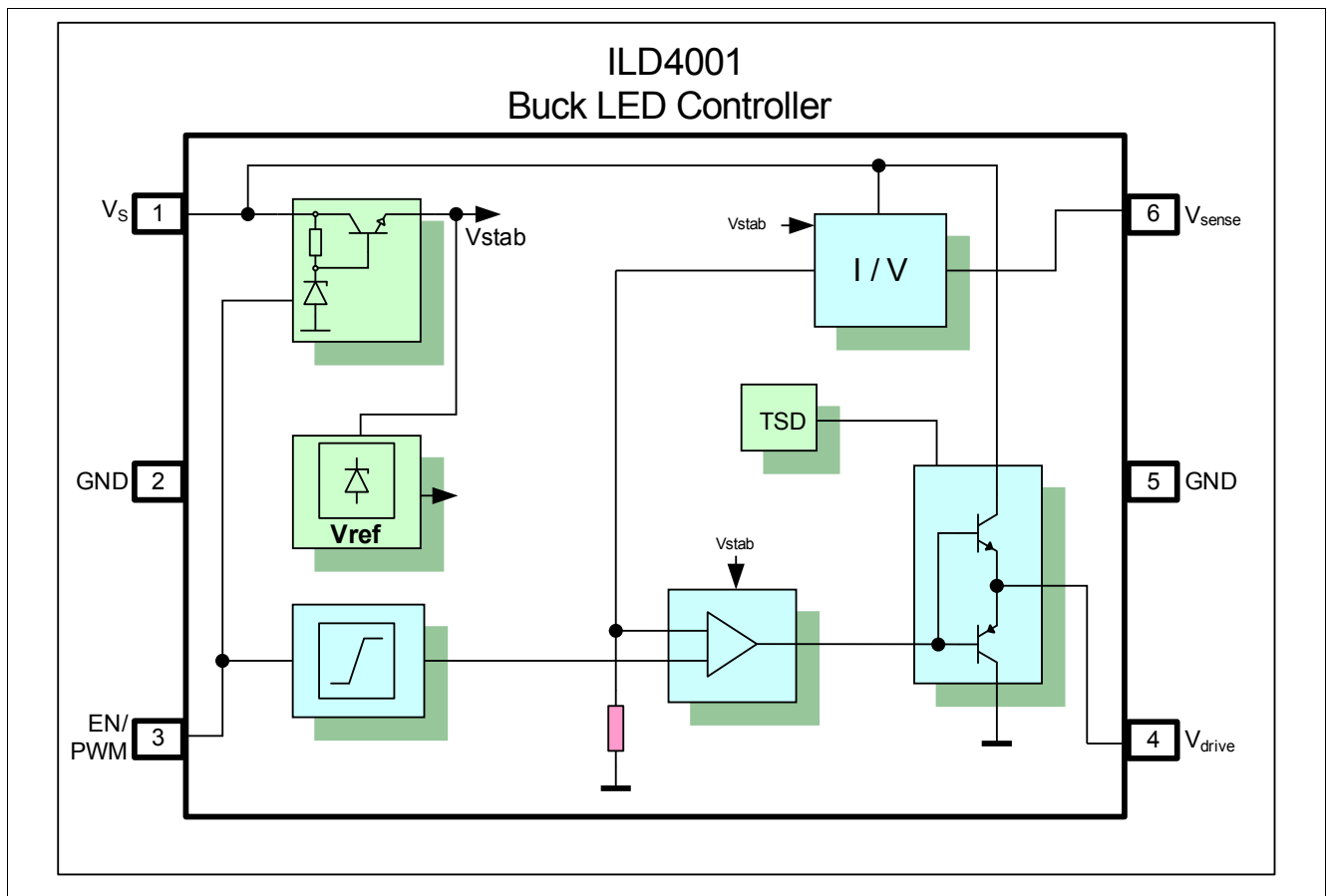


Figure 1 Block Diagram

Pin Definition

Table 1 Pin Definition and Function

Pin No.	Name	Pin Type	Buffer Type	Function
1	V_s	Input	–	Supply voltage
2	GND	GND	–	IC ground
3	EN / PWM	Input	–	Multifunctional pin: <ul style="list-style-type: none">• Chip enable signal• Analog dimming signal• PWM dimming signal
4	V_{drive}	Output	–	Push-pull switch output pin
5	GND	GND	–	IC ground
6	V_{sense}	Input	–	LED current sense pin

3 Maximum Ratings

Table 2 Maximum Ratings

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Supply voltage	V_S	–	–	45	V	–
Peak output current	I_{drive}	–	–	50	mA	–
Total power dissipation, $T_s \leq 115^\circ\text{C}$	P_{tot}	–	–	500	mW	–
Junction temperature	T_J	–	–	150	$^\circ\text{C}$	–
Solder temperature of GND pins	T_{SGND}	–	–	125	$^\circ\text{C}$	–
Storage temperature range	T_{STG}	-65	–	150	$^\circ\text{C}$	–
ESD capability at pin 4	$V_{ESD\ HBM}$	–	–	1	kV	HBM acc. to JESD22-A114
at all other pins		–	–	4		

Attention: Stresses above the max. values listed here may cause permanent damage to the device. Exposure to absolute maximum rating conditions for extended periods may affect device reliability. Maximum ratings are absolute ratings; exceeding only one of these values may cause irreversible damage to the integrated circuit.

4 Thermal Characteristics

Table 3 Maximum Thermal Resistance

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Junction - soldering point ¹⁾	R_{thJS}	–	–	70	K/W	–

1) For calculation of R_{thJA} please refer to application note AN077, "Thermal Resistance Calculation"

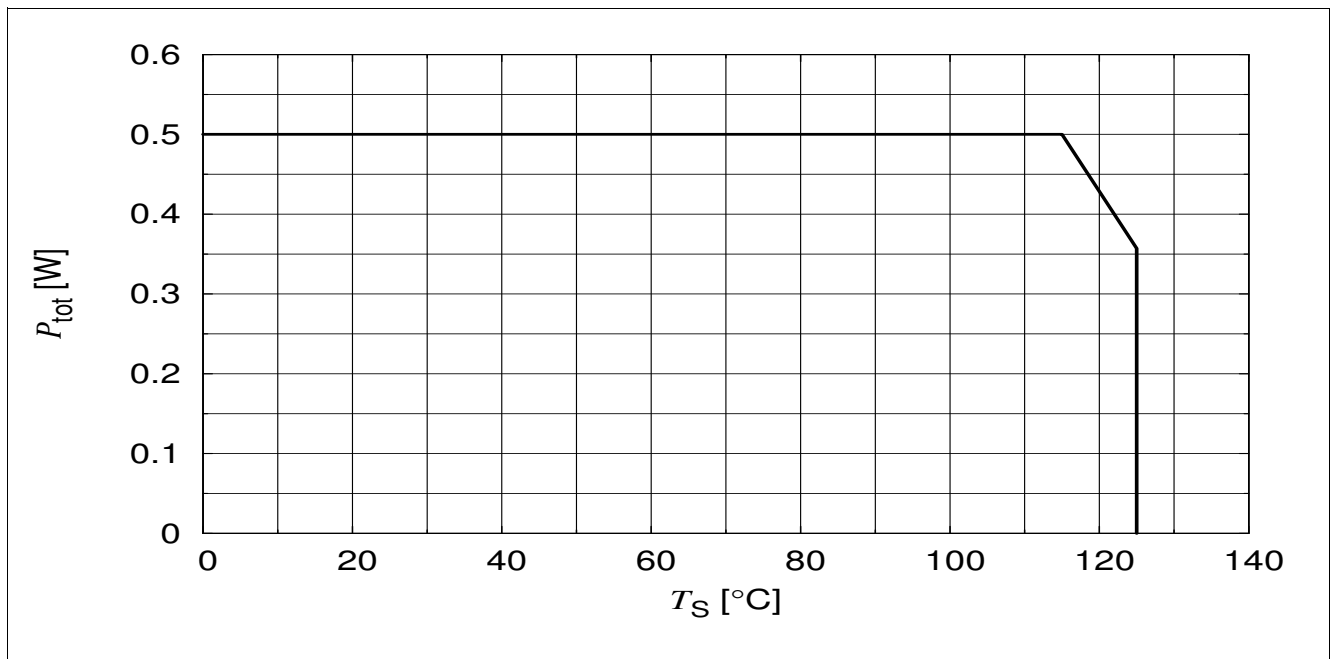


Figure 2 Total Power Dissipation vs. Soldering Point Temperature T_S

Equation (1) is a first estimation to calculate the power dissipation of the IC:

$$P_{tot} = V_S \cdot I_S + f_{Switch} \cdot C_{drive} \cdot V_S \cdot 5V \quad (1)$$

5 Electrical Characteristics

5.1 DC Characteristics

All parameters at $T_A = 25\text{ }^\circ\text{C}$, $V_S = 12\text{ V}$, $V_{EN} = 3\text{ V}$, unless otherwise specified.

Table 4 DC Characteristics

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Supply voltage	V_S	4.5	–	42	V	–
Overall current consumption open load	$I_{S\ open\ load}$	–	4.2	–	mA	$V_S = 4.5\text{ V}$, $I_{LED} = 0\text{ mA}$
Overall current consumption open load	$I_{S\ open\ load}$	–	5.1	–	mA	$V_S = 40\text{ V}$, $I_{LED} = 0\text{ mA}$
Overall current consumption open load	$I_{S\ open\ load}$	–	5.3	–	mA	$V_S = 42\text{ V}$, $I_{LED} = 0\text{ mA}$
Overall standby current consumption	$I_{S\ standby}$	–	–	260	nA	$V_S = 4.5\text{ V}$, $V_{EN} = 0.4\text{ V}$
Overall standby current consumption	$I_{S\ standby}$	–	–	360	nA	$V_S = 40\text{ V}$, $V_{EN} = 0.4\text{ V}$
Enable voltage for standby mode ¹⁾	V_{EN}	0	–	0.4	V	
Enable voltage for analog dimming ²⁾	V_{EN}	1	–	42	V	
Enable voltage for linear analog dimming	V_{EN}	1	–	2	V	linear dimming range
Input current of multifunctional control pin	I_{EN}	–	150	270	μA	$V_{EN} = 3\text{ V}$
Current of Sense input	I_{sense}	–	20	–	μA	At any LED current
Temperature shut down threshold	T_{TSD}	–	120	–	$^\circ\text{C}$	V_{drive} gets pulled low, refers to T_J

- 1) In standby mode ILD4001 doesn't pull low the V_{drive} signal. Depending on gate capacitance driven a 10 - 100 k Ω shunt resistor to GND is required to avoid a floating gate of the MOSFET. A discharge time of about 1 μs is recommended.
- 2) V_{drive} line requires a shunt resistor to GND to avoid a floating gate of the MOSFET for a V_{EN} voltage below the min. specified limit

5.2 Switching Characteristics

All parameters at $T_A = 25\text{ °C}$, unless otherwise specified.

$V_S = 12\text{ V}$, $R_{\text{sense}} = 158\text{ m}\Omega$ ($I_{\text{LED}} = 730\text{ mA}$), $L = 68\text{ }\mu\text{H}$, $V_{\text{EN}} = 3\text{ V}$

Table 5 Switching Characteristics

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Switching frequency	f_{Switch}	–	200	–	kHz	3 LEDs in series
Maximum switching frequency	$f_{\text{Switch max}}$	–		500	kHz	for any coil value
Output voltage in push-high condition	$V_{\text{drive high}}$	–	5	–	V	$I_{\text{drive}} = 10\text{ mA}$
Output voltage in pull-low condition	$V_{\text{drive low}}$	–	250	–	mV	$I_{\text{drive}} = -10\text{ mA}$
Voltage offset of V_{sense} input ¹⁾	V_{sense}	–	116	–	mV	3 LEDs in series, $V_S - V_{\text{fLED}} \geq 3\text{ V}$
Sense threshold hysteresis	$V_{\text{sense hys}}$	–	± 15	–	%	At any LED current
Output current accuracy	I_{outacc}	–	± 3	–	%	3 LEDs in series
Output current drift over supply voltage	I_{outaccVs}	–	6	–	%	3 LEDs in series $V_S = 12 \dots 42\text{ V}$

1) Voltage offset below supply voltage V_S

5.3 Digital Signals

All parameters at $T_A = 25\text{ °C}$, unless otherwise specified.

Table 6 Digital Control Parameter at Pin EN/PWM

Parameter	Symbol	Values			Unit	Note / Test Condition
		Min.	Typ.	Max.		
Input voltage for power on	V_{On}	2.5	3	42	V	Full LED current
Input voltage for power off ¹⁾	V_{Off}	-0.3	–	0.4	V	–
Min. power on puls duration	t_{On}	10			µs	

1) During power off ILD4001 doesn't pull low the V_{drive} signal. Depending on gate capacitance driven a 10 - 100 kΩ shunt resistor to GND is required to avoid a floating gate of the MOSFET. A discharge time of about 1 µs is recommended.

6 Basic Application Information

This section covers the basic information required for calculating the parameters for a certain LED application.

For detailed application information please check the application note **AN213** (Driving 2 - 5 W LEDs with ILD4001) or visit our web site <http://www.infineon.com/lowcostledrivers>

6.1 External MOSFET

An external MOSFET is required to drive the LEDs in the ILD4001 application. There are a few factors to be considered while choosing the suitable external MOSFET. First, choose the correct voltage and current rating of the MOSFET. Please ensure the V_{DS} breakdown voltage and I_{DS} current capability is sufficient and ensure that the external MOSFET is working within the safe operating area region of DC mode. Second, the logic high level from ILD4001 is 5 V and the external MOSFET must be able to be driven with a 5 V gate voltage. Third, choose a low $R_{DS(ON)}$ MOSFET to improve the efficiency of the system.

The BSR302N is recommended for supply voltages up to 30 V and an output current up to 3.7 A. For higher supply voltages up to 42 V, the BSP318S is recommended with an output current of up to 2.6 A.

For an overview of all suitable MOSFETs please visit <http://www.infineon.com/smallsignalmosfets>

6.2 Setting the Average LED Current

The average output current for the LEDs is set by the external sense resistor R_{sense} . To calculate the value of this resistor a first approximation can be calculated using [Equation \(2\)](#).

V_{sense} is dependent on the supply voltage V_s and the number of LEDs in series.

$$R_{sense} = \frac{V_{sense}}{I_{LED}} \tag{2}$$

Example Calculation

$V_s = 12\text{ V}$, $I_{LED} = 730\text{ mA}$, $L = 68\text{ }\mu\text{H}$, $V_{fLED} = 3\text{ V}$, 3 LEDs in series

For this configuration V_{sense} will settle at 116 mV.

→ $R_{sense} = 158\text{ m}\Omega$ according to [Equation \(2\)](#)

An easy way to achieve this resistor value is to connect several standard resistors in parallel.

6.3 Dimming of the LEDs

Analog Voltage Dimming

The voltage level of the EN/PWM pin can be used for analog dimming of the LED current. The analog dimming characteristic graph is shown in [Figure 3](#). To achieve a linear change in LED current versus control voltage the recommended voltage range at the EN/PWM pin is 1 V to 2 V. The maximum achievable LED current is defined by resistor R_{sense} . The maximum LED current will be achieved for $V_{EN} \geq 2.5$ V as shown in [Figure 4](#).

Below 0.4 V the ILD4001 is set to standby mode and the output is switched off. In standby mode ILD4001 doesn't pull low the V_{drive} output signal driving the gate of the external MOSFET. Depending on gate capacitance driven a 10 - 100 k Ω shunt resistor to GND is required to avoid a floating gate of the MOSFET. Furthermore a gate discharge time of about 1 μ s is recommended.

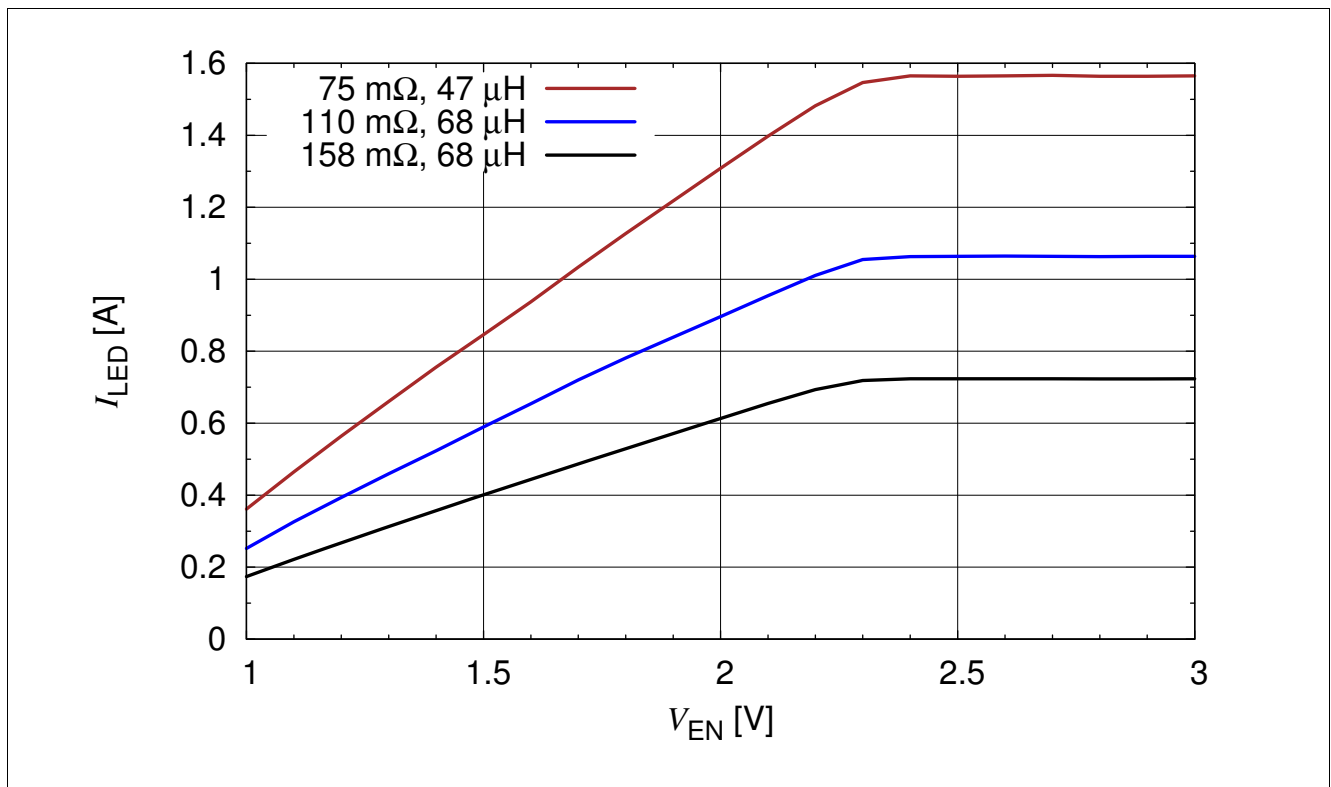


Figure 3 Analog Voltage Dimming (12V, 3 LEDs, $T_A=25^\circ\text{C}$) vs. R_{sense}

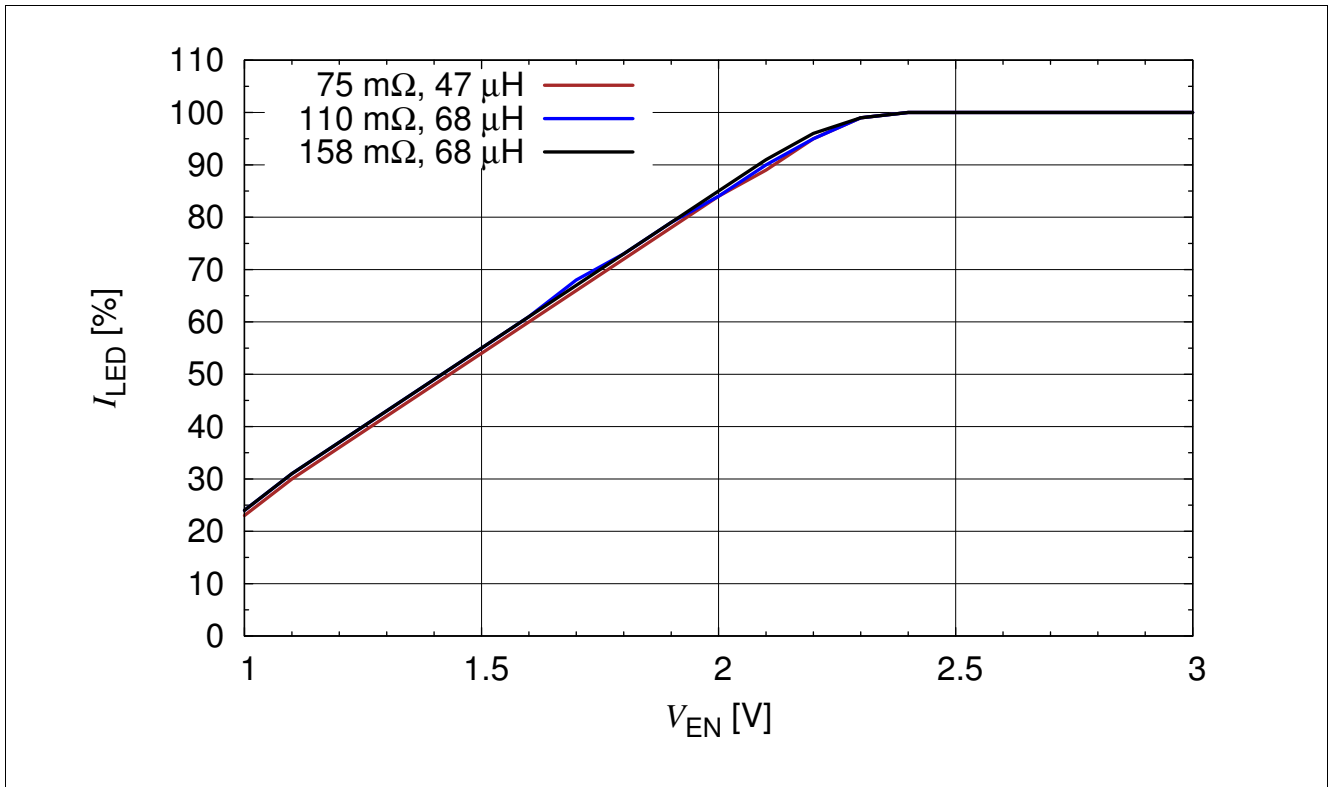


Figure 4 Analog Voltage Dimming (Relative) vs. R_{sense}

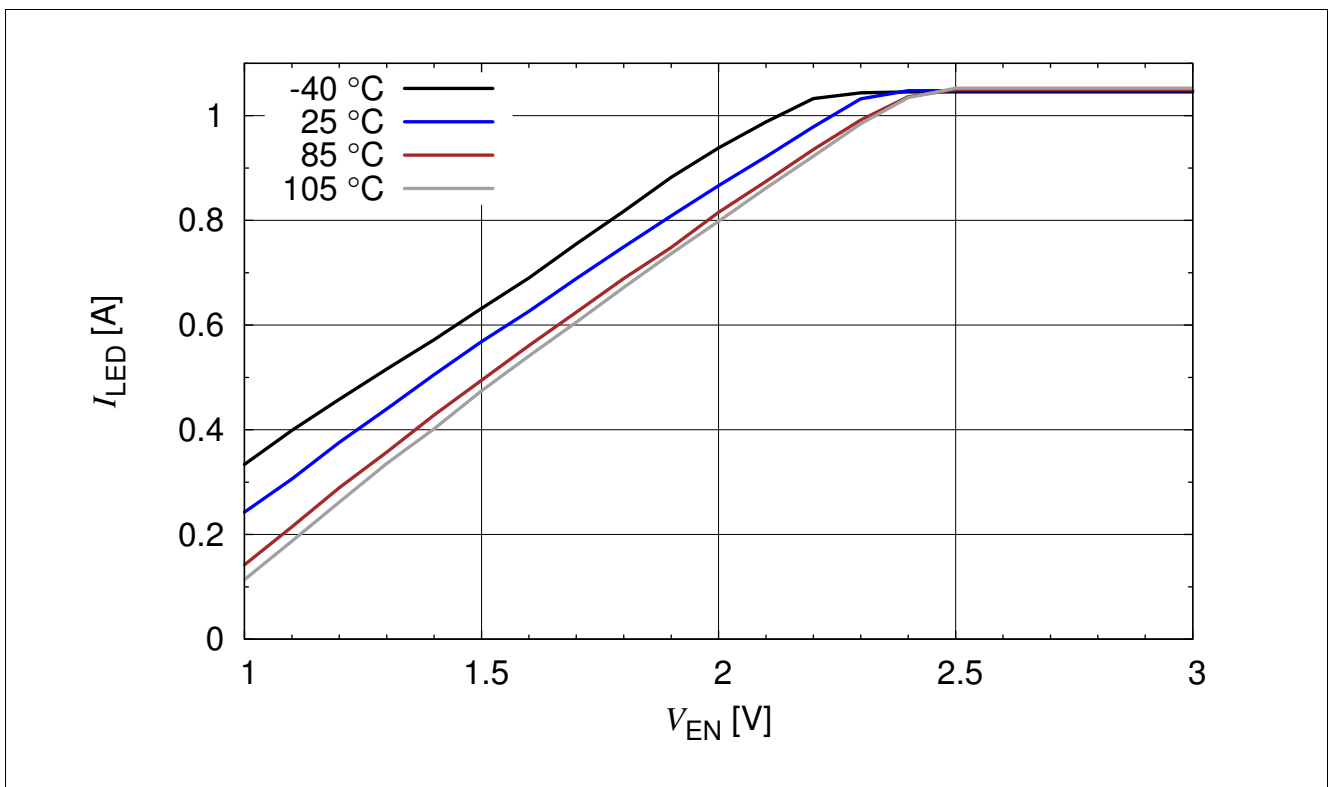


Figure 5 Analog Voltage Dimming vs. T_A (12V, 3 LEDs, 110 mΩ, 47 μH)

PWM Dimming

Besides the analog dimming functionality the EN/PWM pin acts as input for a pulse width modulated (PWM) signal to control the dimming of the LED string. For PWM dimming the signal's logic high level should be at least 2.5 V and the PWM frequency should be lower than 5 kHz. For the ILD4035/4001 demo board a dimming frequency less than 330 Hz is recommended to maintain a maximum contrast ratio of 100:1. The achievable contrast ratio is shown on **Figure 6** based on the measured average LED current deviating 3 dB from the linear reference. The maximum contrast ratio depends mainly on the rise time of the inductor current and is thus dependent on supply voltage, inductor size and LED string forward voltage.

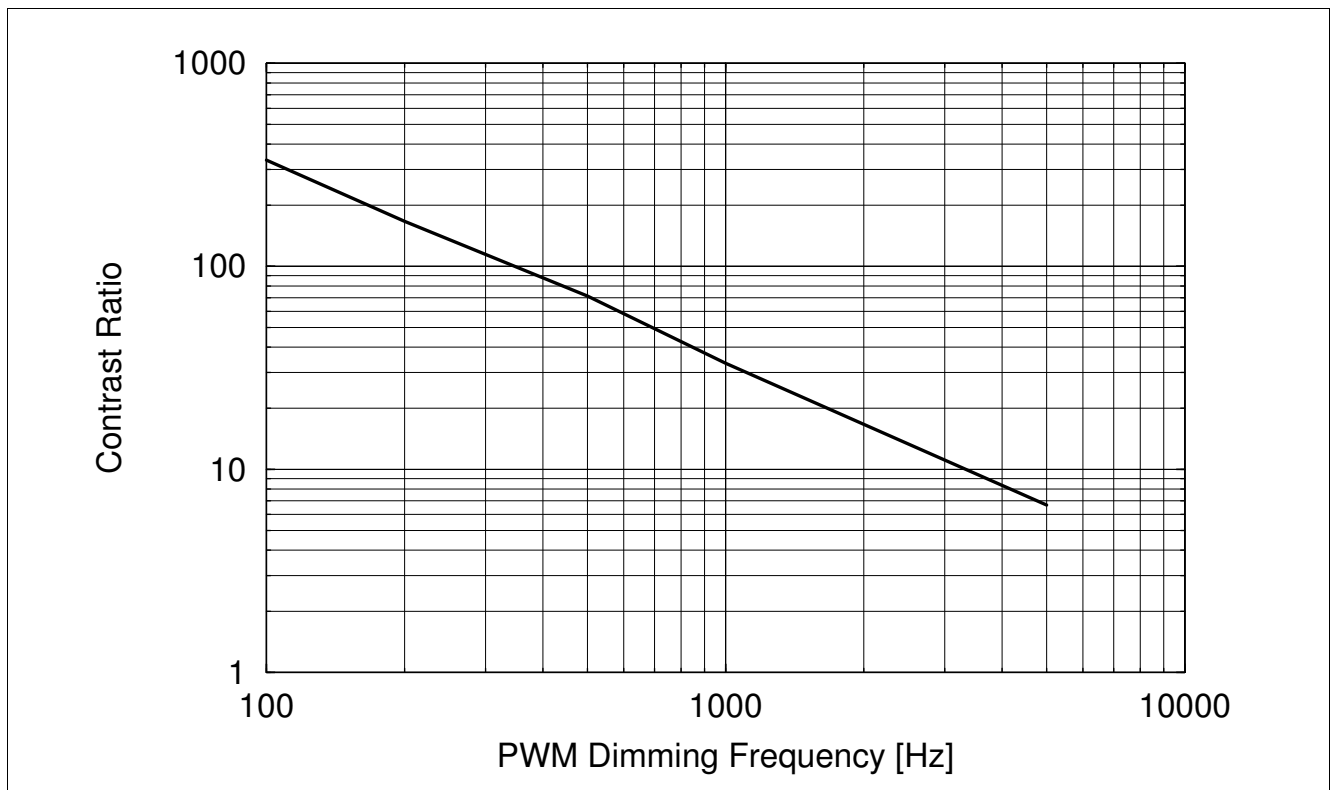


Figure 6 PWM Dimming: 3 dB Deviation of Contrast Ratio to Linear Dimming (12 V, 68 μ H, 3 LEDs)

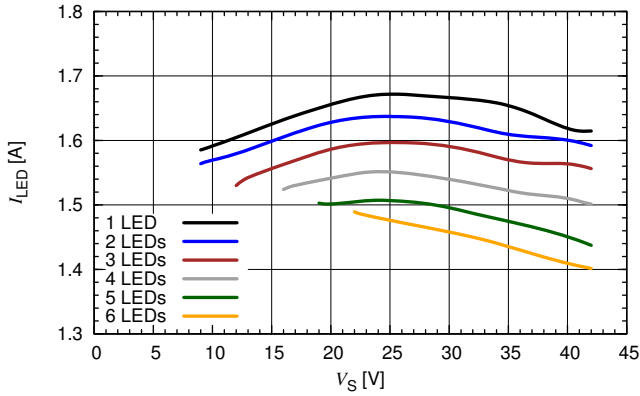
During the low state of the PWM signal ILD4001 doesn't pull low the V_{drive} signal. Depending on gate capacitance driven and intended gate discharge time (about 1 μ s is recommended) a 10 - 100 k Ω shunt resistor to GND is required to avoid a floating gate of the MOSFET.

6.4 Switching Parameters

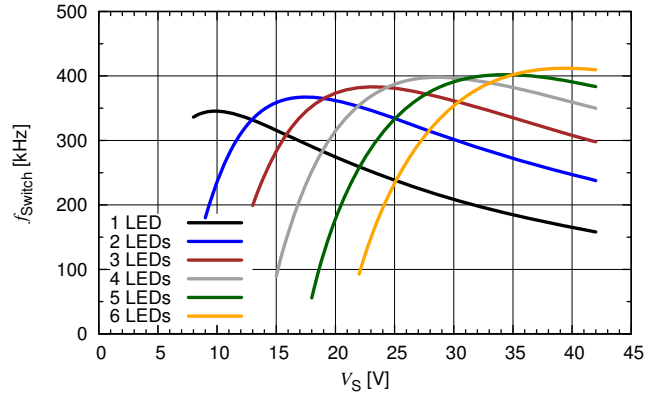
For all shown switching parameters ILD4001 has been measured on evaluation board ILD4035/4001 using a BSP318S N-channel MOSFET at $T_A = 25$ °C. Used LEDs have a typical V_{fLED} of 3 V. See application note **AN213** for further details.

Performance vs. supply voltage and number of LEDs: $R_{sense} = 75\text{ m}\Omega$, $L = 33\text{ }\mu\text{H}$, $V_{fLED} = 3\text{ V}$

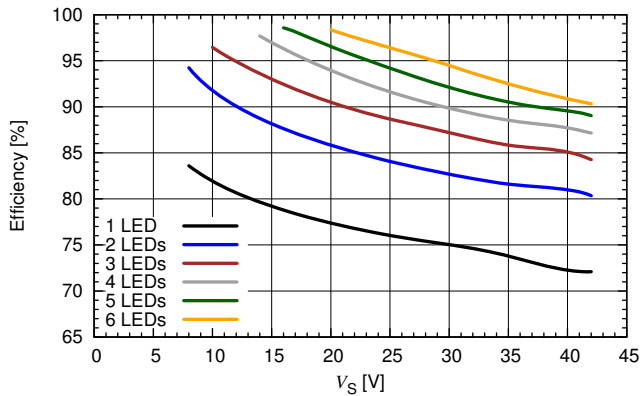
I_{LED} versus V_S and Number of LEDs



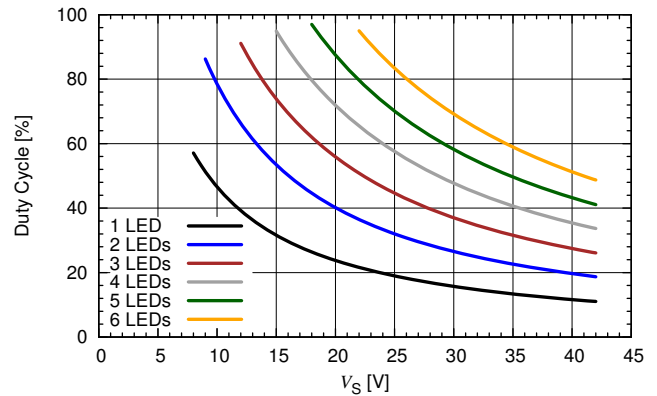
f_{Switch} versus V_S and Number of LEDs



Efficiency versus V_S and Number of LEDs

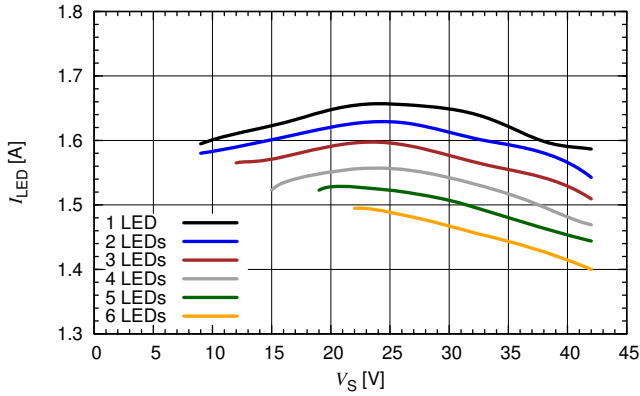


Duty Cycle versus V_S and Number of LEDs

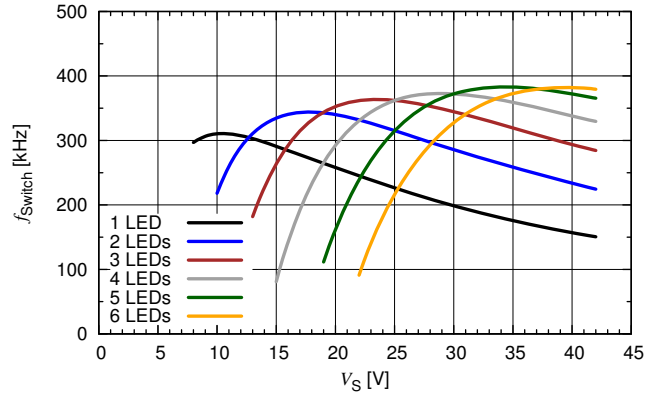


Performance vs. supply voltage and number of LEDs: $R_{sense} = 75\text{ m}\Omega$, $L = 47\text{ }\mu\text{H}$, $V_{fLED} = 3\text{ V}$

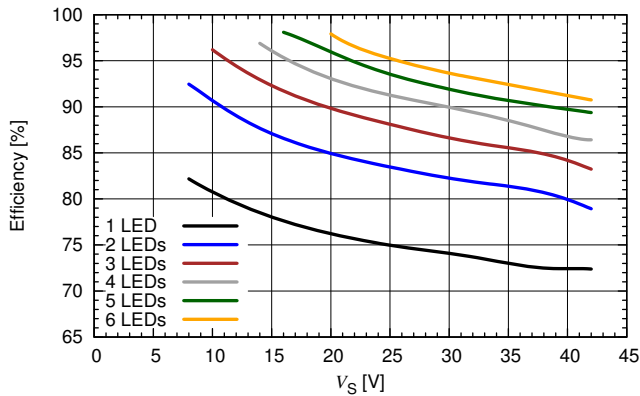
I_{LED} versus V_S and Number of LEDs



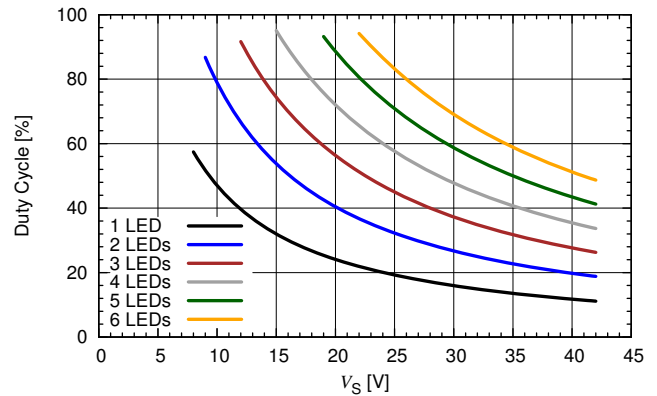
f_{Switch} versus V_S and Number of LEDs



Efficiency versus V_S and Number of LEDs

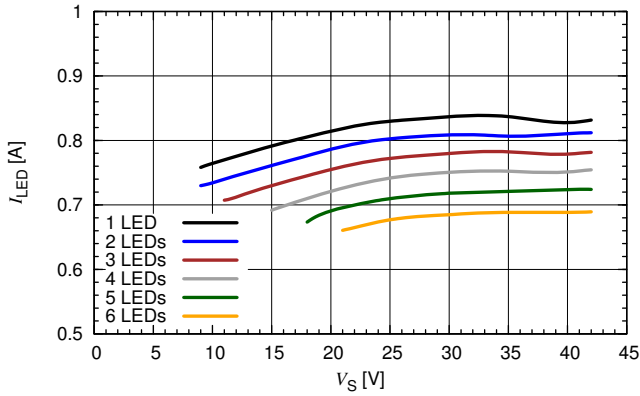


Duty Cycle versus V_S and Number of LEDs

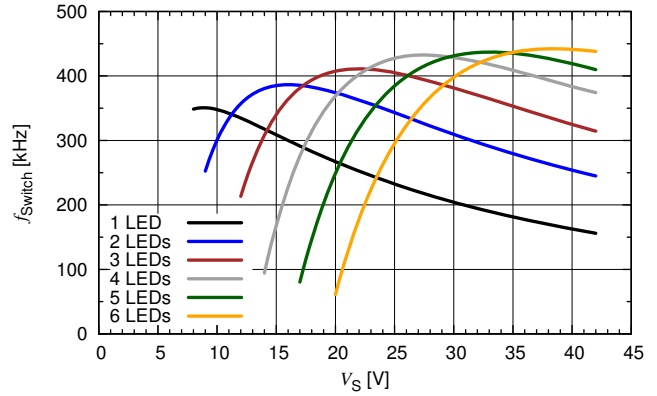


Performance vs. supply voltage and number of LEDs: $R_{sense} = 158\text{ m}\Omega$, $L = 47\text{ }\mu\text{H}$, $V_{fLED} = 3\text{ V}$

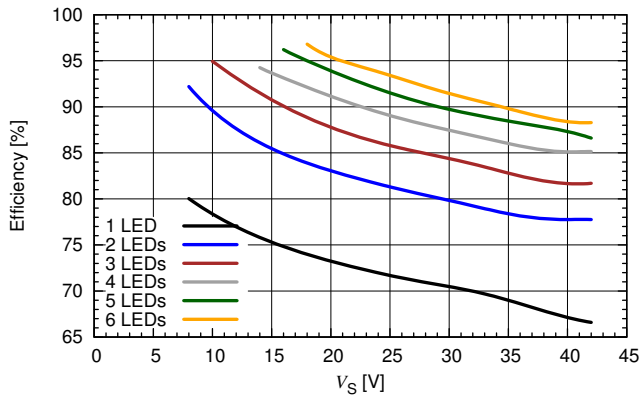
I_{LED} versus V_S and Number of LEDs



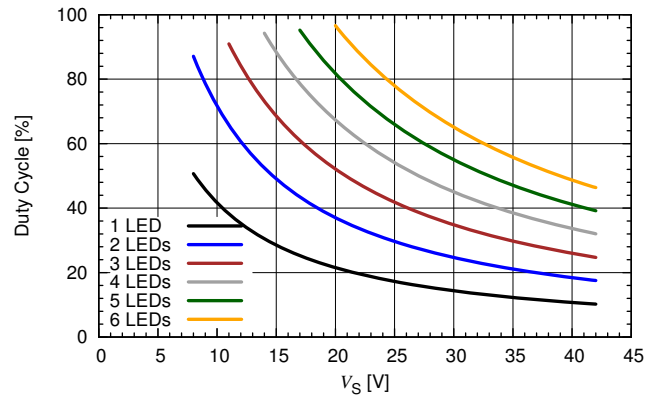
f_{Switch} versus V_S and Number of LEDs



Efficiency versus V_S and Number of LEDs

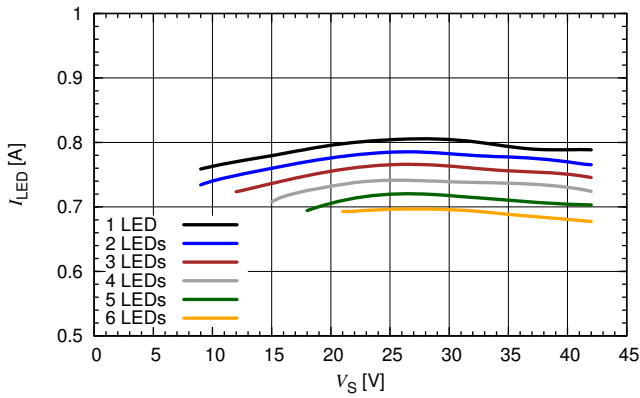


Duty Cycle versus V_S and Number of LEDs

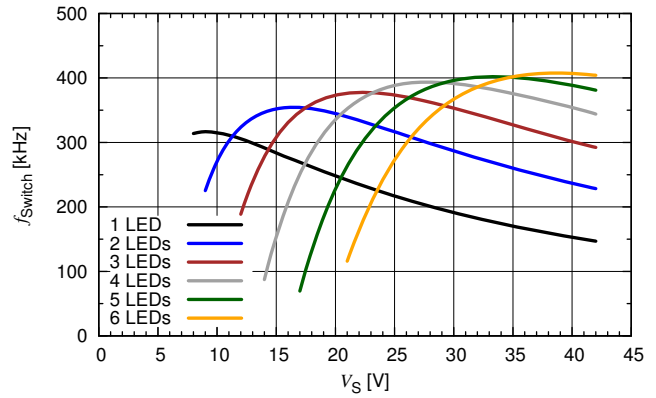


Performance vs. supply voltage and number of LEDs: $R_{\text{sense}} = 158 \text{ m}\Omega$, $L = 68 \text{ }\mu\text{H}$, $V_{\text{fLED}} = 3 \text{ V}$

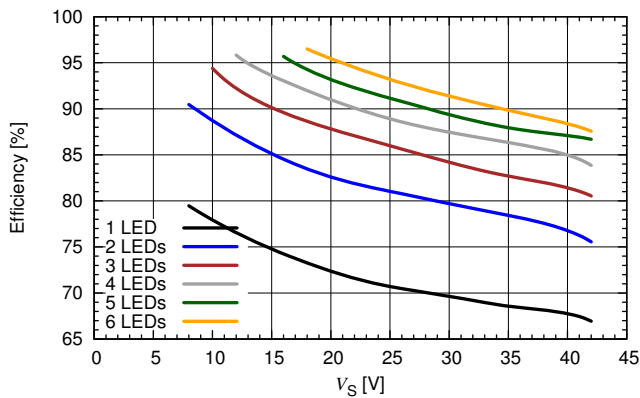
I_{LED} versus V_{S} and Number of LEDs



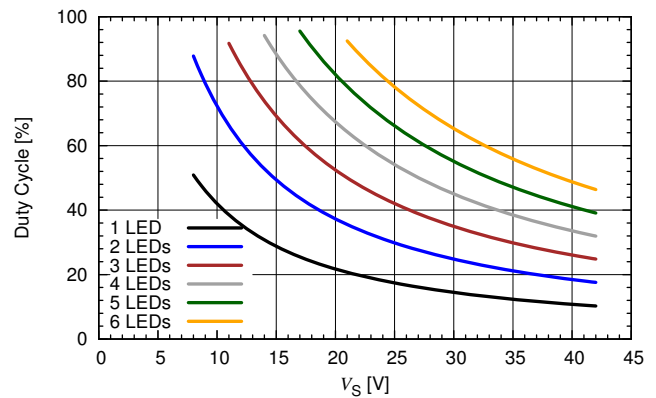
f_{Switch} versus V_{S} and Number of LEDs



Efficiency versus V_{S} and Number of LEDs

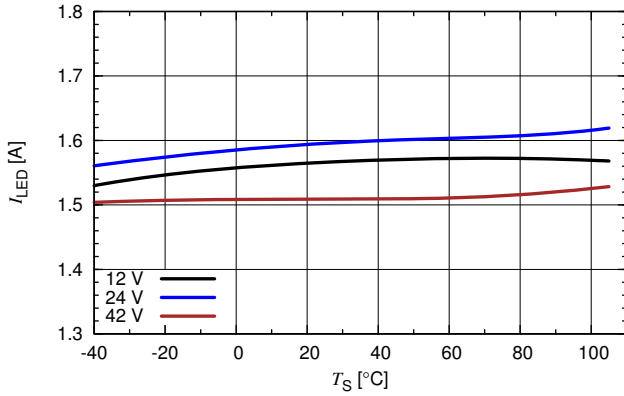


Duty Cycle versus V_{S} and Number of LEDs

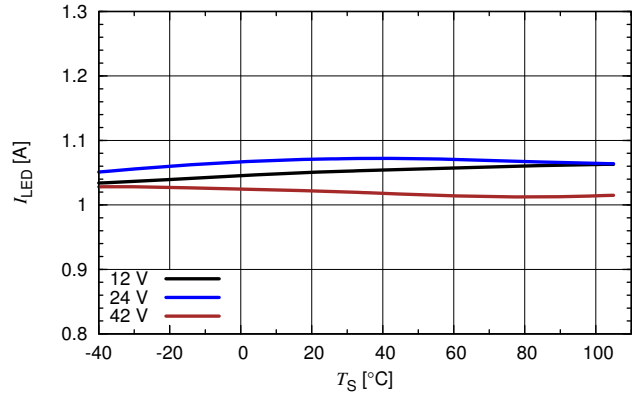


LED current vs. soldering point temperature T_S ($V_S = 12V$, $V_{fLED} = 3 V$, 3 LEDs)

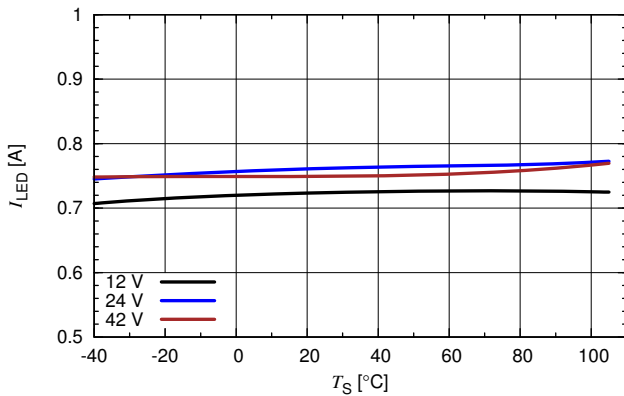
I_{LED} versus T_S ($R_{sense} = 75 m\Omega$, $L = 47 \mu H$)



I_{LED} versus T_S ($R_{sense} = 110 m\Omega$, $L = 68 \mu H$)



I_{LED} versus T_S ($R_{sense} = 158 m\Omega$, $L = 68 \mu H$)



9 Package Information

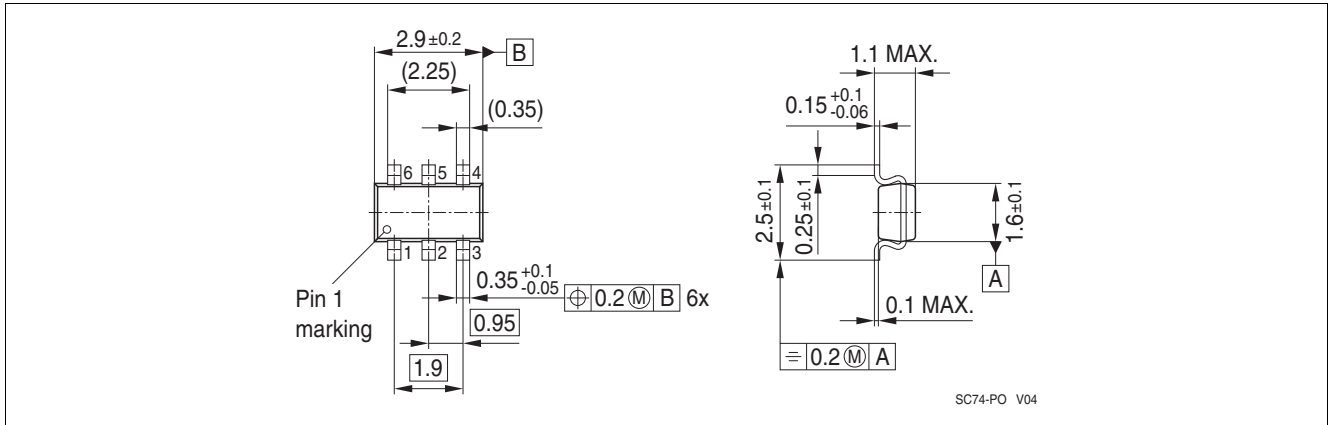


Figure 9 Package Outline SC74

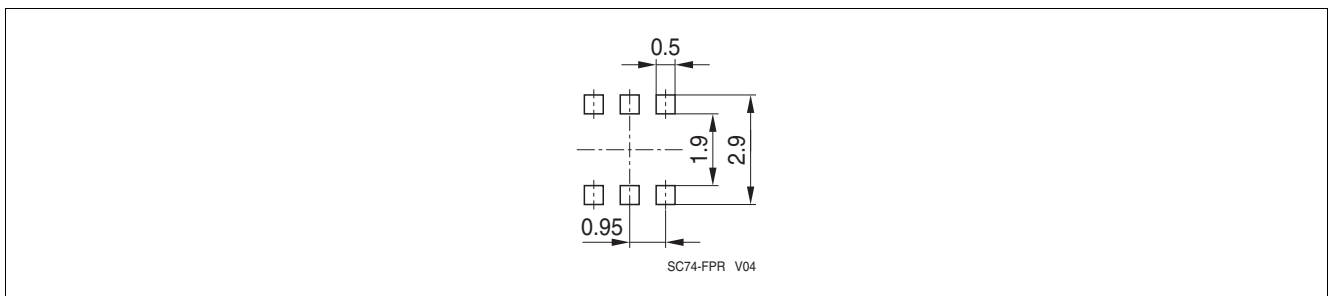


Figure 10 Recommended PCB Footprint for Reflow Soldering

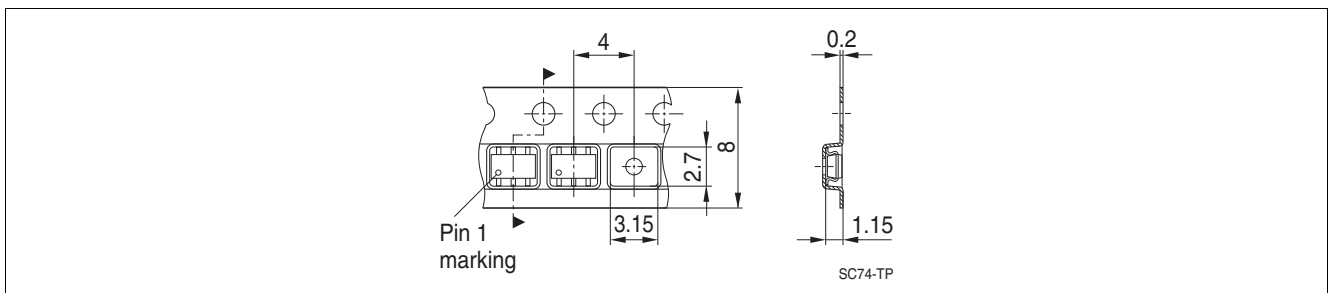


Figure 11 Tape Loading

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