Micropower, Single-Supply, Rail-to-Rail, Precision Instrumentation Amplifiers

General Description

The MAX4194 is a variable-gain precision instrumentation amplifier that combines Rail-to-Rail[®] single-supply operation, outstanding precision specifications, and a high gain bandwidth. This amplifier is also offered in three fixed-gain versions: the MAX4195 (G = +1V/V), the MAX4196 (G = +10V/V), and the MAX4197 (G = +100V/V). The fixed-gain instrumentation amplifiers feature a shutdown function that reduces the quiescent current to 8µA. A traditional three operational amplifier configuration is used to achieve maximum DC precision.

The MAX4194–MAX4197 have rail-to-rail outputs and inputs that can swing to 200mV below the negative rail and to within 1.1V of the positive rail. All parts draw only 93 μ A and operate from a single +2.7V to +7.5V supply or from dual ±1.35V to ±3.75V supplies. These amplifiers are offered in 8-pin SO packages and are specified for the extended temperature range (-40°C to +85°C).

See the MAX4198/MAX4199 data sheet for single-supply, precision differential amplifiers.

Applications

Medical Equipment

- Thermocouple Amplifier
- 4-20mA Loop Transmitters
- Data-Acquisition Systems
- Battery-Powered/Portable Equipment
- Transducer Interface
- Bridge Amplifier

Pin Configurations

Benefits and Features

- Low Power Consumption Is Ideal for Remote-Sensing and Battery-Powered Applications
 - +2.7V Single-Supply Operation
 - Low Power Consumption
 - 93µA Supply Current
 - 8µA Shutdown Current
 - (MAX4195/MAX4196/MAX4197)
- Precision Specifications Maximize Sensor Peformance
 - High Common-Mode Rejection: 115dB (G = +10V/V)
 - Input Common-Mode Range Extends 200mV Below GND
 - Low 50µV Input Offset Voltage (G ≥ +100V/V)
 - Low ±0.01% Gain Error (G = +1V/V)
 - 250kHz -3dB Bandwidth (G = +1V/V, MAX4194)
 - Rail-to-Rail Outputs

Ordering Information

PART	TEMP RANGE	PIN-PACKAGE
MAX4194ESA	-40°C to +85°C	8 SO
MAX4195ESA	-40°C to +85°C	8 SO
MAX4196ESA	-40°C to +85°C	8 SO
MAX4197ESA	-40°C to +85°C	8 SO

Selector Guide

PART	SHUTDOWN	GAIN (V/V)	CMRR (dB)
MAX4194	No	Variable	95 (G = $+1V/V$)
MAX4195	Yes	+1	95
MAX4196	Yes	+10	115
MAX4197	Yes	+100	115



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Operating Temperature Range	40°C to +85°C
Junction Temperature	+150°C
Storage Temperature Range	65°C to +150°C
Lead Temperature (soldering, 10s)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Electrical Characteristics

 $(V_{CC} = +5V, V_{EE} = 0V, R_L = 25k\Omega$ tied to $V_{CC}/2$, $V_{REF} = V_{CC}/2$, $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.)

PARAMETER	SYMBOL	COND	ITIONS	MIN	ТҮР	МАХ	UNITS	
		Inferred by PSR test	Single supply	2.7		7.5	V	
Supply Voltage Range	Vcc	Interfed by For lest	Dual supplies	±1.35		±3.75		
Quiescent Current	ICC	$V_{IN+} = V_{IN-} = V_{CC}/2, V$	DIFF = 0V		93	110	μA	
Shutdown Current	ISHDN	ISHDN = VIL, MAX4195/	MAX4196/MAX4197 only		8	12	μA	
		$G = +1V/V, V_{CM} = V_{CC}$	c/2, T _A = +25°C		±100	±450		
		$G = +10V/V, V_{CM} = V_{C}$	C/2, T _A = +25°C		±75	±225		
Input Offset Voltage		$G = +100V/V, V_{CM} = V$	$T_{CC}/2, T_{A} = +25^{\circ}C$		±50	±225		
	Maa	G = +1000V/V, V _{CM} =	$V_{CC}/2, T_{A} = +25^{\circ}C$		±50			
	Vos	$G = +1V/V, V_{CM} = V_{CC}$	$_{\rm C}/2$, T _A = T _{MIN} to T _{MAX}		±100	±690	μV	
		$G = +1V/V$, $V_{CM} = V_{CC}/2$, $T_A = T_{MIN}$ to T_{MAX}			±75	±345		
		$G = +100V/V$, $V_{CM} = V_{CC}/2$, $T_A = T_{MIN}$ to T_{MAX}			±50	±345		
		$G = +1000V/V$, $V_{CM} = V_{CC}/2$, $T_A = T_{MIN}$ to T_{MAX}			±50			
Input Offset Voltage Drift	TC _{VOS}	G = +1V/V			±1.0	±4.0	uV/°C	
(Note 1)	10008	$G \ge +10V/V$			±0.5	±2.0	μν/ Ο	
Input Resistance	Rin	$V_{CM} = V_{CC}/2$	Differential		1000		MΩ	
Input nesistance	T UN		Common mode		1000			
Input Capacitance	CIN	Vcm = Vcc/2	Differential		1		- pF	
Input Capacitance	CIN	VCM - VCC/2	Common mode		4		ρr	
Input Voltage Range	VIN	Inferred from CMR test	t	V _{EE} - 0.2		V _{CC} - 1.1	V	
		$V_{CM} = V_{EE} - 0.2V$	G = +1V	66	78		dB	
		to $V_{CC} - 1.1V$, $T_A = +25^{\circ}C$,	G = +10V	80	94			
DC Common-Mode	0145	$\Delta R_{\rm S} = 1 k \Omega (\text{Note 1})$	G = +100V	86	99			
Rejection	CMR _{DC}	$V_{CM} = V_{EE} - 0.2V$	G = +1V	60	78			
		to V_{CC} - 1.1V, T _A = T _{MIN} to T _{MAX} ,	G = +10V	74	94			
		$\Delta R_{\rm S} = 1 k \Omega$, (Note 1)	G = +100V	77	99			

Micropower, Single-Supply, Rail-to-Rail, Precision Instrumentation Amplifiers

Electrical Characteristics (continued)

 $(V_{CC} = +5V, V_{EE} = 0V, R_L = 25k\Omega$ tied to $V_{CC}/2$, $V_{REF} = V_{CC}/2$, $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.)

PARAMETER	SYMBOL	CONI	MIN	TYP	MAX	UNITS	
			G = +1V	78	95		
		$V_{CM} = V_{EE} + 0.2V$ to $V_{CC} - 1.1V$,	G = +10V	93	115		-
		$T_{A} = +25^{\circ}C,$ $\Delta R_{S} = 1k\Omega$	G = +100V/V	95	115		
DC Common-Mode			G = +1000V/V		115		
Rejection	CMR _{DC}		G = +1V	73	95		dB
·		$V_{CM} = V_{EE} + 0.2V$	G = +10V	88	115		-
		to V_{CC} - 1.1V, T _A = T _{MIN} to T _{MAX} ,		-			
		$\Delta R_{\rm S} = 1 k \Omega$	G = +100V/V	90	115		-
			G = +1000V/V		115		
AC Common-Mode	0145	$V_{CM} = V_{EE} + 0.2V$	G = +1V		85		
Rejection	CMR _{AC}	to V_{CC} - 1.1V, f = 120Hz	G = +10V		101		dB
			G = +100V		106		
Power-Supply Rejection	PSR	$+2.7V \le V_{CC} \le +7.5V;$ $V_{OUT} = +1.5V; V_{REF} =$ +1.5V; G = +1V/V, +1	= +1.5V; R_{L} = 25k Ω to	90	120		dB
Input Bias Current	Ι _Β	$V_{CM} = V_{CC}/2$			6	20	nA
Input Bias Current Drift	TCIB	$V_{CM} = V_{CC}/2$			15		pA/°C
Input Offset Current	los	$V_{CM} = V_{CC}/2$			±1.0	±3.0	nA
Input Offset Current Drift	TCIOS	$V_{CM} = V_{CC}/2$			15		pA/°C
		G = +1V/V	f = 10Hz		85		nV/√Hz
			f = 100Hz		75		
			f = 10 kHz		72		
			f = 0.1Hz to $10Hz$		1.4		μVRMS
			f = 10Hz		35		nV/√Hz
Input Naina Valtaga			f = 100Hz		32		
Input Noise Voltage	en	G = +10V/V	f = 10 kHz		31		
			f = 0.1Hz to $10Hz$		0.7		μV _{RMS}
			f = 10Hz		32		
			f = 100Hz		31		nV/√Hz
		G = +100V/V	f = 10 kHz		8.7		
			f = 0.1Hz to $10Hz$		0.6		μVRMS
		f = 10Hz			2.4		
Input Noine Current		f = 100Hz			0.76		pA/√Hz
Input Noise Current	in	f = 10 kHz			0.1		7
		f = 0.1Hz to 10Hz			16		pARMS
			V _{CC} - V _{OH}		30	100	
Outout Valtage Outer		$R_L = 25k\Omega$ to $V_{CC}/2$	V _{OL}		30	100	mV
Output Voltage Swing	V _{OH} , V _{OL}	$P_{\rm L} = 5kO + 2\lambda/22/2$	V _{CC} - V _{OH}		100	200	
		$R_L = 5k\Omega$ to $V_{CC}/2$	VOL		100	200	

Micropower, Single-Supply, Rail-to-Rail, Precision Instrumentation Amplifiers

Electrical Characteristics (continued)

 $(V_{CC} = +5V, V_{EE} = 0V, R_L = 25k\Omega$ tied to $V_{CC}/2$, $V_{REF} = V_{CC}/2$, $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.)

PARAMETER	SYMBOL	c	ONDITIC	ONS		MIN	ТҮР	MAX	UNITS	
Short-Circuit Current (Note 2)	ISC						±4.5		mA	
Gain Equation		MAX4194 only					1 + (50kΩ/R _G)			
		T _A = +25°C,	G = +	+1V			±0.01 ±0.1			
		$V_{CM} = V_{CC}/2,$	G = +	+10V			±0.03	±0.3		
		$\begin{aligned} R_L &= 25 k \Omega, \\ V_EE &+ 0.1 V \leq V_OU \end{aligned}$	_{гт} G = +	+100V/	/		±0.05	±0.5		
		$\leq V_{CC} - 0.1V$		-1000V/	V, MAX4194		±0.5			
Gain Error		T _A = +25°C,	G = +	+1V			±0.01	±0.1	%	
		$V_{CM} = V_{CC}/2,$	G = +	+10V			±0.03	±0.3		
		R _L = 5kΩ, V _{EE} + 0.2V ≤ V _{OU}	_{гт} G = +	+100V/	/		±0.05	±0.5		
		$\leq V_{CC} - 0.2V$	-	-1000V/	V, MAX4194		±0.5			
Gain Temperature		MAX4194/MAX41					±1	±8	100	
Coefficient (Note 1)		MAX4196/MAX4197			±1	±15	ppm/°C			
50kΩ Resistance Temperature Coefficient (Note 3)	TC _{50kΩ}	MAX4194				±16		ppm/°C		
Nonlinearity		$V_{EE} + 0.1V \le V_{OUT} \le V_{CC} - 0.1V,$ $V_{CM} = V_{CC}/2, G = +1V/V, +10V/V,$ +100V/V, +1000V/V				±0.001		%		
Capacitive-Load Stability	CL						300		pF	
			G = +1V/V MAX4194			250				
			G = +1V	// V	MAX4195		220			
			G = +10V/V G = +100V/V		MAX4194		17		kHz	
-3dB Bandwidth	BW-3dB	$V_{OUT} \le 0.1 V_{P-P},$ $V_{CM} = V_{CC}/2$			MAX4196		34			
					MAX4194		1.5			
			G = 110		MAX4197		3.1			
			G = +10	V/V000	MAX4194		0.147			
Slew Rate	SR	$V_{OUT} = 2V_{P-P}, G$					0.06		V/µs	
				i = +1V			0.05			
Settling Time	ts	0.1%, V _{OUT} = 2V	P-P	i = +10			0.04		ms	
3	0	,	G	i = +10			5			
			G = +1000V/V			7				
Total Harmonic Distortion	THD	$V_{OUT} = 2V_{P-P}, G$	= +1V/V,	t = 1kł	ΗZ) / · · -	0.001		%	
Input Logic Voltage High	VIH					V _{CC} - 1.5			V	
Input Logic Voltage Low	VIL	V _{EE} < V <u>SHDN</u> < V		IAX419 IAX419	5/MAX4196/			V _{CC} - 2.5 ±0.1	V µA	

Micropower, Single-Supply, Rail-to-Rail, Precision Instrumentation Amplifiers

Electrical Characteristics (continued)

(V_{CC} = +5V, V_{EE} = 0V, R_L = 25k Ω tied to V_{CC}/2, V_{REF} = V_{CC}/2, T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at T_A = +25°C.)

PARAMETER	SYMBOL	CONDITIONS		MIN	ТҮР	MAX	UNITS
Time to Shutdown	t <u>SHDN</u>	G = +1V/V, 0.1%, $V_{OUT} = +3V$	MAX4195/MAX4196/ MAX4197 only		0.5		ms
Enable Time From Shutdown	^t ENABLE	G = +1V/V, 0.1%, V _{OUT} = +3.5V	MAX4195/MAX4196/ MAX4197 only		0.5		ms
Power-Up Delay		$G = +1V/V, 0.1\%, V_{OUT}$	G = +1V/V, 0.1%, V _{OUT} = +3.5V		1		ms
On/Off Settling Time	ton/off	$V_{SHDN} = V_{CC} - 2.5V$ to $V_{CC} - 1.5V$, G = +100V/V, 0.1%, $V_{OUT} = +3.5V$			0.5		ms

Note 1: Guaranteed by design.

Note 2: Maximum output current (sinking/sourcing) in which the gain changes by less than 0.1%.

Note 3: This specification represents the typical temperature coefficient of an on-chip thin film resistor. In practice, the temperature coefficient of the gain for the MAX4194 will be dominated by the temperature coefficient of the external gain-setting resistor.

Typical Operating Characteristics

(V_{CC} = +5V, V_{EE} = 0, R_L = 25k\Omega tied to V_CC/2, T_A = +25°C, unless otherwise noted.)



Micropower, Single-Supply, Rail-to-Rail, Precision Instrumentation Amplifiers

Typical Operating Characteristics (continued)

 $(V_{CC} = +5V, V_{EE} = 0, R_L = 25k\Omega$ tied to $V_{CC}/2, T_A = +25^{\circ}C$, unless otherwise noted.)



Micropower, Single-Supply, Rail-to-Rail, Precision Instrumentation Amplifiers

Typical Operating Characteristics (continued)

(V_{CC} = +5V, V_{EE} = 0, R_L = 25k\Omega tied to V_{CC}/2, T_A = +25^{\circ}C, unless otherwise noted.)











MAX4195/MAX4196 TOTAL HARMONIC DISTORTION PLUS NOISE vs. FREQUENCY



SUPPLY CURRENT vs. TEMPERATURE







Micropower, Single-Supply, Rail-to-Rail, Precision Instrumentation Amplifiers

Ν

Typical Operating Characteristics (continued)

(V_{CC} = +5V, V_{EE} = 0, R_L = 25k Ω tied to V_{CC}/2, T_A = +25°C, unless otherwise noted.)





Pin Description

Р	IN			
MAX4194	MAX4195 MAX4196 MAX4197	NAME	FUNCTION	
1, 8	_	RG-, RG+	Connection for Gain-Setting Resistor	
5	1	REF	Reference Voltage. Offsets output voltage.	
2	2	IN-	Inverting Input	
3	3	IN+	Noninverting Input	
4	4	VEE	Negative Supply Voltage	
	5	FB	Feedback. Connects to OUT.	
6	6	OUT	Amplifier Output	
7	7	Vcc	Positive Supply Voltage	
	8	SHDN	Shutdown Control	

Micropower, Single-Supply, Rail-to-Rail, Precision Instrumentation Amplifiers

Detailed Description

Input Stage

The MAX4194–MAX4197 family of low-power instrumentation amplifiers implements a three-amplifier topology (Figure 1). The input stage is composed of two operational amplifiers that together provide a fixed-gain differential and a unity common-mode gain. The output stage is a conventional differential amplifier that provides an overall common-mode rejection of 115dB (G =



Figure 1. MAX4194 Simplified Block Diagram



Figure 2. MAX4195/MAX4196/MAX4197 Simplified Block Diagram

+10V/V). The MAX4194's gain can be externally set between +1V/V and +10,000V/V (Table 1). The MAX4195/MAX4196/MAX4197 have on-chip gain-setting resistors (Figure 2), and their gains are fixed at +1V/V, +10V/V, and +100V/V, respectively.

Input Voltage Range and Detailed Operation

The common-mode input range for all of these amplifiers is VEE - 0.2V to V_{CC} - 1.1V. Ideally, the instrumentation amplifier (Figure 3) responds only to a differential voltage applied to its inputs, IN+ and IN-. If both inputs are at the same voltage, the output is V_{REF}. A differential voltage at IN+ (V_{IN+}) and IN- (V_{IN-}) develops an identical voltage across the gain-setting resistor, causing a current (IG) to flow. This current also flows through the feedback resistors of the two input amplifiers A1 and A2, generating a differential voltage of:

$$VOUT2 - VOUT1 = IG \cdot (R1 + RG + R1)$$

where V_{OUT1} and V_{OUT2} are the output voltages of A1 and A2, R_G is the gain-setting resistor (internal or external to the part), and R1 is the feedback resistor of the input amplifiers.

IG is determined by the following equation:

$$I_G = (V_{IN+} - V_{IN-}) / R_G$$

The output voltage (V_{OUT}) for the instrumentation amplifier is expressed in the following equation:

$$V_{OUT} = (V_{IN+} - V_{IN-}) \cdot [(2 \cdot R1) / R_G] + 1$$

The common-mode input range is a function of the amplifier's output voltage and the supply voltage. With a power supply of V_{CC}, the largest output signal swing can be obtained with REF tied to V_{CC}/2. This results in an output voltage swing of \pm V_{CC}/2. An output voltage swing less than full-scale increases the common-mode input range.



Figure 3. Instrumentation Amplifier Configuration

Micropower, Single-Supply, Rail-to-Rail, Precision Instrumentation Amplifiers

Table 1. MAX4194 External GainResistor Selection

GAIN (V/V)	CLOSEST R _G (1%) (Ω)	CLOSEST R _G (5%) (Ω)
+1	∞ *	∞ *
+2	49.9k	51k
+5	12.4k	12k
+10	5.62k	5.6k
+20	2.61k	2.7k
+50	1.02k	1.0k
+100	511	510
+200	249	240
+500	100	100
+1,000	49.9	51
+2,000	24.9	24
+5,000	10	10
+10,000	4.99	5.1

*Leave pins 1 and 8 open for G = +1V/V.

VCM vs. VOUT Characterization

Figure 4 illustrates the MAX4194 typical common-mode input voltage range over output voltage swing at unitygain (pins 1 and 8 left floating), with a single-supply voltage of $V_{CC} = +5V$ and a bias reference voltage of $V_{REF} = V_{CC}/2 = +2.5V$. Points A and D show the full input voltage range of the input amplifiers ($V_{EE} - 0.2V$ to $V_{CC} - 1.1V$) since, with +2.5V output, there is zero input differential swing. The other points (B, C, E, and F) are determined by the input voltage range of the input amplitude necessary to produce the associated V_{OUT} . For the higher gain configurations, the V_{CM} range will increase at the endpoints (B, C, E, and F) since a smaller differential voltage is necessary for the given output voltage.

Rail-to-Rail Output Stage

The MAX4194–MAX4197's output stage incorporates a common-source structure that maximizes the dynamic range of the instrumentation amplifier.

The output can drive up to a $25 k\Omega$ (tied to $V_{CC}/2$) resistive load and still typically swing within 30mV of the rails. With an output load of $5 k\Omega$ tied to $V_{CC}/2$, the output voltage swings within 100mV of the rails.

Shutdown Mode

The MAX4195–MAX4197 feature a low-power shutdown mode. When the shutdown pin (SHDN) is pulled low, the internal amplifiers are switched off and the supply current drops to 8µA typically (Figures 5a, 5b, and 5c).

This disables the instrumentation amplifier and puts its output in a high-impedance state. Pulling \overline{SHDN} high enables the instrumentation amplifier.

Applications Information

Setting the Gain (MAX4194)

The MAX4194's gain is set by connecting a single, external gain resistor between the two RG pins (pin 1 and pin 8), and can be described as:

$$G = 1 + 50k\Omega / R_G$$

where G is the instrumentation amplifier's gain and R_G is the gain-setting resistor.

The 50k Ω resistor of the gain equation is the sum of the two resistors internally connected to the feedback loops of the IN+ and IN- amplifiers. These embedded feedback resistors are laser trimmed, and their accuracy and temperature coefficients are included in the gain and drift specification for the MAX4194.



Figure 4. Common-Mode Input Voltage vs. Output Voltage



Figure 5a. MAX4195 Shutdown Mode

Micropower, Single-Supply, Rail-to-Rail, Precision Instrumentation Amplifiers



Figure 5b. MAX4196 Shutdown Mode



Figure 5c. MAX4197 Shutdown Mode

The accuracy and temperature drift of the R_G resistors also influence the IC's precision and gain drift, and can be derived from the equation above. With low R_G values, which are required for high-gain operation, parasitic resistances may significantly increase the gain error.

Capacitive-Load Stability

The MAX4194–MAX4197 are stable for capacitive loads up to 300pF (Figure 6a). Applications that require greater capacitive-load driving capability can use an isolation resistor (Figure 6b) between the output and the capacitive load to reduce ringing on the output signal. However, this alternative reduces gain accuracy because RISO (Figure 6c) forms a potential divider with the load resistor.



Figure 6a. Using a Resistor to Isolate a Capacitive Load from the Instrumentation Amplifier (G = +1V/V)



Figure 6b. Small-Signal Pulse Response with Excessive Capacitive Load ($R_L = 25k\Omega$, $C_L = 1000pF$)



Figure 6c. Small-Signal Pulse Response with Excessive Capacitive Load and Isolating Resistor ($R_{ISO} = 75\Omega$, $R_L = 25k\Omega$, $C_L = 1000$ pF)

Micropower, Single-Supply, Rail-to-Rail, Precision Instrumentation Amplifiers

Power-Supply Bypassing and Layout

Good layout technique optimizes performance by decreasing the amount of stray capacitance at the instrumentation amplifier's gain-setting pins. Excess capacitance will produce peaking in the amplifier's frequency response. To decrease stray capacitance, minimize trace lengths by placing external components as close to the instrumentation amplifier as possible. For best performance, bypass each power supply to ground with a separate 0.1µF capacitor.

Transducer Applications

The MAX4194–MAX4197 instrumentation amplifiers can be used in various signal-conditioning circuits for thermocouples, PT100s, strain gauges (displacement sensors), piezoresistive transducers (PRTs), flow sensors, and bioelectrical applications. Figure 7 shows a simplified example of how to attach four strain gauges (two identical two-element strain gauges) to the inputs of the MAX4194. The bridge contains four resistors, two of which increase and two of which decrease by the same ratio.

With a fully balanced bridge, points A (IN+) and B (IN-) see half the excitation voltage (V_{BRIDGE}). The low impedance (120Ω to 350Ω) of the strain gauges, however, could cause significant voltage drop contributions by the wires leading to the bridge, which would cause excitation variations. Output voltage V_{OUT} can be calculated as follows:

$VOUT = VAB \cdot G$

where G = (1 + 50k Ω / $R_G)$ is the gain of the instrumentation amplifier.

Since V_{AB} is directly proportional to the excitation, gain errors may occur.



Figure 7. Strain Gauge Connection to the MAX4194

Chip Information

TRANSISTOR COUNT: 432

Package Information

For the latest package outline information and land patterns (footprints), go to **www.maximintegrated.com/packages**. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE	PACKAGE	OUTLINE	LAND
TYPE	CODE	NO.	PATTERN NO.
8 SO	S8-5	<u>21-0041</u>	<u>90-0096</u>

Micropower, Single-Supply, Rail-to-Rail, Precision Instrumentation Amplifiers

Revision History

REVISION	REVISION	DESCRIPTION	PAGES
NUMBER	DATE		CHANGED
2	5/15	Updated Benefits and Features section	1

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Вы можете разместить у нас заказ для любого Вашего проекта, будь то серийное производство или разработка единичного прибора.

В нашем ассортименте представлены ведущие мировые производители активных и пассивных электронных компонентов.

Нашей специализацией является поставка электронной компонентной базы двойного назначения, продукции таких производителей как XILINX, Intel (ex.ALTERA), Vicor, Microchip, Texas Instruments, Analog Devices, Mini-Circuits, Amphenol, Glenair.

Сотрудничество с глобальными дистрибьюторами электронных компонентов, предоставляет возможность заказывать и получать с международных складов практически любой перечень компонентов в оптимальные для Вас сроки.

На всех этапах разработки и производства наши партнеры могут получить квалифицированную поддержку опытных инженеров.

Система менеджмента качества компании отвечает требованиям в соответствии с ГОСТ Р ИСО 9001, ГОСТ РВ 0015-002 и ЭС РД 009

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