

Single/Dual/Quad, SOT23, Single-Supply, High-Speed, Low-Power Comparators

General Description

The MAX976/MAX978/MAX998 dual/quad/single, high-speed, low-power comparators are optimized for +3V/+5V single-supply applications. They achieve a 20ns propagation delay while consuming only 225μA supply current per comparator. The MAX998 features a low-power shutdown mode that places the output in a high-impedance state and reduces supply current to 1nA.

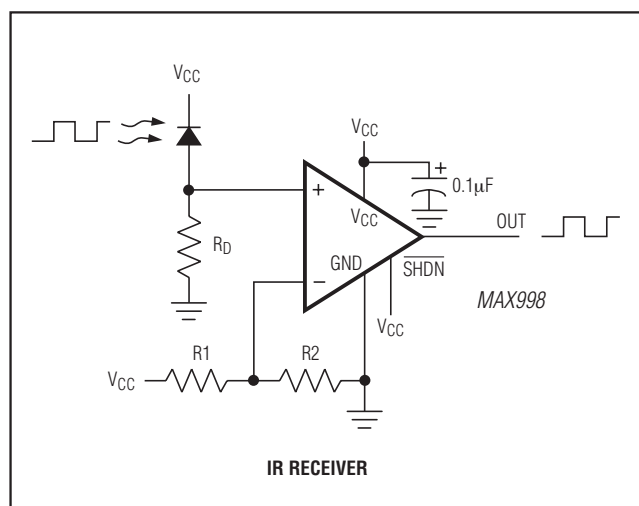
The MAX976/MAX978/MAX998 inputs have a common-mode voltage range that extends 200mV below ground. Their outputs are capable of rail-to-rail operation without external pullup circuitry, making these devices ideal for interface with CMOS/TTL logic. All inputs and outputs can tolerate a continuous short-circuit fault condition to either rail. The comparators' internal hysteresis ensures clean output switching, even with slow-moving input signals.

For space-critical applications, the single MAX998 is available in a 6-pin SOT23 package, the dual MAX976 is available in an 8-pin μMAX® package, and the quad MAX978 is available in a 16-pin QSOP package.

Applications

Battery-Powered Systems
 Threshold Detectors/Discriminators
 3V Systems
 IR Receivers
 Digital Line Receivers

Typical Operating Circuit



μMAX a registered trademark of Maxim Integrated Products, Inc.

Features

- ◆ Single-Supply Operation Down to 2.7V
- ◆ 20ns Propagation Delay
- ◆ 225μA Supply Current
- ◆ 1nA Shutdown Supply Current
- ◆ Rail-to-Rail Outputs
- ◆ Ground-Sensing Inputs
- ◆ Internal Hysteresis Ensures Clean Switching
- ◆ Available in Space-Saving Packages
 - SOT23 (MAX998)
 - μMAX (MAX976)
 - QSOP (MAX978)

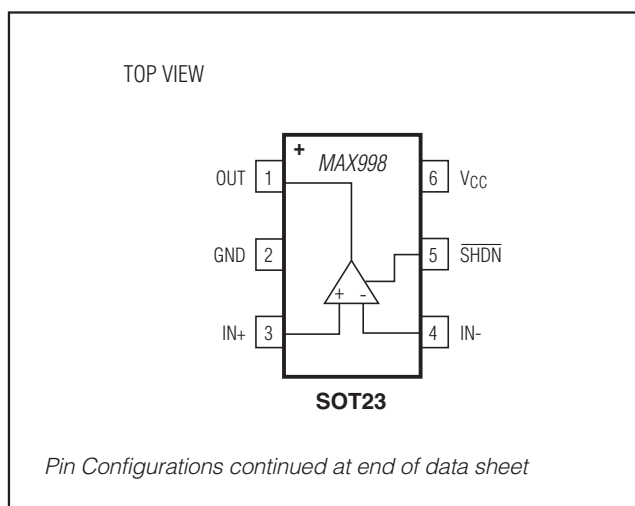
Ordering Information

PART	PIN-PACKAGE	SOT23 TOP MARK
MAX976ESA+	8 SO	—
MAX976EUA+	8 μMAX	—
MAX978ESE+	16 Narrow SO	—
MAX978EEE+	16 QSOP	—
MAX998ESA+	8 SO	—
MAX998EUT+T	6 SOT23	AAAO

Note: All devices are specified over the -40°C to +85°C temperature range.

+ Denotes a lead(Pb)-free/RoHS-compliant package.

Pin Configurations



MAX976/MAX978/MAX998

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ABSOLUTE MAXIMUM RATINGS

Supply Voltage (V_{CC}).....+6V
 SHDN (MAX998).....-0.3V to 6V
 All Other Pins.....-0.3V to ($V_{CC} + 0.3$ V)
 Current into Input Pins..... ± 20 mA
 Duration of Output Short Circuit to GND or V_{CC}Continuous
 Continuous Power Dissipation ($T_A = +70^\circ\text{C}$)
 6-Pin SOT23-6 (derate 7.1mW/ $^\circ\text{C}$ above $+70^\circ\text{C}$)571mW
 8-Pin μMAX (derate 4.10mW/ $^\circ\text{C}$ above $+70^\circ\text{C}$)330mW

8-Pin SO (derate 5.88mW/ $^\circ\text{C}$ above $+70^\circ\text{C}$).....471mW
 16-Pin Narrow SO (derate 8.70mW/ $^\circ\text{C}$ above $+70^\circ\text{C}$) ..696mW
 16-Pin QSOP (derate 8.33mW/ $^\circ\text{C}$ above $+70^\circ\text{C}$).....667mW
 Operating Temperature Range -40°C to $+85^\circ\text{C}$
 Storage Temperature Range -65°C to $+160^\circ\text{C}$
 Lead Temperature (soldering, 10s) $+300^\circ\text{C}$
 Soldering Temperature (reflow) $+260^\circ\text{C}$

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

($V_{CC} = +2.7\text{V}$ to $+5.5\text{V}$, $V_{CM} = 0\text{V}$, $T_A = -40^\circ\text{C}$ to $+85^\circ\text{C}$, unless otherwise noted. Typical values are at $T_A = +25^\circ\text{C}$.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Supply Voltage Range	V_{CC}	Inferred from PSRR test		2.7		5.5	V
Supply Current per Comparator	I_{CC}	$V_{CC} = 5.5\text{V}$			300	650	μA
		$V_{CC} = 2.7\text{V}$			225		
Shutdown Supply Current	I_{SD}	MAX998 only, $\overline{\text{SHDN}} = \text{GND}$			1	500	nA
Power-Supply Rejection Ratio	PSRR	$2.7\text{V} < V_{CC} < 5.5\text{V}$		63	100		dB
Common-Mode Voltage Range	V_{CMR}	(Note 2)		-0.2	$V_{CC} - 1.2$		V
Common-Mode Rejection Ratio	CMRR	$-0.2\text{V} \leq V_{CM} \leq (V_{CC} - 1.2\text{V})$		66	95		dB
Input Offset Voltage	V_{OS}	$V_{CC} = 5\text{V}$ (Note 3)	$T_A = +25^\circ\text{C}$		0.2	± 2	mV
			$T_A = T_{MIN}$ to T_{MAX}			± 3	
Input-Referred Hysteresis	V_{HYS}	$V_{CC} = 5\text{V}$ (Note 4)	MAX976EUA, MAX998EUT	0.3	1.5	5.0	mV
			All others	0.5	1.5	4.0	
Input Bias Current	I_B				75	300	nA
Input Offset Current	I_{OS}				± 5	± 100	nA
OUT Output-Voltage High	V_{OH}	$I_{SOURCE} = 2\text{mA}$, $V_{CC} - V_{OH}$			0.1	0.4	V
OUT Output-Voltage Low	V_{OL}	$I_{SINK} = 2\text{mA}$			0.1	0.4	V
OUT Short-Circuit Current	I_{SH}	$V_{CC} = 5.5\text{V}$	Sinking		74		mA
			Sourcing		90		
Input Capacitance	C_{IN}				3		pF
$\overline{\text{SHDN}}$ Input-Voltage High	V_{IH}	MAX998 only		$0.65 \times V_{CC}$			V
$\overline{\text{SHDN}}$ Input-Voltage Low	V_{IL}	MAX998 only		$0.2 \times V_{CC}$			V
OUT Leakage Current	I_{OUT}	MAX998 only, $\overline{\text{SHDN}} = \text{GND}$, $V_{OUT} = 0\text{V}$ to V_{CC}			1	200	nA
$\overline{\text{SHDN}}$ Input Current	$I_{\overline{\text{SHDN}}}$	MAX998 only			1	200	nA
Propagation Delay	t_{PD}	$C_{LOAD} = 10\text{pF}$, $V_{CC} = 5\text{V}$ (Note 5)	Overdrive = 5mV		28		ns
			Overdrive = 50mV		20	40	
Propagation-Delay Skew	t_{SKEW}	$C_{LOAD} = 10\text{pF}$ (Note 6)			2		ns
Propagation-Delay Matching Between Channels	Δt_{PD}	MAX976/MAX978 only			1		ns
Output Rise/Fall Time	t_R/t_F	$C_{LOAD} = 10\text{pF}$			1.6		ns
Shutdown Delay Time	t_{SD}	MAX998 only, $V_{CC} = 5\text{V}$, $I_{CC} = 10\%$ of typical			5		μs

MAX976/MAX978/MAX998

Single/Dual/Quad, SOT23, Single-Supply, High-Speed, Low-Power Comparators

ELECTRICAL CHARACTERISTICS (continued)

($V_{CC} = +2.7V$ to $+5.5V$, $V_{CM} = 0V$, $T_A = -40^{\circ}C$ to $+85^{\circ}C$, unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$.) (Note 1)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Wake-Up from Shutdown	t_{EN}	MAX998 only, $V_{CC} = 5V$, $I_{CC} = 90\%$ of typical (Note 7)		15		μs
Power-Up Delay	t_{PU}	$V_{CC} = 0V$ to $5V$ step, output valid		3		μs

Note 1: The MAX998EUT specifications are 100% tested at $T_A = +25^{\circ}C$. Limits over the extended temperature range are guaranteed by design, not production tested.

Note 2: Inferred from CMRR test. Either input can be driven to the absolute maximum limit without false output inversion, as long as the other input is within the common-mode voltage range.

Note 3: V_{OS} is defined as the mean of trip points. The trip points are the extremities of the differential input voltage required to make the comparator output change state (Figure 1).

Note 4: The difference between the upper and lower trip points is equal to the width of the input-referred hysteresis zone (Figure 1).

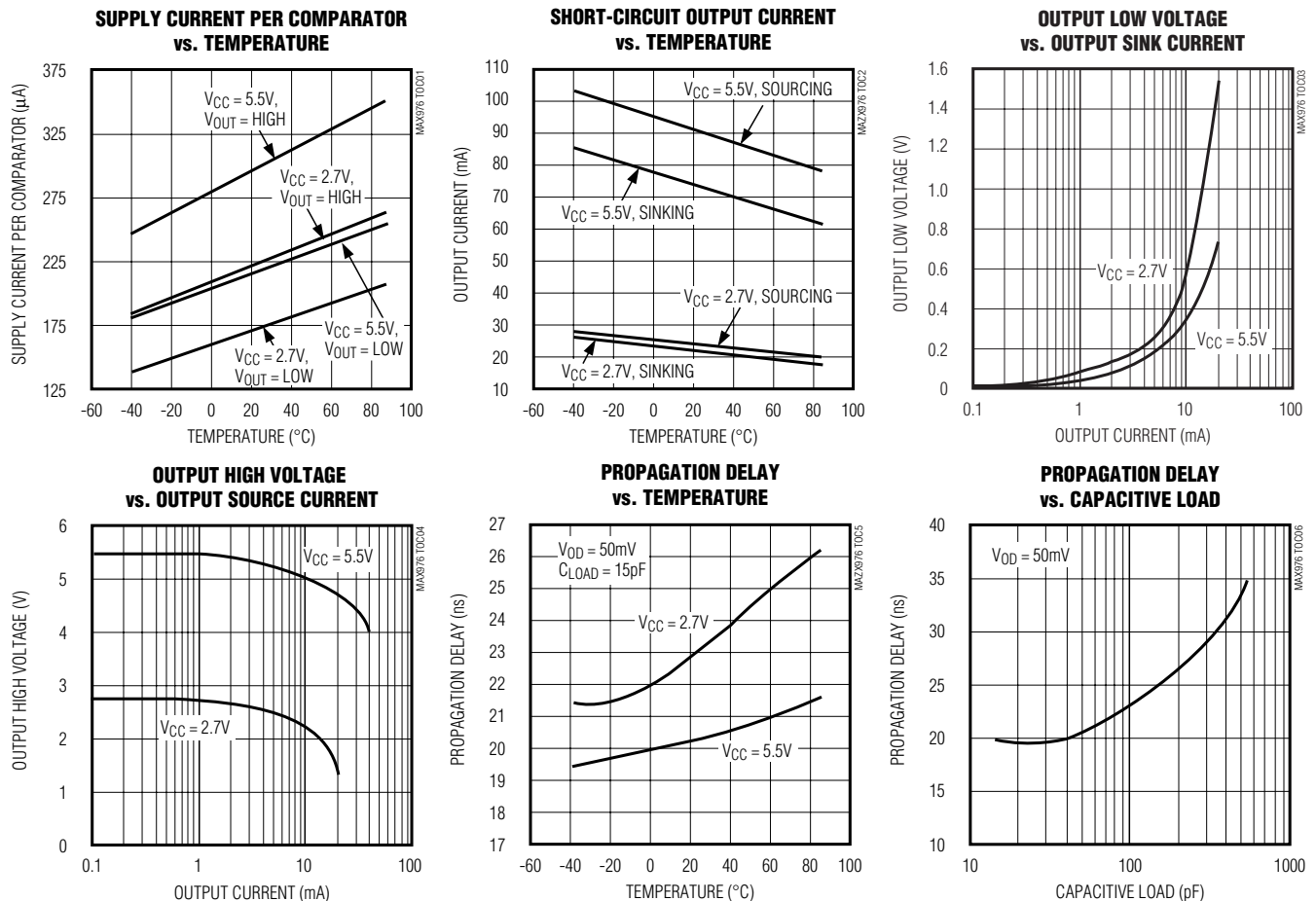
Note 5: Propagation Delay is guaranteed by design. For low overdrive conditions, V_{TRIP} (Figure 1) is added to the overdrive.

Note 6: Propagation-Delay Skew is the difference between the positive-going and the negative-going propagation delay.

Note 7: For design purposes, the t_{EN} can be as high as $60\mu s$.

Typical Operating Characteristics

($V_{CC} = +5V$, $V_{CM} = 0V$, $T_A = +25^{\circ}C$, unless otherwise noted.)

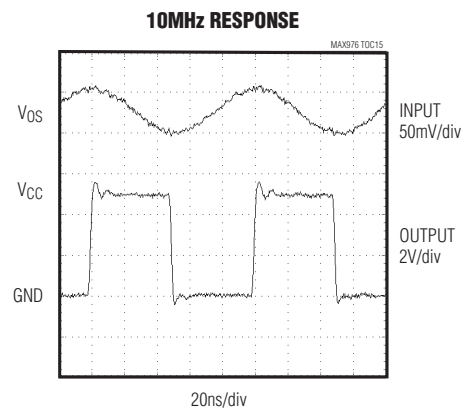
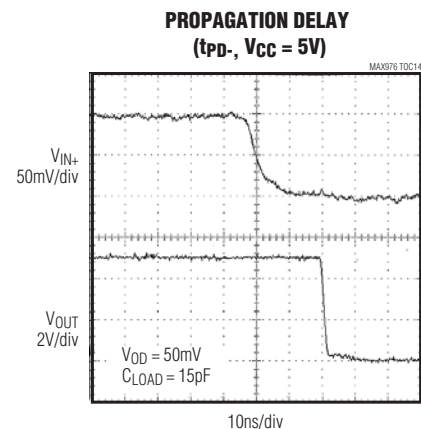
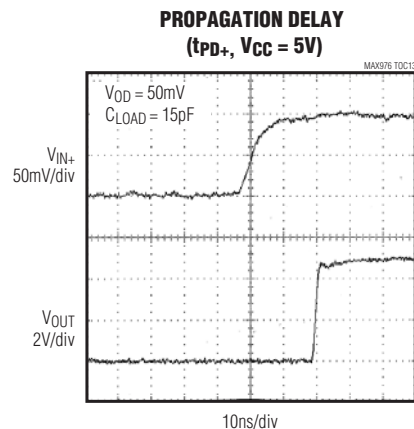
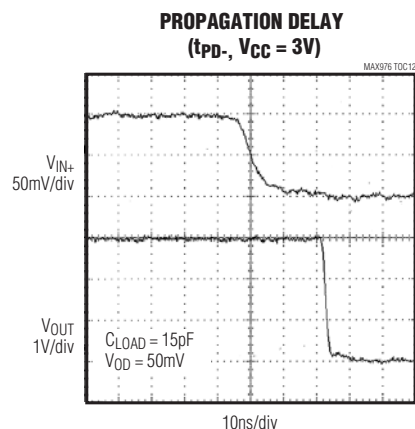
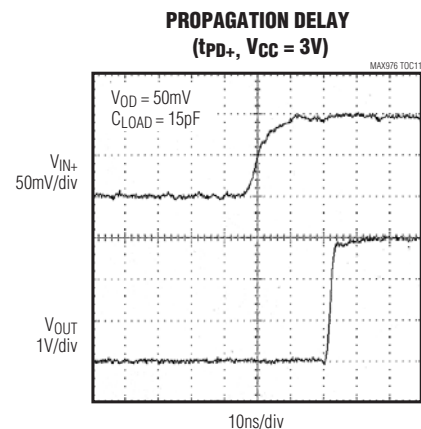
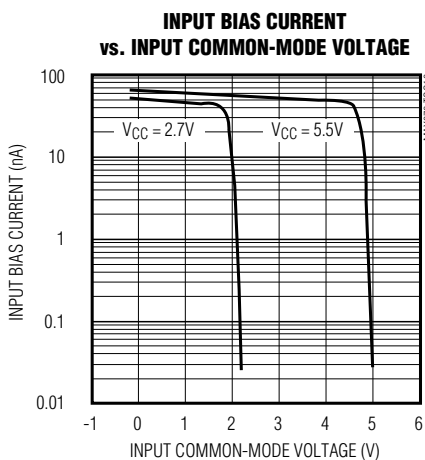
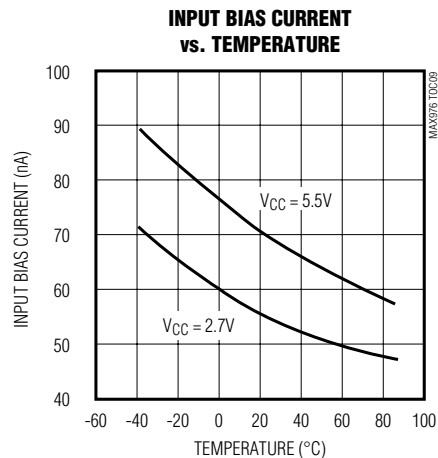
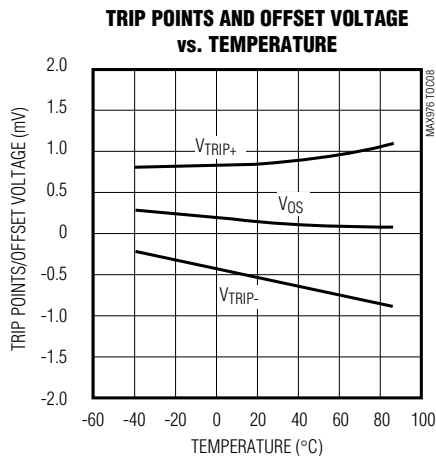
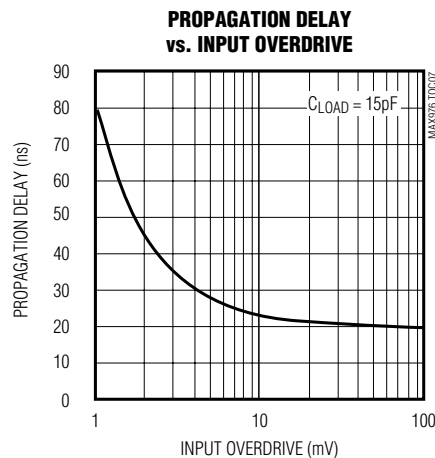


MAX976/MAX978/MAX998

Single/Dual/Quad, SOT23, Single-Supply, High-Speed, Low-Power Comparators

Typical Operating Characteristics (continued)

($V_{CC} = +5V$, $V_{CM} = 0V$, $T_A = +25^\circ C$, unless otherwise noted.)

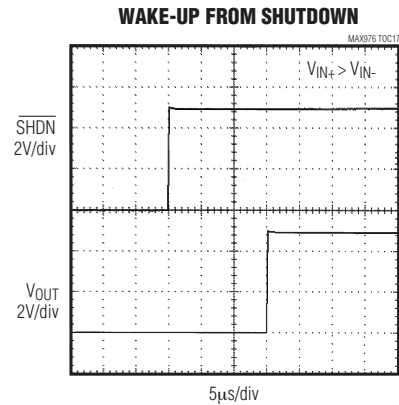
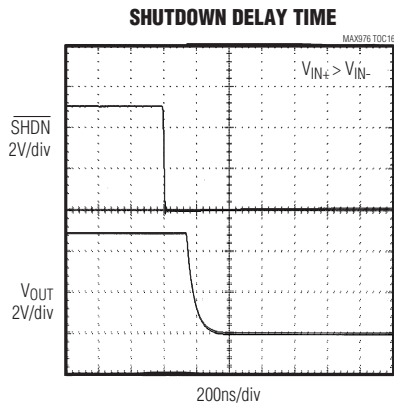


MAX976/MAX978/MAX998

Single/Dual/Quad, SOT23, Single-Supply, High-Speed, Low-Power Comparators

Typical Operating Characteristics (continued)

(VCC = +5V, VCM = 0V, TA = +25°C, unless otherwise noted.)



Pin Description

PIN				NAME	FUNCTION
MAX976	MAX978	MAX998			
SO/μMAX	SO/QSOP	SOT23-6	SO		
1, 3	1, 3, 5, 7	3	3	IN_+	Comparator Noninverting Input
2, 4	2, 4, 6, 8	4	2	IN_-	Comparator Inverting Input
5	9, 13	2	4	GND	Ground
6, 7	10, 11, 14, 15	1	6	OUT_	Comparator Output
8	12, 16	6	7	VCC	Supply Voltage, +2.7V to +5.5V
—	—	—	1,5	N.C.	No Connection. Not internally connected.
—	—	5	8	SHDN	Shutdown Input. Drive low for shutdown mode. Drive high or connect to VCC for normal operation.

Detailed Description

The MAX976/MAX978/MAX998 dual/quad/single comparators operate from a single +2.7V to +5.5V supply. They achieve a 20ns propagation delay while consuming only 225μA of supply current per comparator. The MAX998 features a low-power shutdown mode that places the output in a high-impedance state and reduces supply current to 1nA. Activate shutdown mode by driving SHDN low.

The MAX976/MAX978/MAX998 comparator inputs have a common-mode voltage range of -0.2V to (VCC - 1.2V).

Either input can be driven to the Absolute Maximum Ratings limit without false output inversion, as long as the other input is within the Common-Mode Voltage Range. Their push/pull output structure is capable of rail-to-rail operation without external pull-up circuitry, making these devices ideal for interfacing with CMOS/TTL logic. All inputs and outputs can tolerate a continuous short-circuit fault condition to either supply. The comparator's internal hysteresis ensures clean output switching, even with slow-moving input signals.

MAX976/MAX978/MAX998

Single/Dual/Quad, SOT23, Single-Supply, High-Speed, Low-Power Comparators

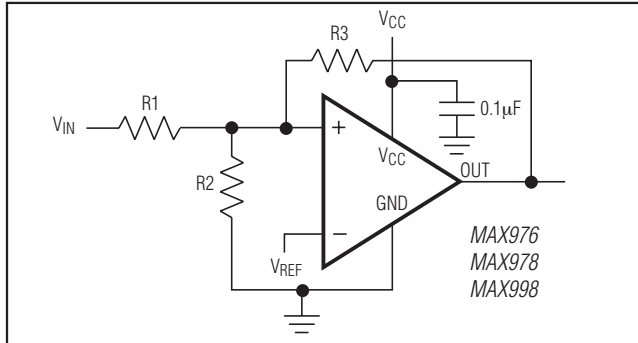


Figure 2. Additional Hysteresis

- 2) Choose the hysteresis band required (V_{HB}). For this example, choose 100mV.
- 3) Calculate R1. $R1 = R3 \times (V_{HB} / V_{CC})$. Plugging in the values for this example,

$$R1 = 1.2M\Omega \times (100mV / 5.0V) = 24k\Omega$$

- 4) Choose the trip point for V_{IN} rising. This is the threshold voltage at which the comparator switches from low to high as V_{IN} rises above the trip point. In this example, choose 3.0V.
- 5) Calculate R2 as follows:

$$R2 = \frac{1}{\left(\frac{V_{THR}}{V_{REF} \times R1}\right) - \frac{1}{R1} - \frac{1}{R3}}$$

$$R2 = \frac{1}{\left(\frac{3.0V}{1.2 \times 24k\Omega}\right) - \frac{1}{24k\Omega} - \frac{1}{1.2M}} = 16.2k\Omega$$

Choose a standard value for R2 of 16kΩ.

- 6) Verify the trip voltage and hysteresis as follows:

$$V_{IN} \text{ rising: } V_{THR} = V_{REF} \times R1 \times \left(\frac{1}{R1} + \frac{1}{R2} + \frac{1}{R3}\right)$$

$$V_{IN} \text{ falling: } V_{THF} = V_{THR} - \left(\frac{R1 \times V_{CC}}{R3}\right)$$

$$\text{Hysteresis} = V_{THR} - V_{THF}$$

IR Receiver

The *Typical Operating Circuit* shows an application using the MAX998 as an infrared receiver. The infrared photodiode creates a current relative to the amount of infrared light present. This current creates a voltage across R_D . When this voltage level crosses the voltage applied by the voltage divider to the inverting input, the output transitions.

Window Comparator

The MAX976 is ideal for making a window detector (undervoltage/overvoltage detector). The schematic shown in Figure 3 uses a MAX6120 reference and component values selected for a 2.0V undervoltage threshold and a 2.5V overvoltage threshold. Choosing different thresholds by changing the values of R1, R2, and R3. OUTA provides an active-low undervoltage indication, and OUTB gives an active-low overvoltage indication. ANDing the two outputs provides an active-high, power-good signal. The design procedure is as follows:

- 1) Select R1. The leakage current into INB- is normally 75nA, so the current through R1 should exceed 1.0µA for the thresholds to be accurate. R1 values in the 50kΩ to 100kΩ range are typical.
- 2) Choose the overvoltage threshold (V_{OTH}) when V_{IN} is rising, and calculate R2 and R3 with the following formula:

$$R_{SUM} = R2 + R3 = R1 \times [V_{OTH} / (V_{REF} + V_H) - 1]$$

where $V_H = 1/2V_{HYST}$.

- 3) Choose the undervoltage threshold (V_{UTH}) when V_{IN} is falling, and calculate R2 with the following formula:

$$R2 = (R1 + R_{SUM}) \times [(V_{REF} - V_H) / V_{UTH}] - R1$$

where $V_H = 1/2V_{HYST}$.

- 4) Calculate R3 with the following formula:

$$R3 = (R_{SUM}) - R2$$

- 5) Verify the resistor values. The equations are as follows:

$$V_{OTH} = (V_{REF} + V_H) \times (R1 + R2 + R3) / R1$$

$$V_{UTH} = (V_{REF} - V_H) \times (R1 + R2 + R3) / (R1 + R2)$$

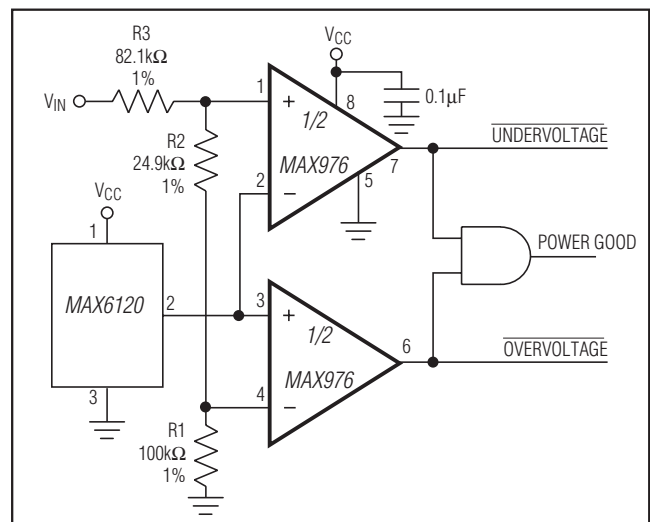
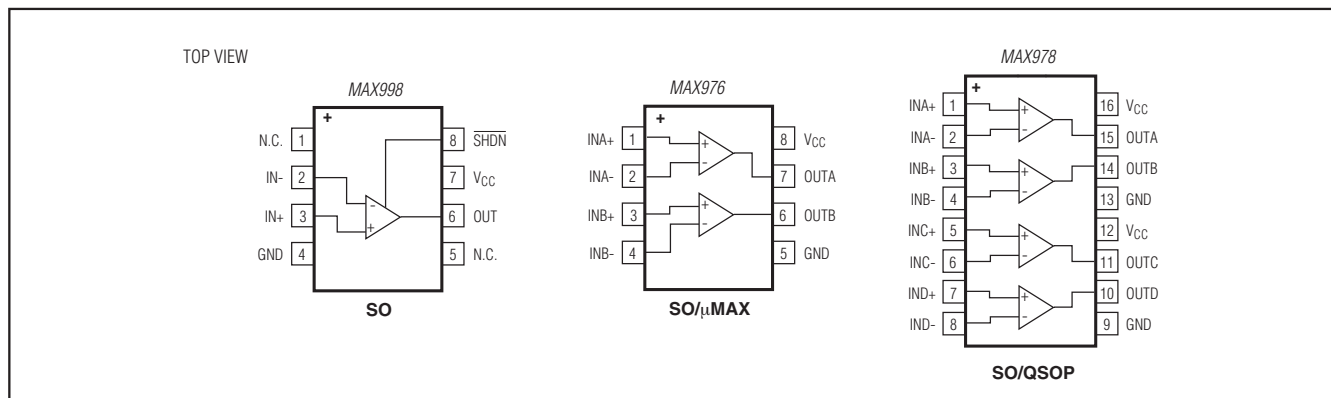


Figure 3. Window Comparator

MAX976/MAX978/MAX998

Single/Dual/Quad, SOT23, Single-Supply, High-Speed, Low-Power Comparators

Pin Configurations (continued)



Chip Information

PROCESS: CMOS

MAX976/MAX978/MAX998

Single/Dual/Quad, SOT23, Single-Supply, High-Speed, Low-Power Comparators

Package Information

For the latest package outline information and land patterns (footprints), go to www.maximintegrated.com/packages. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE TYPE	PACKAGE CODE	OUTLINE NO.	LAND PATTERN NO.
6 SOT23	U6+2	21-0058	90-0175
8 SO	S8+2	21-0041	90-0096
8 μ MAX	U8+1	21-0036	90-0092
16 SO	S16M+3	21-0041	90-0097
16 QSOP	E16M+1	21-0055	90-0167

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Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	10/97	Initial release	—
1	1/98	Adding specs for MAX998	—
2	1/07	Adding input current ratings to Abs Max	—
3	3/09	Update <i>Chip Information</i> , Package Info, correct unit measurement in TOC 8, style changes	1, 3, 4, 8
4	5/14	Added lead-free information to <i>Ordering Information</i> , revised <i>Absolute Maximum Ratings</i> and <i>Package Information</i>	1, 2, 9



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Офис по работе с юридическими лицами:

105318, г.Москва, ул.Щербаковская д.3, офис 1107, 1118, ДЦ «Щербаковский»

Телефон: +7 495 668-12-70 (многоканальный)

Факс: +7 495 668-12-70 (доб.304)

E-mail: info@moschip.ru

Skype отдела продаж:

moschip.ru

moschip.ru_4

moschip.ru_6

moschip.ru_9