

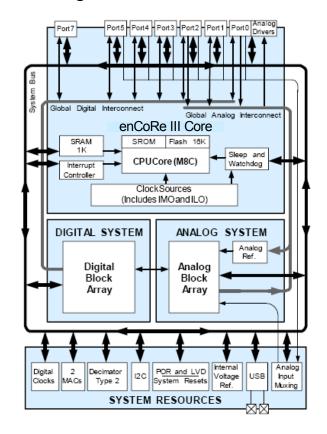
# enCoRe™ III Full-Speed USB Controller

#### **Features**

- Powerful Harvard-architecture processor
  - M8C processor speeds up to 24 MHz
  - □ Two 8 × 8 multiply, 32-bit accumulate
  - □ 3.15 to 5.25-V operating voltage
  - □ USB 2.0 USB-IF certified. TID# 40000110
  - □ Commercial operating temperature range: 0 °C to +70 °C
  - □ Industrial operating temperature range: -40 °C to +85 °C
- Advanced peripherals (enCoRe<sup>™</sup> III blocks)
  - □ Six analog enCoRe III blocks provide:
    - Up to 14-bit incremental and delta sigma analog-to-digital converters (ADCs)
  - □ Programmable threshold comparator
  - □ Four digital enCoRe III blocks provide:
    - 8-bit and 16-bit pulse width modulators (PWMs), timers, and counters
    - I<sup>2</sup>C master
    - · SPI master or slave
    - Full-duplex universal asynchronous receiver-transmitter (UART)
    - CYFISNP modules to talk to Cypress CYFI™ radio
- Complex peripherals by combining blocks
- Full-speed USB (12 Mbps)
  - □ Four unidirectional endpoints
  - □ One bidirectional control endpoint
  - □ Dedicated 256-byte buffer
  - □ No external crystal required
  - Operational at 3.15 V to 3.5 V or 4.35 V to 5.25 V
- Flexible on-chip memory
  - □ 16 KB flash program storage 50,000 erase/write cycles
  - □ 1 KB SRAM data storage
  - □ In-system serial programming (ISSP)
  - □ Partial flash updates
  - ☐ Flexible protection modes
  - □ EEPROM emulation in flash
- Programmable pin configurations
  - □ 25 mA sink on all general purpose I/Os (GPIOs)
  - □ Pull-up, Pull-down, high Z, strong, or open drain drive modes on all GPIOs
  - Configurable interrupt on all GPIOs
- Precision, programmable clocking
  - □ Internal ±4% 24- and 48-MHz oscillator with support for external clock oscillator
  - □ Internal oscillator for watchdog and sleep
  - □ .25% accuracy for USB with no external components

- Additional system resources
  - □ Inter-integrated circuit (I<sup>2</sup>C) slave, master, and multimaster to 400 kHz
  - Watchdog and sleep timers
  - □ User-configurable low-voltage detection (LVD)
  - □ Integrated supervisory circuit
  - □ On-chip precision voltage reference
- Complete development tools
  - ☐ Free development software (PSoC® Designer™)
  - ☐ Full-featured, in-circuit emulator and programmer
  - □ Full-speed emulation
  - □ Complex breakpoint structure
  - □ 128 KB trace memory

## **Block Diagram**



Errata: For information on silicon errata, see "Errata" on page 40. Details include trigger conditions, devices affected, and proposed workaround.



## **Contents**

Applications	3
enCoRe III Functional Overview	3
enCoRe III Core	3
The Digital System	3
The Analog System	4
Additional System Resources	4
enCoRe III Device Characteristics	
Getting Started	5
Application Notes	
Development Kits	5
Training	5
CYPros Consultants	5
Solutions Library	5
Technical Support	5
Development Tools	5
PSoC Designer Software Subsystems	5
Designing with PSoC Designer	6
Select Components	
Configure Components	6
Organize and Connect	6
Generate, Verify, and Debug	6
Pin Information	
56-Pin Part Pinout	7
28-Pin Part Pinout	8
Register Reference	9
Register Mapping Tables	
Register Map Bank 0 Table: User Space	10
Register Map Bank 1 Table: Configuration Space	11
Electrical Specifications	12
Absolute Maximum Ratings	13

Operating Temperature	13
DC Electrical Characteristics	
AC Electrical Characteristics	24
Packaging Information	
Package Diagrams	
Thermal Impedance	
Solder Reflow Peak Temperature	32
Ordering Information	
Ordering Code Definitions	
Acronyms	34
Acronyms Used	34
Reference Documents	34
Document Conventions	35
Units of Measure	35
Numeric Conventions	35
Glossary	35
Errata	
Part Numbers Affected	
CY7C64215 Qualification Status	40
CY7C64215 Errata Summary	40
Document History Page	
Sales, Solutions, and Legal Information	
Worldwide Sales and Design Support	
Products	45
PSoC® Solutions	
Cypress Developer Community	45
Technical Support	45



## **Applications**

- PC human interface devices
  - □ Mouse (optomechanical, optical, trackball)
  - □ Keyboards
  - □ Joysticks
- Gaming
  - □ Game pads
  - □ Console keyboards
- General purpose
  - □ Barcode scanners
  - □ POS terminal
  - □ Consumer electronics
  - □ Toys
  - □ Remote controls
  - □ USB to serial

#### enCoRe III Functional Overview

The enCoRe III is based on the flexible PSoC architecture and is a full-featured, full-speed (12-Mbps) USB part. Configurable analog, digital, and interconnect circuitry enable a high level of integration in a host of consumer, and communication applications.

This architecture enables the user to create customized peripheral configurations that match the requirements of each individual application. Additionally, a fast CPU, flash program memory, SRAM data memory, and configurable I/O are included in both 28-pin SSOP and 56-pin QFN packages.

enCoRe III architecture, as illustrated in the "Block Diagram" on page 1, is comprised of four main areas: enCoRe III core, digital system, analog system, and system resources including a full-speed USB port. Configurable global busing enables all the device resources to combine into a complete custom system. The enCoRe III CY7C64215 can have up to seven I/O ports that connect to the global digital and analog interconnects, providing access to four digital blocks and six analog blocks.

#### enCoRe III Core

The enCoRe III Core is a powerful engine that supports a rich feature set. The core includes a CPU, memory, clocks, and configurable GPIOs.

The M8C CPU core is a powerful processor with speeds up to 24 MHz, providing a four-million instructions per second (MIPS) 8-bit Harvard-architecture microprocessor. The CPU uses an interrupt controller with up to 20 vectors, to simplify programming of real-time embedded events. Program execution is timed and protected using the included sleep and watchdog timers (WDT).

Memory encompasses 16 KB of flash for program storage, 1 KB of SRAM for data storage, and up to 2 KB of EEPROM emulated using the flash. Program flash uses four protection levels on blocks of 64 bytes, enabling customized software IP protection.

enCoRe III incorporates flexible internal clock generators, including a 24-MHz internal main oscillator (IMO) accurate to 8% over temperature and voltage as well as an option for an external clock oscillator (ECO). USB operation requires the OSC LOCK bit of the USB\_CR0 register to be set to obtain IMO accuracy to.25%.

The 24-MHz IMO is doubled to 48 MHz for use by the digital system, if needed. The 48-MHz clock is required to clock the USB block and must be enabled for communication. A low-power 32-kHz internal low-speed oscillator (ILO) is provided for the sleep timer and WDT. The clocks, together with programmable clock dividers (system resource), provide flexibility to integrate almost any timing requirement into enCoRe III. In USB systems, the IMO self-tunes to ±0.25% accuracy for USB communication.

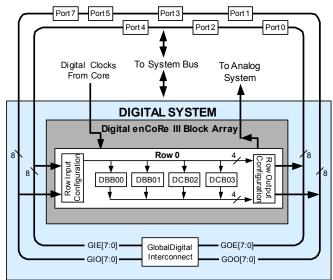
The extended temperature range for the industrial operating range (–40 °C to +85 °C) requires the use of an ECO, which is only available on the 56-pin QFN package.

enCoRe III GPIOs provide connection to the CPU, digital and analog resources of the device. Each pin's drive mode may be selected from eight options, enabling great flexibility in external interfacing. Every pin also has capability to generate a system interrupt on high-level, low-level, and change from last read.

## The Digital System

The digital system is composed of four digital enCoRe III blocks. Each block is an 8-bit resource that is used alone or combined with other blocks to form 8-, 16-, 24-, and 32-bit peripherals, which are called user module references.

Figure 1. Digital System Block Diagram



The following digital configurations can be built from the blocks:

- PWMs, timers, and counters (8-bit and 16-bit)
- UART 8-bit with selectable parity
- SPI master and slave
- I<sup>2</sup>C master
- RF interface: Interface to Cypress CYFI radio

The digital blocks are connected to any GPIO through a series of global buses that can route any signal to any pin. The buses also enable signal multiplexing and performing logic operations. This configurability frees your designs from the constraints of a fixed peripheral controller.

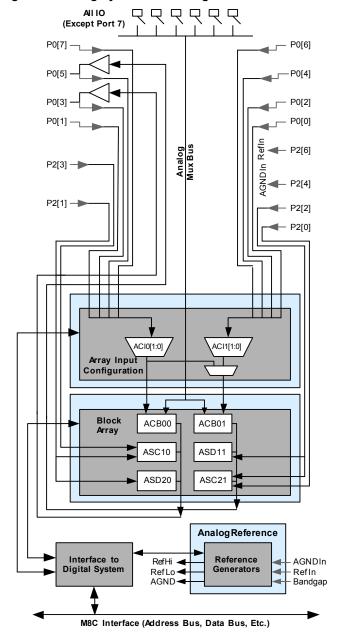


## The Analog System

The analog system is composed of six configurable blocks, comprised of an opamp circuit enabling the creation of complex analog signal flows. Analog peripherals are very flexible and are customized to support specific application requirements. enCoRe III analog function supports the Analog-to-digital converters (with 6- to 14-bit resolution, selectable as incremental, and delta-sigma) and programmable threshold comparator).

Analog blocks are arranged in two columns of three, with each column comprising one continuous time (CT) - AC B00 or AC B01 - and two switched capacitor (SC) - ASC10 and ASD20 or ASD11 and ASC21 - blocks, as shown in Figure 2.

Figure 2. Analog System Block Diagram



#### The Analog Multiplexer System

The analog mux bus can connect to every GPIO pin in ports 0 to 5. Pins are connected to the bus individually or in any combination. The bus also connects to the analog system for analysis with comparators and analog-to-digital converters. It is split into two sections for simultaneous dual-channel processing. An additional 8:1 analog input multiplexer provides a second path to bring Port 0 pins to the analog array.

#### **Additional System Resources**

System resources provide additional capability useful to complete systems. Additional resources include a multiplier, decimator, low voltage detection, and power-on reset. Brief statements describing the merits of each resource follow.

- Full-speed USB (12 Mbps) with five configurable endpoints and 256 bytes of RAM. No external components required except two series resistors. Industrial temperature operating range for USB requires an external clock oscillator.
- Two multiply accumulates (MACs) provide fast 8-bit multipliers with 32-bit accumulate, to assist in both general math and digital filters.
- The decimator provides a custom hardware filter for digital signal processing applications including the creation of delta-sigma ADCs.
- Digital clock dividers provide three customizable clock frequencies for use in applications. The clocks are routed to both the digital and analog systems.
- The I<sup>2</sup>C module provides 100- and 400-kHz communication over two wires. Slave, master, and multimaster modes are all supported.
- LVD interrupts can signal the application of falling voltage levels, while the advanced power-on reset (POR) circuit eliminates the need for a system supervisor.

#### enCoRe III Device Characteristics

enCoRe III devices have four digital blocks and six analog blocks. The following table lists the resources available for specific enCoRe III devices.

Table 1. enCoRe III Device Characteristics

Part Number	Digital I/O	Digital Rows	Digital Blocks	Analog Inputs	Analog Outputs	Analog Columns	Analog Blocks	SRAM Size	Flash Size
CY7C64215 28 Pin	up to 22	1	4	22	2	2	6	1K	16K
CY7C64215 56 Pin	up to 50	1	4	48	2	2	6	1K	16K



## **Getting Started**

The quickest path to understanding the enCoRe III silicon is by reading this datasheet and using the PSoC Designer Integrated Development Environment (IDE). This datasheet is an overview of the enCoRe V integrated circuit and presents specific pin, register, and electrical specifications.

For in-depth information, along with detailed programming details, see the  $PSoC^{\otimes}$  Technical Reference Manual.

For up-to-date ordering, packaging, and electrical specification information, see the latest PSoC device datasheets on the web.

## **Application Notes**

Cypress application notes are an excellent introduction to the wide variety of possible PSoC designs.

## **Development Kits**

PSoC Development Kits are available online from and through a growing number of regional and global distributors, which include Arrow, Avnet, Digi-Key, Farnell, Future Electronics, and Newark.

#### **Training**

Free PSoC technical training (on demand, webinars, and workshops), which is available online via www.cypress.com, covers a wide variety of topics and skill levels to assist you in your designs.

#### **CYPros Consultants**

Certified PSoC consultants offer everything from technical assistance to completed PSoC designs. To contact or become a PSoC consultant go to the CYPros Consultants web site.

#### Solutions Library

Visit our growing library of solution focused designs. Here you can find various application designs that include firmware and hardware design files that enable you to complete your designs quickly.

#### **Technical Support**

Technical support – including a searchable Knowledge Base articles and technical forums – is also available online. If you cannot find an answer to your question, call our Technical Support hotline at 1-800-541-4736.

## **Development Tools**

PSoC Designer™ is the revolutionary integrated design environment (IDE) that you can use to customize PSoC to meet your specific application requirements. PSoC Designer software accelerates system design and time to market. Develop your applications using a library of precharacterized analog and digital peripherals (called user modules) in a drag-and-drop design environment. Then, customize your design by leveraging the dynamically generated application programming interface (API) libraries of code. Finally, debug and test your designs with the integrated debug environment, including in-circuit emulation and standard software debug features. PSoC Designer includes:

- Application editor graphical user interface (GUI) for device and user module configuration and dynamic reconfiguration
- Extensive user module catalog
- Integrated source-code editor (C and assembly)
- Free C compiler with no size restrictions or time limits
- Built-in debugger
- In-circuit emulation
- Built-in support for communication interfaces:
  - ☐ Hardware and software I<sup>2</sup>C slaves and masters
  - □ Full-speed USB 2.0
  - □ Up to four full-duplex universal asynchronous receiver/transmitters (UARTs), SPI master and slave, and wireless

PSoC Designer supports the entire library of PSoC 1 devices and runs on Windows XP, Windows Vista, and Windows 7.

## **PSoC Designer Software Subsystems**

#### Design Entry

In the chip-level view, choose a base device to work with. Then select different onboard analog and digital components that use the PSoC blocks, which are called user modules. Examples of user modules are ADCs, DACs, amplifiers, and filters. Configure the user modules for your chosen application and connect them to each other and to the proper pins. Then generate your project. This prepopulates your project with APIs and libraries that you can use to program your application.

The tool also supports easy development of multiple configurations and dynamic reconfiguration. Dynamic reconfiguration makes it possible to change configurations at run time. In essence, this allows you to use more than 100 percent of PSoC's resources for an application.

## Code Generation Tools

The code generation tools work seamlessly within the PSoC Designer interface and have been tested with a full range of debugging tools. You can develop your design in C, assembly, or a combination of the two.

**Assemblers**. The assemblers allow you to merge assembly code seamlessly with C code. Link libraries automatically use absolute addressing or are compiled in relative mode, and are linked with other software modules to get absolute addressing.

**C Language Compilers**. C language compilers are available that support the PSoC family of devices. The products allow you to create complete C programs for the PSoC family devices. The optimizing C compilers provide all of the features of C, tailored to the PSoC architecture. They come complete with embedded libraries providing port and bus operations, standard keypad and display support, and extended math functionality.

## Debugger

PSoC Designer has a debug environment that provides hardware in-circuit emulation, allowing you to test the program in a physical system while providing an internal view of the PSoC device. Debugger commands allow you to read and program and read and write data memory, and read and write I/O registers. You can read and write CPU registers, set and clear breakpoints,



and provide program run, halt, and step control. The debugger also allows you to create a trace buffer of registers and memory locations of interest.

#### Online Help System

The online help system displays online, context-sensitive help. Designed for procedural and quick reference, each functional subsystem has its own context-sensitive help. This system also provides tutorials and links to FAQs and an online support Forum to aid the designer.

#### In-Circuit Emulator

A low-cost, high-functionality in-circuit emulator (ICE) is available for development support. This hardware can program single devices.

The emulator consists of a base unit that connects to the PC using a USB port. The base unit is universal and operates with all PSoC devices. Emulation pods for each device family are available separately. The emulation pod takes the place of the PSoC device in the target board and performs full-speed (24 MHz) operation.

## **Designing with PSoC Designer**

The development process for the PSoC device differs from that of a traditional fixed function microprocessor. The configurable analog and digital hardware blocks give the PSoC architecture a unique flexibility that pays dividends in managing specification change during development and by lowering inventory costs. These configurable resources, called PSoC Blocks, have the ability to implement a wide variety of user-selectable functions.

The PSoC development process can be summarized in the following four steps:

- 1. Select User Modules
- 2. Configure User Modules
- 3. Organize and Connect
- 4. Generate, Verify, and Debug

#### Select Components

PSoC Designer provides a library of pre-built, pre-tested hardware peripheral components called "user modules". User modules make selecting and implementing peripheral devices, both analog and digital, simple.

## **Configure Components**

Each of the User Modules you select establishes the basic register settings that implement the selected function. They also provide parameters and properties that allow you to tailor their

precise configuration to your particular application. For example, a PWM User Module configures one or more

digital PSoC blocks, one for each 8 bits of resolution. The user module parameters permit you to establish the pulse width and duty cycle. Configure the parameters and properties to correspond to your chosen application. Enter values directly or by selecting values from drop-down menus. All the user modules are documented in datasheets that may be viewed directly in PSoC Designer or on the Cypress website. These user module datasheets explain the internal operation of the User Module and provide performance specifications. Each datasheet describes the use of each user module parameter, and other information you may need to successfully implement your design.

#### **Organize and Connect**

You build signal chains at the chip level by interconnecting user modules to each other and the I/O pins. You perform the selection, configuration, and routing so that you have complete control over all on-chip resources.

## Generate, Verify, and Debug

When you are ready to test the hardware configuration or move on to developing code for the project, you perform the "Generate Configuration Files" step. This causes PSoC Designer to generate source code that automatically configures the device to your specification and provides the software for the system. The generated code provides application programming interfaces (APIs) with high-level functions to control and respond to hardware events at run time and interrupt service routines that you can adapt as needed.

A complete code development environment allows you to develop and customize your applications in C, assembly language, or both.

The last step in the development process takes place inside PSoC Designer's Debugger (access by clicking the Connect icon). PSoC Designer downloads the HEX image to the ICE where it runs at full speed. PSoC Designer debugging capabilities rival those of systems costing many times more. In addition

to traditional single-step, run-to-breakpoint and watch-variable features, the debug interface provides a large trace buffer and allows you to define complex breakpoint events that include monitoring address and data bus values, memory locations and external signals.



## Pin Information

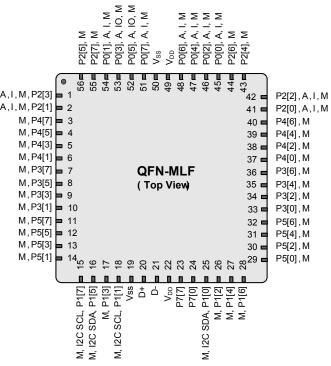
#### **56-Pin Part Pinout**

The CY7C64215 enCoRe III device is available in a 56-pin package which is listed and illustrated in the following table. Every port pin (labeled "P") is capable of digital I/O. However,  $V_{SS}$  and  $V_{DD}$  are not capable of digital I/O.

Table 2. 56-Pin Part Pinout (QFN-MLF SAWN)[1]

Pin	Ty	/pe	Nama	Description		Fig	ure 3.	CY
No.		Analog	Name	Description		•		
1	I/O	I, M	P2[3]	Direct switched capacitor block input.				
2	I/O	I, M	P2[1]	Direct switched capacitor block input.				
3	I/O	М	P4[7]					Σ
4	I/O	М	P4[5]				2[5]	P2[7],
5	I/O	М	P4[3]					
6	I/O	М	P4[1]				99	55=
7	I/O	М	P3[7]			, I , M , P2[	3] 🖨 1	L()
8	I/O	М	P3[5]		Α	, I , M , P2[	-	
9	I/O	М	P3[3]				7] = 3	
10	I/O	М	P3[1]				5] <b>=</b> 4	
11	I/O	М	P5[7]			М, Р4[ М, Р4[	3] <b>=</b> 5	
12	I/O	М	P5[5]			_	7]	
13	I/O	М	P5[3]			_	5] <b>=</b> 8	
14	I/O	М	P5[1]				3] <b>9</b> 9	
15	I/O	М	P1[7]	I <sup>2</sup> C serial clock (SCL).		M, P3[	1] 🗖 10	
16	I/O	М	P1[5]	I <sup>2</sup> C serial data (SDA).			7] 😑 11	
17	I/O	М	P1[3]				5] 🗖 12	
18	I/O	М		I <sup>2</sup> C SCL, ISSP-SCLK.		M, P5[	3] = 13	
19	Po	wer	V <sub>SS</sub>	Ground connection.		M, P5[	1] 14 <sub>5</sub>	16
20	U	SB	D+					0
21	U	SB	D-				⊑	P1[5]
22	Po	wer	$V_{DD}$	Supply voltage.			7	7
23	I/O		P7[7]				ŭ	SDA,
24	I/O		P7[0]				Ö	12C S
25	I/O	М	P1[0]	I <sup>2</sup> C SDA, ISSP-SDATA.			-	M, 12
26	I/O	М	P1[2]				2	2
27	I/O	М	P1[4]	Optional external clock input EXTCLK.				
28	I/O	М	P1[6]					
29	I/O	М	P5[0]		Pin		<i>r</i> ре	Na
30	I/O	М	P5[2]		No.	Digital	Analog	INC
31	I/O	М	P5[4]		44	I/O	М	P
32	I/O	М	P5[6]		45	I/O	I, M	Р
33	I/O	М	P3[0]		46	I/O	I, M	Р
34	I/O	М	P3[2]		47	I/O	I, M	Р
35	I/O	М	P3[4]		48	I/O	I, M	Р
36	I/O	М	P3[6]		49		wer	٧
37	I/O	М	P4[0]		50	Po	wer	\

Figure 3. CY7C64215 56-Pin enCoRe III Device



28	I/O	М	P1[6]						
29	I/O	М	P5[0]		Pin	Type		Name	Description
30	I/O	М	P5[2]		No.	Digital	Analog	Ivallie	Description
31	I/O	М	P5[4]		44	I/O	М	P2[6]	External voltage reference (VREF) input.
32	I/O	М	P5[6]		45	I/O	I, M	P0[0]	Analog column mux input.
33	I/O	М	P3[0]		46	I/O	I, M	P0[2]	Analog column mux input and column output.
34	I/O	М	P3[2]		47	I/O	I, M	P0[4]	Analog column mux input and column output.
35	I/O	М	P3[4]		48	I/O	I, M	P0[6]	Analog column mux input.
36	I/O	М	P3[6]		49	Po	wer	$V_{DD}$	Supply voltage.
37	I/O	М	P4[0]		50	Po	wer	$V_{SS}$	Ground connection.
38	I/O	М	P4[2]		51	I/O	I, M	P0[7]	Analog column mux input.
39	I/O	М	P4[4]		52	I/O	I/O, M	P0[5]	Analog column mux input and column output
40	I/O	М	P4[6]		53	I/O	I/O, M	P0[3]	Analog column mux input and column output.
41	I/O	I, M	P2[0]	Direct switched capacitor block input.	54	I/O	I, M	P0[1]	Analog column mux input.
42	I/O	I, M	P2[2]	Direct switched capacitor block input.	55	I/O	М	P2[7]	
43	I/O	М	P2[4]	External analog ground (AGND) input.	56	I/O	М	P2[5]	

**LEGEND** A = Analog, I = Input, O = Output, and M = Analog Mux Input.

#### Note

Document Number: 38-08036 Rev. \*N Page 7 of 45

The center pad on the QFN-MLF package should be connected to ground (V<sub>SS</sub>) for best mechanical, thermal, and electrical performance. If not connected to ground, it should be electrically floated and not connected to any other signal.



## 28-Pin Part Pinout

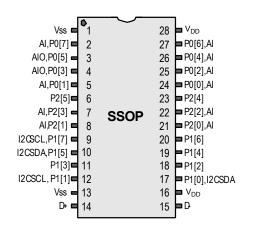
The CY7C64215 enCoRe III device is available in a 28-pin package which is listed and illustrated in the following table. Every port pin (labeled with a "P") is capable of digital I/O. However,  $V_{SS}$  and  $V_{DD}$  are not capable of digital I/O.

Table 3. 28-Pin Part Pinout (SSOP)

Pin	Τ\	ре		
No.		Analog	Name	Description
1		wer	GND	Ground connection.
2	I/O	I, M	P0[7]	Analog column mux input.
3	I/O	I/O,M	P0[5]	Analog column mux input and column output.
4	I/O	I/O,M	P0[3]	Analog column mux input and column output.
5	I/O	I,M	P0[1]	Analog column mux input.
6	I/O	М	P2[5]	
7	I/O	М	P2[3]	Direct switched capacitor block input.
8	I/O	М	P2[1]	Direct switched capacitor block input.
9	I/O	М	P1[7]	I <sup>2</sup> C SCL
10	I/O	М	P1[5]	I <sup>2</sup> C SDA
11	I/O	М	P1[3]	
12	I/O	М	P1[1]	I <sup>2</sup> C SCL, ISSP-SCLK.
13	Power		GND	Ground connection.
14	U	SB	D+	
15	U	SB	D-	
16	Po	wer	$V_{DD}$	Supply voltage.
17	I/O	М	P1[0]	I <sup>2</sup> C SCL, ISSP-SDATA.
18	I/O	М	P1[2]	
19	I/O	М	P1[4]	
20	I/O	М	P1[6]	
21	I/O	М	P2[0]	Direct switched capacitor block input.
22	I/O	М	P2[2]	Direct switched capacitor block input.
23	I/O	М	P2[4]	External analog ground (AGND) input.
24	I/O	М	P0[0]	Analog column mux input.
25	I/O	М	P0[2]	Analog column mux input and column output.
26	I/O	М	P0[4]	Analog column mux input and column output.
27	I/O	М	P0[6]	Analog column mux input.
28	Po	wer	$V_{DD}$	Supply voltage.

**LEGEND** A = Analog, I = Input, O = Output, and M = Analog Mux Input.

Figure 4. CY7C64215 28-Pin enCoRe III Device





# **Register Reference**

The register conventions specific to this section are listed in the following table.

**Table 4. Register Conventions** 

Convention	Description					
R	Read register or bit(s)					
W	Write register or bit(s)					
L	Logical register or bit(s)					
С	Clearable register or bit(s)					
#	Access is bit specific					

## **Register Mapping Tables**

The enCoRe III device has a total register address space of 512 bytes. The register space is referred to as I/O space and is divided into two banks, bank 0 and bank 1. The XOI bit in the Flag register (CPU\_F) determines which bank the user is currently in. When the XOI bit is set to '1', the user is in bank 1.

**Note** In the following register mapping tables, blank fields are reserved and should not be accessed.



# Register Map Bank 0 Table: User Space

Register I	Map Bank	0 Table		ace							
Name	Addr (0,Hex)		Name	Addr (0,Hex)		Name	Addr (0,Hex)	Access	Name	Addr (0,Hex)	Access
PRT0DR	00	RW	PMA0_DR	40	RW	ASC10CR0	80	RW		C0	
PRT0IE	01	RW	PMA1_DR	41	RW	ASC10CR1	81	RW		C1	
PRT0GS	02	RW	PMA2_DR	42	RW	ASC10CR2	82	RW		C2	
PRT0DM2	03	RW	PMA3_DR	43	RW	ASC10CR3	83	RW		C3	
PRT1DR	04	RW	PMA4_DR	44	RW	ASD11CR0	84	RW		C4	
PRT1IE	05	RW	PMA5_DR	45	RW	ASD11CR1	85	RW		C5	
PRT1GS	06	RW	PMA6_DR	46	RW	ASD11CR2	86	RW		C6	
PRT1DM2	07	RW	PMA7_DR	47	RW	ASD11CR3	87	RW		C7	
PRT2DR	08	RW	USB_SOF0	48	R		88			C8	
PRT2IE	09	RW	USB_SOF1	49	R		89			C9	
PRT2GS	0A	RW	USB_CR0	4A	RW		8A			CA	
PRT2DM2	0B	RW	USBIO_CR0	4B	#		8B			СВ	
PRT3DR	0C	RW	USBIO_CR1	4C	RW		8C			CC	
PRT3IE	0D	RW		4D			8D			CD	
PRT3GS	0E	RW	EP1_CNT1	4E	#		8E			CE	
PRT3DM2	0F	RW	EP1_CNT	4F	RW		8F			CF	
PRT4DR	10	RW	EP2_CNT1	50	#	ASD20CR0	90	RW	CUR_PP	D0	RW
PRT4IE	11	RW	EP2_CNT	51	RW	ASD20CR1	91	RW	STK_PP	D1	RW
PRT4GS	12	RW	EP3_CNT1	52	#	ASD20CR2	92	RW		D2	
PRT4DM2	13	RW	EP3_CNT	53	RW	ASD20CR3	93	RW	IDX_PP	D3	RW
PRT5DR	14	RW	EP4_CNT1	54	#	ASC21CR0	94	RW	MVR_PP	D4	RW
PRT5IE	15	RW	EP4_CNT	55	RW	ASC21CR1	95	RW	MVW_PP	D5	RW
PRT5GS	16	RW	EP0_CR	56	#	ASC21CR2	96	RW	I2C_CFG	D6	RW
PRT5DM2	17	RW	EP0_CNT	57	#	ASC21CR3	97	RW	I2C_SCR	D7	#
	18		EP0_DR0	58	RW		98		I2C_DR	D8	RW
	19		EP0_DR1	59	RW		99		I2C_MSCR	D9	#
	1A		EP0_DR2	5A	RW		9A		INT_CLR0	DA	RW
	1B		EP0_DR3	5B	RW		9B		INT_CLR1	DB	RW
PRT7DR	1C	RW	EP0_DR4	5C	RW		9C		INT_CLR2	DC	RW
PRT7IE	1D	RW	EP0_DR5	5D	RW		9D		INT_CLR3	DD	RW
PRT7GS	1E	RW	EP0_DR6	5E	RW		9E		INT_MSK3	DE	RW
PRT7DM2	1F	RW	EP0_DR7	5F	RW		9F		INT_MSK2	DF	RW
DBB00DR0	20	#	AMX IN	60	RW		A0		INT MSK0	E0	RW
DBB00DR1	21	W	AMUXCFG	61	RW		A1		INT MSK1	E1	RW
DBB00DR2	22	RW		62			A2		INT_VC	E2	RC
DBB00CR0	23	#	ARF CR	63	RW		A3		RES WDT	E3	W
DBB01DR0	24	#	CMP CR0	64	#		A4		DEC DH	E4	RC
DBB01DR1	25	W	ASY_CR	65	#		A5		DEC_DL	E5	RC
DBB01DR2	26	RW	CMP_CR1	66	RW		A6		DEC_CR0	E6	RW
DBB01CR0	27	#		67			A7		DEC_CR1	E7	RW
DCB02DR0	28	#		68		MUL1 X	A8	W	MULO X	E8	W
DCB02DR1	29	W		69		MUL1 Y	A9	W	MUL0 Y	E9	W
DCB02DR2	2A	RW		6A		MUL1 DH	AA	R	MUL0 DH	EA	R
DCB02CR0	2B	#		6B		MUL1 DL	AB	R	MUL0 DL	EB	R
DCB03DR0	2C	#	TMP_DR0	6C	RW	ACC1_DR1	AC	RW	ACC0_DR1	EC	RW
DCB03DR1	2D	W	TMP_DR1	6D	RW	ACC1_DR0	AD	RW	ACC0_DR0	ED	RW
DCB03DR2	2E	RW	TMP_DR2	6E	RW	ACC1_DR3	AE	RW	ACC0_DR3	EE	RW
DCB03CR0	2F	#	TMP_DR3	6F	RW	ACC1_DR2	AF	RW	ACC0_DR2	EF	RW
	30		ACB00CR3	70	RW	RDI0RI	В0	RW	_	F0	
	31		ACB00CR0	71	RW	RDI0SYN	B1	RW		F1	
	32		ACB00CR1	72	RW	RDI0IS	B2	RW		F2	
	33		ACB00CR2	73	RW	RDI0LT0	B3	RW		F3	
	34		ACB01CR3	74	RW	RDI0LT1	B4	RW		F4	
	35		ACB01CR0	75	RW	RDI0RO0	B5	RW		F5	
	36		ACB01CR1	76	RW	RDI0RO1	B6	RW		F6	
	37		ACB01CR2	77	RW		B7		CPU F	F7	RL
	38			78			B8		·-	F8	
	39			79			B9			F9	$\vdash$
	3A			7A			BA			FA	$\vdash$
	3B			7B			BB			FB	$\vdash$
	3C			7C			BC			FC	$\vdash$
	3D			7D			BD		DAC D	FD	RW
	3E			7E			BE		CPU SCR1	FE	#
	3F			7F			BF		CPU SCR0	FF	#
	ro Bosoniod on	l	L.,	l		# Access is b			3. 5_00.0	1	L

Blank fields are Reserved and should not be accessed.

# Access is bit specific.



Register Map Bank 1 Table: Configuration Space

	Addr		: Configurati	Addr			Addr			Addr	
Name	(1,Hex)	Access	Name	(1,Hex)	Access	Name	(1,Hex)	Access	Name	(1,Hex)	Access
PRT0DM0	00	RW	PMA0_WA	40	RW	ASC10CR0	80	RW	USBIO_CR2	C0	RW
PRT0DM1	01	RW	PMA1_WA	41	RW	ASC10CR1	81	RW	USB_CR1	C1	#
PRT0IC0	02	RW	PMA2_WA	42	RW	ASC10CR2	82	RW			
PRT0IC1	03	RW	PMA3_WA	43	RW	ASC10CR3	83	RW			
PRT1DM0	04	RW	PMA4_WA	44	RW	ASD11CR0	84	RW	EP1_CR0	C4	#
PRT1DM1	05	RW	PMA5_WA	45	RW	ASD11CR1	85	RW	EP2_CR0	C5	#
PRT1IC0	06	RW	PMA6_WA	46	RW	ASD11CR2	86	RW	EP3_CR0	C6	#
PRT1IC1	07	RW	PMA7_WA	47	RW	ASD11CR3	87	RW	EP4_CR0	C7	#
PRT2DM0	08	RW		48			88			C8	
PRT2DM1	09	RW		49			89			C9	
PRT2IC0	0A	RW		4A			8A			CA	
PRT2IC1	0B	RW		4B			8B			СВ	
PRT3DM0	0C	RW		4C			8C			CC	
PRT3DM1	0D	RW		4D			8D			CD	
PRT3IC0	0E	RW		4E			8E			CE	
PRT3IC1	0F	RW		4F			8F			CF	
PRT4DM0	10	RW	PMA0_RA	50	RW		90		GDI_O_IN	D0	RW
PRT4DM1	11	RW	PMA1_RA	51	RW	ASD20CR1	91	RW	GDI_E_IN	D1	RW
PRT4IC0	12	RW	PMA2_RA	52	RW	ASD20CR2	92	RW	GDI_O_OU	D2	RW
PRT4IC1	13	RW	PMA3_RA	53	RW	ASD20CR3	93	RW	GDI_E_OU	D3	RW
PRT5DM0	14	RW	PMA4_RA	54	RW	ASC21CR0	94	RW		D4	
PRT5DM1	15	RW	PMA5 RA	55	RW	ASC21CR1	95	RW		D5	
PRT5IC0	16	RW	PMA6_RA	56	RW	ASC21CR2	96	RW		D6	<u> </u>
PRT5IC1	17	RW	PMA7_RA	57	RW	ASC21CR3	97	RW		D7	
	18			58			98		MUX CR0	D8	RW
	19			59			99		MUX_CR1	D9	RW
	1A			5A			9A		MUX_CR2	DA	RW
	1B			5B			9B	1	MUX_CR3	DB	RW
PRT7DM0	1C	RW		5C			9C	1		DC	
PRT7DM1	1D	RW		5D			9D		OSC_GO_EN	DD	RW
PRT7IC0	1E	RW		5E			9E	1	OSC CR4	DE	RW
PRT7IC1	1F	RW		5F			9F		OSC_CR3	DF	RW
DBB00FN	20	RW	CLK CR0	60	RW		A0		OSC CR0	E0	RW
DBB00IN	21	RW	CLK_CR1	61	RW		A1		OSC_CR1	E1	RW
DBB00OU	22	RW	ABF_CR0	62	RW		A2		OSC_CR2	E2	RW
	23	100	AMD_CR0	63	RW		A3		VLT_CR	E3	RW
DBB01FN	24	RW	CMP_GO_EN	64	RW		A4		VLT CMP	E4	R
DBB01IN	25	RW	OWI _OO_EN	65	RW		A5		VET_OWN	E5	10
DBB010U	26	RW	AMD_CR1	66	RW		A6	-		E6	
DBB0100	27	IXVV	ALT CR0	67	RW		A7			E7	
DCB02FN	28	RW	ALI_CIN	68	IXVV		A8		IMO TR	E8	W
DCB02FN DCB02IN	29	RW		69			A9		ILO TR	E9	W
DCB02IN DCB02OU	29 2A	RW		6A			AA		BDG_TR	EA	RW
DCB0200		KW		6B			AB		_	EB	W
DCB03FN	2B 2C	RW	TMD DD0	6C	D\A/		AC	ļ	ECO_TR	EC	RW
	2D		TMP_DR0	6D	RW			ļ	MUX_CR4 MUX_CR5		D\M/
DCB03IN DCB03OU	2D 2E	RW RW	TMP_DR1 TMP_DR2	6E	RW		AD AE	ļ	MOV_CK2	EE	RW
DOBUSOU	2E 2F	1500	TMP_DR3	6F	RW		AF			EF	
						DDIADI		D\A/			
	30		ACB00CR3	70	RW	RDI0RI	B0	RW		F0	
	31		ACB00CR0	71	RW	RDIOSYN	B1	RW		F1	
	32		ACB00CR1	72	RW	RDI0IS	B2	RW		F2	
	33		ACB00CR2	73	RW	RDIOLTO	B3	RW		F3	
	34		ACB01CR3	74	RW	RDI0LT1	B4	RW		F4	
	35		ACB01CR0	75 70	RW	RDI0RO0	B5	RW		F5	
	36		ACB01CR1	76	RW	RDI0RO1	B6	RW	ODIL 5	F6	
	37		ACB01CR2	77	RW		B7		CPU_F	F7	RL
	38			78			B8			F8	
	39			79			B9			F9	
	3A			7A			BA			FA	
							BB	1 -		FB	
	3B			7B	<u> </u>						<u> </u>
	3C			7C			BC			FC	
				7C 7D			BC BD		DAC_CR	FC FD	RW
	3C			7C			BC		DAC_CR CPU_SCR1 CPU_SCR0	FC	RW #

Blank fields are Reserved and should not be accessed.

# Access is bit specific.



# **Electrical Specifications**

This section presents the DC and AC electrical specifications of the CY7C64215 enCoRe III. For the most up-to-date electrical specifications, confirm that you have the most recent datasheet by going to the web at http://www.cypress.com/go/usb.

Specifications are valid for –40 °C  $\leq$  T<sub>A</sub>  $\leq$  85 °C and T<sub>J</sub>  $\leq$  100 °C, except where noted. Specifications for devices running at greater than 12 MHz are valid for –40 °C  $\leq$  T<sub>A</sub>  $\leq$  70 °C and T<sub>J</sub>  $\leq$  82 °C.

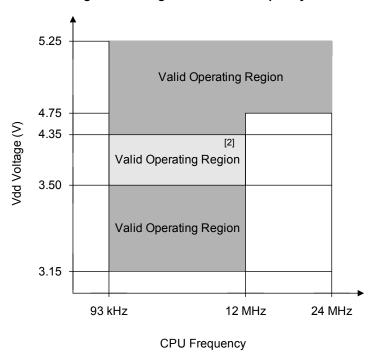


Figure 5. Voltage versus CPU Frequency

#### Note

<sup>2.</sup> This is a valid operating region for the CPU, but USB hardware is non functional in the voltage range from  $3.50\,\mathrm{V}$  to  $4.35\,\mathrm{V}$ .



# **Absolute Maximum Ratings**

# **Table 5. Absolute Maximum Ratings**

Parameter	Description	Min	Тур	Max	Unit	Notes
T <sub>STG</sub>	Storage temperature	<b>–</b> 55	_	+100	°C	Higher storage temperatures reduces data retention time.
T <sub>BAKETEMP</sub>	Bake temperature	_	125	See package label	°C	-
T <sub>BAKETIME</sub>	Bake time	See package label	_	72	Hours	-
T <sub>A</sub>	Ambient temperature with power applied	0	_	+70	°C	_
$V_{DD}$	Supply voltage on V <sub>DD</sub> relative to V <sub>SS</sub>	-0.5	-	+6.0	V	-
$V_{IO}$	DC input voltage	$V_{SS} - 0.5$	_	V <sub>DD</sub> + 0.5	V	_
V <sub>IO2</sub>	DC voltage applied to tristate	V <sub>SS</sub> – 0.5	-	V <sub>DD</sub> + 0.5	V	-
I <sub>MIO</sub>	Maximum current into any port pin	-25	_	+50	mA	_
I <sub>MAIO</sub>	Maximum current into any port pin configured as an analog driver	-50	_	+50	mA	_
ESD	Electrostatic discharge voltage	2000	-	-	V	Human body model ESD.
LU	Latch up current	_	_	200	mA	_

# **Operating Temperature**

# **Table 6. Operating Temperature**

Parameter	Description	Min	Тур	Max	Unit	Notes
T <sub>AC</sub>	Commercial ambient temperature	0	_	+70	°C	_
T <sub>AI</sub>	Industrial ambient temperature	-40	-	+85	°C	USB operation requires the use of an external clock oscillator and the 56-pin QFN package.
T <sub>J</sub>	Junction temperature	<del>-4</del> 0	_	+100	°C	The temperature rise from ambient to junction is package specific. See "Thermal Impedance" on page 32. The user must limit the power consumption to comply with this requirement.



## **DC Electrical Characteristics**

#### DC Chip-Level Specifications

The following table lists guaranteed maximum and minimum specifications for the voltage and temperature ranges: 4.75 V to 5.25 V and  $-40~^{\circ}\text{C} \leq T_A \leq 85~^{\circ}\text{C}$ , or 3.15 V to 3.5 V and  $-40~^{\circ}\text{C} \leq T_A \leq 85~^{\circ}\text{C}$ , respectively. Typical parameters apply to 5 V and 3.3 V at 25  $^{\circ}\text{C}$  and are for design guidance only.

Table 7. DC Chip-Level Specifications

Parameter	Description	Min	Тур	Max	Unit	Notes
V <sub>DD</sub>	Supply voltage	3.0	_	5.25		See DC POR and LVD specifications, Table 15 on page 22. USB hardware is not functional when $V_{DD}$ is between 3.5 V to 4.35 V.
I <sub>DD5</sub>	Supply current, IMO = 24 MHz (5 V)	-	14	27	mA	Conditions are $V_{DD}$ = 5.0 V, $T_A$ = 25 °C, CPU = 3 MHz, SYSCLK doubler disabled, VC1 = 1.5 MHz, VC2 = 93.75 kHz, VC3 = 93.75 kHz, analog power = off.
I <sub>DD3</sub>	Supply current, IMO = 24 MHz (3.3 V)	-	8	14	mA	Conditions are $V_{DD}$ = 3.3 V, $T_A$ = 25 °C, CPU = 3 MHz, SYSCLK doubler disabled, VC1 = 1.5 MHz, VC2 = 93.75 kHz, VC3 = 0.367 kHz, analog power = off.
I <sub>SB</sub>	Sleep [3] (mode) current with POR, LVD, sleep timer, and WDT [4].	-	3	6.5	μА	Conditions are with internal slow speed oscillator, $V_{DD}$ = 3.3 V, 0 °C $\leq$ $T_{A} \leq$ 55 °C, analog power = off.
I <sub>SBH</sub>	Sleep (mode) current with POR, LVD, sleep timer, and WDT at high temperature [4].	_	4	25	•	Conditions are with internal slow speed oscillator, $V_{DD}$ = 3.3 V, 55 °C < $T_A \le 70$ °C, analog power = off.

#### Notes

Document Number: 38-08036 Rev. \*N Page 14 of 45

<sup>3.</sup> Errata: When the device operates at4.75 V to 5.25 V and the 3.3-V regulator is enabled, a short low pulse may be created on the DP signal line during device wakeup. The 15- to 20-µs low pulse of the DP line may be interpreted by the host computer as a deattach or the beginning of a wakeup. For more details refer to Errata on

<sup>4.</sup> Standby current includes all functions (POR, LVD, WDT, sleep time) needed for reliable system operation. This should be compared with devices that have similar functions enabled.



## DC GPIO Specifications

The following table lists guaranteed maximum and minimum specifications for the voltage and temperature ranges: 4.75 V to 5.25 V and  $-40 \text{ °C} \leq T_A \leq 85 \text{ °C}$ , or 3.15 V to 3.5 V and  $-40 \text{ °C} \leq T_A \leq 85 \text{ °C}$ , respectively. Typical parameters apply to 5 V and 3.3 V at 25 °C and are for design guidance only.

Table 8. DC GPIO Specifications

Parameter	Description	Min	Тур	Max	Unit	Notes
R <sub>PU</sub>	Pull-up resistor	4	5.6	8	kΩ	_
R <sub>PD</sub>	Pull-down resistor	4	5.6	8	kΩ	_
V <sub>OH</sub>	High output level	V <sub>DD</sub> – 1.0	1	-	V	$I_{OH}$ = 10 mA, $V_{DD}$ = 4.75 to 5.25 V (8 total loads, 4 on even port pins (for example, P0[2], P1[4]), 4 on odd port pins (for example, P0[3], P1[5])). 80 mA maximum combined $I_{OH}$ budget.
V <sub>OL</sub>	Low output level	-	-	0.75	V	$I_{OL}$ = 25 mA, $V_{DD}$ = 4.75 to 5.25 V (8 total loads, 4 on even port pins (for example, P0[2], P1[4]), 4 on odd port pins (for example, P0[3], P1[5])). 150 mA maximum combined $I_{OL}$ budget.
I <sub>OH</sub>	High-level source current	10	-	-	mA	_
I <sub>OL</sub>	Low-level sink current	25	_	_	mA	_
$V_{IL}$	Input low level	_	-	8.0	V	V <sub>DD</sub> = 3.15 to 5.25.
$V_{IH}$	Input high level	2.1	_		V	V <sub>DD</sub> = 3.15 to 5.25.
$V_{H}$	Input hysteresis	_	60	-	mV	_
I <sub>IL</sub>	Input leakage (absolute value)	_	1	-	nA	Gross tested to 1 μA.
C <sub>IN</sub>	Capacitive load on pins as input	_	3.5	10	pF	Package and pin dependent. Temp = 25 °C.
C <sub>OUT</sub>	Capacitive load on pins as output	_	3.5	10	pF	Package and pin dependent. Temp = 25 °C.

#### DC Full-Speed USB Specifications

The following table lists guaranteed maximum and minimum specifications for the voltage and temperature ranges when the IMO is selected as system clock: 4.75 V to 5.25 V and  $0 ^{\circ}\text{C} \leq \text{T}_{A} \leq 70 ^{\circ}\text{C}$ , or 3.15 V to 3.5 V and  $0 ^{\circ}\text{C} \leq \text{T}_{A} \leq 70 ^{\circ}\text{C}$ , respectively.

The following table lists guaranteed maximum and minimum specifications for the voltage and temperature ranges when an external clock is selected as the system clock: 4.75 V to 5.25 V and  $-40 ^{\circ}\text{C} \leq T_A \leq 85 ^{\circ}\text{C}$ , or 3.15 V to 3.5 V and  $-40 ^{\circ}\text{C} \leq T_A \leq 85 ^{\circ}\text{C}$ .

Typical parameters apply to 5 V and 3.3 V at 25 °C and are for design guidance only.

Table 9. DC Full Speed (12 Mbps) USB Specifications

Parameter	Description	Min	Тур	Max	Unit	Notes
USB Interfa	ace					
$V_{DI}$	Differential input sensitivity	0.2	_	_	V	(D+) – (D–)
$V_{CM}$	Differential input common mode range	8.0	_	2.5	V	_
$V_{SE}$	Single-ended receiver threshold	0.8	_	2.0	V	_
C <sub>IN</sub>	Transceiver capacitance	_	_	20	pF	_
I <sub>IO</sub>	High Z state data line leakage	-10	_	10	μΑ	0 V < V <sub>IN</sub> < 3.3 V.
R <sub>EXT</sub>	External USB series resistor	23	_	25	Ω	In series with each USB pin.
V <sub>UOH</sub>	Static output high, driven	2.8	-	3.6	V	15 k $\Omega$ ± 5% to ground. Internal pull-up enabled.
V <sub>UOHI</sub>	Static output high, idle	2.7	_	3.6	V	15 kΩ ± 5% to ground. Internal pull-up enabled.
V <sub>UOL</sub>	Static output low	_	-	0.3	V	15 k $\Omega$ ± 5% to ground. Internal pull-up enabled.
Z <sub>O</sub>	USB driver output impedance	28	_	44	Ω	Including R <sub>EXT</sub> resistor.
V <sub>CRS</sub>	D+/D- crossover voltage	1.3	-	2.0	V	_



## DC Analog Output Buffer Specifications

The following tables list guaranteed maximum and minimum specifications for the voltage and temperature ranges: 4.75 V to 5.25 V and  $-40~^{\circ}\text{C} \le T_A \le 85~^{\circ}\text{C}$ , or 3.15 V to 3.5 V and  $-40~^{\circ}\text{C} \le T_A \le 85~^{\circ}\text{C}$ , respectively. Typical parameters apply to 5 V and 3.3 V at 25  $^{\circ}\text{C}$  and are for design guidance only.

Table 10. 5 V DC Analog Output Buffer Specifications

Parameter	Description	Min	Тур	Max	Unit	Notes
C <sub>L</sub>	Load Capacitance	ı	-	200	pF	This specification applies to the external circuit that is being driven by the analog output buffer.
V <sub>OSOB</sub>	Input offset voltage (absolute value)	ı	3	12	mV	_
TCV <sub>OSOB</sub>	Average input offset voltage drift	-	+6	_	μV/°C	_
V <sub>CMOB</sub>	Common mode input voltage range	0.5	_	V <sub>DD</sub> – 1.0	V	_
R <sub>OUTOB</sub>	Output resistance Power = low Power = high	1 1	0.6 0.6		W W	_
V <sub>OHIGHOB</sub>	High output voltage swing (Load = 32 ohms to V <sub>DD</sub> /2) Power = low Power = high	0.5 × V <sub>DD</sub> + 1.1 0.5 × V <sub>DD</sub> + 1.1		_ _	V V	_
V <sub>OLOWOB</sub>	Low output voltage swing (Load = 32 ohms to V <sub>DD</sub> /2) Power = low Power = high	- -		0.5 × V <sub>DD</sub> – 1.3 0.5 × V <sub>DD</sub> – 1.3	V V	-
I <sub>SOB</sub>	Supply current including bias cell (no load) Power = low Power = high	- -	1.1 2.6	5.1 8.8	mA mA	_
PSRR <sub>OB</sub>	Supply voltage rejection ratio	53	64	-	dB	$(0.5 \times V_{DD} - 1.3) \le V_{OUT} \le (V_{DD} - 2.3).$

Table 11. 3.3 V DC Analog Output Buffer Specifications

Parameter	Description	Min	Тур	Max	Unit	Notes
C <sub>L</sub>	Load Capacitance	_	ı	200	pF	This specification applies to the external circuit that is being driven by the analog output buffer.
V <sub>OSOB</sub>	Input offset voltage (absolute value)	_	3	12	mV	_
TCV <sub>OSOB</sub>	Average input offset voltage drift	_	+6	_	μV/°C	_
V <sub>CMOB</sub>	Common mode input voltage range	0.5	-	V <sub>DD</sub> – 1.0	V	-
R <sub>OUTOB</sub>	Output resistance Power = low Power = high	_ _	1	_ _	W W	_
V <sub>OHIGHOB</sub>	High output voltage swing (Load = 1 $K\Omega$ to $V_{DD}/2$ ) Power = low Power = high	0.5 × V <sub>DD</sub> + 1.0 0.5 × V <sub>DD</sub> + 1.0	_ _	<del>-</del>	V	-
V <sub>OLOWOB</sub>	Low output voltage swing (Load = 1 $K\Omega$ to $V_{DD}/2$ ) Power = low Power = high	<u>-</u>		0.5 × V <sub>DD</sub> – 1.0 0.5 × V <sub>DD</sub> – 1.0	V	-
I <sub>SOB</sub>	Supply current including bias cell (no load) Power = low Power = high	-	0.8 2.0	2.0 4.3	mA mA	-
PSRR <sub>OB</sub>	Supply voltage rejection ratio	34	64	-	dB	$(0.5 \times V_{DD} - 1.0) \le V_{OUT} \le (0.5 \times V_{DD} + 0.9).$



## DC Analog Reference Specifications

The following tables list guaranteed maximum and minimum specifications for the voltage and temperature ranges: 4.75 V to 5.25 V and  $-40 \,^{\circ}\text{C} \leq \text{T}_{\text{A}} \leq 85 \,^{\circ}\text{C}$ , or  $3.15 \,^{\circ}\text{V}$  to  $3.5 \,^{\circ}\text{V}$  and  $-40 \,^{\circ}\text{C} \leq \text{T}_{\text{A}} \leq 85 \,^{\circ}\text{C}$ , respectively. Typical parameters apply to  $5 \,^{\circ}\text{V}$  and  $3.3 \,^{\circ}\text{V}$  at  $25 \,^{\circ}\text{C}$  and are for design guidance only.

The guaranteed specifications are measured through the analog continuous time PSoC blocks. The power levels for AGND refer to the power of the Analog Continuous Time PSoC block. The power levels for RefHi and RefLo refer to the Analog Reference Control register. The limits stated for AGND include the offset error of the AGND buffer local to the Analog Continuous Time PSoC block. Reference control power is high.

Table 12. 5-V DC Analog Reference Specifications

Reference ARF_CR [5:3]	Reference Power Settings	Symbol	Reference	Description	Min	Тур	Max	Units
0b000	RefPower = high	V <sub>REFHI</sub>	Ref High	V <sub>DD</sub> /2 + Bandgap	V <sub>DD</sub> /2 + 1.229	V <sub>DD</sub> /2 + 1.290	V <sub>DD</sub> /2 + 1.346	V
	Opamp bias = high	$V_{AGND}$	AGND	V <sub>DD</sub> /2	$V_{DD}/2 - 0.038$	V <sub>DD</sub> /2	$V_{DD}/2 + 0.040$	V
		V <sub>REFLO</sub>	Ref Low	V <sub>DD</sub> /2 – Bandgap	V <sub>DD</sub> /2 – 1.356	V <sub>DD</sub> /2 – 1.295	V <sub>DD</sub> /2 – 1.218	V
	RefPower = high	$V_{REFHI}$	Ref High	V <sub>DD</sub> /2 + Bandgap	V <sub>DD</sub> /2 + 1.220	V <sub>DD</sub> /2 + 1.292	V <sub>DD</sub> /2 + 1.348	V
	Opamp bias = low	$V_{AGND}$	AGND	V <sub>DD</sub> /2	V <sub>DD</sub> /2 – 0.036	V <sub>DD</sub> /2	$V_{DD}/2 + 0.036$	V
		$V_{REFLO}$	Ref Low	V <sub>DD</sub> /2 – Bandgap	V <sub>DD</sub> /2 – 1.357	V <sub>DD</sub> /2 – 1.297	V <sub>DD</sub> /2 – 1.225	V
	RefPower = medium	$V_{REFHI}$	Ref High	V <sub>DD</sub> /2 + Bandgap	V <sub>DD</sub> /2 + 1.221	V <sub>DD</sub> /2 + 1.293	V <sub>DD</sub> /2 + 1.351	V
	Opamp bias = high	$V_{AGND}$	AGND	V <sub>DD</sub> /2	$V_{DD}/2 - 0.036$	V <sub>DD</sub> /2	$V_{DD}/2 + 0.036$	V
		$V_{REFLO}$	Ref Low	V <sub>DD</sub> /2 – Bandgap	V <sub>DD</sub> /2 – 1.357	V <sub>DD</sub> /2 – 1.298	V <sub>DD</sub> /2 – 1.228	V
	RefPower = medium	$V_{REFHI}$	Ref High	V <sub>DD</sub> /2 + Bandgap	V <sub>DD</sub> /2 + 1.219	V <sub>DD</sub> /2 + 1.293	V <sub>DD</sub> /2 + 1.353	V
	Opamp bias = low	$V_{AGND}$	AGND	V <sub>DD</sub> /2	$V_{DD}/2 - 0.037$	$V_{DD}/2 - 0.001$	$V_{DD}/2 + 0.036$	V
		$V_{REFLO}$	Ref Low	V <sub>DD</sub> /2 – Bandgap	V <sub>DD</sub> /2 – 1.359	V <sub>DD</sub> /2 – 1.299	V <sub>DD</sub> /2 – 1.229	V
0b001	RefPower = high Opamp bias = high	V <sub>REFHI</sub>	Ref High	P2[4]+P2[6] (P2[4] = V <sub>DD</sub> /2, P2[6] = 1.3 V)	P2[4] + P2[6] - 0.092	P2[4] + P2[6] – 0.011	P2[4] + P2[6] + 0.064	V
		$V_{AGND}$	AGND	P2[4]	P2[4]	P2[4]	P2[4]	_
		V <sub>REFLO</sub>	Ref Low	P2[4]–P2[6] (P2[4] = V <sub>DD</sub> /2, P2[6] = 1.3 V)	P2[4] – P2[6] – 0.031	P2[4] – P2[6] + 0.007	P2[4] – P2[6] + 0.056	V
	RefPower = high Opamp bias = low	V <sub>REFHI</sub>	Ref High	P2[4]+P2[6] (P2[4] = V <sub>DD</sub> /2, P2[6] = 1.3 V)	P2[4] + P2[6] - 0.078	P2[4] + P2[6] - 0.008	P2[4] + P2[6] + 0.063	V
		$V_{AGND}$	AGND	P2[4]	P2[4]	P2[4]	P2[4]	_
		V <sub>REFLO</sub>	Ref Low	P2[4]–P2[6] (P2[4] = V <sub>DD</sub> /2, P2[6] = 1.3 V)	P2[4] – P2[6] – 0.031	P2[4] – P2[6] + 0.004	P2[4]-P2[6]+ 0.043	V
	RefPower = medium Opamp bias = high	V <sub>REFHI</sub>	Ref High	P2[4]+P2[6] (P2[4] = V <sub>DD</sub> /2, P2[6] = 1.3 V)	P2[4] + P2[6] - 0.073	P2[4] + P2[6] - 0.006	P2[4] + P2[6] + 0.062	V
		$V_{AGND}$	AGND	P2[4]	P2[4]	P2[4]	P2[4]	_
		V <sub>REFLO</sub>	Ref Low	P2[4]–P2[6] (P2[4] = V <sub>DD</sub> /2, P2[6] = 1.3 V)	P2[4] – P2[6] – 0.032	P2[4] – P2[6] + 0.003	P2[4] – P2[6] + 0.038	V
	RefPower = medium Opamp bias = low	V <sub>REFHI</sub>	Ref High	P2[4]+P2[6] (P2[4] = V <sub>DD</sub> /2, P2[6] = 1.3 V)	P2[4] + P2[6] - 0.073	P2[4] + P2[6] - 0.006	P2[4] + P2[6] + 0.062	V
		V <sub>AGND</sub>	AGND	P2[4]	P2[4]	P2[4]	P2[4]	_
		V <sub>REFLO</sub>	Ref Low	P2[4]–P2[6] (P2[4] = V <sub>DD</sub> /2, P2[6] = 1.3 V)	P2[4] – P2[6] – 0.034	P2[4] – P2[6] + 0.002	P2[4] – P2[6] + 0.037	V



Table 12. 5-V DC Analog Reference Specifications (continued)

Reference ARF_CR [5:3]	Reference Power Settings	Symbol	Reference	Description	Min	Тур	Max	Units
0b010	RefPower = high	V <sub>REFHI</sub>	Ref High	$V_{DD}$	V <sub>DD</sub> – 0.037	V <sub>DD</sub> – 0.007	V <sub>DD</sub>	V
	Opamp bias = high	V <sub>AGND</sub>	AGND	V <sub>DD</sub> /2	V <sub>DD</sub> /2 – 0.036	V <sub>DD</sub> /2 – 0.001	V <sub>DD</sub> /2 + 0.036	V
		V <sub>REFLO</sub>	Ref Low	V <sub>SS</sub>	$V_{SS}$	V <sub>SS</sub> + 0.005	V <sub>SS</sub> + 0.029	V
	RefPower = high	V <sub>REFHI</sub>	Ref High	$V_{DD}$	V <sub>DD</sub> - 0.034	V <sub>DD</sub> – 0.006	$V_{DD}$	V
	Opamp bias = low	V <sub>AGND</sub>	AGND	V <sub>DD</sub> /2	V <sub>DD</sub> /2 – 0.036	V <sub>DD</sub> /2 – 0.001	V <sub>DD</sub> /2 + 0.035	V
		V <sub>REFLO</sub>	Ref Low	V <sub>SS</sub>	$V_{SS}$	V <sub>SS</sub> + 0.004	V <sub>SS</sub> + 0.024	V
	RefPower = medium	V <sub>REFHI</sub>	Ref High	$V_{DD}$	V <sub>DD</sub> – 0.032	V <sub>DD</sub> – 0.005	$V_{DD}$	V
	Opamp bias = high	V <sub>AGND</sub>	AGND	V <sub>DD</sub> /2	V <sub>DD</sub> /2 – 0.036	V <sub>DD</sub> /2 – 0.001	$V_{DD}/2 + 0.035$	V
		V <sub>REFLO</sub>	Ref Low	V <sub>SS</sub>	$V_{SS}$	V <sub>SS</sub> + 0.003	V <sub>SS</sub> + 0.022	V
	RefPower = medium	V <sub>REFHI</sub>	Ref High	$V_{DD}$	V <sub>DD</sub> – 0.031	V <sub>DD</sub> – 0.005	$V_{DD}$	V
	Opamp bias = low	V <sub>AGND</sub>	AGND	V <sub>DD</sub> /2	V <sub>DD</sub> /2 – 0.037	V <sub>DD</sub> /2 – 0.001	$V_{DD}/2 + 0.035$	V
		V <sub>REFLO</sub>	Ref Low	V <sub>SS</sub>	$V_{SS}$	V <sub>SS</sub> + 0.003	V <sub>SS</sub> + 0.020	V
0b011	RefPower = high	V <sub>REFHI</sub>	Ref High	3 × Bandgap	3.760	3.884	4.006	V
	Opamp bias = high	V <sub>AGND</sub>	AGND	2 × Bandgap	2.522	2.593	2.669	V
		V <sub>REFLO</sub>	Ref Low	Bandgap	1.252	1.299	1.342	V
	RefPower = high	V <sub>REFHI</sub>	Ref High	3 × Bandgap	3.766	3.887	4.010	V
	Opamp bias = low	V <sub>AGND</sub>	AGND	2 × Bandgap	2.523	2.594	2.670	V
		V <sub>REFLO</sub>	Ref Low	Bandgap	1.252	1.297	1.342	V
	RefPower = medium	V <sub>REFHI</sub>	Ref High	3 × Bandgap	3.769	3.888	4.013	V
	Opamp bias = high	V <sub>AGND</sub>	AGND	2 × Bandgap	2.523	2.594	2.671	V
		V <sub>REFLO</sub>	Ref Low	Bandgap	1.251	1.296	1.343	V
	RefPower = medium	V <sub>REFHI</sub>	Ref High	3 × Bandgap	3.769	3.889	4.015	V
	Opamp bias = low	$V_{AGND}$	AGND	2 × Bandgap	2.523	2.595	2.671	V
		V <sub>REFLO</sub>	Ref Low	Bandgap	1.251	1.296	1.344	V
0b100	RefPower = high Opamp bias = high	V <sub>REFHI</sub>	Ref High	2 × Bandgap + P2[6] (P2[6] = 1.3 V)	2.483 – P2[6]	2.582 - P2[6]	2.674 – P2[6]	V
		$V_{AGND}$	AGND	2 × Bandgap	2.522	2.593	2.669	V
		V <sub>REFLO</sub>	Ref Low	2 × Bandgap – P2[6] (P2[6] = 1.3 V)	2.524 – P2[6]	2.600 - P2[6]	2.676 – P2[6]	V
	RefPower = high Opamp bias = low	V <sub>REFHI</sub>	Ref High	2 × Bandgap + P2[6] (P2[6] = 1.3 V)	2.490 - P2[6]	2.586 - P2[6]	2.679 - P2[6]	V
		$V_{AGND}$	AGND	2 × Bandgap	2.523	2.594	2.669	V
		V <sub>REFLO</sub>	Ref Low	2 × Bandgap – P2[6] (P2[6] = 1.3 V)	2.523 – P2[6]	2.598 - P2[6]	2.675 – P2[6]	V
	RefPower = medium Opamp bias = high	V <sub>REFHI</sub>	Ref High	2 × Bandgap + P2[6] (P2[6] = 1.3 V)	2.493 – P2[6]	2.588 – P2[6]	2.682 - P2[6]	V
		V <sub>AGND</sub>	AGND	2 × Bandgap	2.523	2.594	2.670	V
		V <sub>REFLO</sub>	Ref Low	2 × Bandgap – P2[6] (P2[6] = 1.3 V)	2.523 – P2[6]	2.597 – P2[6]	2.675 – P2[6]	V
	RefPower = medium Opamp bias = low	V <sub>REFHI</sub>	Ref High	2 × Bandgap + P2[6] (P2[6] = 1.3 V)	2.494 - P2[6]	2.589 - P2[6]	2.685 - P2[6]	V
		V <sub>AGND</sub>	AGND	2 × Bandgap	2.523	2.595	2.671	V
		V <sub>REFLO</sub>	Ref Low	2 × Bandgap – P2[6] (P2[6] = 1.3 V)	2.522 - P2[6]	2.596 - P2[6]	2.676 - P2[6]	V



Table 12. 5-V DC Analog Reference Specifications (continued)

Reference ARF_CR [5:3]	Reference Power Settings	Symbol	Reference	Description	Min	Тур	Max	Units
0b101	RefPower = high Opamp bias = high	V <sub>REFHI</sub>	Ref High	P2[4] + Bandgap (P2[4] = V <sub>DD</sub> /2)	P2[4] + 1.218	P2[4] + 1.291	P2[4] + 1.354	V
		$V_{AGND}$	AGND	P2[4]	P2[4]	P2[4]	P2[4]	-
		V <sub>REFLO</sub>	Ref Low	P2[4] - Bandgap (P2[4] = $V_{DD}/2$ )	P2[4] - 1.335	P2[4] - 1.294	P2[4] - 1.237	V
	RefPower = high Opamp bias = low	V <sub>REFHI</sub>	Ref High	P2[4] + Bandgap (P2[4] = V <sub>DD</sub> /2)	P2[4] + 1.221	P2[4] + 1.293	P2[4] + 1.358	V
		$V_{AGND}$	AGND	P2[4]	P2[4]	P2[4]	P2[4]	_
		V <sub>REFLO</sub>	Ref Low	P2[4] – Bandgap (P2[4] = V <sub>DD</sub> /2)	P2[4] - 1.337	P2[4] - 1.297	P2[4] - 1.243	V
	RefPower = medium Opamp bias = high	V <sub>REFHI</sub>	Ref High	P2[4] + Bandgap (P2[4] = V <sub>DD</sub> /2)	P2[4] + 1.222	P2[4] + 1.294	P2[4] + 1.360	V
		$V_{AGND}$	AGND	P2[4]	P2[4]	P2[4]	P2[4]	_
		V <sub>REFLO</sub>	Ref Low	P2[4] – Bandgap (P2[4] = V <sub>DD</sub> /2)	P2[4] - 1.338	P2[4] - 1.298	P2[4] - 1.245	V
	RefPower = medium Opamp bias = low	V <sub>REFHI</sub>	Ref High	P2[4] + Bandgap (P2[4] = V <sub>DD</sub> /2)	P2[4] + 1.221	P2[4] + 1.294	P2[4] + 1.362	V
		$V_{AGND}$	AGND	P2[4]	P2[4]	P2[4]	P2[4]	_
		V <sub>REFLO</sub>	Ref Low	P2[4] – Bandgap (P2[4] = V <sub>DD</sub> /2)	P2[4] - 1.340	P2[4] - 1.298	P2[4] - 1.245	V
0b110	RefPower = high	$V_{REFHI}$	Ref High	2 × Bandgap	2.513	2.593	2.672	V
	Opamp bias = high	$V_{AGND}$	AGND	Bandgap	1.264	1.302	1.340	V
		$V_{REFLO}$	Ref Low	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub> + 0.008	V <sub>SS</sub> + 0.038	V
	RefPower = high Opamp bias = low	V <sub>REFHI</sub>	Ref High	2 × Bandgap	2.514	2.593	2.674	V
		V <sub>AGND</sub>	AGND	Bandgap	1.264	1.301	1.340	V
		$V_{REFLO}$	Ref Low	$V_{SS}$	V <sub>SS</sub>	$V_{SS} + 0.005$	V <sub>SS</sub> + 0.028	V
	RefPower = medium	$V_{REFHI}$	Ref High	2 × Bandgap	2.514	2.593	2.676	V
	Opamp bias = high	$V_{AGND}$	AGND	Bandgap	1.264	1.301	1.340	V
		$V_{REFLO}$	Ref Low	$V_{SS}$	V <sub>SS</sub>	$V_{SS} + 0.004$	V <sub>SS</sub> + 0.024	V
	RefPower = medium	$V_{REFHI}$	Ref High	2 × Bandgap	2.514	2.593	2.677	V
	Opamp bias = low	$V_{AGND}$	AGND	Bandgap	1.264	1.300	1.340	V
		V <sub>REFLO</sub>	Ref Low	V <sub>SS</sub>	V <sub>SS</sub>	$V_{SS} + 0.003$	V <sub>SS</sub> + 0.021	V
0b111	RefPower = high	$V_{REFHI}$	Ref High	3.2 × Bandgap	4.028	4.144	4.242	V
	Opamp bias = high	$V_{AGND}$	AGND	1.6 × Bandgap	2.028	2.076	2.125	V
		$V_{REFLO}$	Ref Low	$V_{SS}$	V <sub>SS</sub>	$V_{SS} + 0.008$	$V_{SS} + 0.034$	V
	RefPower = high	$V_{REFHI}$	Ref High	3.2 × Bandgap	4.032	4.142	4.245	V
	Opamp bias = low	$V_{AGND}$	AGND	1.6 × Bandgap	2.029	2.076	2.126	V
		V <sub>REFLO</sub>	Ref Low	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub> + 0.005	V <sub>SS</sub> + 0.025	V
	RefPower = medium	1121111	Ref High	3.2 × Bandgap	4.034	4.143	4.247	V
	Opamp bias = high	V <sub>AGND</sub>	AGND	1.6 × Bandgap	2.029	2.076	2.126	V
		V <sub>REFLO</sub>	Ref Low	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub> + 0.004	V <sub>SS</sub> + 0.021	V
	RefPower = medium	V <sub>REFHI</sub>	Ref High	3.2 × Bandgap	4.036	4.144	4.249	V
	Opamp bias = low	$V_{AGND}$	AGND	1.6 × Bandgap	2.029	2.076	2.126	V
		V <sub>REFLO</sub>	Ref Low	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub> + 0.003	V <sub>SS</sub> + 0.019	V



Table 13. 3.3-V DC Analog Reference Specifications

Reference ARF_CR [5:3]	Reference Power Settings	Symbol	Reference	Description	Min	Тур	Max	Units
0b000	RefPower = high	V <sub>REFHI</sub>	Ref High	V <sub>DD</sub> /2 + Bandgap	V <sub>DD</sub> /2 + 1.200	V <sub>DD</sub> /2 + 1.290	V <sub>DD</sub> /2 + 1.365	V
	Opamp bias = high	V <sub>AGND</sub>	AGND	V <sub>DD</sub> /2	V <sub>DD</sub> /2 – 0.030	V <sub>DD</sub> /2	V <sub>DD</sub> /2 + 0.034	V
		V <sub>REFLO</sub>	Ref Low	V <sub>DD</sub> /2 – Bandgap	V <sub>DD</sub> /2 – 1.346	V <sub>DD</sub> /2 – 1.292	V <sub>DD</sub> /2 – 1.208	V
	RefPower = high	V <sub>REFHI</sub>	Ref High	V <sub>DD</sub> /2 + Bandgap	V <sub>DD</sub> /2 + 1.196	V <sub>DD</sub> /2 + 1.292	V <sub>DD</sub> /2 + 1.374	V
	Opamp bias = low	$V_{AGND}$	AGND	V <sub>DD</sub> /2	V <sub>DD</sub> /2 – 0.029	V <sub>DD</sub> /2	V <sub>DD</sub> /2 + 0.031	V
		V <sub>REFLO</sub>	Ref Low	V <sub>DD</sub> /2 – Bandgap	V <sub>DD</sub> /2 – 1.349 V <sub>DD</sub> /2 – 1.295		V <sub>DD</sub> /2 – 1.227	V
	RefPower = medium	$V_{REFHI}$	Ref High	V <sub>DD</sub> /2 + Bandgap	V <sub>DD</sub> /2 + 1.204	V <sub>DD</sub> /2 + 1.293	V <sub>DD</sub> /2 + 1.369	V
	Opamp bias = high	$V_{AGND}$	AGND	V <sub>DD</sub> /2	V <sub>DD</sub> /2 – 0.030	V <sub>DD</sub> /2	$V_{DD}/2 + 0.030$	V
		V <sub>REFLO</sub>	Ref Low	V <sub>DD</sub> /2 – Bandgap	V <sub>DD</sub> /2 – 1.351	V <sub>DD</sub> /2 – 1.297	V <sub>DD</sub> /2 – 1.229	V
	RefPower = medium	V <sub>REFHI</sub>	Ref High	V <sub>DD</sub> /2 + Bandgap	V <sub>DD</sub> /2 + 1.189	V <sub>DD</sub> /2 + 1.294	V <sub>DD</sub> /2 + 1.384	V
	Opamp bias = low	$V_{AGND}$	AGND	V <sub>DD</sub> /2	$V_{DD}/2 - 0.032$	V <sub>DD</sub> /2	$V_{DD}/2 + 0.029$	V
		$V_{REFLO}$	Ref Low	V <sub>DD</sub> /2 – Bandgap	V <sub>DD</sub> /2 – 1.353	V <sub>DD</sub> /2 – 1.297	V <sub>DD</sub> /2 – 1.230	V
0b001	RefPower = high Opamp bias = high	V <sub>REFHI</sub>	Ref High	P2[4]+P2[6] (P2[4] = V <sub>DD</sub> /2, P2[6] = 0.5 V)	P2[4] + P2[6] – 0.105	P2[4] + P2[6] – 0.008	P2[4] + P2[6] + 0.095	V
		V <sub>AGND</sub>	AGND	P2[4]	P2[4]	P2[4]	P2[4]	_
		V <sub>REFLO</sub>	Ref Low	P2[4]–P2[6] (P2[4] = V <sub>DD</sub> /2, P2[6] = 0.5 V)	P2[4] – P2[6] – 0.035	P2[4]-P2[6]+ 0.006	P2[4] – P2[6] + 0.053	V
	RefPower = high Opamp bias = low	V <sub>REFHI</sub>	Ref High	P2[4]+P2[6] (P2[4] = V <sub>DD</sub> /2, P2[6] = 0.5 V)	P2[4] + P2[6] - 0.094	P2[4] + P2[6] – 0.005	P2[4] + P2[6] + 0.073	V
		$V_{AGND}$	AGND	P2[4]	P2[4]	P2[4]	P2[4]	-
		V <sub>REFLO</sub>	Ref Low	P2[4]–P2[6] (P2[4] = V <sub>DD</sub> /2, P2[6] = 0.5 V)	P2[4] – P2[6] – 0.033	P2[4]-P2[6]+ 0.002	P2[4] – P2[6] + 0.042	V
	RefPower = medium Opamp bias = high	V <sub>REFHI</sub>	Ref High	P2[4]+P2[6] (P2[4] = V <sub>DD</sub> /2, P2[6] = 0.5 V)	P2[4] + P2[6] - 0.094	P2[4] + P2[6] - 0.003	P2[4] + P2[6] + 0.075	V
		$V_{AGND}$	AGND	P2[4]	P2[4]	P2[4]	P2[4]	-
		V <sub>REFLO</sub>	Ref Low	P2[4]–P2[6] (P2[4] = V <sub>DD</sub> /2, P2[6] = 0.5 V)	P2[4] – P2[6] – 0.035	P2[4] – P2[6]	P2[4]-P2[6]+ 0.038	V
	RefPower = medium Opamp bias = low	V <sub>REFHI</sub>	Ref High	P2[4]+P2[6] (P2[4] = V <sub>DD</sub> /2, P2[6] = 0.5 V)	P2[4] + P2[6] - 0.095	P2[4] + P2[6] - 0.003	P2[4] + P2[6] + 0.080	V
		V <sub>AGND</sub>	AGND	P2[4]	P2[4]	P2[4]	P2[4]	_
		V <sub>REFLO</sub>	Ref Low	P2[4]–P2[6] (P2[4] = V <sub>DD</sub> /2, P2[6] = 0.5 V)	P2[4] – P2[6] – 0.038	P2[4] - P2[6]	P2[4] – P2[6] + 0.038	V
0b010	RefPower = high	$V_{REFHI}$	Ref High	$V_{DD}$	V <sub>DD</sub> – 0.119	V <sub>DD</sub> – 0.005	$V_{DD}$	V
	Opamp bias = high	$V_{AGND}$	AGND	V <sub>DD</sub> /2	V <sub>DD</sub> /2 – 0.028	V <sub>DD</sub> /2	V <sub>DD</sub> /2 + 0.029	V
		V <sub>REFLO</sub>	Ref Low	$V_{SS}$	V <sub>SS</sub>	V <sub>SS</sub> + 0.004	V <sub>SS</sub> + 0.022	V
	RefPower = high	$V_{REFHI}$	Ref High	$V_{DD}$	V <sub>DD</sub> – 0.131	V <sub>DD</sub> – 0.004	$V_{DD}$	V
	Opamp bias = low	$V_{AGND}$	AGND	V <sub>DD</sub> /2	V <sub>DD</sub> /2 – 0.028	V <sub>DD</sub> /2	$V_{DD}/2 + 0.028$	V
		$V_{REFLO}$	Ref Low	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub> + 0.003	V <sub>SS</sub> + 0.021	V
	RefPower = medium	$V_{REFHI}$	Ref High	$V_{DD}$	V <sub>DD</sub> – 0.111	V <sub>DD</sub> – 0.003	$V_{DD}$	V
	Opamp bias = high	$V_{AGND}$	AGND	V <sub>DD</sub> /2	$V_{DD}/2 - 0.029$	V <sub>DD</sub> /2	$V_{DD}/2 + 0.028$	V
		V <sub>REFLO</sub>	Ref Low	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub> + 0.002	V <sub>SS</sub> + 0.017	V
	RefPower = medium	$V_{REFHI}$	Ref High	$V_{DD}$	V <sub>DD</sub> – 0.128	V <sub>DD</sub> – 0.003	$V_{DD}$	V
	Opamp bias = low	$V_{AGND}$	AGND	V <sub>DD</sub> /2	$V_{DD}/2 - 0.029$	V <sub>DD</sub> /2	$V_{DD}/2 + 0.029$	V
		$V_{REFLO}$	Ref Low	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub> + 0.002	V <sub>SS</sub> + 0.019	V
0b011	All power settings. Not allowed for 3.3 V.	_	_	_	_	_	_	_



Table 13. 3.3-V DC Analog Reference Specifications (continued)

Reference ARF_CR [5:3]	Reference Power Settings	Symbol	Reference	Description	Min	Тур	Max	Units
0b100	All power settings. Not allowed for 3.3 V.	_	-	-	_	_	_	_
0b101	RefPower = high Opamp bias = high	5 KEIIII 5 II		P2[4] + 1.214	P2[4] + 1.291	P2[4] + 1.359	V	
		V <sub>AGND</sub>	AGND	P2[4]	P2[4]	P2[4]	P2[4]	-
		V <sub>REFLO</sub>	Ref Low	P2[4] - Bandgap (P2[4] = $V_{DD}/2$ )	P2[4] – 1.335	P2[4] – 1.292	P2[4] - 1.200	V
	RefPower = high Opamp bias = low	V <sub>REFHI</sub>	Ref High	P2[4] + Bandgap (P2[4] = $V_{DD}/2$ )	P2[4] + 1.219	P2[4] + 1.293	P2[4] + 1.357	V
		$V_{AGND}$	AGND	P2[4]	P2[4]	P2[4]	P2[4]	-
		V <sub>REFLO</sub>	Ref Low	P2[4] - Bandgap (P2[4] = $V_{DD}/2$ )	P2[4] – 1.335	P2[4] - 1.295	P2[4] - 1.243	V
	RefPower = medium Opamp bias = high	V <sub>REFHI</sub>	Ref High	P2[4] + Bandgap (P2[4] = V <sub>DD</sub> /2)	P2[4] + 1.222	P2[4] + 1.294	P2[4] + 1.356	V
		V <sub>AGND</sub>	AGND	P2[4]	P2[4]	P2[4]	P2[4]	-
		V <sub>REFLO</sub>	Ref Low	P2[4] - Bandgap (P2[4] = $V_{DD}/2$ )	P2[4] – 1.337	P2[4] - 1.296	P2[4] - 1.244	V
	RefPower = medium Opamp bias = low	V <sub>REFHI</sub>	Ref High	P2[4] + Bandgap (P2[4] = V <sub>DD</sub> /2)	P2[4] + 1.224	P2[4] + 1.295	P2[4] + 1.355	V
		$V_{AGND}$	AGND	P2[4]	P2[4]	P2[4]	P2[4]	_
		V <sub>REFLO</sub>	Ref Low	P2[4] - Bandgap (P2[4] = $V_{DD}/2$ )	P2[4] - 1.339	P2[4] - 1.297	P2[4] - 1.244	V
0b110	RefPower = high	$V_{REFHI}$	Ref High	2 × Bandgap	2.510	2.595	2.655	V
	Opamp bias = high	$V_{AGND}$	AGND	Bandgap	1.276	1.301	1.332	V
		V <sub>REFLO</sub>	Ref Low	$V_{SS}$	V <sub>SS</sub>	V <sub>SS</sub> + 0.006	V <sub>SS</sub> + 0.031	V
	RefPower = high	$V_{REFHI}$	Ref High	2 × Bandgap	2.513	2.594	2.656	V
	Opamp bias = low	$V_{AGND}$	AGND	Bandgap	1.275	1.301	1.331	V
		V <sub>REFLO</sub>	Ref Low	$V_{SS}$	V <sub>SS</sub>	V <sub>SS</sub> + 0.004	V <sub>SS</sub> + 0.021	V
	RefPower = medium	$V_{REFHI}$	Ref High	2 × Bandgap	2.516	2.595	2.657	V
	Opamp bias = high	$V_{AGND}$	AGND	Bandgap	1.275	1.301	1.331	V
		V <sub>REFLO</sub>	Ref Low	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub> + 0.003	V <sub>SS</sub> + 0.017	V
	RefPower = medium	$V_{REFHI}$	Ref High	2 × Bandgap	2.520	2.595	2.658	V
	Opamp bias = low	$V_{AGND}$	AGND	Bandgap	1.275	1.300	1.331	V
		V <sub>REFLO</sub>	Ref Low	V <sub>SS</sub>	V <sub>SS</sub>	V <sub>SS</sub> + 0.002	V <sub>SS</sub> + 0.015	V
0b111	All power settings. Not allowed for 3.3 V.	_	_	_	_	_	_	_

## DC Analog enCoRe III Block Specifications

The following table lists guaranteed maximum and minimum specifications for the voltage and temperature ranges: 4.75 V to 5.25 V and  $-40 \text{ °C} \leq T_A \leq 85 \text{ °C}$ , or 3.15 V to 3.5 V and  $-40 \text{ °C} \leq T_A \leq 85 \text{ °C}$ , respectively. Typical parameters apply to 5 V and 3.3 V at 25 °C and are for design guidance only.

Table 14. DC Analog enCoRe III Block Specifications

Parameter	Description	Min	Тур	Max	Unit	Notes
R <sub>CT</sub>	Resistor unit value (CT)	_	12.2	_	kΩ	_
C <sub>SC</sub>	Capacitor unit value (SC)	-	80	-	fF	_

Document Number: 38-08036 Rev. \*N Page 21 of 45



## DC POR and LVD Specifications

The following table lists guaranteed maximum and minimum specifications for the voltage and temperature ranges: 4.75 V to 5.25 V and  $-40~^{\circ}\text{C} \le T_{A} \le 85~^{\circ}\text{C}$ , or 3.15 V to 3.5 V and  $-40~^{\circ}\text{C} \le T_{A} \le 85~^{\circ}\text{C}$ , respectively. Typical parameters apply to 5 V or 3.3 V at 25  $^{\circ}\text{C}$  and are for design guidance only.

**Note** The bits PORLEV and VM in the following table refer to bits in the VLT\_CR register. See the *PSoC® Technical Reference Manual* for more information on the VLT\_CR register.

Table 15. DC POR and LVD Specifications

Parameter	Description	Min	Тур	Max	Unit	Notes
	V <sub>DD</sub> value for PPOR trip (positive ramp) PORLEV[1:0] = 00b PORLEV[1:0] = 01b PORLEV[1:0] = 10b	-	2.91 4.39 4.55	ı	V V V	_
V <sub>PPOR0</sub> V <sub>PPOR1</sub> V <sub>PPOR2</sub>	V <sub>DD</sub> value for PPOR trip (negative ramp) PORLEV[1:0] = 00b PORLEV[1:0] = 01b PORLEV[1:0] = 10b	-	2.82 4.39 4.55	ı	V V V	_
V <sub>PH0</sub> V <sub>PH1</sub> V <sub>PH2</sub>	PPOR hysteresis PORLEV[1:0] = 00b PORLEV[1:0] = 01b PORLEV[1:0] = 10b	1 1	92 0 0	1 1 1	mV mV mV	_
VLVD0 VLVD1 VLVD2 VLVD3 VLVD4 VLVD5 VLVD6 VLVD7	V <sub>DD</sub> value for LVD trip VM[2:0] = 000b VM[2:0] = 001b VM[2:0] = 010b VM[2:0] = 011b VM[2:0] = 100b VM[2:0] = 101b VM[2:0] = 110b VM[2:0] = 111b	2.86 2.96 3.07 3.92 4.39 4.55 4.63 4.72	2.92 3.02 3.13 4.00 4.48 4.64 4.73 4.81	2.98 <sup>[6]</sup> 3.08 3.20 4.08 4.57 4.74 <sup>[7]</sup> 4.82 4.91	V V V V V	_

#### Notes

<sup>5.</sup> Errata: When VDD of the device is pulled below ground just before power on, the first read from each 8K Flash page may be corrupted. This issue does not affect Flash page 0 because it is the selected page upon reset. For more details in Errata on page 40.

<sup>6.</sup> Always greater than 50 mV above PPOR (PORLEV = 00) for falling supply.

<sup>7.</sup> Always greater than 50 mV above PPOR (PORLEV = 10) for falling supply.



#### DC Programming Specifications

The following table lists guaranteed maximum and minimum specifications for the voltage and temperature ranges: 4.75 V to 5.25 V and  $-40 \text{ °C} \leq T_A \leq 85 \text{ °C}$ , or 3.15 V to 3.5 V and  $-40 \text{ °C} \leq T_A \leq 85 \text{ °C}$ , respectively. Typical parameters apply to 5 V and 3.3 V at 25 °C and are for design guidance only.

**Table 16. DC Programming Specifications** 

Parameter	Description	Min	Тур	Max	Unit	Notes
$V_{DDP}$	V <sub>DD</sub> for programming and erase	4.5	5.0	5.5	V	This specification applies to the functional requirements of external programmer tools
V <sub>DDLV</sub>	Low V <sub>DD</sub> for verify	3.0	3.1	3.2	٧	This specification applies to the functional requirements of external programmer tools
$V_{DDHV}$	High V <sub>DD</sub> for verify	5.1	5.2	5.3	V	This specification applies to the functional requirements of external programmer tools
V <sub>DDIWRITE</sub>	Supply voltage for flash write operation	3.15	_	5.25	V	This specification applies to this device when it is executing internal flash writes
$I_{DDP}$	Supply current during programming or verify	-	15	30	mΑ	_
$V_{ILP}$	Input low voltage during programming or verify	_	_	0.8	V	_
$V_{IHP}$	Input high voltage during programming or Verify	2.1	_	_	V	_
I <sub>ILP</sub>	Input current when applying Vilp to P1[0] or P1[1] during programming or verify	_	-	0.2	mA	Driving internal pull-down resistor.
I <sub>IHP</sub>	Input current when applying Vihp to P1[0] or P1[1] during programming or verify	-	_	1.5	mA	Driving internal pull-down resistor.
$V_{OLV}$	Output low voltage during programming or verify	-	_	$V_{SS} + 0.75$	V	_
V <sub>OHV</sub>	Output high voltage during programming or verify	V <sub>DD</sub> – 1.0	_	V <sub>DD</sub>	V	_
Flash <sub>ENPB</sub>	Flash endurance (per block)	50,000 <sup>[8]</sup>	_	_	-	Erase/write cycles per block.
Flash <sub>ENT</sub>	Flash endurance (total) <sup>[9]</sup>	1,800,000	_	-	-	Erase/write cycles.
Flash <sub>DR</sub>	Flash data retention	10	_	_	Years	_

## DC I<sup>2</sup>C Specifications

The following table lists guaranteed maximum and minimum specifications for the voltage and temperature ranges: 4.75 V to 5.25 V and  $-40 \text{ °C} \leq T_A \leq 85 \text{ °C}$ , or 3.15 V to 3.5 V and  $-40 \text{ °C} \leq T_A \leq 85 \text{ °C}$ , respectively. Typical parameters apply to 5 V and 3.3 V at 25 °C and are for design guidance only.

Table 17. DC I<sup>2</sup>C Specifications [10]

Symbol	Description	Min	Тур	Max	Units	Notes
$V_{ILI2C}$	Input low level	_	_	$0.3 \times V_{DD}$	V	$3.15 \text{ V} \le \text{V}_{DD} \le 3.6 \text{ V}$
		_	_	0.25 × V <sub>DD</sub>	V	$4.75 \text{ V} \le \text{V}_{DD} \le 5.25 \text{ V}$
V <sub>IHI2C</sub>	Input high level	0.7 × V <sub>DD</sub>	_	_	V	$3.15 \text{ V} \le \text{V}_{DD} \le 5.25 \text{ V}$

#### Notes

- 8. The 50,000 cycle Flash endurance per block will only be guaranteed if the Flash is operating within one voltage range. Voltage ranges are 3.0V to 3.6V and 4.75V to 5.25V.
- 9. A maximum of 36 x 50,000 block endurance cycles is allowed. This may be balanced between operations on 36x1 blocks of 50,000 maximum cycles each, 36x2 blocks of 25,000 maximum cycles each, or 36x4 blocks of 12,500 maximum cycles each (to limit the total number of cycles to 36x50,000 and that no single block ever sees more than 50,000 cycles).
  - ever sees more than 50,000 cycles).
    For the full industrial range, the user must employ a temperature sensor user module (FlashTemp) and feed the result to the temperature argument before writing. Refer to the Flash APIs application note AN2015 for more information.
- 10. All GPIOs meet the DC GPIO V<sub>IL</sub> and V<sub>IH</sub> specifications found in the DC GPIO Specifications sections. The I<sup>2</sup>C GPIO pins also meet the mentioned specifications.

Document Number: 38-08036 Rev. \*N Page 23 of 45



## **AC Electrical Characteristics**

AC Chip-Level Specifications

The following table lists guaranteed maximum and minimum specifications for the voltage and temperature ranges: 4.75 V to 5.25 V and -40 °C  $\leq$  T<sub>A</sub>  $\leq$  85 °C, or 3.15 V to 3.5 V and -40 °C  $\leq$  T<sub>A</sub>  $\leq$  85 °C, respectively. Typical parameters apply to 5 V and 3.3 V at 25 °C and are for design guidance only.

Table 18. AC Chip-Level Specifications

Parameter	Description	Min	Тур	Max	Unit	Notes
F <sub>IMO245V</sub>	IMO frequency for 24 MHz (5 V)	23.04	24	24.96 <sup>[11, 12]</sup>	MHz	Trimmed for 5 V operation using factory trim values.
F <sub>IMO243V</sub>	IMO frequency for 24 MHz (3.3 V)	22.08	24	25.92 <sup>[11,13]</sup>	MHz	Trimmed for 3.3 V operation using factory trim values.
F <sub>IMOUSB</sub>	IMO frequency with USB frequency locking enabled and USB traffic present	23.94	24	24.06 <sup>[12]</sup>	MHz	USB operation for system clock source from the IMO is limited to $0^{\circ}\text{C} \leq \text{T}_{\text{A}} \leq 70^{\circ}\text{C}$ .
F <sub>CPU1</sub>	CPU frequency (5 V nominal)	0.090	24	24.96 <sup>[11,12]</sup>	MHz	SLIMO mode = 0.
F <sub>CPU2</sub>	CPU frequency (3.3 V nominal)	0.086	12	12.96 <sup>[12,13]</sup>	MHz	SLIMO mode = 0.
F <sub>BLK5</sub>	Digital PSoC block frequency (5 V nominal)	0	48	49.92 <sup>[11,12,14]</sup>	MHz	Refer to the AC Digital Block Specifications on page 26.
F <sub>BLK3</sub>	Digital PSoC block frequency (3.3 V nominal)	0	24	25.92 <sup>[12,14]</sup>	MHz	_
F <sub>32K1</sub>	ILO frequency	15	32	64	kHz	_
F <sub>32K_U</sub>	ILO untrimmed frequency	5	-	100	kHz	After a reset and before the M8C starts to run, the ILO is not trimmed. See the System Resets section of the PSoC Technical Reference Manual for details on this timing.
DC <sub>ILO</sub>	ILO duty cycle	20	50	80	%	_
DC <sub>24M</sub>	24-MHz duty cycle	40	50	60	%	-
Step24M	24-MH trim step size	_	50	-	kHz	_
Fout48M	48-MHz output frequency	46.08	48.0	49.92 <sup>[11,13]</sup>	MHz	Trimmed. Utilizing factory trim values.
F <sub>MAX</sub>	Maximum frequency of signal on row input or row output	ı	_	12.96	MHz	_
SR <sub>POWER_UP</sub>	Power supply slew rate	_	_	250	V/ms	_
T <sub>POWERUP</sub>	Time from end of POR to CPU executing code	-	16	100	ms	_
T <sub>jit_IMO</sub> <sup>[15]</sup>	24 MHz IMO cycle-to-cycle jitter (RMS)	-	200	1200	ps	
	24 MHz IMO long term N cycle-to-cycle jitter (RMS)	_	900	6000	ps	N = 32.
	24 MHz IMO period jitter (RMS)	_	200	900	ps	

#### Notes

<sup>11. 4.75</sup> V < V<sub>DD</sub> < 5.25 V.
12. Accuracy derived from Internal Main Oscillator with appropriate trim for V<sub>DD</sub> range.
13. 3.0 V < V<sub>DD</sub> < 3.6 V. See application note AN2012 "Adjusting PSoC Microcontroller Trims for Dual Voltage-Range Operation" for information on trimming for operation at 3.3 V.

<sup>14.</sup> See the individual user module data sheets for information on maximum frequencies for user modules.

<sup>15.</sup> Refer to Cypress Jitter Specifications application note, Understanding Datasheet Jitter Specifications for Cypress Timing Products - AN5054 for more information.



## AC GPIO Specifications

The following table lists guaranteed maximum and minimum specifications for the voltage and temperature ranges: 4.75 V to 5.25 V and  $-40 \text{ °C} \leq T_A \leq 85 \text{ °C}$ , or 3.15 V to 3.5 V and  $-40 \text{ °C} \leq T_A \leq 85 \text{ °C}$ , respectively. Typical parameters apply to 5 V and 3.3 V at 25 °C and are for design guidance only.

Table 19. AC GPIO Specifications

Parameter	Description	Min	Тур	Max	Unit	Notes
F <sub>GPIO</sub>	GPIO operating frequency	0	_	12	MHz	Normal Strong Mode
TRiseF	Rise time, normal strong mode, Cload = 50 pF	3	_	18	ns	V <sub>DD</sub> = 4.5 to 5.25 V, 10%–90%
TFallF	Fall time, normal strong mode, Cload = 50 pF	2	_	18	ns	V <sub>DD</sub> = 4.5 to 5.25 V, 10%–90%
TRiseS	Rise time, slow strong mode, Cload = 50 pF	10	27	-	ns	V <sub>DD</sub> = 3 to 5.25 V, 10%–90%
TFallS	Fall time, slow strong mode, Cload = 50 pF	10	22	_	ns	V <sub>DD</sub> = 3 to 5.25 V, 10%–90%

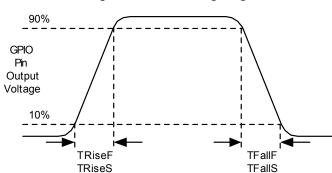


Figure 6. GPIO Timing Diagram

#### AC Full Speed USB Specifications

The following table lists guaranteed maximum and minimum specifications for the voltage and temperature ranges: 4.75 V to 5.25 V and  $-40 \,^{\circ}\text{C} \leq \text{T}_{\text{A}} \leq 85 \,^{\circ}\text{C}$ , or  $3.15 \,^{\circ}\text{V}$  to  $3.5 \,^{\circ}\text{V}$  and  $-40 \,^{\circ}\text{C} \leq \text{T}_{\text{A}} \leq 85 \,^{\circ}\text{C}$ , respectively. Typical parameters apply to  $5 \,^{\circ}\text{V}$  and  $3.3 \,^{\circ}\text{V}$  at  $25 \,^{\circ}\text{C}$  and are for design guidance only.

Table 20. AC Full-Speed (12 Mbps) USB Specifications

Parameter	Description	Min	Тур	Max	Unit	Notes
T <sub>RFS</sub>	Transition rise time	4	-	20	ns	For 50-pF load.
T <sub>FSS</sub>	Transition fall time	4	_	20	ns	For 50-pF load.
T <sub>RFMFS</sub>	Rise/fall time matching: (T <sub>R</sub> /T <sub>F</sub> )	90	-	111	%	For 50-pF load.
T <sub>DRATEFS</sub>	Full-speed data rate	12 – 0.25%	12	12 + 0.25%	Mbps	-



## AC Digital Block Specifications

The following table lists guaranteed maximum and minimum specifications for the voltage and temperature ranges: 4.75 V to 5.25 V and  $-40~^{\circ}\text{C} \leq T_A \leq 85~^{\circ}\text{C}$ , or 3.15 V to 3.5 V and  $-40~^{\circ}\text{C} \leq T_A \leq 85~^{\circ}\text{C}$ , respectively. Typical parameters apply to 5 V and 3.3 V at 25  $^{\circ}\text{C}$  and are for design guidance only.

Table 21. AC Digital Block Specifications

Function	Description	Min	Тур	Max	Unit	Notes
All functions	Block input clock frequency			ı	ı	
	V <sub>DD</sub> ≥ 4.75 V	-	_	49.92	MHz	
	V <sub>DD</sub> < 4.75 V	-	_	25.92	MHz	
Timer	Input clock frequency			I.	I	
	No capture, V <sub>DD</sub> ≥ 4.75 V	_	_	49.92	MHz	
	No capture, V <sub>DD</sub> < 4.75 V	-	_	25.92	MHz	
	With capture	_	_	25.92	MHz	
	Capture pulse width	50 <sup>[16]</sup>	_	-	ns	
Counter	Input clock frequency			I.	I	
	No enable input, V <sub>DD</sub> ≥ 4.75 V	_	_	49.92	MHz	
	No enable input, V <sub>DD</sub> < 4.75 V	_	_	25.92	MHz	
	With enable input	_	_	25.92	MHz	
	Enable input pulse width	50 <sup>[16]</sup>	_	_	ns	
	Kill pulse width		I			
	Asynchronous restart mode	20	_	_	ns	
	Synchronous restart mode	50 <sup>[16]</sup>	_	_	ns	
	Disable mode	50 <sup>[16]</sup>	_	_	ns	
	Input clock frequency		I	ı	ı	
	V <sub>DD</sub> ≥ 4.75 V	-	_	49.92	MHz	
	V <sub>DD</sub> < 4.75 V	-	_	25.92	MHz	
CRCPRS	Input clock frequency			I.	I	
(PRS Mode)	V <sub>DD</sub> ≥ 4.75 V	_	_	49.92	MHz	
inida)	V <sub>DD</sub> < 4.75 V	-	_	25.92	MHz	
CRCPRS (CRC Mode)	Input clock frequency	-	_	24.6	MHz	
SPIM	Input clock frequency	-	-	8.2	MHz	The SPI serial clock (SCLK) frequency is equal to the input clock frequency divided by 2.
SPIS	Input clock (SCLK) frequency	_	_	4.1	MHz	The input clock is the SPI SCLK in SPIS mode.
	Width of SS_negated between transmissions	50 <sup>[16]</sup>	_	-	ns	
Transmitter	Input clock frequency		I	ı	ı	The baud rate is equal to the input clock frequency
	V <sub>DD</sub> ≥ 4.75 V, 2 stop bits	_	_	49.92	MHz	divided by 8.
	V <sub>DD</sub> ≥ 4.75 V, 1 stop bit	_	_	24.6	MHz	
	V <sub>DD</sub> < 4.75 V	-	_	24.6	MHz	
Receiver	Input clock frequency					The baud rate is equal to the input clock frequency
	V <sub>DD</sub> ≥ 4.75 V, 2 stop bits	_	_	49.92	MHz	divided by 8.
	V <sub>DD</sub> ≥ 4.75 V, 1 stop bit	-	-	24.6	MHz	
	V <sub>DD</sub> < 4.75 V	_	_	24.6	MHz	

## Note

<sup>16.50</sup> ns minimum input pulse width is based on the input synchronizers running at 24 MHz (42 ns nominal period).



## AC External Clock Specifications

The following tables list guaranteed maximum and minimum specifications for the voltage and temperature ranges: 4.75 V to 5.25 V and  $-40 \text{ °C} \leq T_A \leq 85 \text{ °C}$ , or 3.15 V to 3.5 V and  $-40 \text{ °C} \leq T_A \leq 85 \text{ °C}$ , respectively. Typical parameters apply to 5 V and 3.3 V at 25 °C and are for design guidance only.

Table 22. AC External Clock Specifications

Parameter	Description	Min	Тур	Max	Unit	Notes
F <sub>OSCEXT</sub>	Frequency for USB applications	23.94	24	24.06		USB operation in the extended Industrial temperature range ( $-40~^{\circ}\text{C} \leq T_{\text{A}} \leq 85~^{\circ}\text{C}$ ) requires that the system clock is sourced from an external clock oscillator.
_	Duty cycle	47	50	53	%	_
_	Power-up to IMO switch	150	-	_	μS	_

## AC Analog Output Buffer Specifications

The following tables list guaranteed maximum and minimum specifications for the voltage and temperature ranges: 4.75 V to 5.25 V and  $-40 \,^{\circ}\text{C} \leq \text{T}_{A} \leq 85 \,^{\circ}\text{C}$ , or  $3.15 \,^{\circ}\text{V}$  to  $3.5 \,^{\circ}\text{V}$  and  $-40 \,^{\circ}\text{C} \leq \text{T}_{A} \leq 85 \,^{\circ}\text{C}$ , respectively. Typical parameters apply to  $5 \,^{\circ}\text{V}$  and  $3.3 \,^{\circ}\text{V}$  at  $25 \,^{\circ}\text{C}$  and are for design guidance only.

Table 23. 5 V AC Analog Output Buffer Specifications

Parameter	Description	Min	Тур	Max	Unit	Notes
T <sub>ROB</sub>	Rising settling time to 0.1%, 1 V Step, 100-pF load Power = low Power = high	_ _	_ _	2.5 2.5	μ <b>s</b> μ <b>s</b>	_
T <sub>SOB</sub>	Falling settling time to 0.1%, 1 V Step, 100-pF load Power = low Power = high	_ _	_ _	2.2 2.2	μ <b>s</b> μ <b>s</b>	_
SR <sub>ROB</sub>	Rising slew rate (20% to 80%), 1 V Step, 100-pF load Power = low Power = high	0.65 0.65	_ _	_ _	V/μs V/μs	_
SR <sub>FOB</sub>	Falling slew rate (80% to 20%), 1 V Step, 100-pF load Power = low Power = high	0.65 0.65	_ _	_ _	V/μs V/μs	_
BW <sub>OBSS</sub>	Small signal bandwidth, 20 mV <sub>pp</sub> , 3-dB BW, 100-pF load Power = low Power = high	0.8 0.8	_ _	_ _	MHz MHz	_
BW <sub>OBLS</sub>	Large signal bandwidth, 1 V <sub>pp</sub> , 3-dB BW, 100-pF load Power = low Power = high	300 300	_ _	_ _	kHz kHz	_

Table 24. 3.3 V AC Analog Output Buffer Specifications

Parameter	Description	Min	Тур	Max	Unit	Notes
T <sub>ROB</sub>	Rising settling time to 0.1%, 1 V Step, 100-pF load Power = low Power = high	_ _	_ _	3.8 3.8	μ <b>S</b> μ <b>S</b>	_
T <sub>SOB</sub>	Falling settling time to 0.1%, 1 V Step, 100-pF load Power = low Power = high		_ _	2.6 2.6	μ <b>s</b> μ <b>s</b>	_
SR <sub>ROB</sub>	Rising slew rate (20% to 80%), 1 V Step, 100-pF load Power = low Power = high	0.5 0.5	_ _	_ _	V/μs V/μs	_
SR <sub>FOB</sub>	Falling slew rate (80% to 20%), 1 V Step, 100-pF load Power = low Power = high	0.5 0.5	_ _	_ _	V/μs V/μs	_
BW <sub>OBSS</sub>	Small signal bandwidth, 20 mV <sub>pp</sub> , 3dB BW, 100-pF load Power = low Power = high	0.7 0.7	_ _	_ _	MHz MHz	_
BW <sub>OBLS</sub>	Large signal bandwidth, 1 V <sub>pp</sub> , 3dB BW, 100-pF load Power = low Power = high	200 200	_ _	_ _	kHz kHz	_



## AC Programming Specifications

The following table lists guaranteed maximum and minimum specifications for the voltage and temperature ranges: 4.75 V to 5.25 V and  $-40~^{\circ}\text{C} \le T_A \le 85~^{\circ}\text{C}$ , or 3.15 V to 3.5 V and  $-40~^{\circ}\text{C} \le T_A \le 85~^{\circ}\text{C}$ , respectively. Typical parameters apply to 5 V and 3.3 V at 25  $^{\circ}\text{C}$  and are for design guidance only.

**Table 25. AC Programming Specifications** 

Parameter	Description	Min	Тур	Max	Unit	Notes
T <sub>RSCLK</sub>	Rise time of SCLK	1	_	20	ns	-
T <sub>FSCLK</sub>	Fall time of SCLK	1	_	20	ns	-
T <sub>SSCLK</sub>	Data setup time to falling edge of SCLK	40	-	_	ns	-
T <sub>HSCLK</sub>	Data hold time from falling edge of SCLK	40	_	_	ns	-
F <sub>SCLK</sub>	Frequency of SCLK	0	_	8	MHz	-
T <sub>ERASEB</sub>	Flash erase time (block)	_	10	_	ms	-
T <sub>WRITE</sub>	Flash block write time	_	40	_	ms	-
T <sub>DSCLK</sub>	Data out delay from falling edge of SCLK	-	_	45	ns	V <sub>DD</sub> > 3.6
T <sub>DSCLK3</sub>	Data out delay from falling edge of SCLK	_	_	50	ns	3.15 ≤ V <sub>DD</sub> ≤ 3.5
T <sub>ERASEALL</sub>	Flash erase time (bulk)	-	40	_	ms	Erase all blocks and protection fields at once.
T <sub>PROGRAM_HOT</sub>	Flash block erase + flash block write time	_	_	100	ms	$0  ^{\circ}\text{C} \le \text{T}_{\text{J}} \le 100  ^{\circ}\text{C}$
T <sub>PROGRAM_COLD</sub>	Flash block erase + flash block write time	_	_	200	ms	$-40~^{\circ}C \le T_{J} \le 0~^{\circ}C$

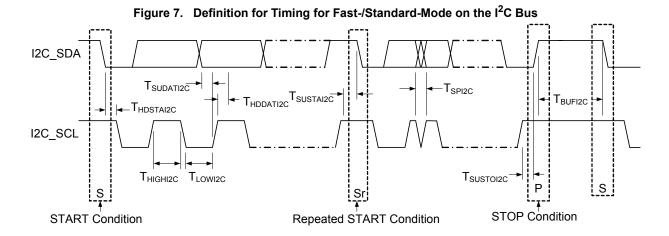


# AC I<sup>2</sup>C Specifications

The following table lists guaranteed maximum and minimum specifications for the voltage and temperature ranges: 4.75 V to 5.25 V and  $-40~^{\circ}\text{C} \le T_A \le 85~^{\circ}\text{C}$ , or 3.15 V to 3.5 V and  $-40~^{\circ}\text{C} \le T_A \le 85~^{\circ}\text{C}$ , respectively. Typical parameters apply to 5 V and 3.3 V at 25  $^{\circ}\text{C}$  and are for design guidance only.

Table 26. AC Characteristics of the I<sup>2</sup>C SDA and SCL Pins for V<sub>DD</sub>

Parameter	Description	Standa	rd-Mode	Fast-l	Mode	Unit	Notes
Parameter	Description	Min	Max	Min	Max	Ullit	Notes
F <sub>SCLI2C</sub>	SCL clock frequency	0	100	0	400	kHz	_
T <sub>HDSTAI2C</sub>	Hold time (repeated) START condition. After this period, the first clock pulse is generated.	4.0	_	0.6	_	μS	_
T <sub>LOWI2C</sub>	LOW period of the SCL clock	4.7	_	1.3	_	μS	_
T <sub>HIGHI2C</sub>	HIGH period of the SCL clock	4.0	_	0.6	_	μS	_
T <sub>SUSTAI2C</sub>	Setup time for a repeated START condition	4.7	_	0.6	_	μS	_
T <sub>HDDATI2C</sub>	Data hold time	0	-	0	_	μS	_
T <sub>SUDATI2C</sub>	Data setup time	250	-	100 <sup>[17]</sup>	_	ns	_
T <sub>SUSTOI2C</sub>	Setup time for STOP condition	4.0	_	0.6	_	μS	_
T <sub>BUFI2C</sub>	Bus free time between a STOP and START condition	4.7	_	1.3	_	μS	_
T <sub>SPI2C</sub>	Pulse width of spikes are suppressed by the input filter.	_	_	0	50	ns	_



Note

<sup>17.</sup> A Fast-Mode I²C-bus device can be used in a Standard-Mode I²C-bus system, but the requirement T<sub>SUDATI2C</sub> ≥ 250 ns must then be met. This automatically is the case if the device does not stretch the LOW period of the SCL signal. If such device does stretch the LOW period of the SCL signal, it must output the next data bit to the SDA line t<sub>rmax</sub> + T<sub>SUDATI2C</sub> = 1000 + 250 = 1250 ns (according to the Standard-Mode I²C-bus specification) before the SCL line is released.

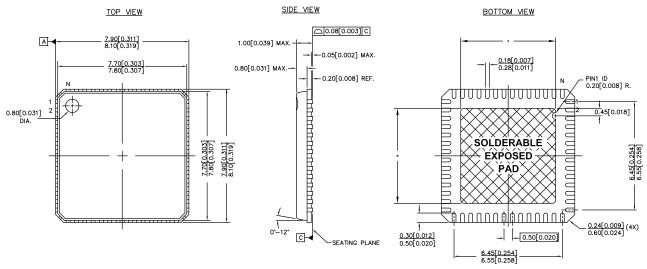


# **Packaging Information**

This section illustrates the package specification for the CY7C64215 enCoRe III, along with the thermal impedance for the package. **Important Note** Emulation tools may require a larger area on the target PCB than the chip's footprint. For a detailed description of the emulation tools' dimensions, refer to the emulator pod drawings at <a href="http://www.cypress.com">http://www.cypress.com</a>.

## **Package Diagrams**

Figure 8. 56-pin QFN (8 × 8 × 1.0 mm) 4.5 × 5.21 E-Pad (Subcon Punch Type Package) Package Outline, 001-12921



#### NOTES:

- 1. MATCH AREA IS SOLDERABLE EXPOSED METAL.
- 2. REFERENCE JEDEC#: MO-220
- 3. PACKAGE WEIGHT: 0.162g
- 4. ALL DIMENSIONS ARE IN MM [MIN/MAX]
- 5. PACKAGE CODE

PART #	DESCRIPTION
LF56A	STANDARD
LY56A	PB-FREE

001-12921 \*C



Figure 9. 56-pin QFN (8 × 8 × 1.0 mm) 4.5 × 5.2 E-Pad (Sawn) Package Outline, 001-53450

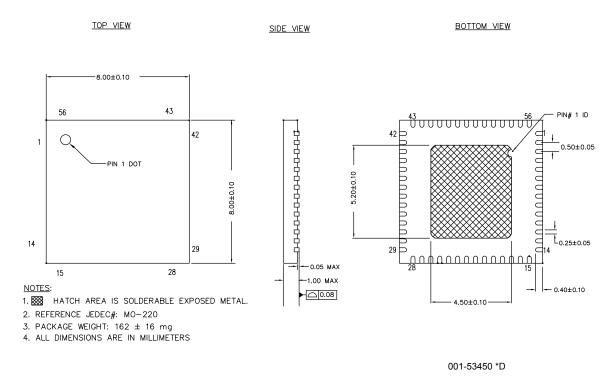
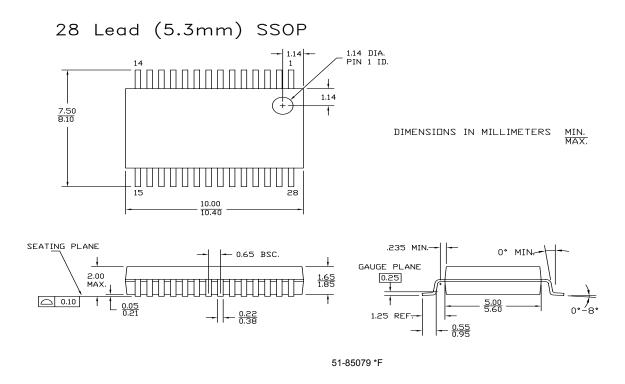


Figure 10. 28-pin SSOP (210 Mils) Package Outline, 51-85079





## Thermal Impedance

Table 27. Thermal Impedance for the Package

Package	Typical θ <sub>JA</sub> <sup>[18]</sup>
56-pin QFN <sup>[19]</sup>	20 °C/W
28-pin SSOP	96 °C/W

## **Solder Reflow Peak Temperature**

Following is the minimum solder reflow peak temperature to achieve good solderability.

Table 28. Solder Reflow Peak Temperature

Package	Maximum Peak Temperature	Time at Maximum Peak Temperature
56-pin QFN	260 °C	20 s
28-pin SSOP	260 °C	20 s

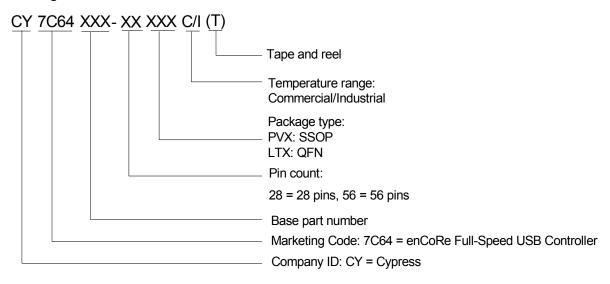
Notes  $18.\,T_J = T_A + POWER\,x\,\theta_{JA}$   $19.\,To\,achieve\,the\,thermal\,impedance\,specified\,for\,the\,QFN\,package,\,the\,center\,thermal\,pad\,should\,be\,soldered\,to\,the\,PCB\,ground\,plane.$ 



# **Ordering Information**

Package	Ordering Code	Flash Size	SRAM (Bytes)	Temperature Range
28-pin SSOP	CY7C64215-28PVXC	16K	1K	Commercial, 0 °C to 70 °C
28-pin SSOP (Tape and Reel)	CY7C64215-28PVXCT	16K	1K	Commercial, 0 °C to 70 °C
28-pin SSOP	CY7C64215-28PVXI	16 K	1K	Industrial, –40 °C to 85 °C
28-pin SSOP (Tape and Reel)	CY7C64215-28PVXIT	16 K	1K	Industrial, –40 °C to 85 °C
56-pin QFN (Sawn)	CY7C64215-56LTXC	16K	1K	Commercial, 0 °C to 70 °C
56-pin QFN (Sawn) (Tape and Reel)	CY7C64215-56LTXCT	16K	1K	Commercial, 0 °C to 70 °C
56-pin QFN (Sawn)	CY7C64215-56LTXI	16K	1K	Industrial, –40 °C to 85 °C
56-pin QFN (Sawn) (Tape and Reel)	CY7C64215-56LTXIT	16K	1K	Industrial, –40 °C to 85 °C

# **Ordering Code Definitions**





## **Acronyms**

## **Acronyms Used**

The following table lists the acronyms that are used in this document.

Acronym	Description	Acronym	Description
AC	alternating current	MIPS	million instructions per second
ADC	analog-to-digital converter	PCB	printed circuit board
API	application programming interface	PGA	programmable gain amplifier
CPU	central processing unit	POR	power-on reset
CRC	cyclic redundancy check	PPOR	precision power-on reset
CT	continuous time	PSoC <sup>®</sup>	Programmable System-on-Chip™
DAC	digital-to-analog converter	PWM	pulse-width modulator
DC	direct current	QFN	quad flat no leads
EEPROM	electrically erasable programmable read-only memory	RF	radio frequency
GPIO	general purpose I/O	SC	switched capacitor
ICE	in-circuit emulator	SLIMO	slow IMO
IDE	integrated development environment	SPI™	serial peripheral interface
ILO	internal low speed oscillator	SRAM	static random-access memory
IMO	internal main oscillator	SROM	supervisory read-only memory
I/O	input/output	SSOP	shrink small-outline package
ISSP	In-System Serial Programming	UART	universal asynchronous receiver / transmitter
LVD	low voltage detect	USB	universal serial bus
MAC	multiply-accumulate	WDT	watchdog timer

## **Reference Documents**

CY8CPLC20, CY8CLED16P01, CY8C29x66, CY8C27x43, CY8C24x94, CY8C24x23, CY8C24x23A, CY8C22x13, CY8C21x34, CY8C21x23, CY7C64215, CY7C603xx, CY8CNP1xx, and CYWUSB6953 PSoC® Programmable System-on-Chip Technical Reference Manual (TRM) (001-14463)

Design Aids – Reading and Writing PSoC<sup>®</sup> Flash – AN2015 (001-40459)

Document Number: 38-08036 Rev. \*N Page 34 of 45



## **Document Conventions**

#### Units of Measure

Symbol	Unit of Measure	Symbol	Unit of Measure
°C	degree Celsius	ms	milli-second
dB	decibels	mV	milli-volts
fF	femto farad	nA	nanoampere
kHz	kilohertz	ns	nanosecond
kΩ	kilohm	Ω	ohm
MHz	megahertz	pF	picofarad
μΑ	microampere	ps	picosecond
μS	microsecond	%	percent
μV	microvolts	V	volts
mA	milli-ampere	W	watts
mm	milli-meter		

#### **Numeric Conventions**

Hexadecimal numbers are represented with all letters in uppercase with an appended lowercase 'h' (for example, '14h' or '3Ah'). Hexadecimal numbers may also be represented by a '0x' prefix, the C coding convention. Binary numbers have an appended lowercase 'b' (for example, '01010100b' or '01000011b'). Numbers not indicated by an 'h' or 'b' are decimal.

active high 5. A logic signal having its asserted state as the logic 1 state.

6. A logic signal having the logic 1 state as the higher voltage of the two states.

analog blocks The basic programmable opamp circuits. These are SC (switched capacitor) and CT (continuous time) blocks.

These blocks can be interconnected to provide ADCs, DACs, multi-pole filters, gain stages, and much more.

API (Application

analog-to-digital

(ADC)

A device that changes an analog signal to a digital signal of corresponding magnitude. Typically, an ADC converts a voltage to a digital number. The digital-to-analog (DAC) converter performs the reverse operation.

Programming Interface)

A series of software routines that comprise an interface between a computer application and lower level services and functions (for example, user modules and libraries). APIs serve as building blocks for programmers that create software applications.

asynchronous

A signal whose data is acknowledged or acted upon immediately, irrespective of any clock signal.

Bandgap reference

A stable voltage reference design that matches the positive temperature coefficient of VT with the negative temperature coefficient of VBE, to produce a zero temperature coefficient (ideally) reference.

bandwidth

- 1. The frequency range of a message or information processing system measured in hertz.
- 2. The width of the spectral region over which an amplifier (or absorber) has substantial gain (or loss); it is sometimes represented more specifically as, for example, full width at half maximum.

bias

- 1. A systematic deviation of a value from a reference value.
- 2. The amount by which the average of a set of values departs from a reference value.
- 3. The electrical, mechanical, magnetic, or other force (field) applied to a device to establish a reference level to operate the device.

block

- 1. A functional unit that performs a single function, such as an oscillator.
- A functional unit that may be configured to perform one of several functions, such as a digital PSoC block or an analog PSoC block.



buffer	1. A storage area for data that is used to compensate for a speed difference, when transferring data from one device to another. Usually refers to an area reserved for IO operations, into which data is read, or from which data is written.
	<ol> <li>A portion of memory set aside to store data, often before it is sent to an external device or as it is received from an external device.</li> </ol>
	3. An amplifier used to lower the output impedance of a system.
bus	1. A named connection of nets. Bundling nets together in a bus makes it easier to route nets with similar routing patterns.
	2. A set of signals performing a common function and carrying similar data. Typically represented using vector notation; for example, address[7:0].
	3. One or more conductors that serve as a common connection for a group of related devices.
clock	The device that generates a periodic signal with a fixed frequency and duty cycle. A clock is sometimes used to synchronize different logic blocks.
comparator	An electronic circuit that produces an output voltage or current whenever two input levels simultaneously satisfy predetermined amplitude requirements.
compiler	A program that translates a high level language, such as C, into machine language.
configuration space	In PSoC devices, the register space accessed when the XIO bit, in the CPU_F register, is set to '1'.
crystal oscillator	An oscillator in which the frequency is controlled by a piezoelectric crystal. Typically a piezoelectric crystal is less sensitive to ambient temperature than other circuit components.
cyclic redundancy check (CRC)	A calculation used to detect errors in data communications, typically performed using a linear feedback shift register. Similar calculations may be used for a variety of other purposes such as data compression.
data bus	A bi-directional set of signals used by a computer to convey information from a memory location to the central processing unit and vice versa. More generally, a set of signals used to convey data between digital functions.
debugger	A hardware and software system that allows the user to analyze the operation of the system under development. A debugger usually allows the developer to step through the firmware one step at a time, set break points, and analyze memory.
dead band	A period of time when neither of two or more signals are in their active state or in transition.
digital blocks	The 8-bit logic blocks that can act as a counter, timer, serial receiver, serial transmitter, CRC generator, pseudo-random number generator, or SPI.
digital-to-analog (DAC)	A device that changes a digital signal to an analog signal of corresponding magnitude. The analog-to-digital (ADC) converter performs the reverse operation.
duty cycle	The relationship of a clock period high time to its low time, expressed as a percent.
emulator	Duplicates (provides an emulation of) the functions of one system with a different system, so that the second system appears to behave like the first system.
external reset (XRES)	An active high signal that is driven into the PSoC device. It causes all operation of the CPU and blocks to stop and return to a pre-defined state.



flash An electrically programmable and erasable, non-volatile technology that provides users with the programmability

and data storage of EPROMs, plus in-system erasability. Non-volatile means that the data is retained when power

is off.

Flash block The smallest amount of Flash ROM space that may be programmed at one time and the smallest amount of Flash

space that may be protected. A Flash block holds 64 bytes.

frequency The number of cycles or events per unit of time, for a periodic function.

gain The ratio of output current, voltage, or power to input current, voltage, or power, respectively. Gain is usually

expressed in dB.

I<sup>2</sup>C A two-wire serial computer bus by Philips Semiconductors (now NXP Semiconductors). I2C is an Inter-Integrated

Circuit. It is used to connect low-speed peripherals in an embedded system. The original system was created in the early 1980s as a battery control interface, but it was later used as a simple internal bus system for building control electronics. I2C uses only two bi-directional pins, clock and data, both running at +5V and pulled high with

resistors. The bus operates at 100 kbits/second in standard mode and 400 kbits/second in fast mode.

ICE The in-circuit emulator that allows users to test the project in a hardware environment, while viewing the debugging

device activity in a software environment (PSoC Designer).

input/output (I/O) A device that introduces data into or extracts data from a system.

interrupt A suspension of a process, such as the execution of a computer program, caused by an event external to that

process, and performed in such a way that the process can be resumed.

interrupt service routine (ISR)

A block of code that normal code execution is diverted to when the M8C receives a hardware interrupt. Many interrupt sources may each exist with its own priority and individual ISR code block. Each ISR code block ends with the RETI instruction, returning the device to the point in the program where it left normal program execution.

jitter 1. A r

- 1. A misplacement of the timing of a transition from its ideal position. A typical form of corruption that occurs on serial data streams.
- The abrupt and unwanted variations of one or more signal characteristics, such as the interval between successive pulses, the amplitude of successive cycles, or the frequency or phase of successive cycles.

low-voltage detect A circuit that senses  $V_{DD}$  and provides an interrupt to the system when  $V_{DD}$  falls lower than a selected threshold. (LVD)

M8C An 8-bit Harvard-architecture microprocessor. The microprocessor coordinates all activity inside a PSoC by

interfacing to the Flash, SRAM, and register space.

master device A device that controls the timing for data exchanges between two devices. Or when devices are cascaded in

width, the master device is the one that controls the timing for data exchanges between the cascaded devices

and an external interface. The controlled device is called the slave device.

microcontroller An integrated circuit chip that is designed primarily for control systems and products. In addition to a CPU, a

microcontroller typically includes memory, timing circuits, and IO circuitry. The reason for this is to permit the realization of a controller with a minimal quantity of chips, thus achieving maximal possible miniaturization. This in turn, reduces the volume and the cost of the controller. The microcontroller is normally not used for

general-purpose computation as is a microprocessor.

mixed-signal The reference to a circuit containing both analog and digital techniques and components.

modulator A device that imposes a signal on a carrier.



noise 1. A disturbance that affects a signal and that may distort the information carried by the signal.

2. The random variations of one or more characteristics of any entity such as voltage, current, or data.

oscillator A circuit that may be crystal controlled and is used to generate a clock frequency.

parity A technique for testing transmitting data. Typically, a binary digit is added to the data to make the sum of all the

digits of the binary data either always even (even parity) or always odd (odd parity).

phase-locked loop (PLL) An electronic circuit that controls an **oscillator** so that it maintains a constant phase angle relative to a reference signal.

pinouts The pin number assignment: the relation between the logical inputs and outputs of the PSoC device and their

schematic and PCB design (both being computer generated files) and may also involve pin names.

physical counterparts in the printed circuit board (PCB) package. Pinouts involve pin numbers as a link between

port A group of pins, usually eight.

power on reset (POR)

A circuit that forces the PSoC device to reset when the voltage is lower than a pre-set level. This is one type of

hardware reset.

PSoC<sup>®</sup> Cypress Semiconductor's PSoC<sup>®</sup> is a registered trademark and Programmable System-on-Chip™ is a trademark

of Cypress.

PSoC Designer™ The software for Cypress' Programmable System-on-Chip technology.

pulse width modulator (PWM)

An output in the form of duty cycle which varies as a function of the applied measurand

RAM An acronym for random access memory. A data-storage device from which data can be read out and new data

can be written in.

register A storage device with a specific capacity, such as a bit or byte.

reset A means of bringing a system back to a know state. See hardware reset and software reset.

ROM An acronym for read only memory. A data-storage device from which data can be read out, but new data cannot

be written in.

serial 1. Pertaining to a process in which all events occur one after the other.

2. Pertaining to the seguential or consecutive occurrence of two or more related activities in a single device or

channel.

settling time The time it takes for an output signal or value to stabilize after the input has changed from one value to another.

shift register A memory storage device that sequentially shifts a word either left or right to output a stream of serial data.

slave device A device that allows another device to control the timing for data exchanges between two devices. Or when

devices are cascaded in width, the slave device is the one that allows another device to control the timing of data exchanges between the cascaded devices and an external interface. The controlling device is called the master

device.

SRAM An acronym for static random access memory. A memory device allowing users to store and retrieve data at a

high rate of speed. The term static is used because, after a value has been loaded into an SRAM cell, it remains

unchanged until it is explicitly altered or until power is removed from the device.



SROM An acronym for supervisory read only memory. The SROM holds code that is used to boot the device, calibrate

circuitry, and perform Flash operations. The functions of the SROM may be accessed in normal user code,

operating from Flash.

stop bit A signal following a character or block that prepares the receiving device to receive the next character or block.

synchronous 1. A signal whose data is not acknowledged or acted upon until the next active edge of a clock signal.

2. A system whose operation is synchronized by a clock signal.

tri-state A function whose output can adopt three states: 0, 1, and Z (high-impedance). The function does not drive any

value in the Z state and, in many respects, may be considered to be disconnected from the rest of the circuit,

allowing another output to drive the same net.

UART A UART or universal asynchronous receiver-transmitter translates between parallel bits of data and serial bits.

user modules Pre-build, pre-tested hardware/firmware peripheral functions that take care of managing and configuring the lower

level Analog and Digital PSoC Blocks. User Modules also provide high level API (Application Programming

Interface) for the peripheral function.

user space The bank 0 space of the register map. The registers in this bank are more likely to be modified during normal

program execution and not just during initialization. Registers in bank 1 are most likely to be modified only during

the initialization phase of the program.

 $V_{DD}$  A name for a power net meaning "voltage drain." The most positive power supply signal. Usually 5 V or 3.3 V.

V<sub>SS</sub> A name for a power net meaning "voltage source." The most negative power supply signal.

watchdog timer A timer that must be serviced periodically. If it is not serviced, the CPU resets after a specified period of time.



#### **Errata**

This section describes the errata for the enCoRe III CY7C64215 device. The information in this document describes hardware issues associated with Silicon Revision A.

Contact your local Cypress sales representative if you have questions.

#### **Part Numbers Affected**

Part Number	Silicon Revision	
CY7C64215	A	

#### CY7C64215 Qualification Status

Product Status: In Production

## CY7C64215 Errata Summary

This table defines the errata applicability to available enCoRe III CY7C64215 family devices.

Items	Part Number	Silicon Revision	Fix Status
USB interface DP line pulses low when the enCoRe III device wakes from sleep.	CY7C64215	А	No silicon fix planned. Use workaround.
Invalid flash reads may occur if V <sub>DD</sub> is pulled to –0.5 V just before power on.	CY7C64215	А	
PMA Index Register fails to auto-increment with CPU_Clock set to SysClk/1 (24 MHz).	CY7C64215	А	

#### 1. USB interface DP line pulses low when the enCoRe III device wakes from sleep

#### ■ Problem Definition

When the device operates at 4.75 V to 5.25 V and the 3.3-V regulator is enabled, a short low pulse may be created on the DP signal line during device wakeup. The 15- to 20-µs low pulse of the DP line may be interpreted by the host computer as a deattach or the beginning of a wakeup.

## ■ Parameters Affected

The bandgap reference voltage used by the 3.3-V regulator decreases during sleep due to leakage. Upon device wakeup, the bandgap is re-enabled and, after a delay for settling, the 3.3-V regulator is enabled. On some devices the 3.3-V regulator used to generate the USB DP signal may be enabled before the bandgap is fully stabilized. This can cause a low pulse on the regulator output and DP signal line until the bandgap stabilizes. In applications where  $V_{DD}$  is 3.3 V, the regulator is not used and, therefore, the DP low pulse is not generated.

## **■ Trigger Condition**

N/A

#### ■ Scope of Impact

N/A

### ■ Workaround

To prevent the DP signal from pulsing low, keep the bandgap enabled during sleep. The most efficient method is to set the No Buzz bit in the OSC\_CR0 register. The No Buzz bit keeps the bandgap powered and output stable during sleep. Setting the No Buzz bit results in a nominal 100  $\mu$ A increase in sleep current. Leaving the analog reference block enabled during sleep also resolves this issue because it forces the bandgap to remain enabled. The following example shows how to disable the No Buzz bit:

#### **Assembly**

```
M8C_SetBank1
  or reg[OSC_CR0], 0x20
  M8C_SetBank0

C
OSC_CR0 |= 0x20;
```



#### **■ Fix Status**

There is no planned silicon fix; use workaround.

## 2. Invalid flash reads may occur if $V_{DD}$ is pulled to $-0.5\ V$ just before power on

#### ■ Problem Definition

When V<sub>DD</sub> of the device is pulled below ground just before power on, the first read from each 8-KB flash page may be corrupted. This issue does not affect flash page 0 because it is the selected page upon reset.

#### ■ Parameters Affected

When  $V_{DD}$  is pulled below ground before power on, an internal flash reference may deviate from its nominal voltage. The reference deviation tends to result in the first flash read from that page returning 0xFF. During the first read from each page, the reference is reset resulting in all future reads returning the correct value. A short delay of 5  $\mu$ s before the first real read provides time for the reference voltage to stabilize.

#### **■ Trigger Condition**

N/A

#### ■ Scope of Impact

N/A

#### **■** Workaround

To prevent an invalid flash read, a dummy read from each flash page must occur before use of the pages. A delay of 5 µs must occur after the dummy read and before a real read. The dummy reads occur as soon as possible and must be located in flash page 0 before a read from any other flash page. An example for reading a byte of memory from each flash page is listed below. Place it in *boot.tpl* and *boot.asm* immediately after the 'start:' label.

#### **■ Fix Status**

There is no planned silicon fix; use workaround.

#### 3. PMA Index Register fails to auto-increment with CPU\_Clock set to SysClk/1 (24 MHz)

## ■ Problem Definition

When the device operates at 4.75 V to 5.25 V and the CPU\_Clock is set to SysClk/1 (24 MHz), the USB PMA Index Register may fail to increment automatically when used in an OUT endpoint configuration at full-speed. When the application program attempts to use the bReadOutEP() function, the first byte in the PMA buffer is always returned.

#### ■ Parameters Affected

An internal flip-flop hold problem is associated with the index register increment function. All reads of the associated RAM originate from the first byte. The hold problem has no impact on other circuits or functions within the device.

#### **■** Trigger Condition

N/A

## ■ Scope of Impact

N/A



#### **■ WORKAROUND**

To make certain that the index register properly increments, set the CPU\_Clock to SysClk/2 (12 MHz) during the read of the PMA buffer. An example for the clock adjustment method follows:

**PSoC Designer 4.3 User Module workaround**: PSoC Designer Release 4.3 and subsequent releases include a revised full-speed USB User Module with the revised firmware workaround included (see the following example).

24-Mhz read PMA workaround

```
;;
M8C SetBank1
mov A, reg[OSC CR0]
push A
and A, Oxf8 ;clear the clock bits (briefly chg the cpu_clk to 3Mhz)
or A, 0x02 ; will set clk to 12Mhz
mov reg[OSC CR0], A ; clk is now set at 12Mhz
M8C SetBank0
.loop:
   mov A, reg[PMA0 DR] ; Get the data from the PMA space
   mov [X], A ; save it in data array
   inc X; increment the pointer
   dec [USB APITemp+1] ; decrement the counter
   jnz .loop ; wait for count to zero out
;;
;; 24Mhz read PMA workaround (back to previous clock speed)
pop A ; recover previous reg[OSC CR0] value
M8C SetBank1
mov reg[OSC CR0], A ; clk is now set at previous value
M8C SetBank0
;; end 24Mhz read PMA workaround
```

## **■** Fix Status

There is no planned silicon fix; use workaround.



# **Document History Page**

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Rev.	ECN No.	Submission Date	Orig. of Change	Description of Change
**	131325	See ECN	XGR	New data sheet.
*A	385256	See ECN	BHA	Changed status from Advance Information to Preliminary. Added standard data sheet items. Changed Part number from CY7C642xx to CY7C64215.
*B	2547630	08/04/2008	AZIEL / PYRS	Operational voltage range for USB specified under "Full Speed USB (12Mbps)". CMP_GO_EN1 register removed as it has no functionality on Radon. Figure "CPU Frequency" adjusted to show invalid operating region for USB with footnote describing reason. DC electrical characteristic, V <sub>DD</sub> . Note added describing where USB hardwa is non-functional.
*C	2620679	12/12/2008	CMCC / PYRS	Added Package Handling information.  Deleted note regarding link to amkor.com for MLF package dimensions.
*D	2717887	06/11/2009	DPT	Added 56 -Pin Sawn QFN (8 X 8 mm) package diagram and added CY7C64215-56LTXC part information in the Ordering Information table.
*E	2852393	01/15/2010		<ul> <li>Added Table of Contents.</li> <li>Added external clock oscillator option and Industrial Temperature information to the Features, Pin Information, Electrical Specifications, Operating Temperature, DC Electrical Characteristics, AC Electrical Characteristics and Ordering Information sections.</li> <li>Updated DC GPIO, AC Chip, and AC Programming Specifications follows:         <ul> <li>Replaced TRAMP (time) with SRPOWER_UP (slew rate) specification.</li> <li>Added I<sub>OH</sub>, I<sub>OL</sub>, DCILO, F32K_U, TPOWERUP, TERASEALL, TPROGRAM_HOT, and TPROGRAM_COLD specifications.</li> <li>Updated V<sub>DD</sub> ranges on Figure 5 and Table 8.</li> <li>Added notes for VM and VDI on Table 10.</li> <li>Removed TR/TF from Table 20.</li> </ul> </li> <li>Update Ordering Information for: CY7C64215-56LFXCT, CY7C64215-28PVXCT, CY7C64215-56LTXIT Tape and Reel.</li> <li>Updated 28-Pin SSOP and 56-Pin QFN PUNCH and SAWN package diagrams.</li> <li>Updated copyright and Sales, Solutions, and Legal Information URLs.</li> </ul>
*F	2892683	03/15/2010	NJF	Updated Cypress website links. Added T <sub>BAKETEMP</sub> and T <sub>BAKETIME</sub> parameters in Absolute Maximum Rating Updated AC Chip-Level Specifications Removed inactive parts from Ordering Information Updated note in Packaging Information.
*G	3070717	10/25/2010	XUT	Removed reference to CYFISPI in Features. Updated datasheet as per Cypress style guide and new datasheet template
*H	3090908	11/19/10	CSAI	Updated AC Chip-Level Specifications table. Added DC I <sup>2</sup> C Specification.
*	3143408	01/17/11	NJF	Added DC I <sup>2</sup> C Specifications table. Added Tjit_IMO specification, removed existing jitter specifications. Updated Analog reference tables. Updated Units of Measure, Acronyms, Glossary, and References sections. Updated solder reflow specifications. No specific changes were made to AC Digital Block Specifications table an I <sup>2</sup> C Timing Diagram. They were updated for clearer understanding.



# **Document History Page** (continued)

Rev.	ECN No.	Submission Date	Orig. of Change	Description of Change
*J	3995635	05/09/2013	CSAI	Updated Packaging Information: spec 001-12921 – Changed revision from *A to *B. spec 001-53450 – Changed revision from *B to *C. spec 51-85079 – Changed revision from *D to *E.  Added Errata.
*K	4080167	07/29/2013	CSAI	Added Errata footnotes (Note 3, 5).  Updated Electrical Specifications: Updated DC Electrical Characteristics: Updated DC Chip-Level Specifications: Added Note 3 and referred the same note in "Sleep Mode" in description of I <sub>SI</sub> parameter in Table 7. Updated DC POR and LVD Specifications: Added Note 5 and referred the same note in V <sub>PPOR0</sub> , V <sub>PPOR1</sub> , V <sub>PPOR2</sub> parameters in Table 15.  Updated Reference Documents: Removed references of spec 001-17397 and spec 001-14503 as these speciare obsolete.  Updated in new template.
*L	4247931	01/16/2014	CSAI	Updated Packaging Information: spec 001-53450 – Changed revision from *C to *D. Completing Sunset Review.
*M	4481449	08/28/2014	MVTA	Updated Packaging Information: spec 001-12921 – Changed revision from *B to *C.  Updated Ordering Information (Updated part numbers).  Updated in new template.
*N	5730242	05/10/2017	MVTA	Updated Packaging Information: spec 51-85079 - Changed revsion from *E to *F. Updated Cypress Logo and Copyright.



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Document Number: 38-08036 Rev. \*N Revised May 19, 2017 Page 45 of 45

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