



## SY88236L/AL

### 2.5Gbps Burst Mode Laser Driver with Integrated Limiting Amplifier

#### General Description

The SY88236L is a single supply 3.3V integrated burst mode laser driver and post amplifier for A-PON, B-PON, EPON, GE-PON, and G-PON applications with data rates from 155Mbps up to 2.5Gbps. The driver can deliver modulation current up to 85mA, and provides a high compliance voltage that makes it suitable for high-current operation with the laser DC-coupled to it. The post amplifier can detect signals with amplitude as low as 5mV<sub>PP</sub>.

The SY88236AL is a version of the SY88236L without 50Ω termination resistors at the inputs of the driver and the post amplifier. The SY88236AL is to be used specially in SFF modules mounted on ONU mother boards which have preinstalled terminations. Removing post amplifier input terminations will allow for receiver gain control.

All support documentation can be found on Micrel's web site at: [www.micrel.com](http://www.micrel.com).

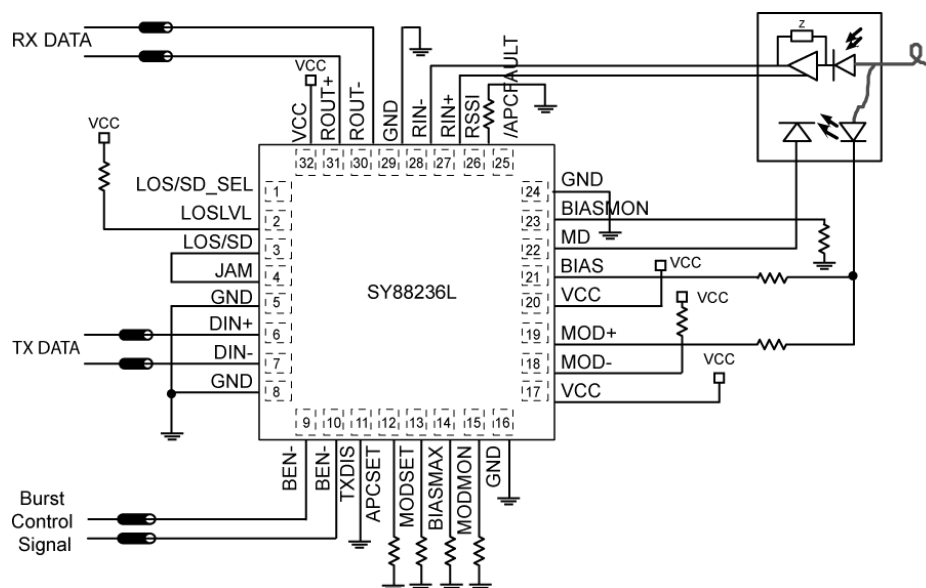
#### Features

- 2.4V minimum laser compliance voltage
- Operation up to 2.5Gbps
- Fast burst mode enable/disable delay
- Modulation current up to 85mA
- Bias current up to 70mA
- Infinite bias current hold time between bursts
- Bias, Modulation, and power monitoring
- High input sensitivity post amplifier, 5mV<sub>PP</sub>
- Programmable LOS level
- Available in 32-pin (5mm x 5mm) QFN package

#### Applications

- Multi-rate burst mode applications: A-PON, B-PON, G-PON, E-PON, GE-PON

#### Typical Application



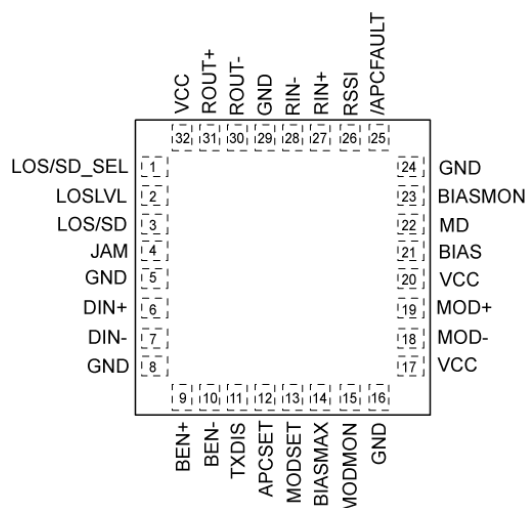
**Ordering Information<sup>(1)</sup>**

Part Number	Voltage	Temperature Range	Package Type	Package Marking	Lead Finish
SY88236LMG	3.3V	−40° to +85°C	QFN-32	SY88236L with Pb-Free bar-line indicator	NiPdAu Pb-Free
SY88236LMGTR <sup>(2)</sup>	3.3V	−40° to +85°C	QFN-32	SY88236L with Pb-Free bar-line indicator	NiPdAu Pb-Free
SY88236ALMG	3.3V	−40° to +85°C	QFN-32	SY88236A with Pb-Free bar-line indicator	NiPdAu Pb-Free
SY88236ALMGTR <sup>(2)</sup>	3.3V	−40° to +85°C	QFN-32	SY88236A with Pb-Free bar-line indicator	NiPdAu Pb-Free

**Notes:**

1. Contact factory for die availability. Dice are guaranteed at T<sub>A</sub> = +25°C, DC Electricals only.
2. Tape and Reel.

## Pin Configuration



32-Pin QFN

## Pin Description

Pin Number	Pin Name	Pin Function
1	LOS/SD_SEL	LOS or SD selection, TTL input. Set high, connect to VCC, or leave open to select LOS. Set low or connect to GND to select SD.
2	LOSLVL	Loss-of-Signal Level Set. A resistor from this pin to V <sub>CC</sub> sets the threshold for the data input amplitude at which LOS will be asserted.
3	LOS/SD	Loss-of-Signal (LOS selected): asserts high when the data input amplitude falls below the threshold set by LOS <sub>LVL</sub> . Signal Detect (SD selected): asserts low when the data input amplitude falls below the threshold set by LOS <sub>LVL</sub> .
4	JAM	Active low TTL/CMOS. Internally pulled-up with 75kΩ. Connect to GND or apply a low level signal (<0.8 V) to enable the post amp output. Can be shorted to LOS/SD (pin 3) to create a squelch function. The polarity of this input follows the polarity of LOS/SD.
6	DIN+	SY88236L: Driver Non-inverting input data. Internally terminated with 50 Ω to a reference voltage. SY88236AL: Driver Non-inverting input data. No internal termination.
7	DIN-	SY88236L: Driver inverting input data. Internally terminated with 50 Ω to a reference voltage. SY88236AL: Driver inverting input data. No internal termination.
9	BEN+	Non-Inverting burst enable. Accepts any input, single-ended or differential: TTL/CMOS, LVPECL, CML, LVDS, and HSTL. BEN requires an external termination. See Figure 2-a-d.
10	BEN-	Inverting burst enable. Accepts any input, single-ended or differential: TTL/CMOS, LVPECL, CML, LVDS, and HSTL. BEN requires an external termination. See Figure 2-a-d.
11	TXDIS	Internally pulled-up. Pull-down with a 22kΩ or lower resistance or apply a low level signal (<0.8 V) to enable bias and modulation. Keep floating or apply a high level (>2V) to disable bias and modulation.

**Pin Description (continued)**

Pin Number	Pin Name	Pin Function
12	APCSET	Bias current setting and control. The bias current is set by installing an external resistor from this pin to ground or using a current source. Connect a 50k $\Omega$ resistor to GND for open loop operation.
13	MODSET	Modulation current setting and control. The modulation current is set by installing an external resistor from this pin to ground or using a current source.
14	BIASMAX	Install a resistor between this pin and GND to set the maximum bias current for the closed loop operation. The APC loop controls the bias current up to the level of BIASMAX. When the bias current reaches the maximum value set through this pin, the driver continues to sink a current equal to this maximum. For open loop operations, this pin sets the bias current.
15	MODMON	Modulation Current Monitor. Provides a current, which represents 1/100 of the modulation current. Install a resistor between this pin and GND to convert that current to a voltage proportional to the modulation current.
18	MOD-	Inverted modulation current output. Provides modulation current when input data is negative.
19	MOD+	Non-inverted modulation current output. Provides modulation current when input data is positive.
21	BIAS	Bias current output, sources current when BEN+ is high. Connect to the cathode of the laser through a resistor.
22	MD	Input from the laser monitoring photodiode. Connect to the anode of the laser photodiode for APC operation.
23	BIASMON	Bias Monitor. Provides a current, which represents 1/50 of the bias current. Install a resistor between this pin and GND to convert that current to a voltage.
24	RSSI	Received Signal Strength Indicator. Install a resistor from this pin to GND to get a voltage proportional to the received signal.
25	/APCFAULT	Indicates APC failure when Low. Active Low TTL/CMOS.
27	RIN+	<b>SY88236L:</b> Post amplifier Non-inverting input data. Internally terminated with 50 $\Omega$ to a reference voltage.
		<b>SY88236AL:</b> Post amplifier Non-inverting input data. No internal termination.
28	RIN-	<b>SY88236L:</b> Driver inverting input data. Internally terminated with 50 $\Omega$ to a reference voltage.
		<b>SY88236AL:</b> Driver inverting input data. No internal termination.
30	ROUT-	Post Amplifier Complementary CML data output.
31	ROUT+	Post Amplifier true CML data output.
5, 8, 16, 24, 29	GND	Ground. Ground and exposed pad must be connected to the plane of the most negative potential.
17, 20, 32		Supply Voltage. Bypass with a 0.1 $\mu$ F//0.01 $\mu$ F low ESR capacitor as close to VCC pin as possible.

## Truth Tables

DIN+	DIN-	TXDIS	MOD+ <sup>(2)</sup>	MOD-	Laser Output Power <sup>(3)</sup>
L	H	L	H	L	L
H	L	L	L	H	H
X	X	X	H	L	L

**Table 1. Modulation Output Truth Table<sup>(1, 3)</sup>**

LOS/SD_SEL	Function Selected	JAM	Output
H	LOS	L	Enabled
H	LOS	H	Disabled
L	SD	L	Disabled
L	SD	H	Enabled

**Table 3. Post Amp Output Truth Table**

TXDIS	BEN+	BEN-	BIAS
L	H	L	ON
L	L	H	OFF
H	X	X	OFF

**Table 2. BIAS Output Truth Table**

### Notes:

1. Assuming BEN+ = H and BEN- = L.
2.  $I_{MOD} = 0$  when MOD+ = H.
3. Assuming that the cathode of the laser is connected to MOD+.

**Absolute Maximum Ratings<sup>(1)</sup>**

Supply Voltage ( $V_{IN}$ )	..... -0.5V to +4.0V
CML Input Voltage ( $V_{IN}$ )	..... $V_{CC}-1.2V$ to $V_{CC}+0.5V$
TTL Control Input Voltage ( $V_{IN}$ )	..... 0V to $V_{CC}$
Lead Temperature (soldering, 20sec.)	..... +260°C
Storage Temperature ( $T_s$ )	..... -65°C to +150°C

**Operating Ratings<sup>(2)</sup>**

Supply Voltage ( $V_{CC}$ )	..... +3.0V to +3.6V
Ambient Temperature ( $T_A$ )	..... -40°C to +85°C
Package Thermal Resistance <sup>(3)</sup>	
QFN	
( $\theta_{JA}$ ) Still-air	..... 60°C/W
( $\psi_{JB}$ )	..... 33°C/W

**DC Electrical Characteristics**

$T_A = -40^\circ\text{C}$  to  $+85^\circ\text{C}$  and  $V_{CC} = +3.0\text{V}$  to  $+3.6\text{V}$ , unless otherwise noted. Typical values are  $V_{CC} = +3.3\text{V}$ ,  $T_A = 25^\circ\text{C}$ ,  $I_{MOD} = 30\text{mA}$ ,  $I_{BIAS} = 30\text{mA}$ .

Symbol	Parameter	Condition	Min	Typ	Max	Units
$I_{CC}$	Power Supply Current	Modulation and Bias currents excluded		90	150 <sup>(4)</sup>	mA
$V_{IL}$	TXDIS, JAM, and LOS/SD_SEL Input Low		-0.3		0.8	V
$V_{IH}$	TXDIS, JAM, and LOS/SD_SEL Input High		2		$V_{CC} + 0.3$	V

**Laser Driver**

$V_{MOD\_MIN}$	Minimum Voltage Required at the Driver Output, MOD+ and MOD-, for Proper Operation		0.6			V
$V_{BIAS\_MIN}$	Minimum Voltage Required at the Driver Output, BIAS pin, for Proper Operation		0.8			V
$I_{BIAS}$	Bias-ON Current	Voltage at Bias pin $\geq 0.8\text{V}$	1		70	mA
$I_{BIAS\_OFF}$	Bias-OFF Current	Current at BIAS pin when TXDIS is high or BEN is low			150	$\mu\text{A}$
$R_{IN}$ (SY88236L only)	Input Resistance at DIN+ and DIN-	Single ended	42.5	50	57.5	$\Omega$
BEN+, BEN-	Burst Mode Enable Signal	Single ended		0.8		V
$V_{IH}(\text{BEN})$	High Voltage	BEN+, BEN-	2		$V_{CC} + 0.3$	V
$V_{IL}(\text{BEN})$	Low Voltage	BEN+, BEN-	-0.3		0.8	V
$V_{OL}$	/APCFAULT Output Low	$I_{OL} = 2\text{mA}$			0.5	V
$I_{OH}$	/APCFAULT Output Leakage	$V_{OH} = V_{CC}$			100	$\mu\text{A}$
$I_{MD}$	Current range at MD pin		50		1500	$\mu\text{A}$

**Notes:**

1. Permanent device damage may occur if absolute maximum ratings are exceeded. This is a stress rating only and functional operation is not implied at conditions other than those detailed in the operational sections of this data sheet. Exposure to absolute maximum ratings conditions for extended periods may affect device reliability.
2. The data sheet limits are not guaranteed if the device is operated beyond the operating ratings.
3. Package Thermal Resistance assumes exposed pad is soldered (or equivalent) to the devices most negative potential on the PCB.  $\theta_{JB}$  uses a 4-layer and  $\theta_{JA}$  in still air unless otherwise stated.
4.  $I_{CC} = 150\text{mA}$  for worst-case conditions with  $I_{MOD} = 85\text{mA}$ ,  $I_{BIAS} = 70\text{mA}$ ,  $T_A = +85^\circ\text{C}$ ,  $V_{CC} = 3.6\text{V}$ .

**Post Amplifier**

Symbol	Parameter	Condition	Min	Typ	Max	Units
LOS <sub>LVL</sub>	LOS <sub>LVL</sub> Voltage		$V_{CC} - 1.3$		$V_{CC}$	V
V <sub>OH</sub>	ROUT+, ROUT- HIGH Voltage		$V_{CC} - 0.020$	$V_{CC} - 0.005$	$V_{CC}$	V
V <sub>OL</sub>	ROUT+, ROUT- LOW Voltage		$V_{CC} - 0.475$	$V_{CC} - 0.400$	$V_{CC} - 0.350$	V
V <sub>OFFSET</sub>	Differential Output Offset				$\pm 80$	mV
Z <sub>0</sub> (ROUT)	Single-Ended Output Impedance		42.5	50	57.5	$\Omega$
Z <sub>i</sub> (RIN) (SY88236L only)	Single-Ended Input Impedance		42.5	50	57.5	$\Omega$
V <sub>OL</sub> (LOS/SD)	LOS/SD Output Low	I <sub>OL</sub> = 2mA			0.5	V
I <sub>OH</sub> (LOS/SD)	LOS/SD Output Leakage	V <sub>OH</sub> = V <sub>CC</sub>			100	$\mu$ A

## AC Electrical Characteristics

$T_A = -40^\circ\text{C}$  to  $+85^\circ\text{C}$  and  $V_{CC} = +3.0\text{V}$  to  $+3.6\text{V}$ , unless otherwise noted. Typical values are  $V_{CC} = +3.3\text{V}$ ,  $T_A = 25^\circ\text{C}$ ,  $I_{MOD} = 30\text{mA}$ ,  $I_{BIAS} = 30\text{mA}$ .

Symbol	Parameter	Condition	Min	Typ	Max	Units
<b>Laser Driver</b>						
	Data Rate	NRZ	0.155		2.5	Gbps
$V_{DIFF-IN}$ (DIN)	Differential Input Voltage Swing		100		2400	mV <sub>PP</sub>
$V_{IN}$ (BEN+/BEN-)	Input Voltage Swing, Single Ended	BEN+, BEN-	100			mV
$V_{DIFF-IN}$ (BEN+/BEN-)	Differential Input Voltage Swing	BEN+, BEN-	200			mV <sub>PP</sub>
$I_{MOD}$	Modulation Current <sup>(5)</sup>	AC-coupled	10		85	mA
		DC-coupled, Voltage at MOD pin $\geq 0.6\text{V}$	10		70 <sup>(6)</sup>	mA
$I_{MOD\_OFF}$	Modulation OFF Current	Current at MOD+ when TXDIS is high or BEN+ is low			150	$\mu\text{A}$
		Current at MOD- when TXDIS is high or BEN+ is low			150	$\mu\text{A}$
$t_r$	Output Current Rise Time	20% to 80%, $I_{MOD} = 60\text{mA}$		60	85	ps
$t_f$	Output Current Fall Time	20% to 80%, $I_{MOD} = 60\text{mA}$		60	85	ps
Jitter	Total Jitter <sup>(7)</sup>	155Mbps data rate			30	ps <sub>PP</sub>
		622Mbps data rate			30	ps <sub>PP</sub>
		1.25Gbps data rate			30	ps <sub>PP</sub>
		2.5Gbps data rate			30	ps <sub>PP</sub>
$t_{INIT}$	APC Loop Initialization Time	Power up with TXDIS low and BEN+ high			12	$\mu\text{s}$
		TXDIS changes from high to low with power ON and BEN+ high			10	$\mu\text{s}$
		BEN changes from low to high with power ON and TXDIS low			2.5	ns
	Burst Enable Delay <sup>(8, 9)</sup>				2.5	ns
	Burst Disable Delay <sup>(8, 10)</sup>				2	ns
	Burst ON-Time	155Mbps	1.9			$\mu\text{s}$
		622Mbps	720			ns
		1.25Gbps	576			ns
		2.5Gbps	576			ns
	Burst OFF-Time	155Mbps	1.9			$\mu\text{s}$
		622Mbps	720			ns
		1.25Gbps	576			ns
		2.5Gbps	576			ns

### Notes:

- Load =  $15\Omega$ .
- Assuming  $V_{CC} = 3.0\text{V}$ , Laser bandgap voltage =  $1\text{V}$ , laser package inductance =  $1\text{nH}$ , laser equivalent series resistor =  $5\Omega$ , and damping resistor =  $10\Omega$ .
- Total jitter is measured using  $2^7 - 1$  PRBS pattern.
- Measured with a laser equivalent resistive load.
- Burst Enable Delay is measured as the time between the instant when the BEN+ signal going from low to high reaches 50% of its amplitude and the instant at which the modulation current or the bias current (whichever takes longer) reaches 90% of its final value.
- Burst Disable Delay is measured as the time between the instant when the BEN+ signal going from high to low reaches 50% of its amplitude and the instant at which the modulation current or the bias current (whichever takes longer) goes below 10% of its final value.



**Post Amplifier**

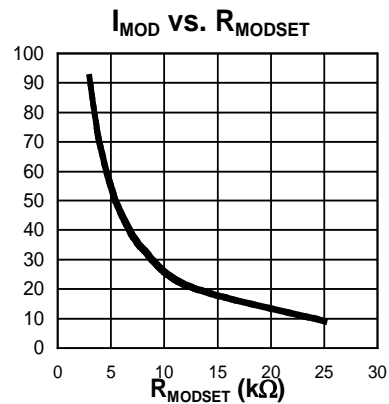
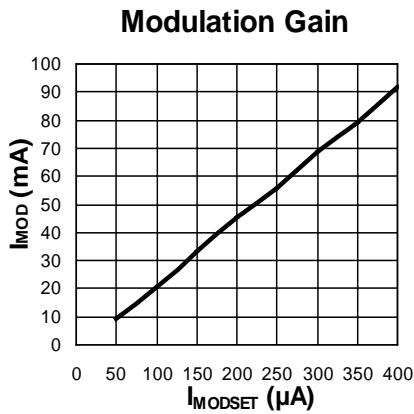
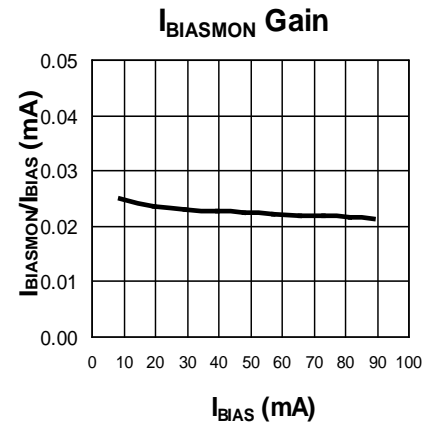
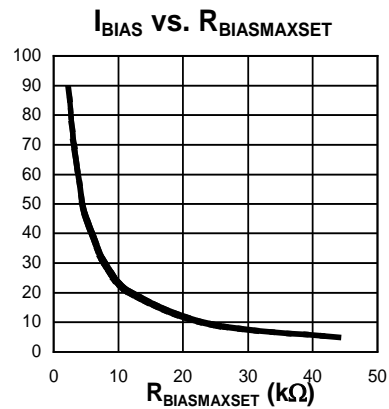
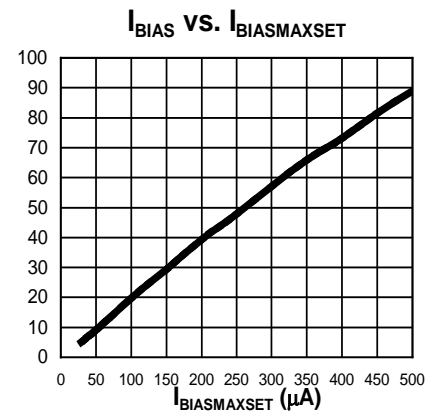
Symbol	Parameter	Condition	Min	Typ	Max	Units
$t_r, t_f$	Output Rise/Fall Time (20% to 80%)	Note 11		60	120	ps
$t_{JITTER}$	Deterministic Random	Note 12 Note 13		15 5		pSP pSRMS
$V_{Diff\_IN}$ (RIN)	Differential Input Voltage Swing		5		1800	mV <sub>PP</sub>
$V_{Diff\_OUT}$ (ROUT)	Differential Output Voltage Swing	Note 11	700	800	950	mV <sub>PP</sub>
$G_{RSSI}$	RSSI Gain = $I_{RSSI} / V_{Diff\_IN}$ (RIN)	$5mV_{PP} \leq V_{Diff\_IN}$ (RIN) $\leq 200mV_{PP}$		5		$\mu A / mV_{PP}$
RSSI Linearity		$5mV_{PP} \leq V_{Diff\_IN}$ (RIN) $\leq 200mV_{PP}$		$\pm 2.5$		%
$LOS_{AL}$	Low LOS Assert Level	$R_{LOSLVL} = 15k\Omega$	2	8		mV <sub>PP</sub>
$LOS_{DL}$	Low LOS De-assert Level	$R_{LOSLVL} = 15k\Omega$		10	20	mV <sub>PP</sub>
$HSY_L$	Low LOS Hysteresis	$R_{LOSLVL} = 15k\Omega$ , Note 14	1.5	2.6	6	dB
$LOS_{AM}$	Medium LOS Assert Level	$R_{LOSLVL} = 5k\Omega$	4	12		mV <sub>PP</sub>
$LOS_{DM}$	Medium LOS De-assert Level	$R_{LOSLVL} = 5k\Omega$		16	30	mV <sub>PP</sub>
$HSY_M$	Medium LOS Hysteresis	$R_{LOSLVL} = 5k\Omega$ , Note 14	1.5	2.8	6	dB
$LOS_{AH}$	High LOS Assert Level	$R_{LOSLVL} = 100\Omega$	15	25		mV <sub>PP</sub>
$LOS_{DH}$	High LOS De-assert Level	$R_{LOSLVL} = 100\Omega$		36	50	mV <sub>PP</sub>
$HSY_H$	High LOS Hysteresis	$R_{LOSLVL} = 100\Omega$	1.5	3.2	6	dB
$T_{OFF}$	LOS Release Time	Note 15		2	10	$\mu s$
$T_{ON}$	LOS Assert Time	Note 15		2	10	$\mu s$
$B_{-3dB}$	3dB Bandwidth			2.0		GHz
$A_{V(Diff)}$	Differential Voltage Gain			38		dB
$S_{21}$	Single-Ended Small-Signal Gain		26	32		dB

**Notes:**

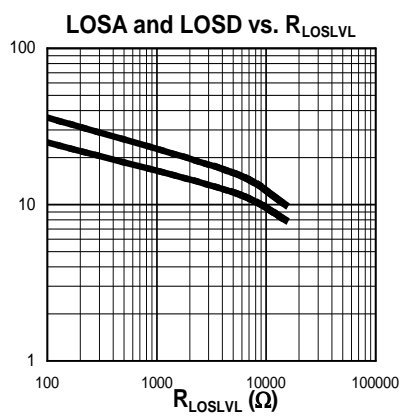
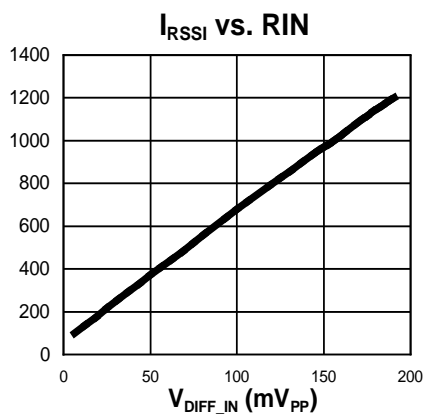
11. Amplifier in limiting mode. Input is a 200MHz square wave.
12. Deterministic jitter measured using 2.5Gbps K28.5 pattern,  $V_{ID} = 10mV_{PP}$ .
13. Random jitter measured using 2.5Gbps K28.7 pattern,  $V_{ID} = 10mV_{PP}$ .
14. This specification defines electrical hysteresis as  $20\log(LOS\ De\text{-}Assert/LOS\ Assert)$ . The ratio between optical hysteresis and electrical hysteresis is found to vary between 1.5 and 2 depending upon the level of received optical power and ROSA characteristics. Based on that ratio, the optical hysteresis corresponding to the electrical hysteresis range 1dB-4.5 dB, shown in the AC characteristics table, will be 0.5dB-3dB Optical Hysteresis.
15. In real world applications, the LOS Release/Assert time can be strongly influenced by the RC time constant of the AC-coupling cap and the 50 $\Omega$  input termination. To keep this time low, use a decoupling cap with the lowest value that is allowed by the data rate and the number of consecutive identical bits in the application (typical values are in the range of 0.001 $\mu F$  to 1.0 $\mu F$ ).

## Typical Characteristics

### Laser Driver



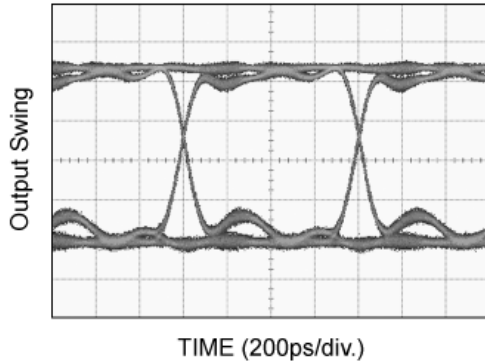
### Post Amplifier



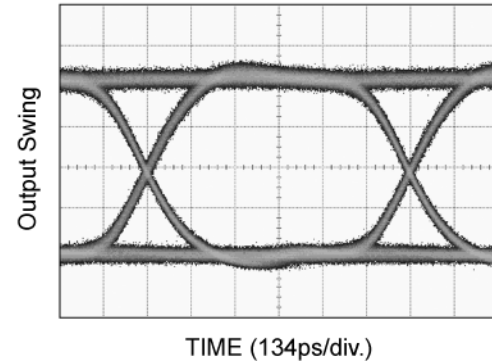
## Functional Characteristics

### Laser Driver

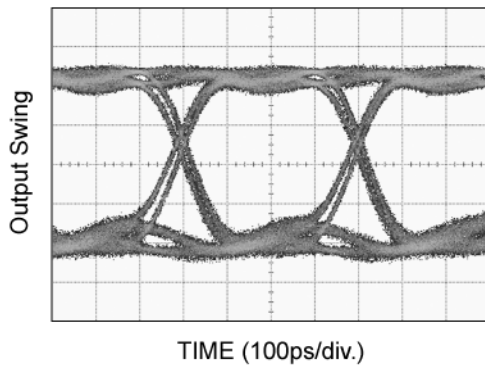
**Electrical Eye Diagram @1.25Gbps**



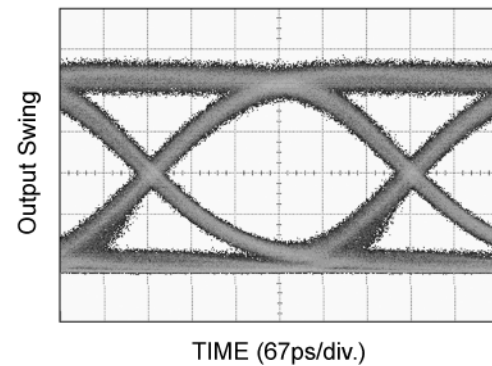
**Optical Eye Diagram @1.25Gbps  
with 2.3G Filter  
Laser DC-Coupled to the Driver**



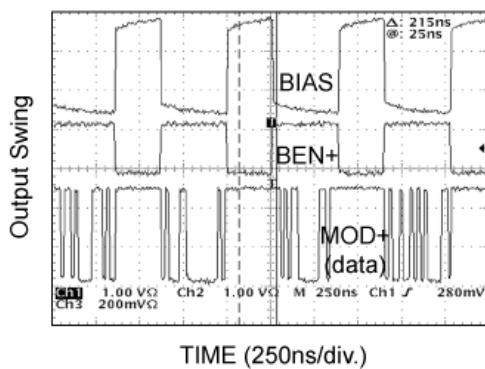
**Electrical Eye Diagram @2.5Gbps**



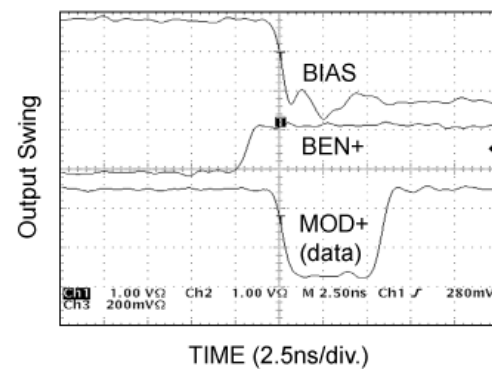
**Optical Eye Diagram @2.5Gbps  
with 2.3G Filter  
Laser DC-Coupled to the Driver**

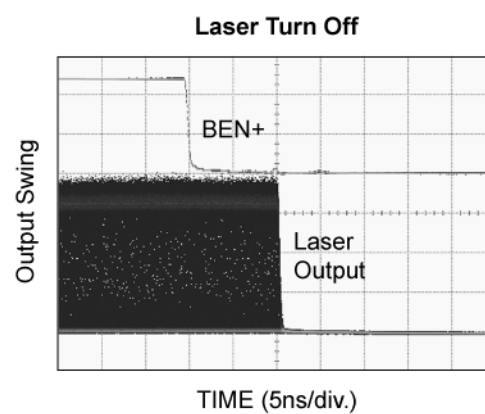
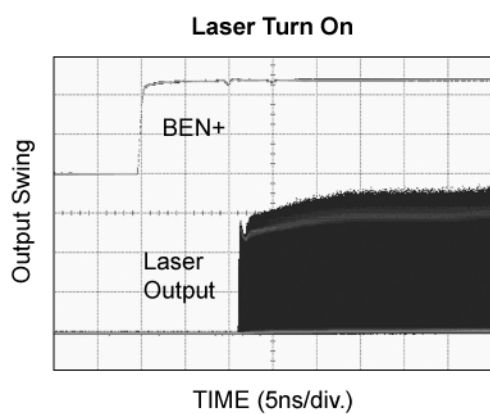
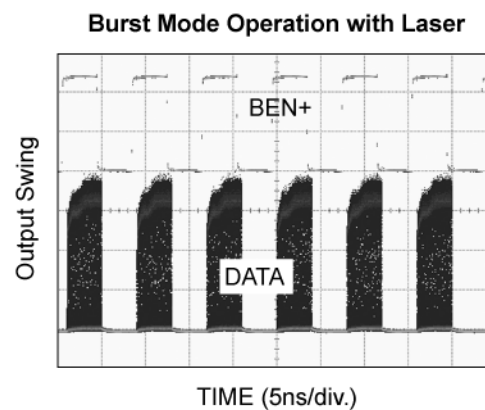
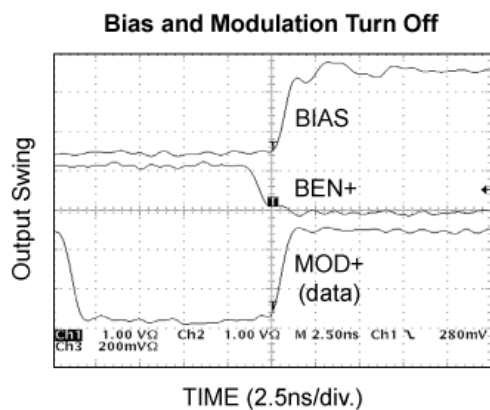


**Driver Burst Mode Operation**



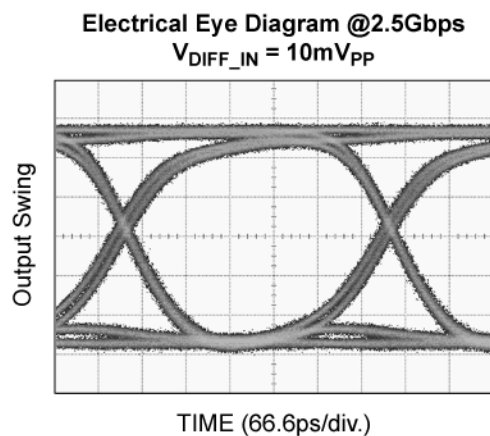
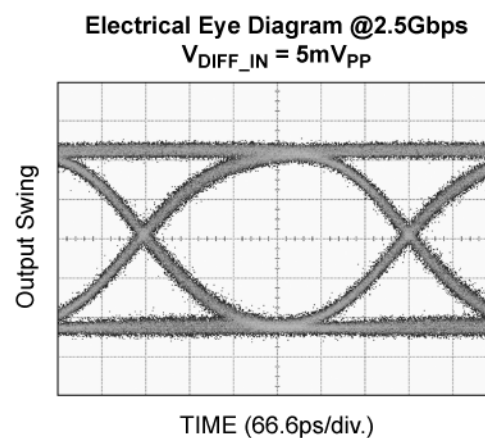
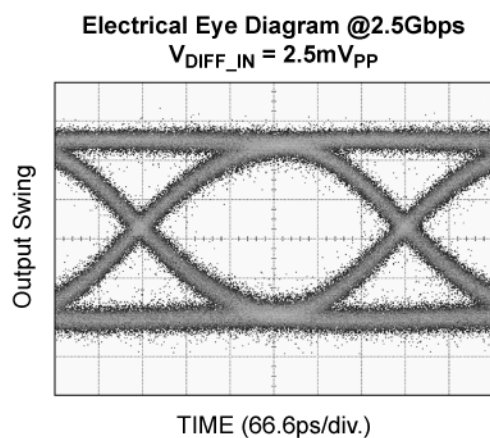
**Bias and Modulation Turn On**



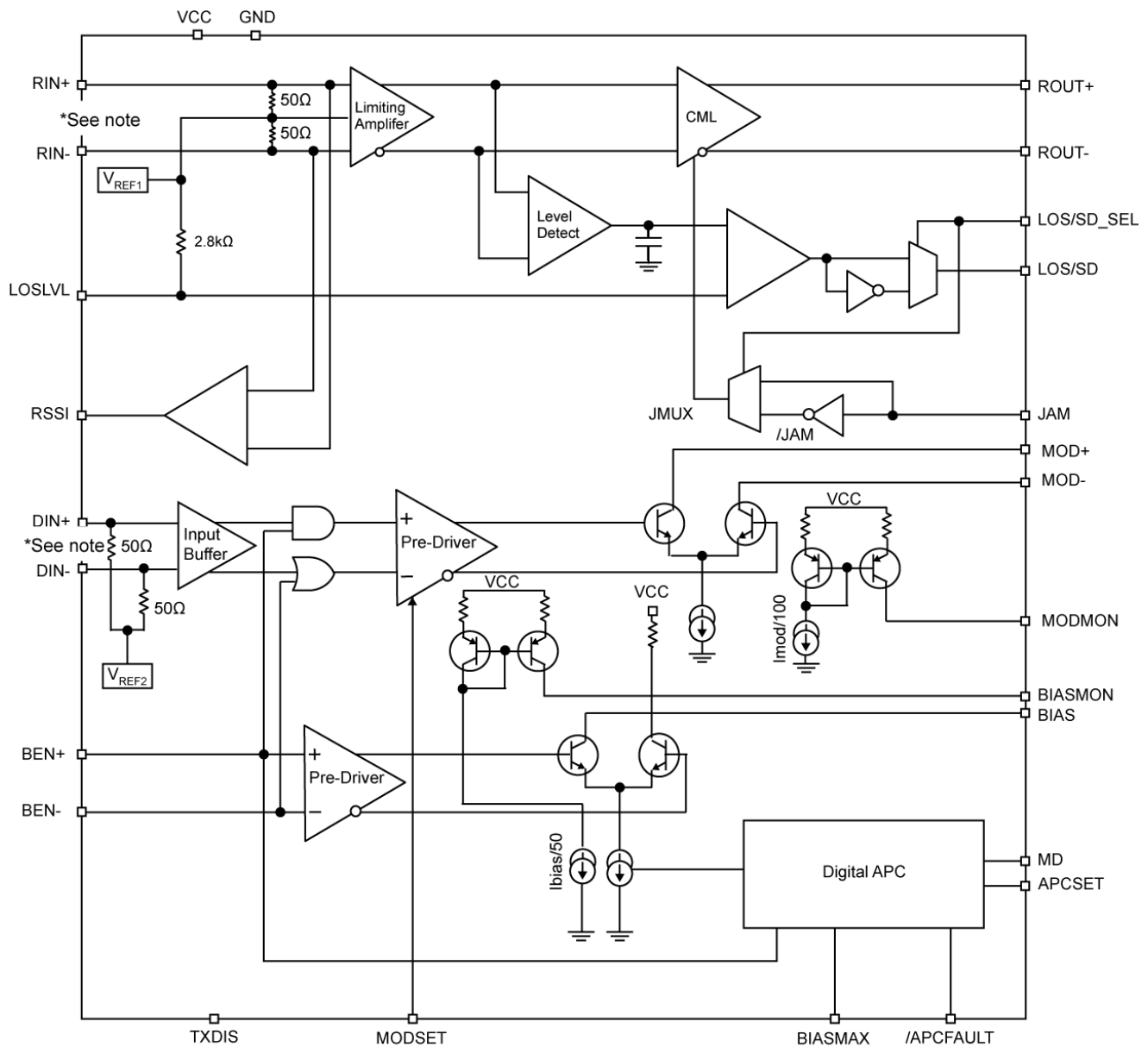


## Functional Characteristics (continued)

### Post Amplifier



## Functional Diagram



\*note: The internal terminations at DIN+, DIN-, RIN+, RIN- apply for SY88236L only.  
SY88236AL doesn't have internal terminations at those inputs

**The SY88236L is shown. For the SY88236AL, 50Ω terminations should be removed from DIN± and RIN±.**

## Functional Description

### Laser Driver

The laser driver is comprised from a modulator, a bias circuit, and a digital APC loop.

The driver features bias and modulation current monitoring functions, which can be configured for optical power monitoring.

#### **BIAS and Modulation Setting**

Bias and modulation currents are set by installing resistors from APCSET to ground and from MODSET to ground respectively or by applying a negative current at those pins.  $I_{BIAS}$  variation versus  $R_{BIASMAXSET}$  resistor and  $I_{BIASMAXSET}$ , and  $I_{MOD}$  variation versus  $R_{MODSET}$  resistor and  $I_{MODSET}$  are shown on page 10.

#### **BIASMAX**

A resistor between BIASMAX pin and ground sets the maximum bias the driver can sink. At normal operation, the bias current tracks the laser optical power through the laser monitoring photodiode and the APC loop to compensate for any power deviation from the nominal value set at the start of operation using APCSET. If for any failure (laser or photodiode degradation, open feedback circuit, etc.) the APC loop keeps increasing the bias current to compensate for the low power indication, the bias current will stop increasing when it reaches BIASMAX value and continues to operate at that maximum value and APCFAULT is asserted.

BIASMAX also sets the bias current when the circuit is operating in the open loop mode.

#### **APC Loop Function**

At start up, with the driver enabled, TXDIS low and BEN+ high, the laser turns ON within a few microseconds and its back facet monitoring photodiode starts to generate a photocurrent proportional to the optical power. The photocurrent is fed back to the MD pin on the driver where it's converted to a voltage. The conversion voltage is compared to APCSET on the driver. At equilibrium, the feedback voltage equals the APCSET voltage and the laser optical power reaches its nominal value. If the laser power deviates from its nominal value, the APC loop brings it back to its nominal setting.

#### **APC Loop Failure**

The APCFAULT is asserted Low if the bias current reaches BIASMAX or if the APC loop counter reaches its minimum or its maximum counts.

#### **Interfacing the Driver with the Laser Diode**

As shown on the "Typical Application" drawing, MOD+ pin is connected to the laser cathode through a 10

resistor and MOD- pin is connected to VCC with a 10 resistor equivalent to 10 (damping resistor) in series with the laser (equivalent resistor of 9). The laser can be driven differentially by connecting MOD- to the anode of the laser through 15 (15Ω pull -up removed) and isolating the anode of the laser from VCC with an inductor.

### Post Amplifier

The post amplifier detects and amplifies signals with data rates from DC up to 3.2Gbps, and amplitude as small as 5mV<sub>PP</sub>. To reduce the noise at the output of the post amplifier when the input signal is absent or lower than the minimum detectable level set by LOS<sub>LVL</sub>, a JAM pin is provided, which can be connected to LOS/SD output to turn off the output buffer when LOSS/SD is asserted.

#### **Input Amplifier/Buffer**

Figure 1-d shows a simplified schematic of the input stage. The high-sensitivity of the input amplifier allows signals as small as 5mV<sub>PP</sub> to be detected and amplified. The input amplifier allows input signals as large as 1800mV<sub>PP</sub>. Small input signals below typically 12mV<sub>PP</sub> are linearly amplified with a typically 38dB differential voltage gain. For input signals larger than 12mV<sub>PP</sub>, the output signal is limited to typically 800mV<sub>PP</sub>.

#### **Output Buffer**

The post amplifier CML output buffer is designed to drive 50Ω lines and is internally terminated with 50Ω to V<sub>CC</sub>. Figure 1e shows a simplified schematic of the output stage.

#### **Loss-of-Signal**

The post amplifier generates a selectable chatter-free loss-of-signal (LOS) or signal detect (SD) open-collector TTL output as shown in Figure 2g. LOS/SD is used to determine that the input amplitude is too small to be considered as a valid input. When the LOSS function is selected (LOS/SD\_SEL=1), LOS/SD asserts high if the input amplitude falls below the threshold set by LOSLVL and de-asserts low otherwise. If SD function is selected (LOS/SD\_SEL=0), LOS/SD asserts low if the input amplitude falls below the threshold set by LOSLVL and de-asserts high otherwise. LOS/SD can be fed back to the JAM input to maintain output stability under a loss of signal condition. Jam de-asserts low the true output signal without removing the input signals. Typically, 3dB LOS hysteresis is provided to prevent chattering.

#### **Loss/Signal Detect Selection**

A pin (LOS/SD\_SEL) is provided to select between LOS (set to high) or SD (set to low) function. It also controls the internal circuitry of JAM input to follow LOS/SD selection.

**Loss-of-Signal-Level Set**

A programmable LOS/SD level set pin ( $LOS_{LVL}$ ) sets the threshold of the input amplitude detection. Connecting an external resistor between  $V_{CC}$  and  $LOS_{LVL}$  sets the voltage at  $LOS_{LVL}$ . This voltage ranges from  $V_{CC}$  to  $V_{CC} - 1.3V$ . The external resistor creates a voltage divider between  $V_{CC}$  and  $V_{CC} - 1.3V$ , as shown in Figure 2f.

**Hysteresis**

The post amplifier provides typically 3dB LOS electrical hysteresis, which is defined as  $20\log (VIN_{LOS-Assert} / VIN_{LOS-De-Assert})$ . Since the relationship between the voltage out of the ROSA to optical power at its input is linear, the optical hysteresis will be typically half of the electrical hysteresis reported in the datasheet, but in practice, the ratio between electrical and optical hysteresis is found to be within the range 1.5 to 1.8. Thus, 3dB electrical hysteresis will correspond to an optical hysteresis within the range 1.6dB to 2dB.

**RSSI Pin**

The post amplifier has an RSSI (Received Signal Strength) pin, which provides a current proportional to the amplitude of the signal at the input of the post amplifier from the ROSA. Install a resistor between this pin and GND to convert the current into a monitoring voltage proportional to the amplitude of the signal at the input of the post amplifier. The value of the resistor should be selected to keep the voltage at the RSSI pin under its limits of 1.2V to maintain RSSI linearity.



## Input and Output Stages (SY88236L)<sup>(1)</sup>

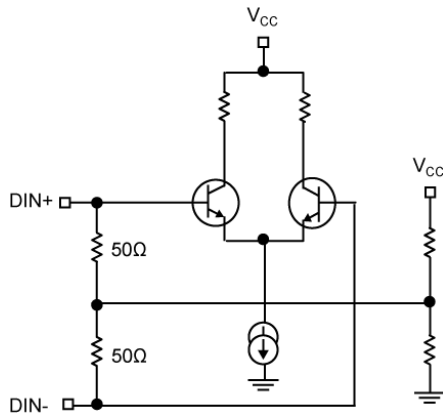


Figure 1a. Simplified Driver Input Stage<sup>(1)</sup>

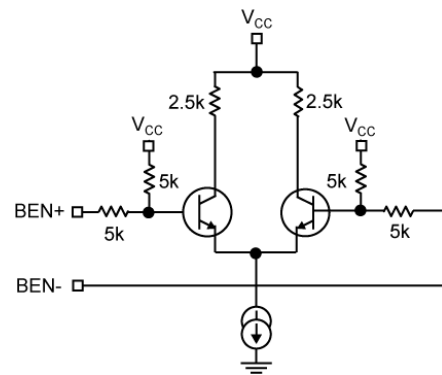


Figure 1b. Simplified BEN Input Stage

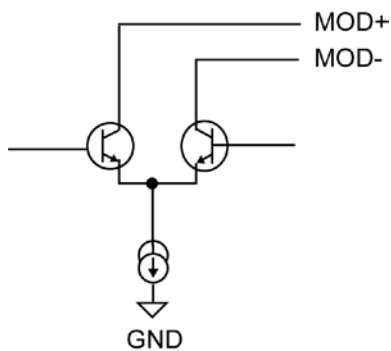


Figure 1c. Simplified Driver Output Stage

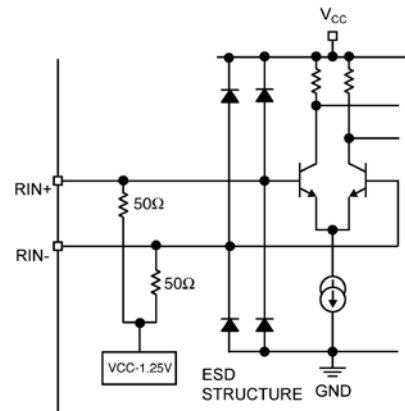


Figure 1d. Post Amplifier Input Stage<sup>(1)</sup>

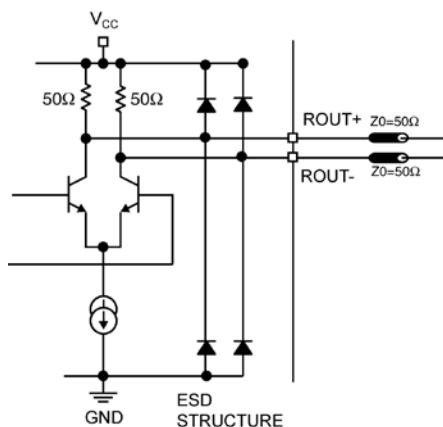


Figure 1e. Post Amplifier Output Stage

### Note:

1. Applies for SY88236L only. For SY88236AL input terminations need to be removed.

## Interfacing DIN and BEN Inputs to Different Logic Drivers (SY88236L)<sup>(1)</sup>

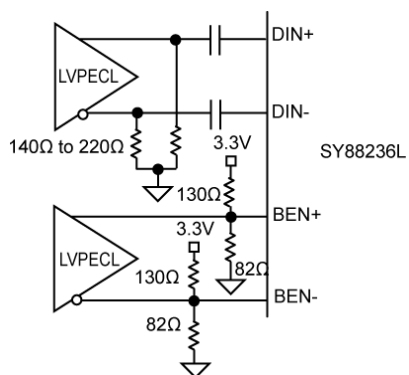


Figure 2a. Driving DIN and BEN with PECL Outputs<sup>(1)</sup>

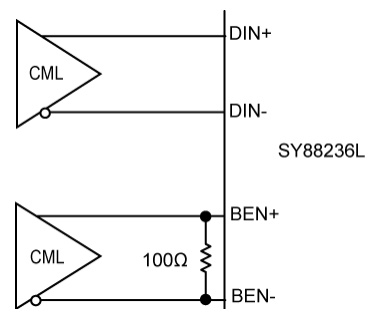


Figure 2b. Driving DIN and BEN with CML Outputs<sup>(1)</sup>

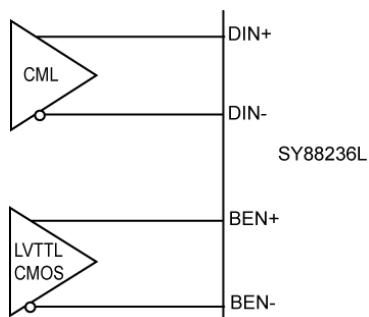


Figure 2c. Driving BEN with LVTTTL/CMOS Outputs<sup>(1)</sup>

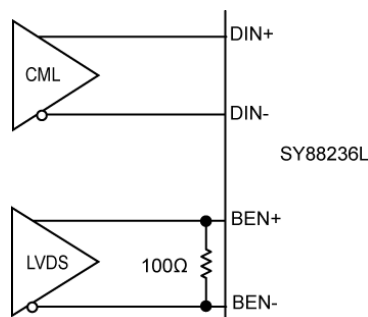


Figure 2d. Driving BEN with LVDS Outputs<sup>(1)</sup>

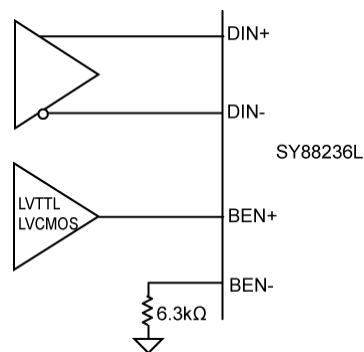


Figure 2e. Driving BEN with Single-Ended LVTTTL/CMOS

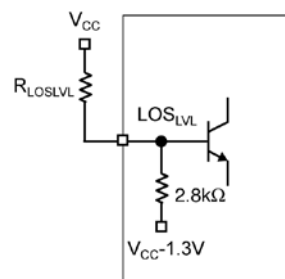


Figure 2f. LOS<sub>LVL</sub> Setting Circuit

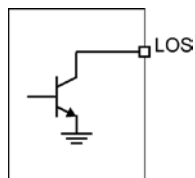
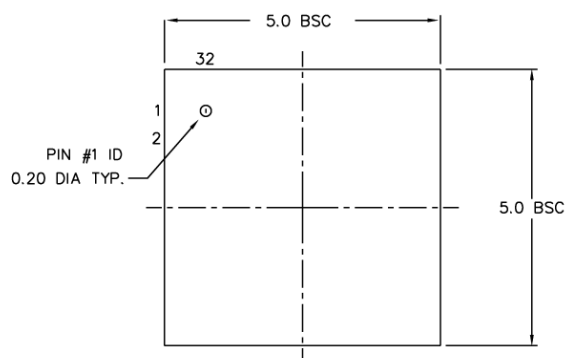


Figure 2g. LOS Output Structure

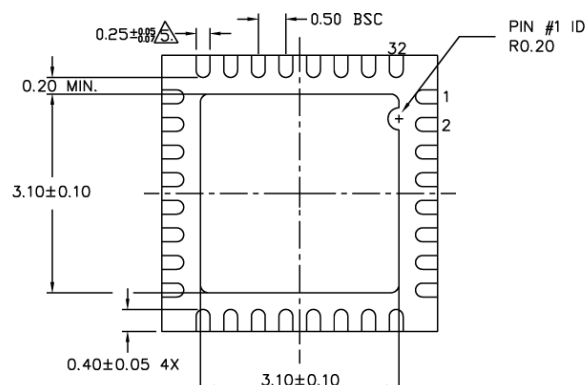
### Note:

1. Applies for SY88236L only. For SY88236AL input terminations need to be added.

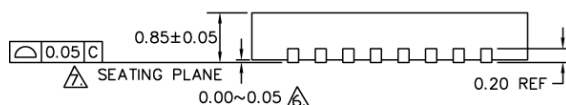
## Package Information



TOP VIEW



BOTTOM VIEW



SIDE VIEW

### NOTE:

1. ALL DIMENSIONS ARE IN MILLIMETERS.
2. MAX. PACKAGE WARPAGE IS 0.05 mm.
3. MAXIMUM ALLOWABLE BURRS IS 0.076 mm IN ALL DIRECTIONS.
4. PIN #1 ID ON TOP WILL BE LASER/INK MARKED.
5. DIMENSION APPLIES TO METALIZED TERMINAL AND IS MEASURED BETWEEN 0.20 AND 0.25 mm FROM TERMINAL TIP.
6. APPLIED ONLY FOR TERMINALS.
7. APPLIED FOR EXPOSED PAD AND TERMINALS.

### 32-Pin (5mm x 5mm) QFN

**MICREL, INC. 2180 FORTUNE DRIVE SAN JOSE, CA 95131 USA**  
 TEL +1 (408) 944-0800 FAX +1 (408) 474-1000 WEB <http://www.micrel.com>

The information furnished by Micrel in this data sheet is believed to be accurate and reliable. However, no responsibility is assumed by Micrel for its use. Micrel reserves the right to change circuitry and specifications at any time without notification to the customer.

Micrel Products are not designed or authorized for use as components in life support appliances, devices or systems where malfunction of a product can reasonably be expected to result in personal injury. Life support devices or systems are devices or systems that (a) are intended for surgical implant into the body or (b) support or sustain life, and whose failure to perform can be reasonably expected to result in a significant injury to the user. A Purchaser's use or sale of Micrel Products for use in life support appliances, devices or systems is a Purchaser's own risk and Purchaser agrees to fully indemnify Micrel for any damages resulting from such use or sale.

© 2009 Micrel, Incorporated.

## Данный компонент на территории Российской Федерации

**Вы можете приобрести в компании MosChip.**

Для оперативного оформления запроса Вам необходимо перейти по данной ссылке:

<http://moschip.ru/get-element>

Вы можете разместить у нас заказ для любого Вашего проекта, будь то серийное производство или разработка единичного прибора.

В нашем ассортименте представлены ведущие мировые производители активных и пассивных электронных компонентов.

Нашей специализацией является поставка электронной компонентной базы двойного назначения, продукции таких производителей как XILINX, Intel (ex.ALTERA), Vicor, Microchip, Texas Instruments, Analog Devices, Mini-Circuits, Amphenol, Glenair.

Сотрудничество с глобальными дистрибьюторами электронных компонентов, предоставляет возможность заказывать и получать с международных складов практически любой перечень компонентов в оптимальные для Вас сроки.

На всех этапах разработки и производства наши партнеры могут получить квалифицированную поддержку опытных инженеров.

Система менеджмента качества компании отвечает требованиям в соответствии с ГОСТ Р ИСО 9001, ГОСТ РВ 0015-002 и ЭС РД 009

### Офис по работе с юридическими лицами:

105318, г.Москва, ул.Щербаковская д.3, офис 1107, 1118, ДЦ «Щербаковский»

Телефон: +7 495 668-12-70 (многоканальный)

Факс: +7 495 668-12-70 (доб.304)

E-mail: [info@moschip.ru](mailto:info@moschip.ru)

Skype отдела продаж:

moschip.ru

moschip.ru\_4

moschip.ru\_6

moschip.ru\_9