

ZXCT1107Q, ZXCT1109Q, ZXCT1110Q
AUTOMOTIVE GRADE MICROPOWER CURRENT MONITOR
Description

The ZXCT1107Q/09Q/10Q are high side unipolar current sense monitors. These devices eliminate the need to disrupt the ground plane when sensing a load current.

The wide common-mode input voltage range and low quiescent current coupled with SOT23 packages make them suitable for a range of automotive applications.

The device is line powered and as such doesn't need a separate supply rail.

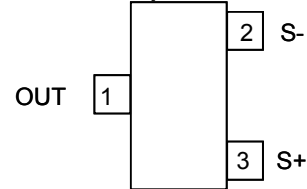
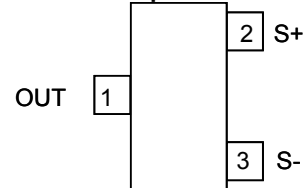
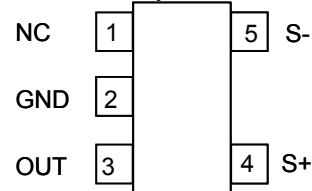
Quiescent current is only 3 μ A thereby minimizing current sensing error. One external gain setting resistor increases versatility by permitting wide gain ranges.

The ZXCT1107Q/09Q/10Q have been qualified to AEC-Q100 Grade 1 and are Automotive Grade supporting PPAPs.

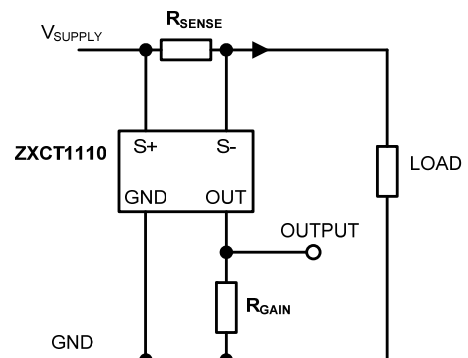
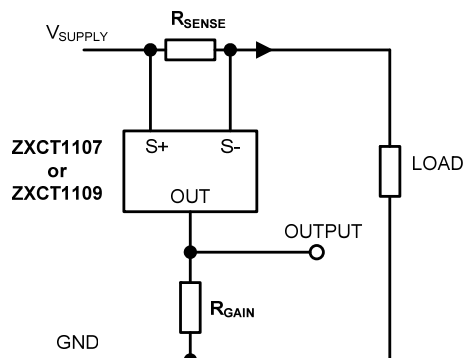
Features

- Wide supply and common-mode voltage range:
 - 2.5V to 36V
- SOT23 packages
 - 3-pin ZXCT1107Q/09Q
 - 5-pin ZXCT1110Q
- Low quiescent current (3 μ A).
- Extended industrial temperature range -40 to +125°C
- Green Molding in SOT23
 - **Totally Lead-Free & Fully RoHS Compliant (Notes 1 & 2)**
 - **Halogen and Antimony Free. "Green" Device (Note 3)**
- Automotive Grade
 - **Qualified to AEC-Q100 Standards for High Reliability**
 - **PPAP Capable (Note 4)**

- Notes:
1. No purposely added lead. Fully EU Directive 2002/95/EC (RoHS) & 2011/65/EU (RoHS 2) compliant.
 2. See http://www.diodes.com/quality/lead_free.html for more information about Diodes Incorporated's definitions of Halogen- and Antimony-free, "Green" and Lead-free.
 3. Halogen- and Antimony-free "Green" products are defined as those which contain <900ppm bromine, <900ppm chlorine (<1500ppm total Br + Cl) and <1000ppm antimony compounds.
 4. Automotive products are AEC-Q100 qualified and are PPAP capable. Automotive, AEC-Q100 and standard products are electrically and thermally the same, except where specified. For more information, please refer to http://www.diodes.com/quality/product_compliance_definitions/.

Pin Assignments
ZXCT1107Q
Top View

ZXCT1109Q
Top View

ZXCT1110Q
Top View

Applications

- Automotive current measurement
- Battery management
- Over current monitor

Typical Application Circuit


ZXCT1107Q, ZXCT1109Q, ZXCT1110Q

Pin Descriptions

Package	SOT23		SOT25	Description
Name	ZXCT1107Q	ZXCT1109Q	ZXCT1110Q	
OUT	1	1	3	Output pin (Note 5). Current output.
S+	3	2	4	This is the positive input of the current monitor and has a wide common-mode input range. This also acts as the analog supply and provides power to internal circuitry. The current through this pin varies with differential sense voltage.
S-	2	3	5	This is the negative input of the current monitor and has a wide common-mode input range.
GND	-	-	2	Ground pin and substrate connection.
NC	-	-	1	No connection

Note: 5. For the ZXCT1107Q and ZXCT1109Q pin 1 (OUT) acts as both the output pin and substrate connection. This means that for the ZXCT1107Q and ZXCT1109Q the minimum rail voltage that can be used is 2.5V + V_{OUT} (see applications section for more details).

Absolute Maximum Ratings

Description		Rating	Unit
Voltage on S- and S+ relative to OUT (ZXCT1107/9)		-0.3 to 40	V
Voltage on S- and S+ relative to GND (ZXCT1110)		-0.3 to 40	V
Voltage on OUT to GND (ZXCT1110)		-0.3 to V _{S+}	V
Differential Sense Voltage, V _{SENSE} (Notes 6 and 7)		-0.3 to +0.8	V
Maximum Current into S+ and S- (Note 7)		±8.5	mA
Storage Temperature		-55 to +150	°C
Maximum Junction Temperature		+150	°C
Package Power Dissipation (T _A = +25°C) SOT23 SOT25		300 (De-rate to Zero at +150°C) 300 (De-rate to Zero at +150°C)	mW
ESD Ratings			
ZXCT1107Q/09Q			
HBM	Human Body Model	2500	V
MM	Machine Model	200	V
CDM	Charged Device Model	1000	V
ZXCT1110Q			
HBM	Human Body Model	1000	V
MM	Machine Model	150	V
CDM	Charged Device Model	TBC	V

Notes: 6. V_{SENSE} = V_{S+} - V_{S-}.

7. The differential input voltage limit, V_{S+} - V_{S-}, may be exceeded provided that the input current limit into S+ or S- is not exceeded.

Caution: Stresses greater than the 'Absolute Maximum Ratings' specified above, may cause permanent damage to the device. These are stress ratings only; functional operation of the device at conditions between maximum recommended operating conditions and absolute maximum ratings is not implied. Device reliability may be affected by exposure to absolute maximum rating conditions for extended periods of time.

Semiconductor devices are ESD sensitive and may be damaged by exposure to ESD events. Suitable ESD precautions should be taken when handling and transporting these devices.)

ZXCT1107Q, ZXCT1109Q, ZXCT1110Q
Recommended Operating Conditions

Symbol	Parameter		Min	Max	Unit
V _{S+}	Common-mode input range (Note 5)	ZXCT1110Q	2.5	36	V
		ZXCT1107Q/9Q	V _{OUT(MAX)} + 2.5	36	
V _{SENSE}	Differential Sense Input voltage range (V _{S+} -V _{S-})		0	0.5	V
V _{OUT}	Output voltage range (0 ≤ V _{SENSE} ≤ 100mV)	ZXCT1110Q	0	V _{S-} - 1	V
		ZXCT1107Q/9Q	0	V _{S+} - 2.5	
T _A	Ambient temperature range		-40	+125	°C

Note: 5. For the ZXCT1107Q and ZXCT1109Q pin 1 (OUT) acts as both the output pin and substrate connection. This means that for the ZXCT1107Q and ZXCT1109Q the minimum rail voltage that can be used is $2.5V + V_{OUT}$ (see applications section for more details).

Electrical Characteristics (@ $T_A = +25^{\circ}C$, $V_{S+} = 20V$, $V_{SENSE} = 100mV$ (Note 6), $R_{GAIN} = 0$, unless otherwise stated)

ZXCT1107Q, ZXCT1109Q

Symbol	Parameter	Conditions	T_A	Min	Typ	Max	Units
I_{S-}	S- input current	$V_{SENSE} = 0V$ (Note 6)	+25°C	—	19	100	nA
			-40°C	—	16	—	
			+125°C	—	35	—	
I_{OUT}	Output current (Note 8)	$V_{SENSE} = 0mV$ (Note 6)	+25°C	1	3	10	μA
			-40°C	—	2.2	—	
			+125°C	—	5.5	—	
		$V_{SENSE} = 10mV$ (Note 6)	+25°C	37	45	54	
			-40°C	—	42	—	
			+125°C	—	49	—	
		$V_{SENSE} = 30mV$ (Note 6)	+25°C	115	124	134	
			-40°C	—	119	—	
			+125°C	—	129	—	
		$V_{SENSE} = 100mV$ (Note 6)	+25°C	394	408	422	
			-40°C	—	396	—	
			+125°C	—	420	—	
		$V_{SENSE} = 200mV$ (Note 6)	+25°C	787	810	832	
			-40°C	—	785	—	
			+125°C	—	832	—	
		$V_{SENSE} = 500mV$ (Note 6)	+25°C	1.965	2.015	2.064	mA
			-40°C	—	1.965	—	
			+125°C	—	2.065	—	
I_{OUT-TC}	Output current temperature coefficient		full range	—	370	—	ppm/°C
CMSR	Common-Mode Sense rejection	$V_{S+} = 2.5V$ to 36V	+25°C	—	0.1	0.4	$\mu A/V$
			-40°C	—	0.13	—	
			+125°C	—	0.05	—	
BW	-3dB Small Signal Bandwidth	$V_{SENSE(AC)} = 10mV_{PP}$ (Note 6) $R_{GAIN} = 2.5k\Omega$	+25°C	—	0.65	—	MHz

Note: 6. $V_{SENSE} = "V_{SENSE+}" - "V_{SENSE-}"$
8. Output current characteristic measured with low impedance ammeter connected to GND

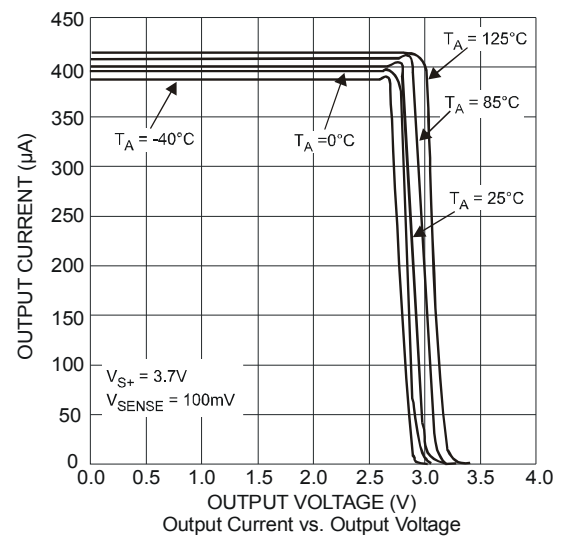
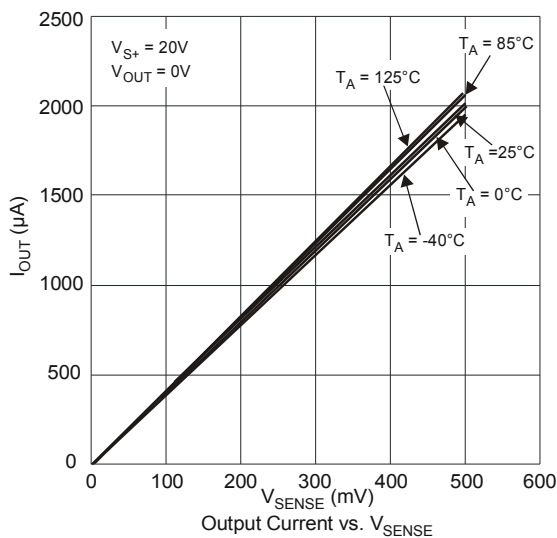
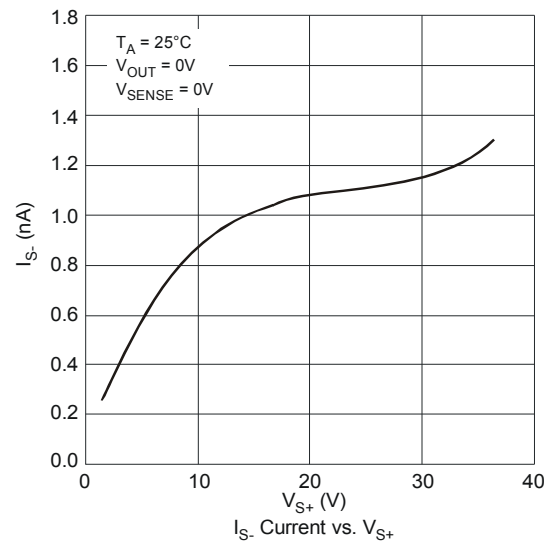
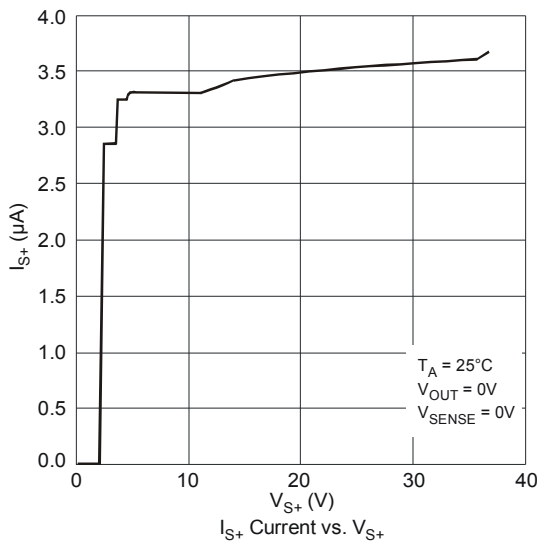
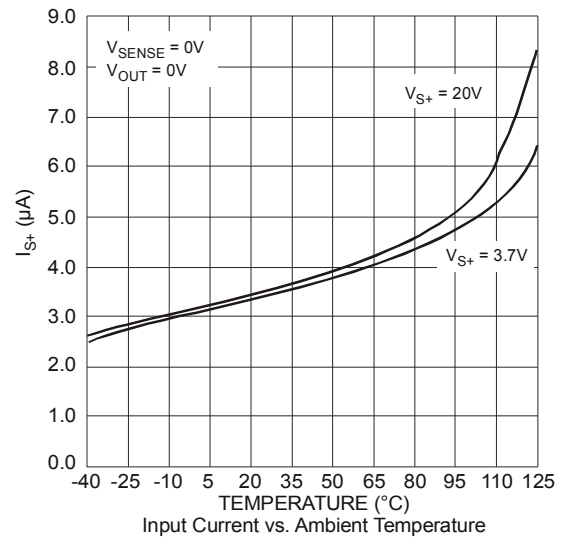
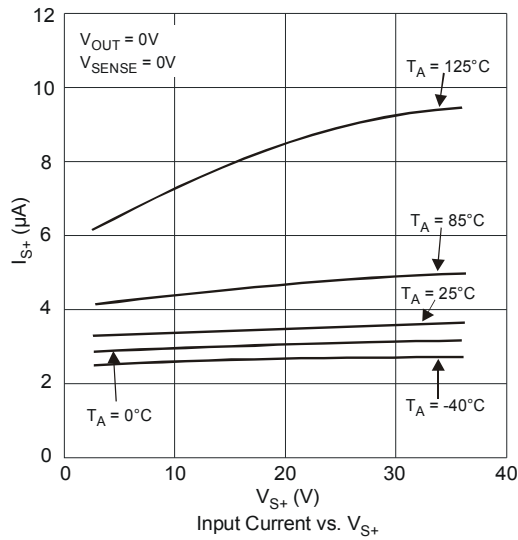
ZXCT1107Q, ZXCT1109Q, ZXCT1110Q
Electrical Characteristics (@ $T_A = +25^\circ\text{C}$, $V_{S+} = 20\text{V}$, $V_{\text{SENSE}} = 100\text{mV}$ (Note 6), $R_{\text{GAIN}} = 0$ unless otherwise stated)

ZXCT1110Q

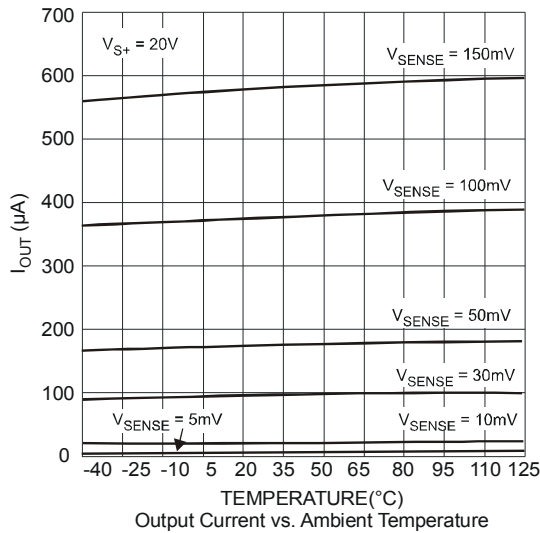
Symbol	Parameter	Conditions	T _A	Min	Typ	Max	Units
I _Q	GND pin current	V _{SENSE} = 0V (Note 6)	+25°C	—	3	5	μA
			-40°C	—	2.2	—	
			+125°C	—	5.8	—	
I _{S-}	S- input current		+25°C	—	19	100	nA
			-40°C	—	16	—	
			+125°C	—	35	—	
CMSR	Common-Mode Sense rejection	V _{S+} = 2.5V to 36V	+25°C	—	0.1	0.4	μA/V
			-40°C	—	0.13	—	
			+125°C	—	0.05	—	
I _{OO}	Output Offset current (Notes 8 and 9)	V _{SENSE} = 10mV (Note 6)	+25°C	—	0	±4	μA
			-40°C	—	-1.8	—	
			+125°C	—	+2.5	—	
G _T	Transconductance	V _{SENSE} = 10mV to 150mV (Note 6)	+25°C	3.928	4	4.072	mA/V
			-40°C	—	3.9	—	
			+125°C	—	4.08	—	
GE	Transconductance error (Note 10)	V _{SENSE} = 10mV to 150mV (Note 6)	+25°C	-1.8	—	1.8	%
I _{OUT} G _{T-TC}	Transconductance temp.co	—	+25°C	—	265	—	ppm/°C
V _{OUTH}	Output relative to V _{S-}	—	+25°C	-1	-0.78	—	V
			-40°C	—	-0.88	—	
			+125°C	—	-0.63	—	
BW	-3dB Small Signal Bandwidth	V _{SENSE} (AC) = 10mV _{PP} , (Note 6) R _{GAIN} = 2.5kΩ	+25°C	—	0.65	—	MHz

- Notes:
- $V_{\text{SENSE}} = "V_{\text{SENSE}+}" - "V_{\text{SENSE}-}"$
 - Output current characteristic measured with low impedance ammeter connected to GND
 - Output Offset Current is defined as difference between actual output current and $40\mu\text{A}$; measured at $V_{\text{SENSE}} = 10\text{mV}$. This will include an error due to bias currents of the device.
 - For $V_{\text{SENSE}} > 10\text{mV}$, the internal voltage-current converter is fully linear. This enables a true offset to be defined and used.

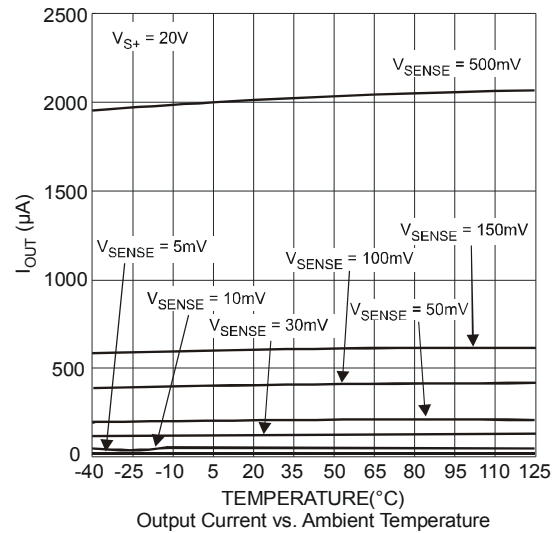
Typical Characteristics



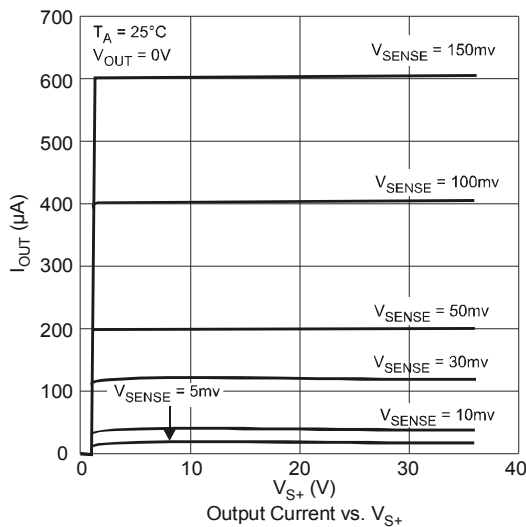
Typical Characteristics (cont.)



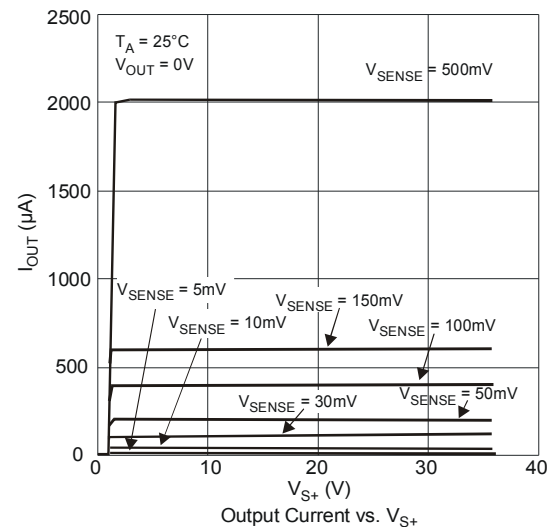
Output Current vs. Ambient Temperature



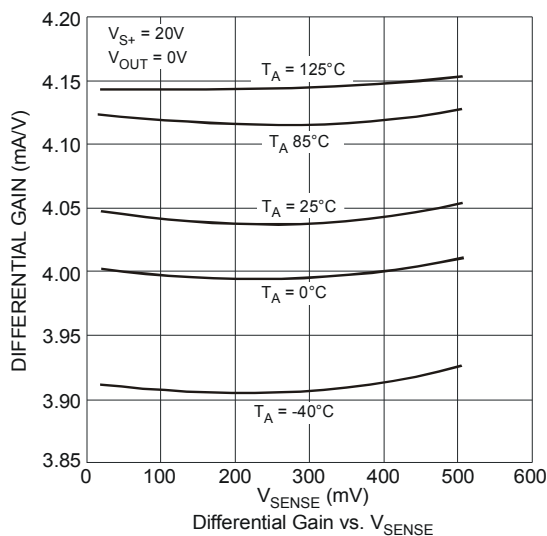
Output Current vs. Ambient Temperature



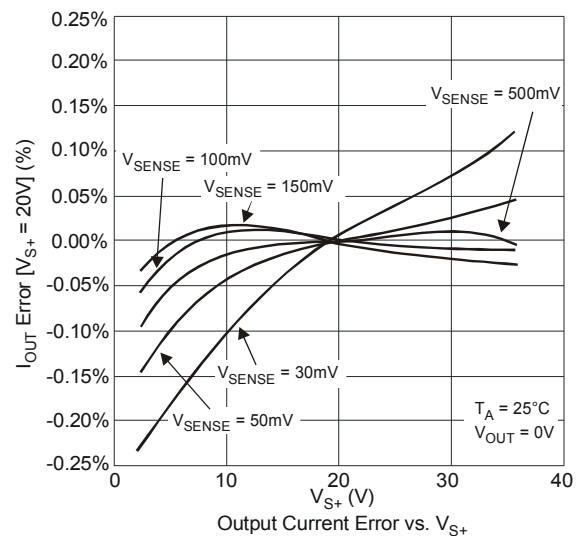
Output Current vs. V_{S+}



Output Current vs. V_{S+}

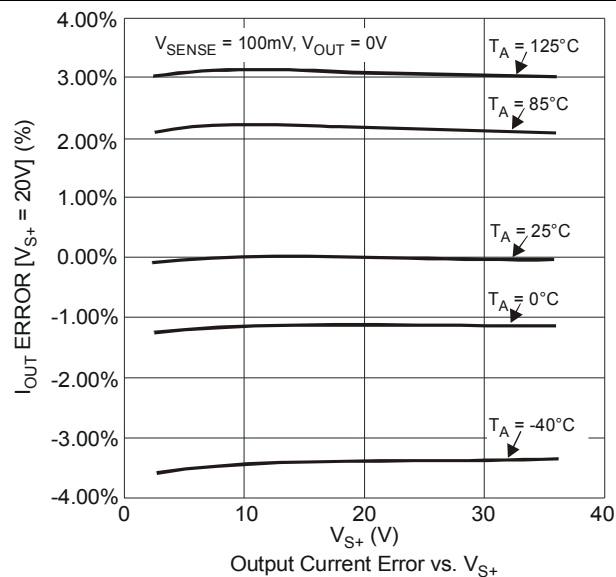


Differential Gain vs. V_{SENSE}

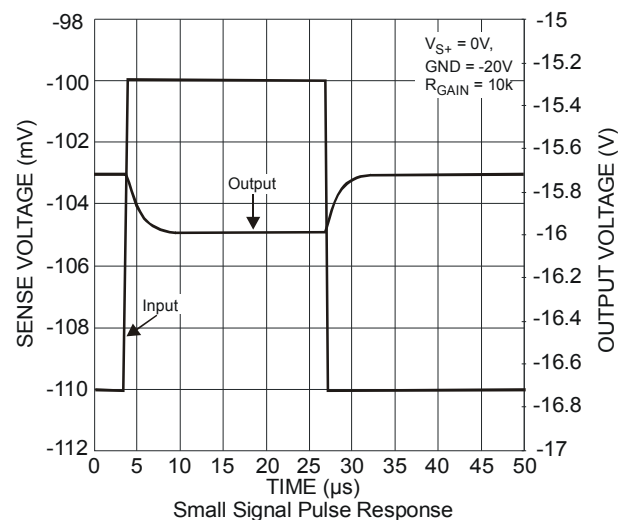
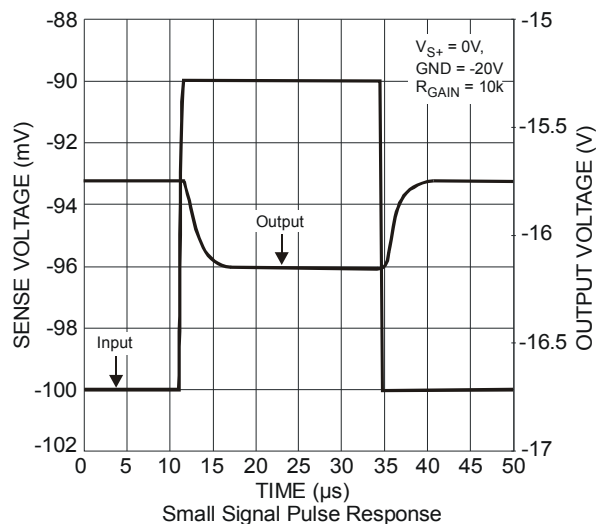
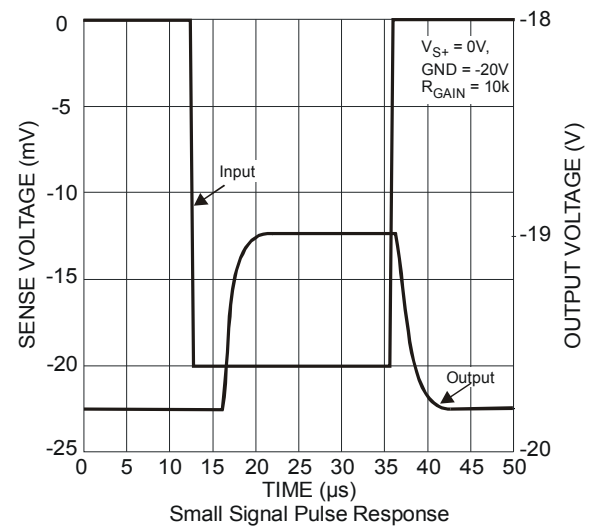
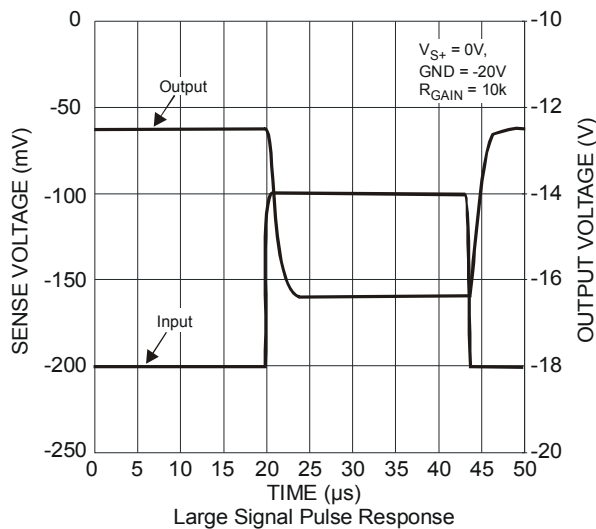
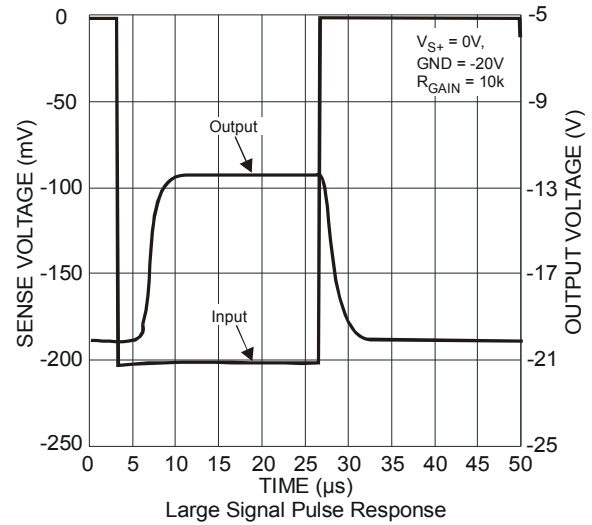
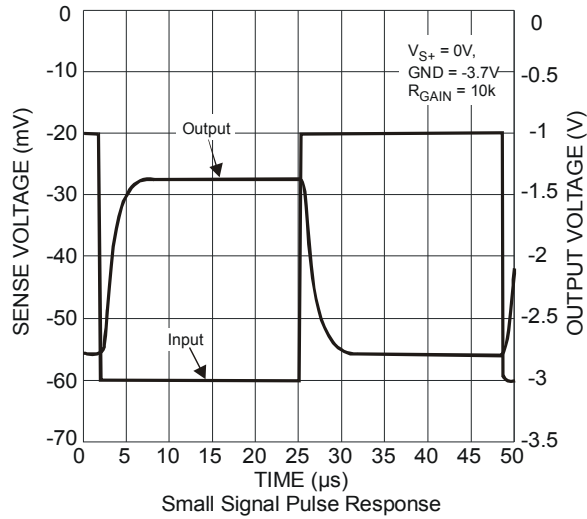


Output Current Error vs. V_{S+}

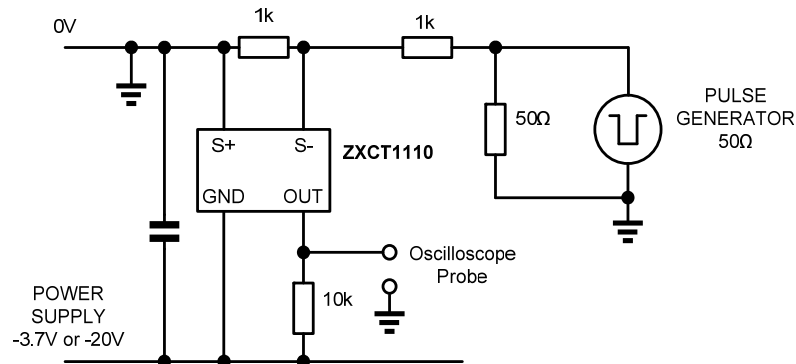
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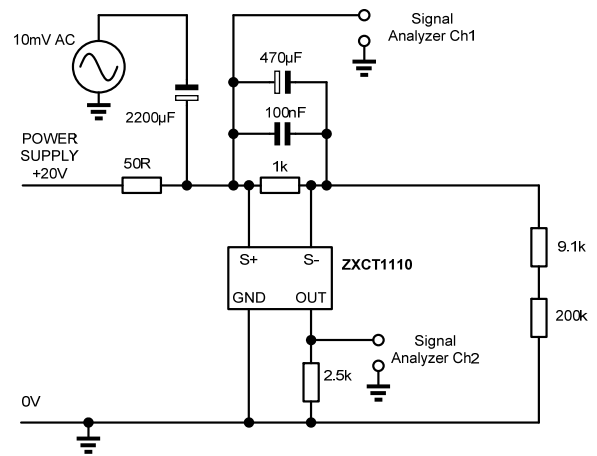
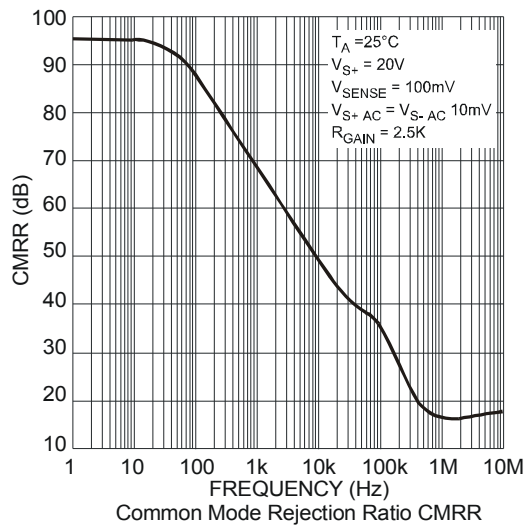
Typical AC Characteristics



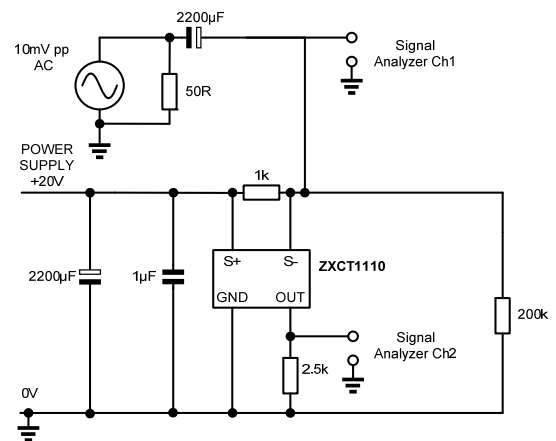
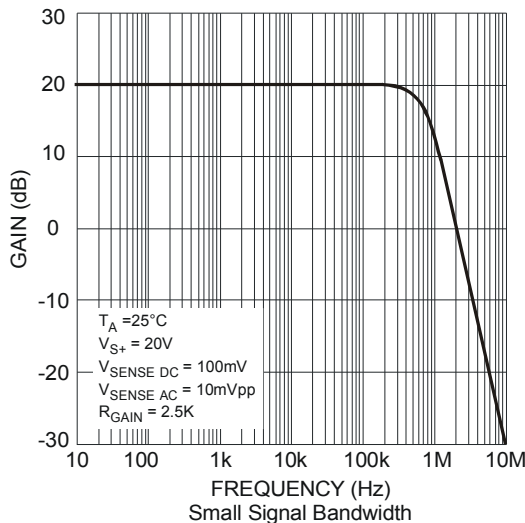
Typical AC Characteristics (cont.)



Test Circuit for Pulse Response



Test circuit for CMRR



Test circuit for Small signal bandwidth

Application Information

Description

The current monitor ICs ZXCT1107Q, ZXCT1109Q and ZXCT1110Q all use a similar application circuit topology for high-side current sensing, with small differences. The ZXCT1110Q has a separate ground pin whereas the ZXCT1107Q and ZXCT1109Q do not. The use of ZXCT1110Q allows reduction of the absolute current measurement error in some applications by providing a reduced output offset current. The ZXCT1107Q provides a mirror image pin assignment of the ZXCT1109Q to ease PCB layout in very small equipment designs.

The basic application circuit for each device is shown in Figure 1. Two external resistors are required. The resistor R_{SENSE} is connected in the path of the current to be monitored. The resistor R_{GAIN} converts the device output current to a voltage for convenient processing by a further device, such as a comparator, amplifier or analog-to-digital converter within a microcontroller system.

The current monitor output current is defined through the nominal transconductance of 4mA/V.

$$I_{OUT} = 0.004 * V_{SENSE} \quad \text{Amp} \quad \text{Equation 1}$$

Then the resistors determine the output voltage as described below.

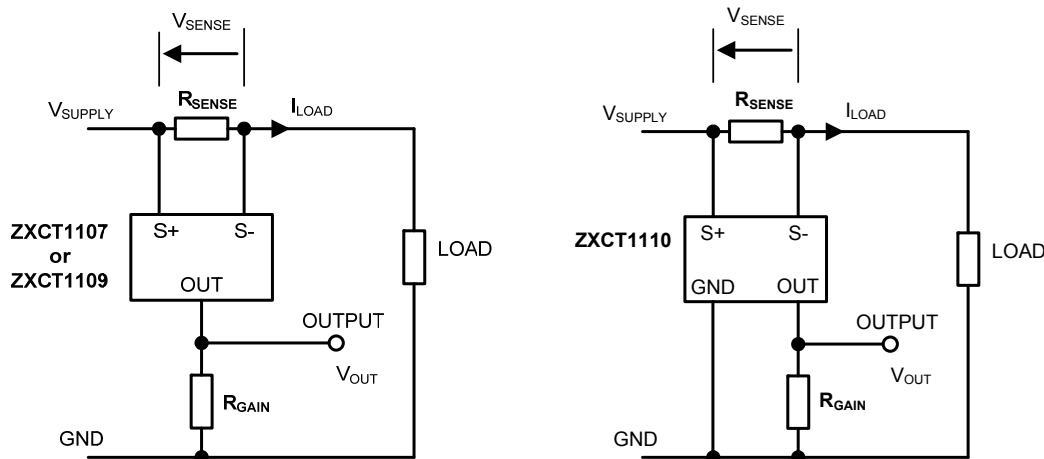


Figure 1 Basic Application Circuit

Calculation of Resistor Values

In order to select R_{SENSE} , a choice of sense voltage is required. This often involves a compromise between power efficiency and accuracy for the given temperature range. The resistor must be small enough to avoid excessive volt drop between the power supply and the load.

However, the resistor must be large enough to avoid excessive current measurement error, particularly random errors.

In a typical application, a digital system, perhaps a microcontroller, is set up to monitor the current. At a certain threshold current level, I_{LOADT} , the system is required to disconnect the load or report a fault. At this current level, the current measurement error must be limited to a known value. The total percentage error comprises the inherent error in the ZXCT1107Q/1109Q/1110Q device and the tolerances of the two resistors R_{SENSE} and R_{GAIN} . The ZXCT1107Q/09Q absolute error is shown in Table 1 with error limits drawn from the table of Electrical Characteristics above. This error varies with V_{SENSE} .

The ZXCT1110Q is highly linear and has a transconductance of 4mA/V $\pm 1.8\%$ and an output of 40 μ A $\pm 4\mu$ A at $V_{SENSE}=10$ mV. The output current can therefore be calculated (Standard International Units) as:

$$I_{OUT} = (40 \pm 4) * 10^{-6} + (V_{SENSE} - 0.01) * (0.004 \pm 0.000072) \text{ A}$$

The worst-case current error is then

$$I_{OUTERROR} = I_{OUT} - I_{OUTIDEAL} = \pm \{ 4 * 10^{-6} + (V_{SENSE} - 0.01) * 0.000072 \} \text{ A}$$

$$\text{or } I_{OUTERROR} = \pm \{ 3.28 * 10^{-6} + V_{SENSE} * 0.000072 \} \text{ A}$$

The percentage error, ϵ , is

$$\epsilon = \frac{I_{OUTERROR}}{I_{OUTIDEAL}} * 100\% = \frac{I_{OUTERROR}}{V_{SENSE} * 0.004} * 100\%$$

Application Information (cont.)

Calculation of Resistor Values (cont.)

$$\text{Then } \varepsilon = \pm \frac{3.28 \times 10^{-4} + V_{\text{SENSE}} \cdot 0.0072}{V_{\text{SENSE}} \cdot 0.004} \%$$

$$\text{or } \varepsilon = \pm \frac{0.082}{V_{\text{SENSE}}} \pm 1.8\% \quad \text{Equation 2}$$

This shows that the error is reduced with increasing V_{SENSE} . Then the minimum V_{SENSE} required to give the error $\pm \varepsilon$ % is

$$V_{\text{SENSE}} = \frac{0.082}{\varepsilon - 1.8} \text{ V} \quad [\text{only for ZXCT1110}] \quad \text{Equation 3}$$

In the application, the effect of the external resistor tolerances must also be taken into account.

Table 1: ZXCT1107Q/ZXCT1109Q Error

V_{SENSE}	Mean I_{OUT}	Error Band	Percent Error
10mV	45.5 μA	$\pm 8.5\mu\text{A}$	$\pm 18.7\%$
30mV	124.5 μA	$\pm 9.5\mu\text{A}$	$\pm 7.6\%$
100mV	408 μA	$\pm 14\mu\text{A}$	$\pm 3.4\%$
200mV	809.5 μA	$\pm 22.5\mu\text{A}$	$\pm 2.8\%$

Table 2: ZXCT1110Q Error

V_{SENSE}	Mean I_{OUT}	Error Band	Percent Error
10mV	40 μA	$\pm 4\mu\text{A}$	$\pm 10\%$
30mV	120 μA	$\pm 5.44\mu\text{A}$	$\pm 4.53\%$
100mV	400 μA	$\pm 10.48\mu\text{A}$	$\pm 2.62\%$
200mV	800 μA	$\pm 17.68\mu\text{A}$	$\pm 2.21\%$

Note that in order to avoid additional error, V_{SUPPLY} must be at least 1.5V greater than V_{OUT} . This margin depends on the value of V_{SENSE} . For $V_{\text{SENSE}} = 100\text{mV}$, this margin can be reduced to 1.0V.

R_{SENSE} is chosen based on the threshold V_{SENSET} of sense voltage selected as above, for the chosen threshold load current, I_{LOADT} .

$$R_{\text{SENSE}} = \frac{V_{\text{SENSET}}}{I_{\text{LOADT}}} \quad \text{Equation 4}$$

R_{GAIN} is chosen to give the required threshold output voltage, V_{OUTT} . From Equation 1, the threshold output current is

$$I_{\text{OUTT}} = 0.004 \cdot V_{\text{SENSET}} \quad \text{Equation 1a}$$

$$R_{\text{GAIN}} = \frac{V_{\text{OUTT}}}{I_{\text{OUTT}}}$$

$$\text{or } R_{\text{GAIN}} = \frac{V_{\text{OUTT}}}{0.004 \cdot V_{\text{SENSET}}} \quad \text{Equation 5}$$

Application Information (cont.)

Design Example

A current monitor is required to provide an output voltage of $2V \pm 6\%$ when the load current of a 12V supply is 2A. Resistors are available with 1% tolerance.

The sense and output resistors will contribute a total possible error of 2%, so the ZXCT11xx error must not exceed 4%. From Table 1, the error of ZXCT1107/1109 will be 3.4% if the sense voltage is 100mV. This satisfies the total error requirement of 6%.

Equation 4 gives

$$R_{\text{SENSE}} = 100\text{mV}/2\text{A} = 50\text{m}\Omega$$

Equation 5 gives

$$R_{\text{GAIN}} = \frac{2}{0.004 * 0.1} = 5\text{k}\Omega$$

A suitable preferred value is 5.1kΩ. This gives a systematic error of +2%, which can be compensated elsewhere in the system if desired. Figure 2 shows the resulting schematic.

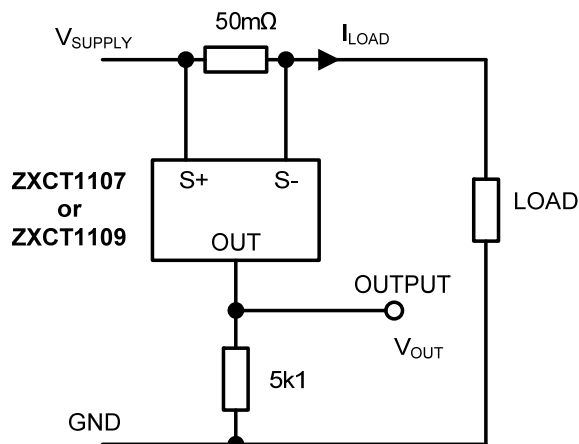


Figure 2. Example Circuit using ZXCT1107/ZXCT1109

Alternatively, the ZXCT1110 may be used with a reduced sense voltage if desired. From Equation 3, with $\varepsilon=4\%$,

$$V_{\text{SENSE}} = \frac{0.082}{\varepsilon - 1.8} = 0.082/(4 - 1.8) = 37.2\text{mV}$$

Equation 4 gives

$$R_{\text{SENSE}} = 37.2\text{mV}/2\text{A} = 18.6\text{m}\Omega$$

A suitable preferred value is 20mΩ. Then the sense voltage will be $2\text{A} * 20\text{m}\Omega = 40\text{mV}$.

Equation 5 gives

$$R_{\text{GAIN}} = \frac{2}{0.004 * 0.04} = 12500\Omega$$

A suitable preferred value is 13kΩ.

Application Information (cont.)

The alternative example circuit using the ZXCT1110 with a sense voltage of 40mV is shown in Figure 3, giving the required overall accuracy of $\pm 6\%$.

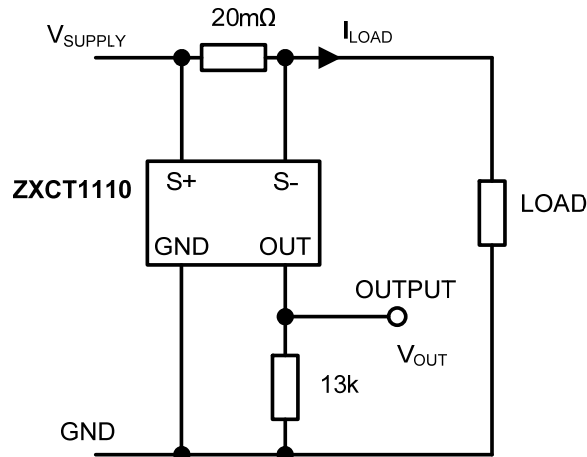


Figure 3 Example Circuit using ZXCT1110

Protection Against Load Short Circuit

In the event of a load short circuit or overload, a large proportion of the supply voltage may appear between the sense terminals. The supply may be current limited, but there is normally a large reservoir capacitor which can deliver enough energy to damage the ZXCT11xx before the supply voltage falls to a safe level due to current limit activation.

The ZXCT11xx is rated for a maximum sense voltage of +0.8V, but is safe if the input current is limited to $\pm 8.5\text{mA}$. In Figure 4, the resistor R_{PROT} limits the current and therefore protects the current monitor device against load short circuit without introducing significant current measurement error.

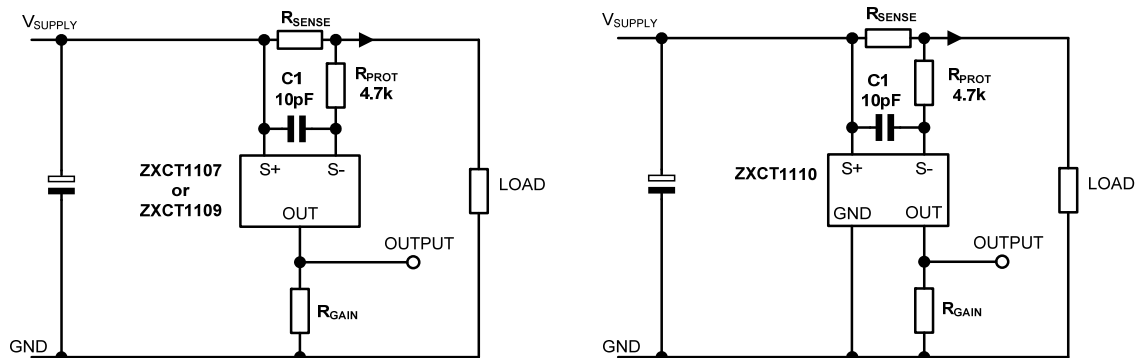


Figure 4 Protection and EMC Filtering

EMC Susceptibility

In many applications, the circuit is required to operate in the presence of RF radiation. This radiation is coupled into the circuit via the supply or load connections and will often cause significant bias shift due to rectification within the integrated circuit. The bias shift may lead to erroneous operation of the integrated circuit. In Figure 4, a low pass filter, R_{PROT} and $C1$, provides significant attenuation in the VHF and UHF regions. The value of $C1$ is recommended to be in the range of 10pF to 47pF, leadless ceramic type.

ZXCT1107Q, ZXCT1109Q, ZXCT1110Q
Application Information (cont.)

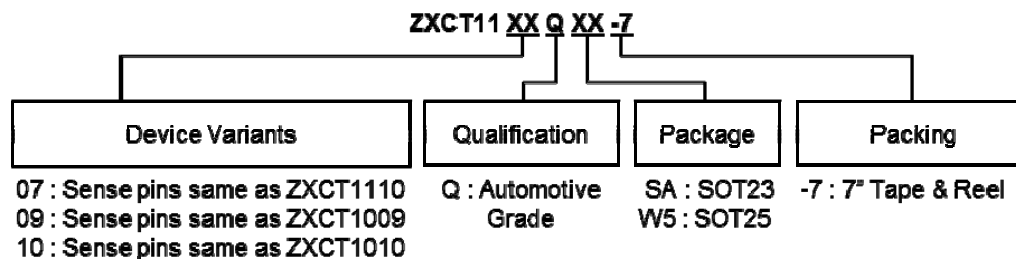
PCB Layout Considerations

It is necessary to take into account a small volt drop in the PCB copper and solder joints to R_{SENSE} . Such volt drops can create noticeable error at currents of the order of 1A and greater. The PCB must be designed to provide the maximum possible copper carrying current via the sense resistor, and the traces from the S+ and S- pins must be connected only to the resistor pads. In this way, the effect of the copper is minimized.

If used, the capacitor C1 should be placed very close the input pins S+ and S-.

Application: High Voltage Current Monitor

The schematic of Figure 5 shows how the current monitor may be used in a high voltage application such as an Electric Vehicle system. In this configuration the 500V PNP transistor Q1 drops a large proportion of the supply voltage. Diode D2 limits the differential input voltage to a safe level under overload conditions. The voltage across U1 is limited by the zener diode D1. The output current of U1 is limited to about 3mA by the D2. As R_{GAIN} is typically a few k Ω , the output voltage is therefore limited to a few volts. An additional output clamping diode may be required for higher gains. R1 is chosen to give a suitable bias current in D1 for the given supply voltage range.

Figure 5 400V Current Monitor
Ordering Information


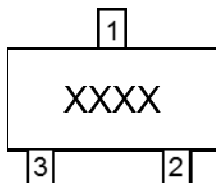
Order Reference	Package (Note 11)	Package Code	Identification Code	Reel Size	Quantity Per Reel	Tape Width (mm)	Qualification Grade (Note 12)
ZXCT1107QSA-7	SOT23	SA	1107	7", 180mm	3000	8	Automotive Grade
ZXCT1109QSA-7	SOT23	SA	1109	7", 180mm	3000	8	Automotive Grade
ZXCT1110QW5-7	SOT25	W5	1110	7", 180mm	3000	8	Automotive Grade

Notes: 11. Pad layout as shown on Diodes Inc. suggested pad layout document AP02001, which can be found on our website at <http://www.diodes.com/datasheets/ap02001.pdf>

12. ZXCT1107Q/09Q/10Q have been qualified to AEC-Q100 grade 1 and is classified as "Automotive Grade" which supports PPAP documentation. See ZXCT1107/09/10 datasheet for commercial qualified versions.

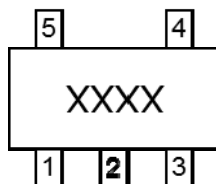
Marking Information

SOT 23



XXXX : Identification code

SOT25

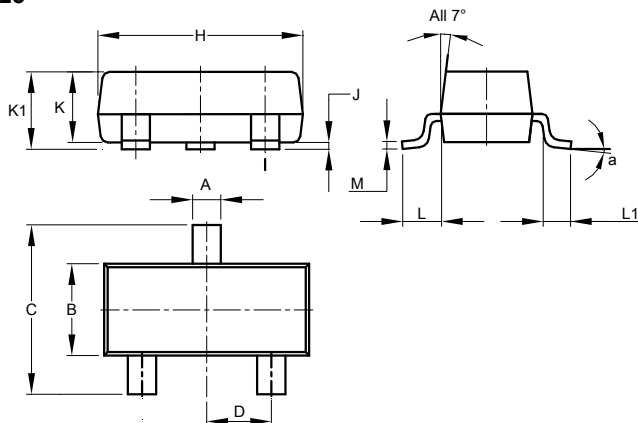


XXXX : Identification code

Package Outline Dimensions

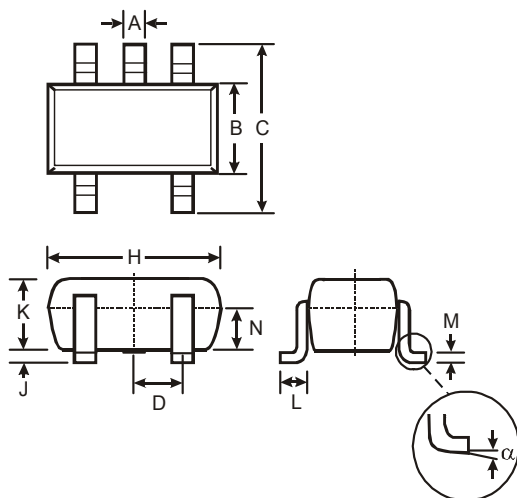
Please see AP02002 at <http://www.diodes.com/datasheets/ap02002.pdf> for latest version.

SOT23



SOT23			
Dim	Min	Max	Typ
A	0.37	0.51	0.40
B	1.20	1.40	1.30
C	2.30	2.50	2.40
D	0.89	1.03	0.915
F	0.45	0.60	0.535
G	1.78	2.05	1.83
H	2.80	3.00	2.90
J	0.013	0.10	0.05
K	0.890	1.00	0.975
K1	0.903	1.10	1.025
L	0.45	0.61	0.55
L1	0.25	0.55	0.40
M	0.085	0.150	0.110
a	8°		
All Dimensions in mm			

SOT25

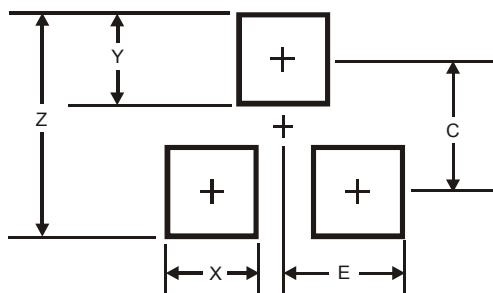


SOT25			
Dim	Min	Max	Typ
A	0.35	0.50	0.38
B	1.50	1.70	1.60
C	2.70	3.00	2.80
D	—	—	0.95
H	2.90	3.10	3.00
J	0.013	0.10	0.05
K	1.00	1.30	1.10
L	0.35	0.55	0.40
M	0.10	0.20	0.15
N	0.70	0.80	0.75
α	0°	8°	—
All Dimensions in mm			

Suggested Pad Layout

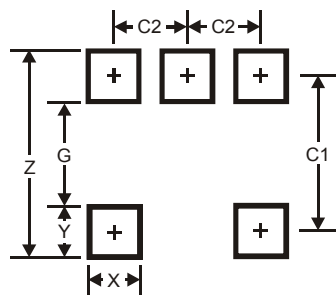
Please see AP02001 at <http://www.diodes.com/datasheets/ap02001.pdf> for the latest version.

SOT23



Dimensions	Value (in mm)
Z	2.9
X	0.8
Y	0.9
C	2.0
E	1.35

SOT25



Dimensions	Value (in mm)
Z	3.20
G	1.60
X	0.55
Y	0.80
C1	2.40
C2	0.95

ZXCT1107Q, ZXCT1109Q, ZXCT1110Q**IMPORTANT NOTICE**

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